

2 MEG VRAM

256K x 8 DRAM WITH 512 x 8 SAM

FEATURES

- Industry standard pinout, timing, and functions
- High-performance, CMOS silicon-gate process
- Single +5V ±10% power supply
- Inputs and outputs are fully TTL compatible
- Refresh modes: RAS ONLY, CAS-BEFORE-RAS, and HIDDEN
- 512-cycle refresh within 8ms
- FAST PAGE MODE
- Dual port organization: 256K x 8 DRAM port
512 x 8 SAM port
- No refresh required for serial access memory
- Low power: 10mW standby; 300mW active, typical
- Fast access times - 70ns random, 22ns serial

SPECIAL FUNCTIONS

- NONPERSISTENT MASKED WRITE
- BLOCK WRITE
- SPLIT READ TRANSFER

OPTIONS

- Timing [DRAM, SAM (Cycle/ Access)]

70ns, 25/22ns	- 7
80ns, 30/25ns	- 8
100ns, 30/27ns	-10
- Packages

Plastic SOJ (400 mil)	DJ
Plastic TSOP (400 mil)	TG
Plastic TSOP (400 mil) reverse pinout	RG

MARKING

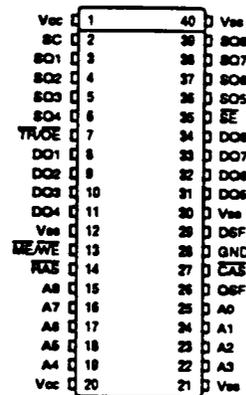
GENERAL DESCRIPTION

The MT42C8255 is a high speed, dual port CMOS dynamic random access memory or video RAM (VRAM) containing 2,097,152 bits. These bits may be accessed by an 8-bit wide DRAM port or by a 512 x 8 bit serial access memory (SAM) port. Data may be transferred from the DRAM to the SAM.

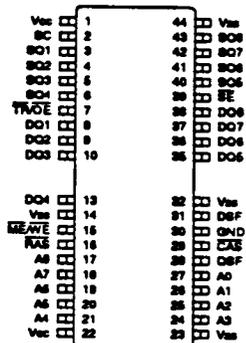
The DRAM portion of the VRAM is functionally identical to the MT4C4256 (256K x 4-bit DRAM), with the addition of MASKED WRITE and BLOCK WRITE. Eight 512-bit data registers make up the serial access memory portion of the VRAM. Data I/O and internal data transfer

PIN ASSIGNMENT (Top View)

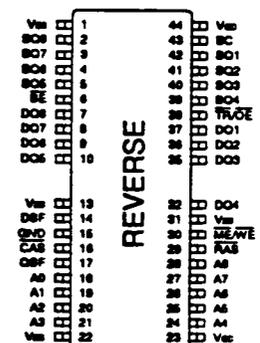
40-Pin SOJ



40/44-Pin TSOP



40/44-Pin TSOP*



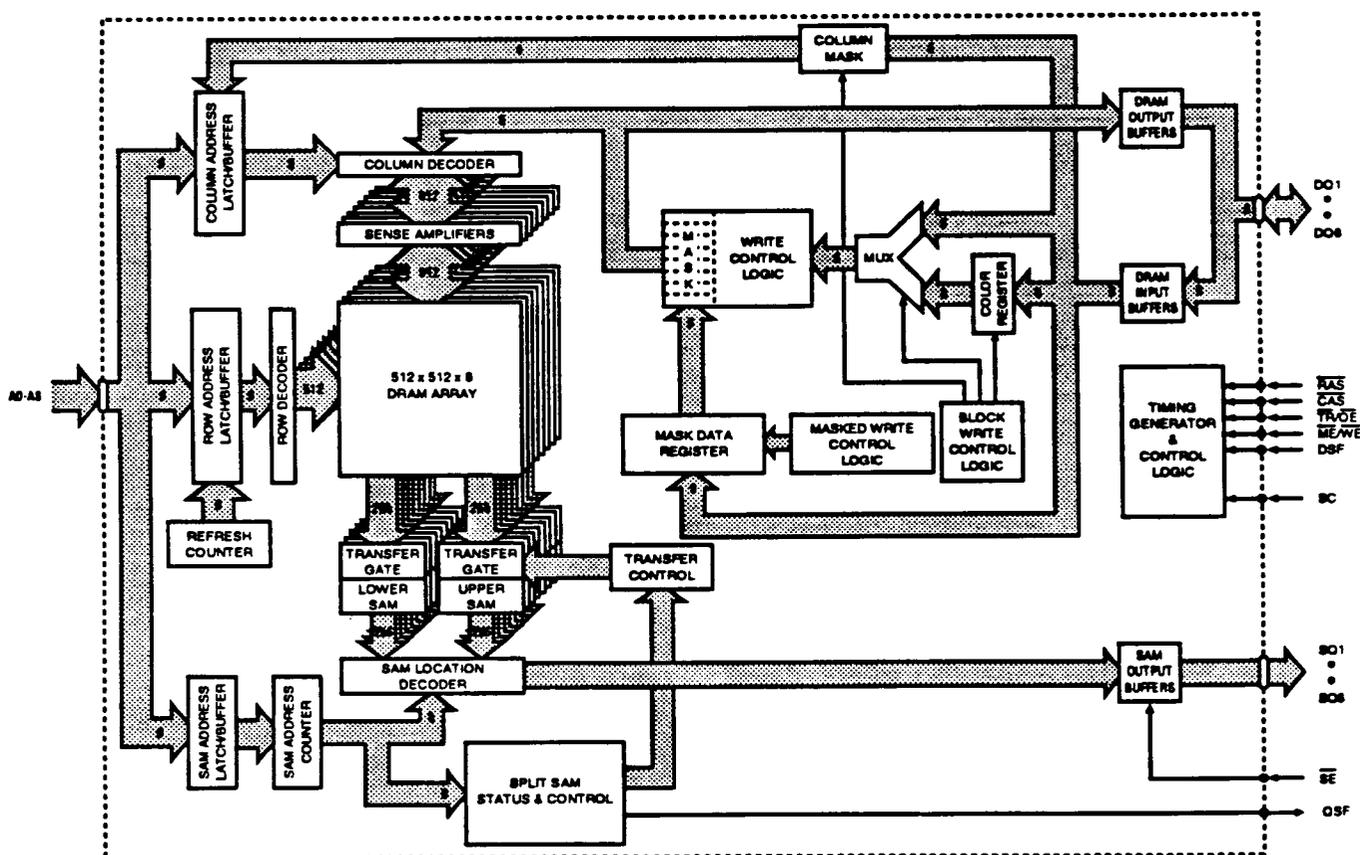
* Consult factory for availability

are accomplished using three separate data paths: the 8-bit random access I/O port, the eight internal 512-bit wide paths between the DRAM and the SAM, and the 8-bit serial output port for the SAM. The rest of the circuitry consists of the control, timing, and address decoding logic.

Each port may be operated asynchronously and independently of the other except when data is being transferred internally. As with all DRAMs, the VRAM must be refreshed to maintain data. The refresh cycles must be timed so that all 512 combinations of RAS addresses are executed at least every 8ms (regardless of sequence). Micron recommends evenly spaced refresh cycles for maximum data integrity. An internal transfer between the DRAM and the SAM counts as a refresh cycle. The SAM portion of the VRAM is fully static and does not require any refresh.

The operation and control of the MT42C8255 are optimized for high performance graphics and communication designs. The dual port architecture is well suited to buffering the sequential data types used in raster graphics display, serial/parallel networking and data communications. Special features such as SPLIT READ TRANSFER and BLOCK WRITE allow further enhancements to system performance.

FUNCTIONAL BLOCK DIAGRAM



PIN DESCRIPTIONS

SOJ PIN NUMBERS	TSOP(TG) PIN NUMBERS	SYMBOL	TYPE	DESCRIPTION
2	2	SC	Input	Serial Clock: Clock input to the serial address counter for the SAM registers.
7	7	TR/OE	Input	Transfer Enable: Enables an internal TRANSFER operation at \overline{RAS} (H \rightarrow L), or Output Enable: Enables the DRAM output buffers when taken LOW after \overline{RAS} goes LOW (\overline{CAS} must also be LOW), otherwise the output buffers are in a High-Z state.
13	15	$\overline{ME/WE}$	Input	Mask Enable: If $\overline{ME/WE}$ is LOW at the falling edge of \overline{RAS} , a MASKED WRITE cycle is performed, or Write Enable: $\overline{ME/WE}$ is also used to select a READ ($\overline{ME/WE} = H$) or WRITE ($\overline{ME/WE} = L$) cycle when accessing the DRAM and READ TRANSFER ($\overline{ME/WE} = H$) to the SAM.
35	39	\overline{SE}	Input	Serial Port Enable: \overline{SE} enables the serial output buffers and allows a serial READ operation to occur, otherwise the output buffers are in a High-Z state. The SAM address count will be incremented by the rising edge of SC when \overline{SE} is inactive (HIGH).
29	31	DSF	Input	Special Function Select: DSF is used to indicate which special functions (BLOCK WRITE, MASKED WRITE, SPLIT TRANSFER, etc.) are used for a particular access cycle (see Truth Table).
14	16	\overline{RAS}	Input	Row Address Strobe: \overline{RAS} is used to clock-in the 9 row-address bits and strobe the $\overline{ME/WE}$, TR/OE, DSF, \overline{SE} , \overline{CAS} and DQ inputs. It also acts as the master chip enable, and must fall for initiation of any DRAM or TRANSFER cycle.
27	29	\overline{CAS}	Input	Column Address Strobe: \overline{CAS} is used to clock-in the 9 column-address bits and as a strobe for the DSF input (BLOCK WRITE only).
25, 24, 23 22, 19, 18 17, 16, 15	27, 26, 25 24, 21, 20 19, 18, 17	A0 to A8	Input	Address Inputs: For the DRAM operation, these inputs are multiplexed and clocked by \overline{RAS} and \overline{CAS} to select one 8-bit word out of the 262,144 available. During TRANSFER operations, A0 to A8 indicate the DRAM row being accessed (when \overline{RAS} goes LOW) and A0-A8 indicate the SAM start address (when \overline{CAS} goes LOW). A8 = "don't care" for the start address during SPLIT READ TRANSFER.
8, 9, 10, 11 31, 32, 33 34	8, 9, 10, 13 35, 36, 37 38	DQ1-DQ8	Input/ Output	DRAM Data I/O: Data input/output for DRAM access cycles: These pins also act as inputs for Color Register load cycles, Bit Masks and Column Mask for BLOCK WRITE.
3, 4, 5, 6, 36 37, 38, 39	3, 4, 5, 6, 40 41, 42, 43	SQ1-SQ8	Output	Serial Data Out: Output or High-Z.
26	28	QSF	Output	Split SAM Status: QSF indicates which half of the SAM is being accessed. LOW if address is 0-255, HIGH if address is 256-511.
28	30	GND	-	No Connect/GND: This pin must be tied to ground to allow for upward functional compatibility with future VRAM feature sets.
1, 20	1, 22	Vcc	Supply	Power Supply: +5V \pm 10%
12, 21 30, 40	14, 23 32, 44	Vss	Supply	Ground

FUNCTIONAL DESCRIPTION

The MT42C8255 can be divided into three functional blocks (see Figure 1): the DRAM, the transfer circuitry, and the SAM. All of the operations described below are shown in the AC Timing Diagrams section of this data sheet and summarized in the Truth Table.

Note: For dual-function pins, the function not being discussed will be surrounded by parentheses. For example, the TR/OE pin will be shown as TR/(OE) in references to transfer operations.

DRAM OPERATION

DRAM REFRESH

Like any DRAM-based memory, the MT42C8255 VRAM must be refreshed to retain data. All 512 row address combinations must be accessed within 8ms. The MT42C8255 supports CAS-BEFORE-RAS, RAS-ONLY and HIDDEN types of refresh cycles.

For the CAS-BEFORE-RAS REFRESH (CBR) cycle, the row addresses are generated and stored in an internal address counter. The user need not supply any address data, and simply must perform 512 CAS-BEFORE-RAS cycles within the 8ms time period.

The refresh address must be generated externally and applied to A0-A8 inputs for RAS-ONLY REFRESH cycles. The DQ pins remain in a High-Z state for both the RAS-ONLY and CAS-BEFORE-RAS cycles.

HIDDEN REFRESH cycles are performed by toggling RAS (and keeping CAS LOW) after a READ or WRITE cycle. This performs CAS-BEFORE-RAS cycles but does not disturb the DQ lines.

Any DRAM READ, WRITE, or TRANSFER cycle also refreshes the DRAM row being accessed. The SAM portion of the MT42C8255 is fully static and does not require any refreshing.

DRAM ACCESS CYCLES (RW)

The DRAM portion of the VRAM is nearly identical to standard 256K x 4 DRAMs. However, because several of the DRAM control pins are used for additional functions on this part, several conditions that were undefined or in "don't care" states for the DRAM are specified for the VRAM. These conditions are highlighted in the following discussion. In addition, the VRAM has special functions that can be used when writing to the DRAM.

The 18 address bits that are used to select an 8-bit word from the 262,144 available are latched into the chip using

the A0-A8, RAS and CAS inputs. First, the nine row-address bits are set up on the address inputs and clocked into the part when RAS transitions from HIGH-to-LOW. Next, the 9 column address bits are set up on the address inputs and clocked-in when CAS goes from HIGH-to-LOW.

Note: RAS also acts as a "master" chip enable for the VRAM. If RAS is inactive, HIGH, all other DRAM control pins (CAS, TR/OE, ME/WE, etc.) are "don't care" and may change state without effect. No DRAM or TRANSFER cycles will be initiated without RAS falling.

For single port DRAMS, the OE pin is a "don't care" when RAS goes LOW. However, for the VRAM, when RAS goes LOW, TR/(OE) selects between DRAM access or TRANSFER cycles. TR/(OE) must be HIGH at the RAS HIGH-to-LOW transition for all DRAM operations (except CAS-BEFORE-RAS).

A DRAM READ operation is performed if (ME)/WE is HIGH when CAS goes LOW and remains HIGH until CAS goes HIGH. The data from the memory cells selected will appear at the DQ1-DQ8 port. The (TR)/OE input must transition from HIGH-to-LOW some time after RAS falls to enable the DRAM output port.

For single port DRAMS, WE is a "don't care" when RAS goes LOW. For the VRAM, ME/WE performs two functions; write mask enable and data write enable. ME/(WE) is used, when RAS goes LOW, to select between a MASKED WRITE cycle and a normal WRITE cycle. If ME/(WE) is LOW at the RAS HIGH-to-LOW transition, a MASKED WRITE operation is selected. For any non-masked DRAM access cycle (READ or WRITE), ME/(WE) must be HIGH at the RAS HIGH-to-LOW transition. If (ME)/WE is LOW before CAS goes LOW, a DRAM EARLY-WRITE operation is performed and the data present on the DQ1-DQ8 data port will be written into the selected memory cells. If (ME)/WE goes LOW after CAS goes LOW, a DRAM LATE-WRITE operation is performed (refer to the AC timing diagrams).

The VRAM can perform all the normal DRAM cycles including READ, EARLY-WRITE, LATE-WRITE, READ-MODIFY-WRITE, FAST-PAGE-MODE READ, FAST-PAGE-MODE WRITE (Late or Early), and FAST-PAGE-MODE READ-MODIFY-WRITE. Refer to the AC timing parameters and diagrams in the data sheet for more details on these operations.

MASKED WRITE (RWM)

The MASKED WRITE feature eliminates the need to do a READ-MODIFY-WRITE cycle when changing individual bits within the 8-bit word. When $\overline{ME}/\overline{WE}$ and DSF are LOW at the RAS HIGH-to-LOW transition, a MASKED WRITE is performed.

The MT42C8255 supports only the nonpersistent mode of MASKED WRITE. In this mode, mask data must be entered with every RAS falling edge. The data (mask data) present on the DQ1-DQ8 inputs will be written into the mask data register (see Figure 1). The mask data acts as an individual write enable for each of the eight DQ1-DQ8 pins. If a LOW (logic "0") is written to a mask data register bit, the input port for that bit is disabled during the subsequent WRITE operation and no new data will be written to that

DRAM cell location. A HIGH (logic "1") on a mask data register bit enables the input port and allows normal WRITE operation to proceed. Note that \overline{CAS} is still HIGH. When \overline{CAS} goes LOW, the bits present on the DQ1-DQ8 inputs will be either written to the DRAM (if the mask data bit is HIGH) or ignored (if the mask data bit is LOW). The DRAM contents that correspond to masked input bits will not be changed during the WRITE cycle. The mask data register is cleared at the end of every nonpersistent MASKED WRITE.

FAST PAGE MODE can be used with MASKED WRITE to write several column locations in an addressed row. The same mask is used during the entire FAST-PAGE-MODE RAS cycle.

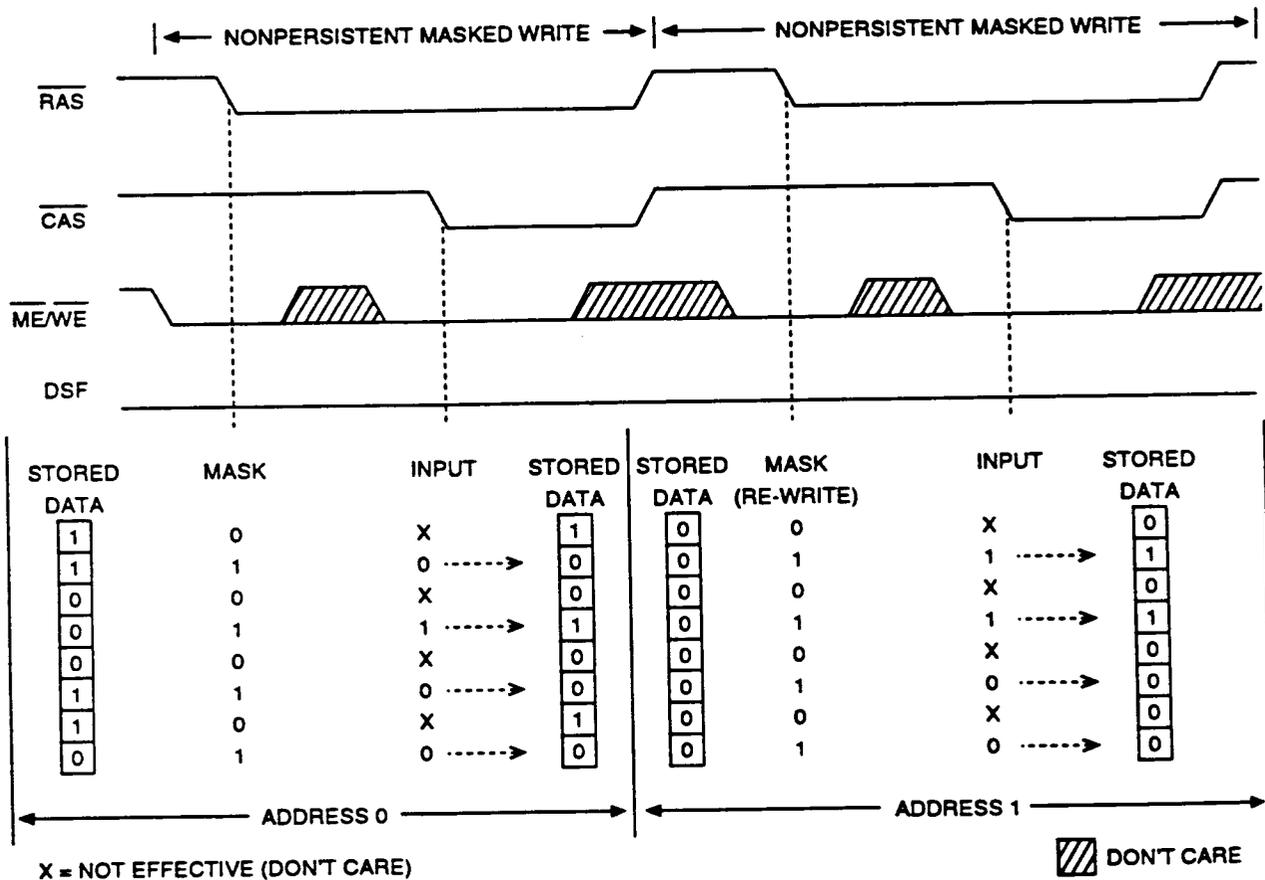


Figure 1
NONPERSISTENT MASKED WRITE EXAMPLE

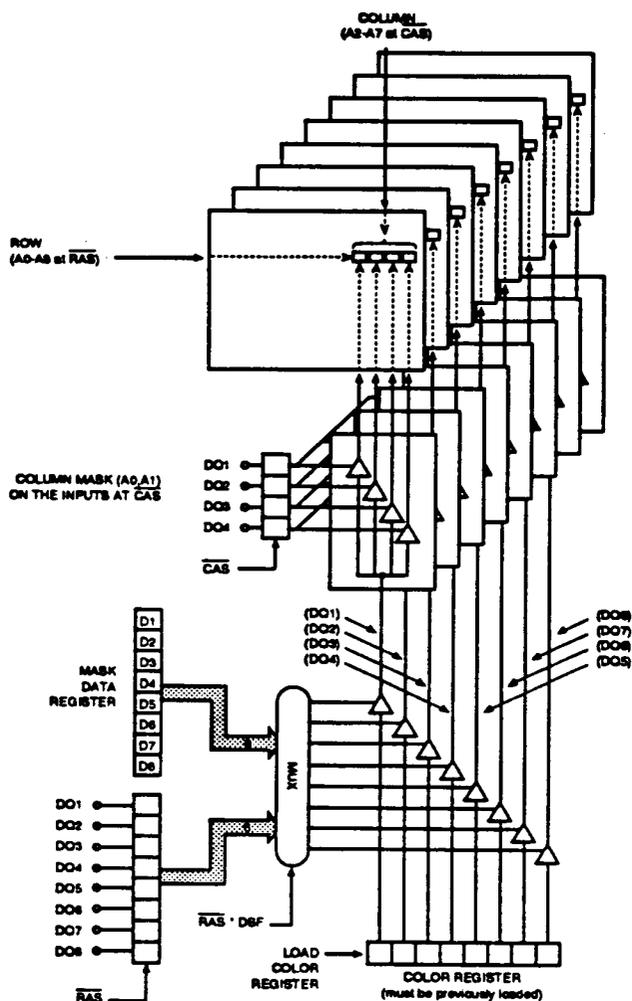


Figure 2
BLOCK WRITE EXAMPLE

BLOCK WRITE (BW)

If DSF is HIGH when $\overline{\text{CAS}}$ goes LOW, the MT42C8255 will perform a BLOCK WRITE cycle instead of a normal WRITE cycle. In BLOCK WRITE cycles, the contents of the color register are directly written to four adjacent column locations (see Figure 2). The color register must be loaded prior to beginning BLOCK WRITE cycles (see LOAD COLOR REGISTER). Each DQ location of the color register is written to the four column locations (or any of the four that are enabled) in the corresponding DQ bit plane.

The row is addressed as in a normal DRAM WRITE cycle. However when $\overline{\text{CAS}}$ goes LOW, only the A2-A8 inputs are used. A2-A8 specify the "block" of four adjacent column locations that will be accessed. The DQ inputs (DQ1, 2, 3, and 4) are then used to determine what combination of the four column locations will be changed. The DQ inputs are "written" at the falling edge of $\overline{\text{CAS}}$ or WE, whichever occurs later (see the WRITE cycle waveforms). The table on this page illustrates how each of the DQ inputs is used to selectively enable or disable individual column locations within the block. The write enable controls are active HIGH; a logic "1" enables the WRITE function and a logic "0" disables the WRITE function.

MASKED BLOCK WRITE (BWM)

The MASKED WRITE functions may be used during BLOCK WRITE cycles. MASKED BLOCK WRITE operates exactly like the normal MASKED WRITE except the mask is now applied to the 8 bit-planes of 4 column locations instead of just one column location.

The combination of $\overline{\text{ME}}/\overline{\text{WE}}$ LOW and DSF LOW when $\overline{\text{RAS}}$ goes LOW initiates a nonpersistent MASKED WRITE cycle. To perform a MASKED BLOCK WRITE, the DSF pin must be HIGH when $\overline{\text{CAS}}$ goes LOW. By using both the column mask input and the MASKED WRITE function of BW, any combination of the eight bit planes may be masked, along with any combination of the four column locations.

INPUTS	COLUMN ADDRESS CONTROLLED	
	A0	A1
DQ1	0	0
DQ2	1	0
DQ3	0	1
DQ4	1	1

LOAD COLOR REGISTER (LCR)

A LOAD COLOR REGISTER cycle is identical to the LOAD MASK REGISTER cycle except DSF is HIGH when $\overline{\text{CAS}}$ goes LOW. The contents of the 8-bit color register are retained until changed by another LOAD COLOR REGISTER cycle (or the part loses power) and are used as data inputs during BLOCK WRITE cycles.

TRANSFER OPERATIONS

TRANSFER operations are initiated when $\overline{\text{TR}}/\overline{\text{OE}}$ is LOW at the falling edge of $\overline{\text{RAS}}$. The state of $\overline{\text{ME}}/\overline{\text{WE}}$ when $\overline{\text{RAS}}$ goes LOW indicates the direction of the TRANSFER (to or from the DRAM), and DSF is used to select between NORMAL TRANSFER and SPLIT TRANSFER cycles. Each of the TRANSFER cycles is described in this section.

READ TRANSFER (RT)

If $\overline{\text{ME}}/\overline{\text{WE}}$ is HIGH and DSF is LOW when $\overline{\text{RAS}}$ goes LOW, a READ TRANSFER cycle is selected. The row address bits indicate which eight 512-bit DRAM row planes are transferred to the eight SAM data register planes. The column address bits indicate the start address (or Tap address) of the serial output cycle from the SAM data registers. $\overline{\text{CAS}}$ must fall for every RT in order to load a valid Tap address. An RT may be accomplished in two ways. If the transfer is to be synchronized with the serial clock, SC, (REAL-TIME READ TRANSFER), $\overline{\text{TR}}/\overline{\text{OE}}$ is taken HIGH after $\overline{\text{CAS}}$ goes LOW. The TRANSFER will be made when $\overline{\text{TR}}/\overline{\text{OE}}$ goes HIGH. If the transfer does not have to be synchronized with SC (READ TRANSFER), $\overline{\text{TR}}/\overline{\text{OE}}$ may go HIGH before $\overline{\text{CAS}}$ goes LOW and the actual data TRANS-

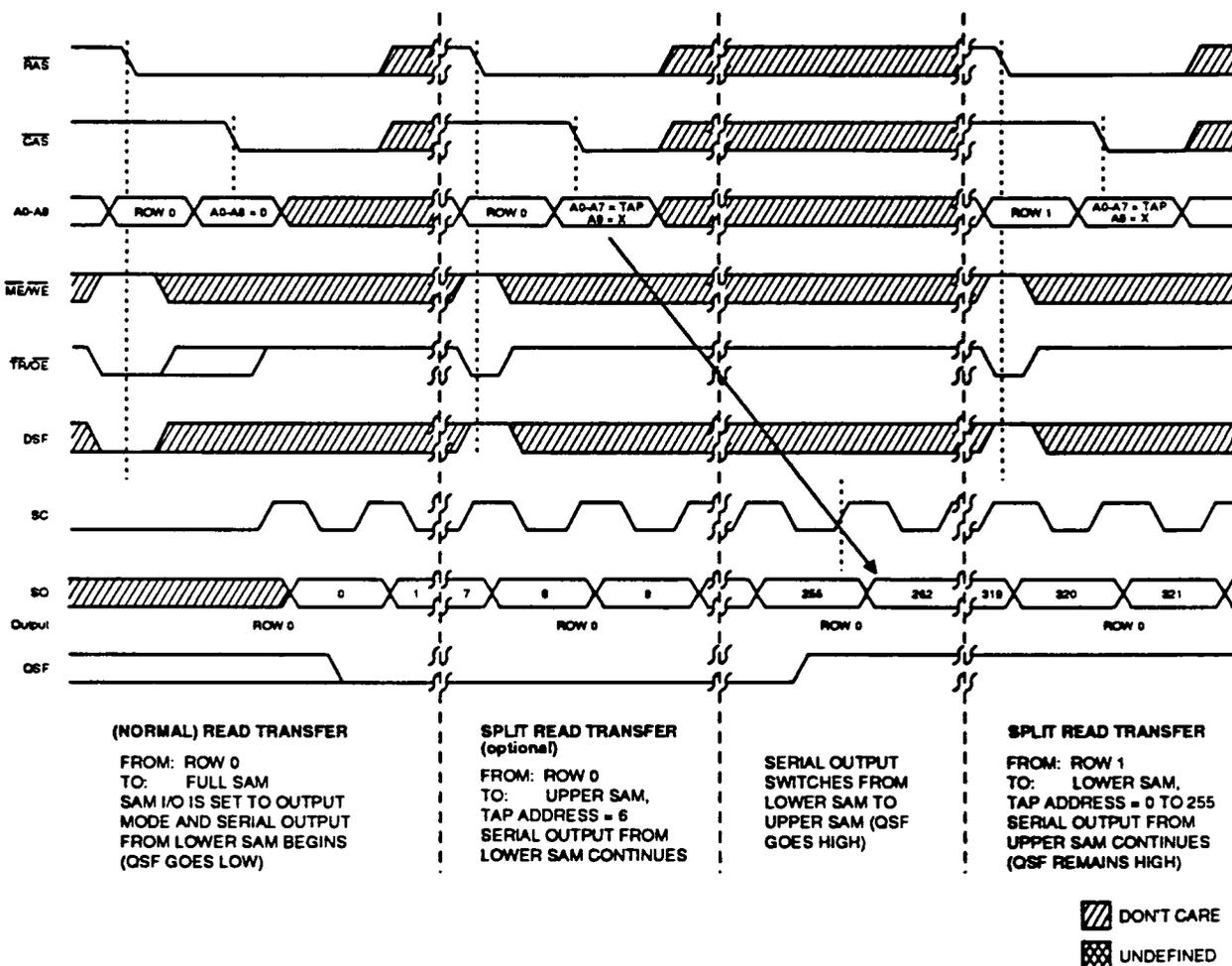


Figure 3
TYPICAL SPLIT-READ-TRANSFER INITIATION SEQUENCE

FER will be timed internally (refer to the AC Timing Diagrams). During the TRANSFER, 4096 bits of DRAM data are written into the SAM data registers and the Tap address is stored in an internal 8-bit register. The split SAM status pin (QSF) will be LOW if the Tap is in the lower half (addresses 0 through 255), and HIGH if it is in the upper half (256 through 511). If \overline{SE} is LOW, the first bits of the new row data will appear at the serial outputs with the first SC clock pulse. \overline{SE} enables the serial outputs and may be either HIGH or LOW during this operation. The SAM address pointer will increment with the SC LOW-to-HIGH transition, regardless of the state of \overline{SE} .

SPLIT READ TRANSFER (SRT)

The SPLIT READ TRANSFER (SRT) cycle eliminates the critical transfer timing required to maintain a continuous serial output data stream. When using normal TRANSFER cycles to do midline reloads, a REAL-TIME READ TRANSFER must be done. The REAL-TIME READ TRANSFER has to occur between the last clock of "old" data and first clock of the "new" data of the SAM port.

When using the SPLIT TRANSFER mode, the SAM is divided into an upper half and a lower half. While data is being serially read from one half of the SAM, new DRAM data may be transferred to the other half. The transfer is not synchronized with the serial clock and may occur at any time while the other half is outputting data.

The $\overline{TR}/(\overline{OE})$ timing is also relaxed for SPLIT TRANSFER cycles. The rising edge of $\overline{TR}/(\overline{OE})$ is not used to complete the TRANSFER cycle and therefore is independent of the falling edge of \overline{CAS} or the rising edge of SC. The transfer timing is generated internally for SPLIT TRANSFER cycles.

A "full" READ TRANSFER cycle must precede any sequence of SRT cycles to provide a reference to which half of the SAM the access will begin (the state of QSF). Then an SRT may be initiated by taking DSF HIGH when \overline{RAS} goes LOW during the TRANSFER cycle. As in nonsplit transfers, the row address is used to specify the DRAM row to be transferred. The column address, A0-A7, is used to input the SAM Tap address. Address pin A8 is a "don't care" when the Tap address is loaded at the HIGH-to-LOW transition of \overline{CAS} . It is internally generated in such a manner that the SPLIT TRANSFER will automatically be to the SAM half not being accessed.

Figure 3 shows a typical SRT initiation sequence. The normal READ TRANSFER is performed first, followed by an SRT of the same row to the upper half of the SAM. The SRT to the upper half is optional, and need only be done if the Tap for the upper half is $\neq 0$. Serial access continues, and when the SAM address counter reaches 255 ("A8" = 0, A0-A7 = 1) the QSF output goes HIGH, and if an SRT was done for the upper half, the new Tap address is loaded for the next half ("A8" = 1, A0-A7 = Tap). Once the serial access has switched to the upper SAM (QSF has gone HIGH), new data may be transferred to the lower SAM. For example, the next step in Figure 3 would be to wait until QSF went LOW (indicating that row-1 data is shifting out of the lower SAM) and execute an SRT of the upper half of row 1 to the upper SAM. If the half boundary is reached before an SRT is done for the next half, the device will leave split mode and the access will start from address 256 if going to the upper half or at 0 if going to the lower half (see Figure 4).

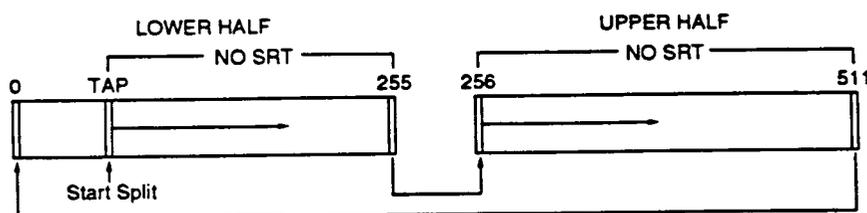


Figure 4
SPLIT SAM TRANSFER

SERIAL OUTPUT

The control inputs for serial output are SC and \overline{SE} . The rising edge of SC increments the serial address counter and provides access to the next SAM location. \overline{SE} enables or disables the serial output buffers.

Serial output of the SAM contents will start at the serial start address that was loaded in the SAM address counter during a READ or SPLIT READ TRANSFER cycle. The SC input increments the address counter and presents the contents of the next SAM location to the 8-bit port. \overline{SE} is used as an output enable during the SAM output operation. The serial address is automatically incremented with every SC LOW-to-HIGH transition, regardless of whether \overline{SE} is HIGH or LOW. The address progresses through the SAM and will wrap around (after count 255 or 511) to the Tap address of the next half for split modes. If an SRT was not performed before the half boundary is reached, the count will progress as illustrated in Figure 4. Address count will wrap around (after count 511) to Tap address 0 if in the "full" SAM modes.

POWER-UP AND INITIALIZATION

When Vcc is initially supplied or when refresh is interrupted for more than 8ms, the MT42C8255 must be initialized.

After Vcc is at specified operating conditions, for 100 μ s minimum, eight \overline{RAS} cycles must be executed to initialize the dynamic memory array. Micron recommends that $\overline{RAS} = \overline{TR}/\overline{OE} \geq V_{IH}$ during power up to ensure that the DRAM I/O pins (DQs) are in a High-Z state. The DRAM array will contain random data.

The SAM portion of the MT42C8255 is completely static in operation and does not require refresh or initialization. The SAM port will power-up with the Output pins (SQs) in High-Z, regardless of the state of \overline{SE} . QSF initializes in the LOW state. The color register will contain random data after power-up.

TRUTH TABLE

CODE	FUNCTION	RAS FALLING EDGE				CAS FALL DSF	A0-A8 ¹		DQ1-DQ8 ²		REGISTER COLOR
		CAS	TR/OE	ME/WE	DSF		RAS	CAS	RAS	CAS/WE ³	
DRAM OPERATIONS											
CBR	CAS-BEFORE-RAS REFRESH	0	1 ⁴	1 ⁴	X	—	X	X	—	X	X
ROR	RAS ONLY REFRESH	1	1	X	X	—	ROW	—	X	—	X
RW	NORMAL DRAM READ OR WRITE	1	1	1	0	0	ROW	COLUMN	X	VALID DATA	X
RWM	MASKED WRITE TO DRAM (NEW MASK)	1	1	0	0	0	ROW	COLUMN	WRITE MASK	VALID DATA	X
BW	BLOCK WRITE TO DRAM	1	1	1	0	1	ROW	COLUMN (A2-A8)	X	COLUMN MASK	USE
BWM	MASKED BLOCK WRITE TO DRAM (NEW MASK)	1	1	0	0	1	ROW	COLUMN (A2-A8)	WRITE MASK	COLUMN MASK	USE
REGISTER OPERATIONS											
LCR	LOAD COLOR REGISTER	1	1	1	1	1	ROW ⁴	X	X	REG DATA	LOAD
TRANSFER OPERATIONS											
RT	READ TRANSFER (DRAM-TO-SAM TRANSFER)	1	0	1	0	X	ROW	TAP ⁵	X	X	X
SRT	SPLIT READ TRANSFER (SPLIT DRAM-TO-SAM TRANSFER)	1	0	1	1	X	ROW	TAP ⁵	X	X	X

- NOTE:**
1. These columns show what must be present on the A0-A8 inputs when RAS falls and when CAS falls.
 2. These columns show what must be present on the DQ1-DQ8 inputs when RAS falls and when CAS falls.
 3. During WRITE (including BLOCK WRITE) cycles, the input data is latched at the falling edge of CAS or ME/WE, whichever is later. Similarly, on READ cycles, the output data is valid after the falling edge of CAS or TR/OE, whichever is later.
 4. The ROW that is addressed will be refreshed, but a ROW address is required.
 5. This is the first SAM address location that the first SC cycle will access. For Split SAM transfers, the Tap will be the first address location accessed of the "new" SAM half after the boundary of the current half is reached (255 for lower half, 511 for upper half).
 6. The MT42C8255 does not require a "1" on these pins, but to ensure compatibility with all 2 Meg VRAM function sets, it is recommended.

ABSOLUTE MAXIMUM RATINGS*

Voltage on Vcc Supply Relative to Vss-1V to +7V
 Operating Temperature, T_A (Ambient)0°C to +70°C
 Storage Temperature (Plastic) -55°C to +150°C
 Power Dissipation 1W
 Short Circuit Output Current 50mA

*Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

RECOMMENDED DC OPERATING CONDITIONS

(0°C ≤ T_A ≤ 70°C)

PARAMETER/CONDITION	SYMBOL	MIN	MAX	UNITS	NOTES
Supply Voltage	V _{CC}	4.5	5.5	V	1
Input High (Logic 1) Voltage, All Inputs	V _{IH}	2.4	V _{CC} +1	V	1
Input Low (Logic 0) Voltage, All Inputs	V _{IL}	-1.0	0.8	V	1

DC ELECTRICAL CHARACTERISTICS

(0°C ≤ T_A ≤ 70°C; V_{CC} = 5V ±10%)

PARAMETER/CONDITION	SYMBOL	MIN	MAX	UNITS	NOTES
INPUT LEAKAGE CURRENT (Any input (0V ≤ V _{IN} ≤ V _{CC}); all other pins not under test = 0V)	I _L	-10	10	μA	
OUTPUT LEAKAGE CURRENT (DQ, SQ disabled, 0V ≤ V _{OUT} ≤ V _{CC})	I _{OZ}	-10	10	μA	
OUTPUT LEVELS					
Output High Voltage (I _{OUT} = -2.5mA)	V _{OH}	2.4		V	
Output Low Voltage (I _{OUT} = 2.5mA)	V _{OL}		0.4	V	1

CAPACITANCE

PARAMETER	SYMBOL	MIN	MAX	UNITS	NOTES
Input Capacitance: A0-A8	C _{I1}		5	pF	2
Input Capacitance: RAS, CAS, ME/WE, TR/OE, SC, SE, DSF	C _{I2}		8	pF	2
Input/Output Capacitance: DQ, SQ	C _{I/O}		9	pF	2
Output Capacitance: QSF	C _O		9	pF	2

CURRENT DRAIN, SAM IN STANDBY

(0°C ≤ T_A ≤ 70°C; V_{CC} = 5V ±10%)

PARAMETER/CONDITION	SYMBOL	MAX			UNITS	NOTES
		-7	-8	-10		
OPERATING CURRENT (\overline{RAS} and \overline{CAS} = Cycling; $t_{RC} = t_{RC}$ (MIN))	Icc1	120	110	100	mA	3, 4 25
OPERATING CURRENT: FAST PAGE MODE ($\overline{RAS} = V_{IL}$; \overline{CAS} = Cycling; $t_{PC} = t_{PC}$ (MIN))	Icc2	110	100	90	mA	3, 4 26
STANDBY CURRENT: TTL INPUT LEVELS Power supply standby current ($\overline{RAS} = \overline{CAS} = V_{IH}$ after 8 \overline{RAS} cycles (MIN))	Icc3	8	8	8	mA	
REFRESH CURRENT: \overline{RAS} -ONLY (\overline{RAS} = Cycling; $\overline{CAS} = V_{IH}$)	Icc4	120	110	100	mA	3, 25
REFRESH CURRENT: \overline{CAS} -BEFORE- \overline{RAS} (\overline{RAS} and \overline{CAS} = Cycling)	Icc5	120	110	100	mA	3, 5
SAM/DRAM DATA TRANSFER	Icc6	130	120	110	mA	3

CURRENT DRAIN, SAM ACTIVE ($t_{SC} = \text{MIN}$)

(0°C ≤ T_A ≤ 70°C; V_{CC} = 5V ±10%)

PARAMETER/CONDITION	SYMBOL	MAX			UNITS	NOTES
		-7	-8	-10		
OPERATING CURRENT (\overline{RAS} and \overline{CAS} = Cycling; $t_{RC} = t_{RC}$ (MIN))	Icc7	160	150	135	mA	3, 4 25
OPERATING CURRENT: FAST PAGE MODE ($\overline{RAS} = V_{IL}$; \overline{CAS} = Cycling; $t_{PC} = t_{PC}$ (MIN))	Icc8	150	140	125	mA	3, 4 26
STANDBY CURRENT: TTL INPUT LEVELS Power supply standby current ($\overline{RAS} = \overline{CAS} = V_{IH}$ after 8 \overline{RAS} cycles (MIN))	Icc9	50	50	45	mA	3, 4
REFRESH CURRENT: \overline{RAS} -ONLY (\overline{RAS} = Cycling; $\overline{CAS} = V_{IH}$)	Icc10	160	150	135	mA	3, 4 25
REFRESH CURRENT: \overline{CAS} -BEFORE- \overline{RAS} (\overline{RAS} and \overline{CAS} = Cycling)	Icc11	160	150	135	mA	3, 4, 5
SAM/DRAM DATA TRANSFER	Icc12	170	160	145	mA	3, 4

DRAM TIMING PARAMETERS

ELECTRICAL CHARACTERISTICS AND RECOMMENDED AC OPERATING CONDITIONS

(Notes: 6, 7, 8, 9, 10, 11, 12, 13) ($0^{\circ}\text{C} \leq T_A \leq +70^{\circ}\text{C}$; $V_{CC} = 5\text{V} \pm 10\%$)

AC CHARACTERISTICS		-7		-8		-10			
PARAMETER	SYM	MIN	MAX	MIN	MAX	MIN	MAX	UNITS	NOTES
Random READ or WRITE cycle time	t_{RC}	120		150		180		ns	
READ-MODIFY-WRITE cycle time	t_{RWC}	165		190		230		ns	
FAST-PAGE-MODE READ or WRITE cycle time	t_{PC}	45		50		55		ns	
FAST-PAGE-MODE READ-MODIFY-WRITE cycle time	t_{PRWC}	90		95		110		ns	
Access time from RAS	t_{RAC}		70		80		100	ns	14
Access time from CAS	t_{CAC}		20		25		25	ns	15, 28
Access time from (TR)/OE	t_{OE}		20		20		25	ns	
Access time from column address	t_{AA}		35		40		45	ns	
Access time from CAS precharge	t_{CPA}		40		45		50	ns	
RAS pulse width	t_{RAS}	70	20,000	80	20,000	100	20,000	ns	
RAS pulse width (FAST PAGE MODE)	t_{RASP}	70	100,000	80	100,000	100	100,000	ns	
RAS hold time	t_{RSH}	20		25		25		ns	
RAS precharge time	t_{RP}	40		60		70		ns	
CAS pulse width	t_{CAS}	20	10,000	25	10,000	25	10,000	ns	
CAS hold time	t_{CSH}	70		80		100		ns	
CAS precharge time	t_{CP}	10		10		10		ns	16
RAS to CAS delay time	t_{RCD}	20	45	20	55	25	70	ns	17
CAS to RAS precharge time	t_{CRP}	10		10		10		ns	
Row address setup time	t_{ASR}	0		0		0		ns	
Row address hold time	t_{RAH}	10		10		15		ns	
RAS to column address delay time	t_{RAD}	15	35	15	45	20	60	ns	18
Column address setup time	t_{ASC}	0		0		0		ns	
Column address hold time	t_{CAH}	15		15		15		ns	
Column address hold time (referenced to RAS)	t_{AR}	55		55		70		ns	
Column address to RAS lead time	t_{RAL}	35		40		50		ns	
Read command setup time	t_{RCS}	0		0		0		ns	
Read command hold time (referenced to CAS)	t_{RCH}	0		0		0		ns	19
Read command hold time (referenced to RAS)	t_{RRH}	0		0		0		ns	19
CAS to output in Low-Z	t_{CLZ}	3		3		3		ns	
Output buffer turn-off delay from CAS	t_{OFF}	3	20	3	20	3	20	ns	20,23
Output disable delay from (TR)/OE	t_{OD}	3	10	3	10	3	20	ns	20,23
Output disable hold time from start of WRITE	t_{OEH}	10		10		20		ns	27
Output Enable to RAS delay	t_{ROH}	0		0		0		ns	

DRAM TIMING PARAMETERS (Continued)**ELECTRICAL CHARACTERISTICS AND RECOMMENDED AC OPERATING CONDITIONS**(Notes: 6, 7, 8, 9, 10, 11, 12, 13) ($0^{\circ}\text{C} \leq T_A \leq +70^{\circ}\text{C}$; $V_{CC} = 5\text{V} \pm 10\%$)

AC CHARACTERISTICS PARAMETER	SYM	-7		-8		-10		UNITS	NOTES
		MIN	MAX	MIN	MAX	MIN	MAX		
Write command setup time	t^1_{WCS}	0		0		0		ns	21
Write command hold time	t^1_{WCH}	15		15		15		ns	
Write command hold time (referenced to $\overline{\text{RAS}}$)	t^1_{WCR}	50		55		70		ns	
Write command pulse width	t^1_{WP}	15		15		15		ns	
Write command to $\overline{\text{RAS}}$ lead time	t^1_{RWL}	20		20		20		ns	
Write command to $\overline{\text{CAS}}$ lead time	t^1_{CWL}	15		20		20		ns	
Data-in setup time	t^1_{DS}	0		0		0		ns	22
Data-in hold time	t^1_{DH}	15		15		15		ns	22
Data-in hold time (referenced to $\overline{\text{RAS}}$)	t^1_{DHR}	50		55		65		ns	
$\overline{\text{RAS}}$ to $\overline{\text{WE}}$ delay time	t^1_{RWD}	90		100		130		ns	21
Column address to $\overline{\text{WE}}$ delay time	t^1_{AWD}	55		65		75		ns	21
$\overline{\text{CAS}}$ to $\overline{\text{WE}}$ delay time	t^1_{CWD}	40		45		55		ns	21
Transition time (rise or fall)	t^1_{τ}	3	35	3	35	3	35	ns	9, 10
Refresh period (512 cycles)	t^1_{REF}		8		8		8	ms	
$\overline{\text{RAS}}$ to $\overline{\text{CAS}}$ precharge time	t^1_{RPC}	0		0		0		ns	
$\overline{\text{CAS}}$ setup time ($\overline{\text{CAS}}$ -BEFORE- $\overline{\text{RAS}}$ REFRESH)	t^1_{CSR}	10		10		10		ns	5
$\overline{\text{CAS}}$ hold time ($\overline{\text{CAS}}$ -BEFORE- $\overline{\text{RAS}}$ REFRESH)	t^1_{CHR}	10		10		10		ns	5
$\overline{\text{ME/WE}}$ to $\overline{\text{RAS}}$ setup time	t^1_{WSR}	0		0		0		ns	
$\overline{\text{ME/WE}}$ to $\overline{\text{RAS}}$ hold time	t^1_{RWH}	15		15		15		ns	
Mask data to $\overline{\text{RAS}}$ setup time	t^1_{MS}	0		0		0		ns	
Mask data to $\overline{\text{RAS}}$ hold time	t^1_{MH}	15		15		15		ns	

TRANSFER AND MODE CONTROL TIMING PARAMETERS
ELECTRICAL CHARACTERISTICS AND RECOMMENDED AC OPERATING CONDITIONS
(Notes 6, 7, 8, 9, 10) ($0^{\circ}\text{C} \leq T_A \leq +70^{\circ}\text{C}$; $V_{CC} = 5\text{V} \pm 10\%$)

AC CHARACTERISTICS		-7		-8		-10			
PARAMETER	SYM	MIN	MAX	MIN	MAX	MIN	MAX	UNITS	NOTES
TR/(OE) LOW to RAS setup time	^t TL _S	0		0		0		ns	
TR/(OE) LOW to RAS hold time	^t TL _H	15	10,000	15	10,000	15	10,000	ns	
TR/(OE) LOW to RAS hold time (REAL-TIME READ-TRANSFER only)	^t R _{TH}	65	10,000	70	10,000	80	10,000	ns	
TR/(OE) LOW to CAS hold time (REAL-TIME READ-TRANSFER only)	^t CT _H	25		25		25		ns	
TR/(OE) HIGH to RAS precharge time	^t TR _P	40		60		70		ns	
TR/(OE) precharge time	^t TR _W	20		25		30		ns	
TR/(OE) HIGH to SC lead time	^t TS _L	5		5		5		ns	
First SC edge to TR/(OE) HIGH delay time	^t TS _D	15		15		15		ns	
SC to RAS setup time	^t SR _S	25		30		30		ns	
TR/(OE) HIGH to RAS setup time	^t YS	0		0		0		ns	
TR/(OE) HIGH to RAS hold time	^t Y _H	15		15		15		ns	
DSF to RAS setup time	^t FS _R	0		0		0		ns	
DSF to RAS hold time	^t RF _H	15		15		15		ns	
SC to QSF delay time	^t SQ _D		25		30		30	ns	
SPLIT TRANSFER setup time	^t ST _S	25		30		30		ns	
SPLIT TRANSFER hold time	^t ST _H	0		0		0		ns	
DSF (at CAS LOW) to RAS hold time	^t F _H _R	50		55		70		ns	
DSF to CAS setup time	^t FS _C	0		0		0		ns	
DSF to CAS hold time	^t CF _H	15		15		20		ns	
TR/(OE) to QSF delay time	^t TO _D		20		25		30	ns	
RAS to QSF delay time	^t RO _D		65		75		85	ns	
CAS to QSF delay time	^t CO _D		40		40		40	ns	
RAS to first SC delay	^t RS _D	70		80		100		ns	
CAS to first SC delay	^t CS _D	30		30		30		ns	

SAM TIMING PARAMETERS

ELECTRICAL CHARACTERISTICS AND RECOMMENDED AC OPERATING CONDITIONS

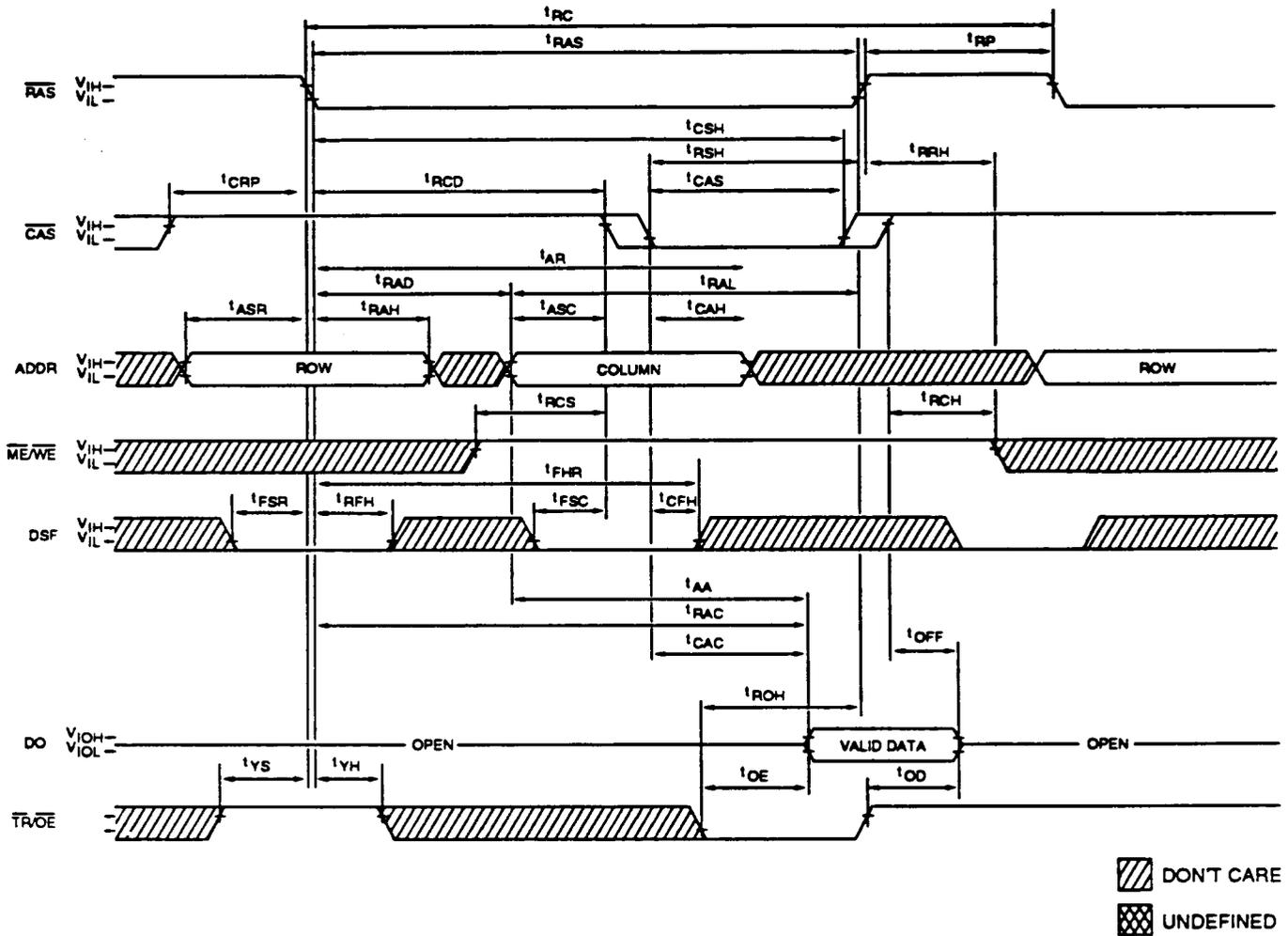
(Notes 6, 7, 8, 9, 10) ($0^{\circ}\text{C} \leq T_A \leq +70^{\circ}\text{C}$; $V_{CC} = 5\text{V} \pm 10\%$)

AC CHARACTERISTICS	PARAMETER	-7		-8		-10		UNITS	NOTES
		MIN	MAX	MIN	MAX	MIN	MAX		
	Serial clock cycle time	¹ SC	25		30		30	ns	
	Access time from SC	¹ SAC		22		25		ns	24, 28
	SC precharge time (SC LOW time)	¹ SP	8		10		10	ns	
	SC pulse width (SC HIGH time)	¹ SAS	8		10		10	ns	
	Access time from SE	¹ SEA		12		15		ns	24
	SE precharge time	¹ SEP	10		10		15	ns	
	SE pulse width	¹ SE	10		10		15	ns	
	Serial data-out hold time after SC high	¹ SOH	5		5		5	ns	24, 28
	Serial output buffer turn-off delay from $\overline{\text{SE}}$	¹ SEZ	3	10	3	12	3	ns	20, 24

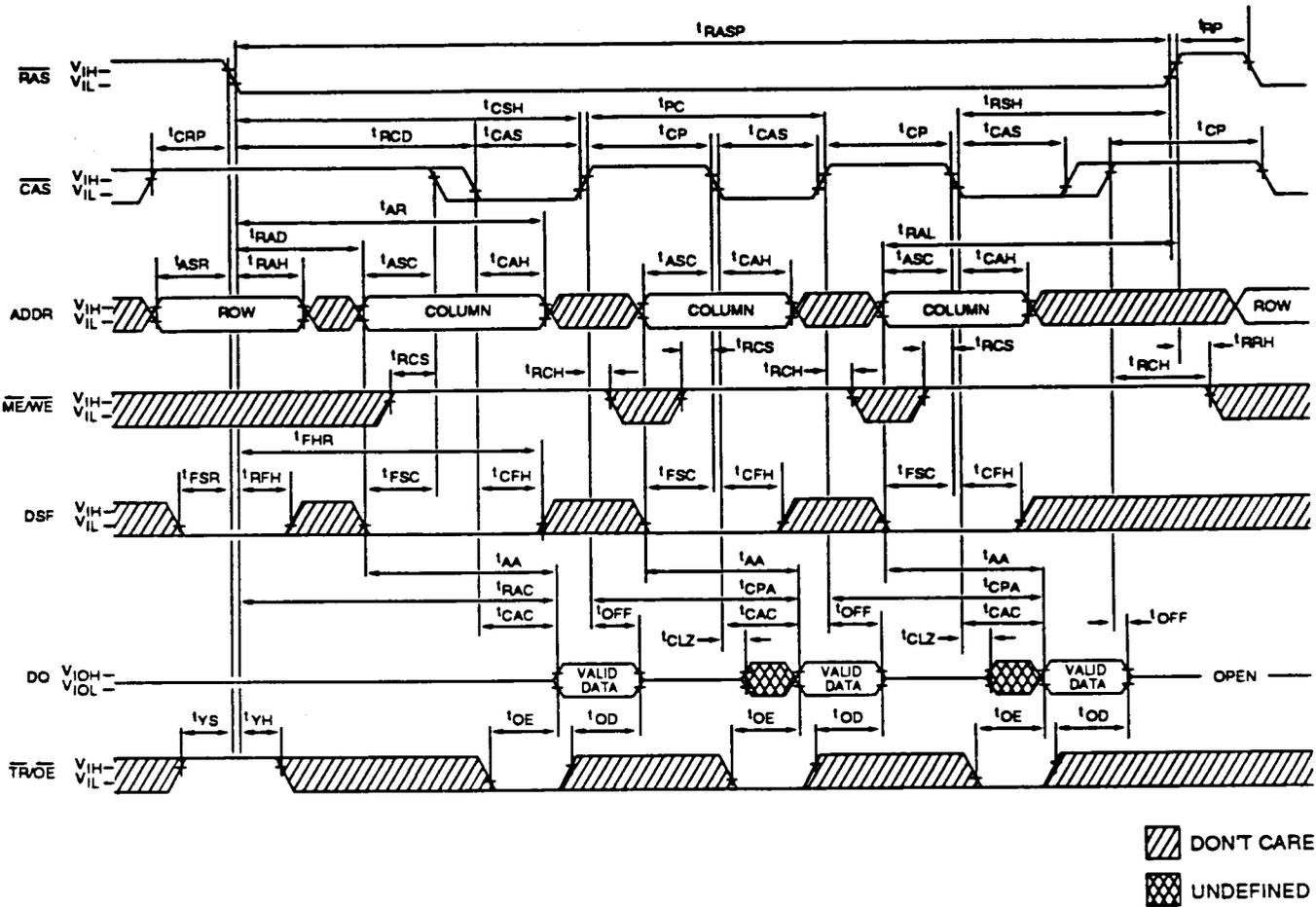
NOTES

1. All voltages referenced to V_{SS}.
2. This parameter is sampled. V_{CC} = 5V ±10%, f = 1 MHz.
3. I_{CC} is dependent on cycle rates.
4. I_{CC} is dependent on output loading. Specified values are obtained with minimum cycle time and the output open.
5. Enables on-chip refresh and address counters.
6. The minimum specifications are used only to indicate cycle time at which proper operation over the full temperature range (0°C ≤ T_A ≤ 70°C) is assured.
7. An initial pause of 100μs is required after power-up followed by any eight RAS cycles before proper device operation is assured. The eight RAS cycle wake-up should be repeated any time the 8ms refresh requirement is exceeded.
8. AC characteristics assume t_T = 5ns.
9. V_{IH} (MIN) and V_{IL} (MAX) are reference levels for measuring timing of input signals. Transition times are measured between V_{IH} and V_{IL} (or between V_{IL} and V_{IH}). Input signals transition from 0 to 3V for AC testing.
10. In addition to meeting the transition rate specification, all input signals must transit between V_{IH} and V_{IL} (or between V_{IL} and V_{IH}) in a monotonic manner.
11. If $\overline{\text{CAS}} = V_{IH}$, DRAM data output (DQ1-DQ8) is High-Z.
12. If $\overline{\text{CAS}} = V_{IL}$, DRAM data output (DQ1-DQ8) may contain data from the last valid READ cycle.
13. DRAM output timing measured with a load equivalent to 1 TTL gate and 50pF. Output reference levels: V_{OH} = 2.0V; V_{OL} = 0.8V.
14. Assumes that t_{RCD} < t_{RCD} (MAX). If t_{RCD} is greater than the maximum recommended value shown in this table, t_{RAC} will increase by the amount that t_{RCD} exceeds the value shown.
15. Assumes that t_{RCD} ≥ t_{RCD} (MAX).
16. If $\overline{\text{CAS}}$ is LOW at the falling edge of RAS, DQ will be maintained from the previous cycle. To initiate a new cycle and clear the data out buffer, $\overline{\text{CAS}}$ must be pulsed HIGH for t_{CP}.
17. Operation within the t_{RCD} (MAX) limit ensures that t_{RAC} (MAX) can be met. t_{RCD} (MAX) is specified as a reference point only; if t_{RCD} is greater than the specified t_{RCD} (MAX) limit, then access time is controlled exclusively by t_{CAC}.
18. Operation within the t_{RAD} (MAX) limit ensures that t_{RCD} (MAX) can be met. t_{RAD} (MAX) is specified as a reference point only; if t_{RAD} is greater than the specified t_{RAD} (MAX) limit, then access time is controlled exclusively by t_{AA}.
19. Either t_{RCH} or t_{RRH} must be satisfied for a READ cycle.
20. t_{OD}, t_{OFF} and t_{SEZ} define the time when the output achieves open circuit (V_{OH} -200mV, V_{OL} +200mV). This parameter is sampled and not 100% tested.
21. t_{WCS}, t_{RWD}, t_{AWD} and t_{CWD} are restrictive operating parameters in LATE-WRITE, READ-WRITE and READ-MODIFY-WRITE cycles only. If t_{WCS} ≥ t_{WCS} (MIN), the cycle is an EARLY-WRITE cycle and the data output will remain an open circuit throughout the entire cycle, regardless of t_{TR/OE}. If t_{WCS} < t_{WCS} (MIN), the cycle is a LATE-WRITE and t_{TR/OE} must control the output buffers during the write to avoid data contention. If t_{RWD} ≥ t_{RWD} (MIN), t_{AWD} ≥ t_{AWD} (MIN) and t_{CWD} ≥ t_{CWD} (MIN), the cycle is a READ-WRITE, and the data output will contain data read from the selected cell. If neither of the above conditions is met, the state of the output buffers (at access time and until $\overline{\text{CAS}}$ goes back to V_{IH}) is indeterminate but the WRITE will be valid, if t_{OD} and t_{OE} are met. See the LATE-WRITE AC Timing diagram.
22. These parameters are referenced to $\overline{\text{CAS}}$ leading edge in EARLY-WRITE cycles and $\overline{\text{ME/WE}}$ leading edge in LATE-WRITE or READ-WRITE cycles.
23. During a READ cycle, if t_{TR/OE} is LOW then taken HIGH, DQ goes open. The DQs will go open with t_{OE} or $\overline{\text{CAS}}$, whichever goes HIGH first.
24. SAM output timing is measured with a load equivalent to 1 TTL gate and 30pF. Output reference levels: V_{OH} = 2.0V; V_{OL} = 0.8V.
25. Address (A0-A8) may be changed two times or less while RAS = V_{IL}.
26. Address (A0-A8) may be changed once or less while $\overline{\text{CAS}} = V_{IH}$ and RAS = V_{IL}.
27. LATE-WRITE and READ-MODIFY-WRITE cycles must have t_{OD} and t_{OE} met (t_{OE} HIGH during WRITE cycle) in order to ensure that the output buffers will be open during the WRITE cycle. The DQs will provide previously read data if $\overline{\text{CAS}}$ remains LOW and t_{OE} is taken LOW after t_{OE} is met. If $\overline{\text{CAS}}$ goes HIGH prior to t_{OE} going back LOW, the DQs will remain open.
28. t_{SAC} is MAX at 70° C and 4.5V V_{CC}; t_{SOH} is MIN at 0°C and 5.5V V_{CC}. These limits will not occur simultaneously at any given voltage or temperature. (t_{SOH} = t_{SAC} - output transition time), this is guaranteed by design.

DRAM READ CYCLE



DRAM FAST-PAGE-MODE READ CYCLE



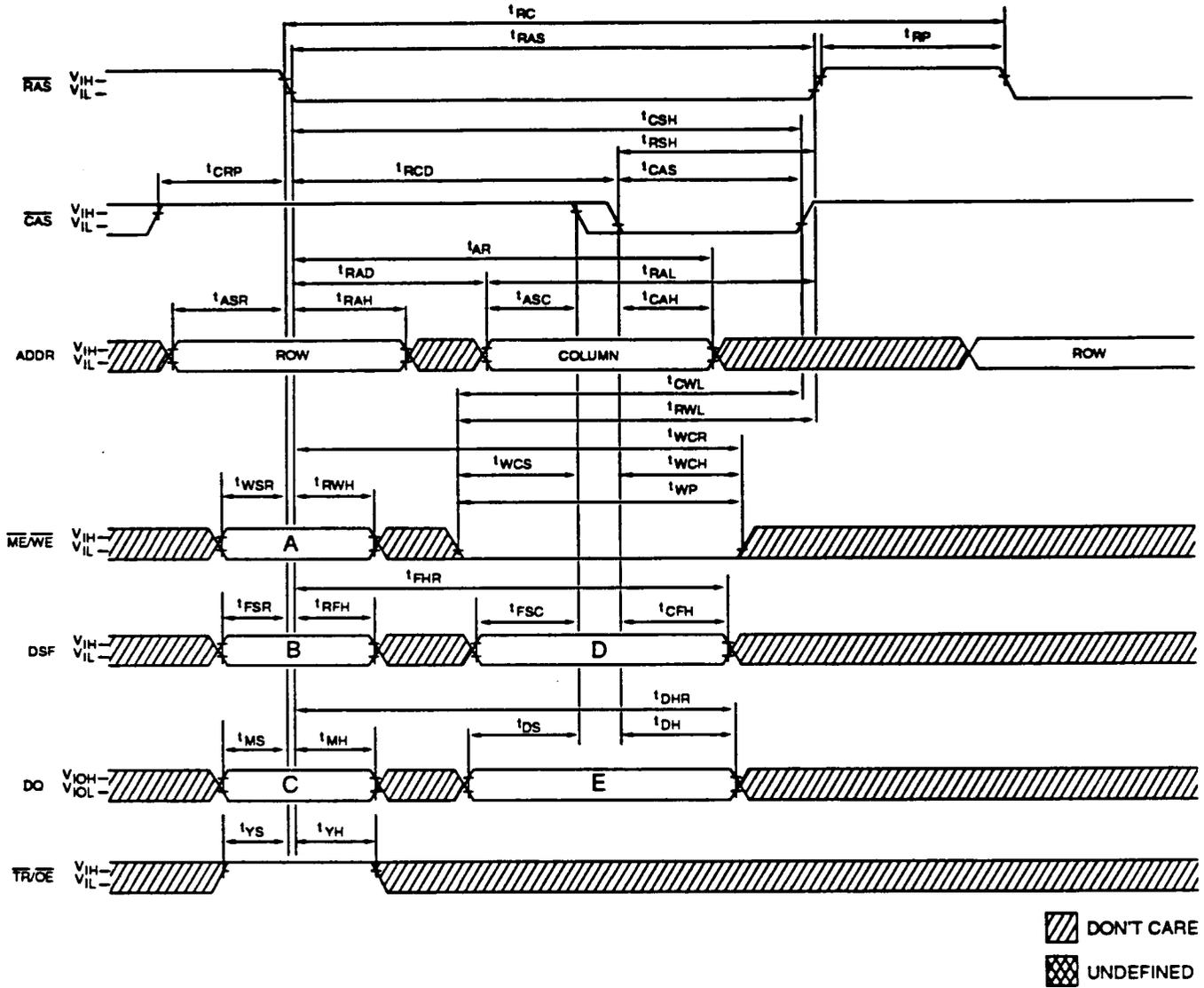
NOTE: 1. WRITE cycles or READ-MODIFY-WRITE cycles may be mixed with READ cycles while in FAST PAGE MODE.

WRITE CYCLE FUNCTION TABLE¹

FUNCTION	LOGIC STATES				
	RAS Falling Edge			CAS Falling Edge	
	A ME/WE	B DSF	C DQ (Input)	D DSF	E ² DQ (Input)
Normal DRAM WRITE	1	0	X	0	DRAM Data
MASKED WRITE to DRAM	0	0	Write Mask	0	DRAM Data (Masked)
BLOCK WRITE to DRAM (No Bit-Plane Mask)	1	0	X	1	Column Mask
MASKED BLOCK WRITE to DRAM	0	0	Write Mask	1	Column Mask
Load Color Register	1	1	X	1	Color Data

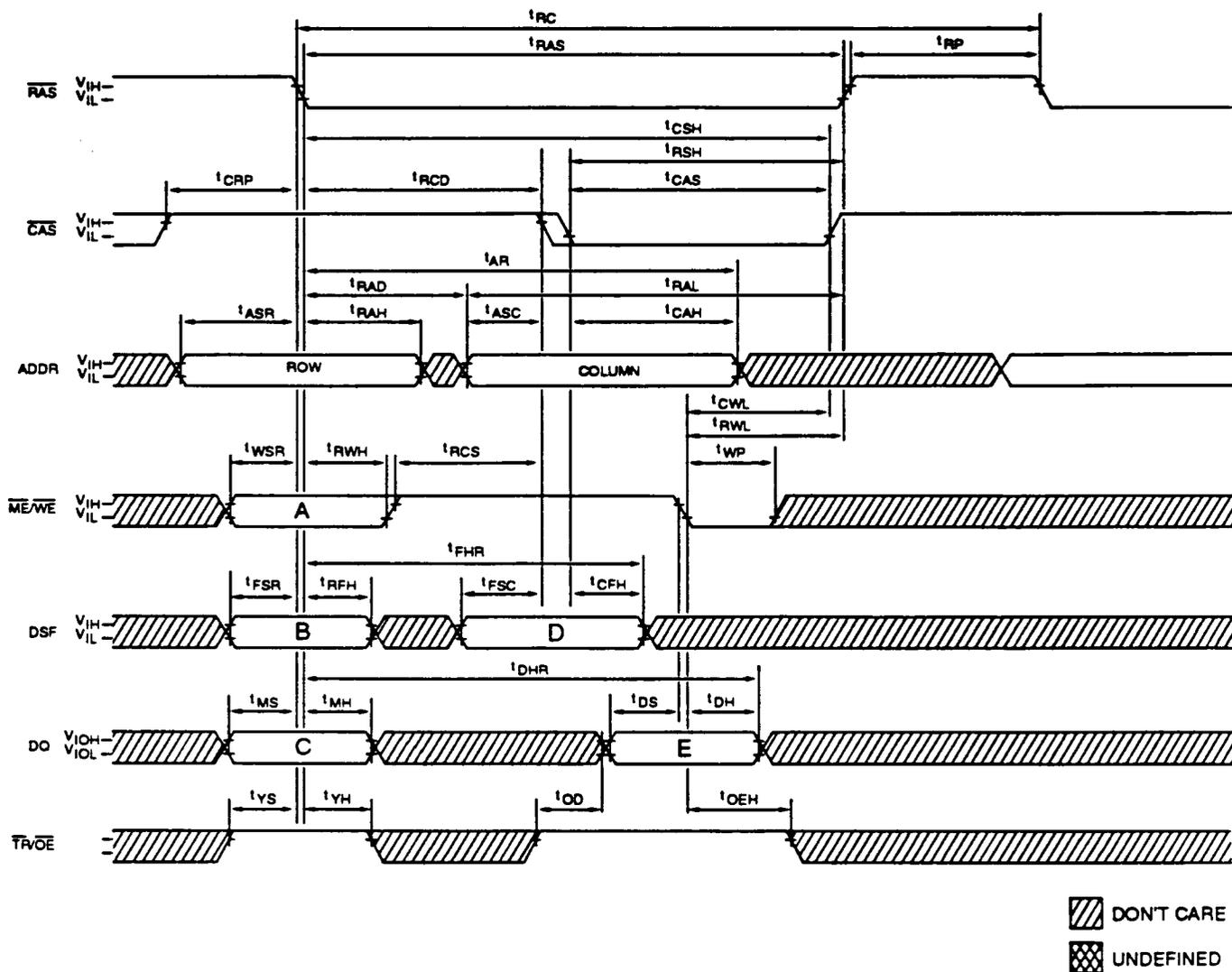
- NOTE:**
1. Refer to this function table to determine the logic states of "A", "B", "C", "D" and "E" for the WRITE cycle timing diagrams on the following pages.
 2. CAS or ME/WE falling edge, whichever occurs later.

DRAM EARLY-WRITE CYCLE 1



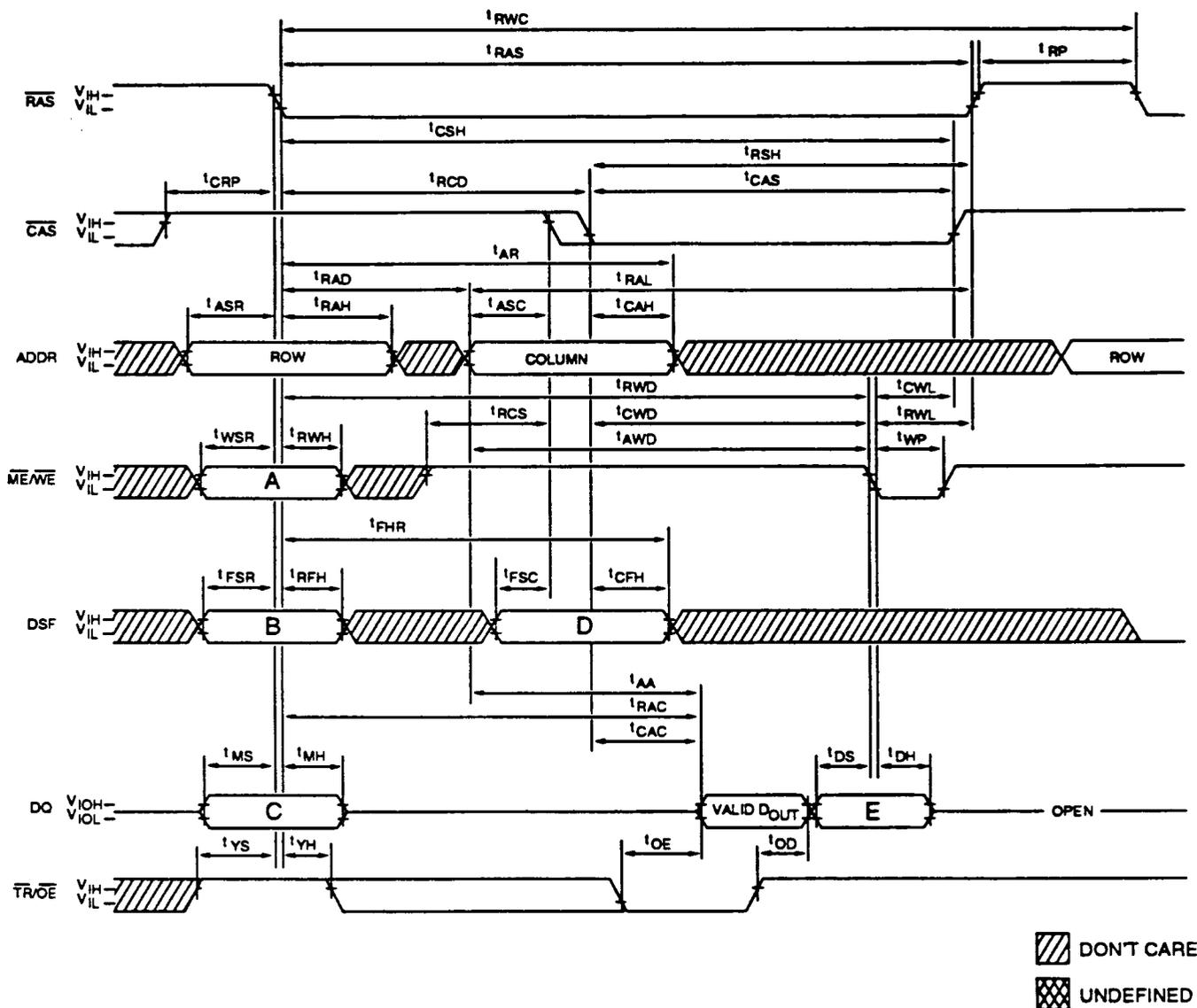
NOTE: 1. The logic states of "A", "B", "C", "D" and "E" determine the type of WRITE operation performed. See the Write Cycle Function Table for a detailed description.

DRAM LATE-WRITE CYCLE



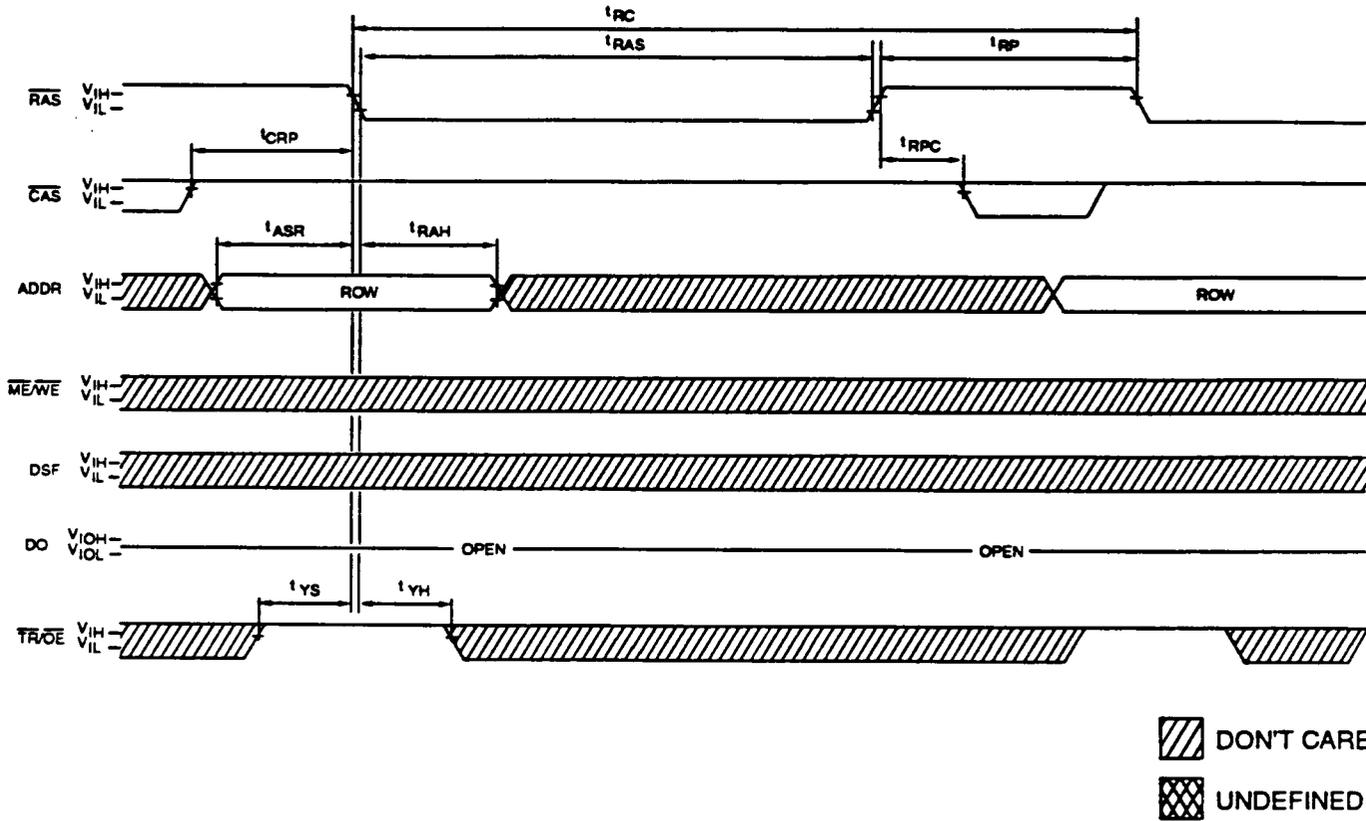
NOTE: 1. The logic states of "A", "B", "C", "D" and "E" determine the type of WRITE operation performed. See the Write Cycle Function Table for a detailed description.

**DRAM READ-WRITE CYCLE
(READ-MODIFY-WRITE CYCLE)**

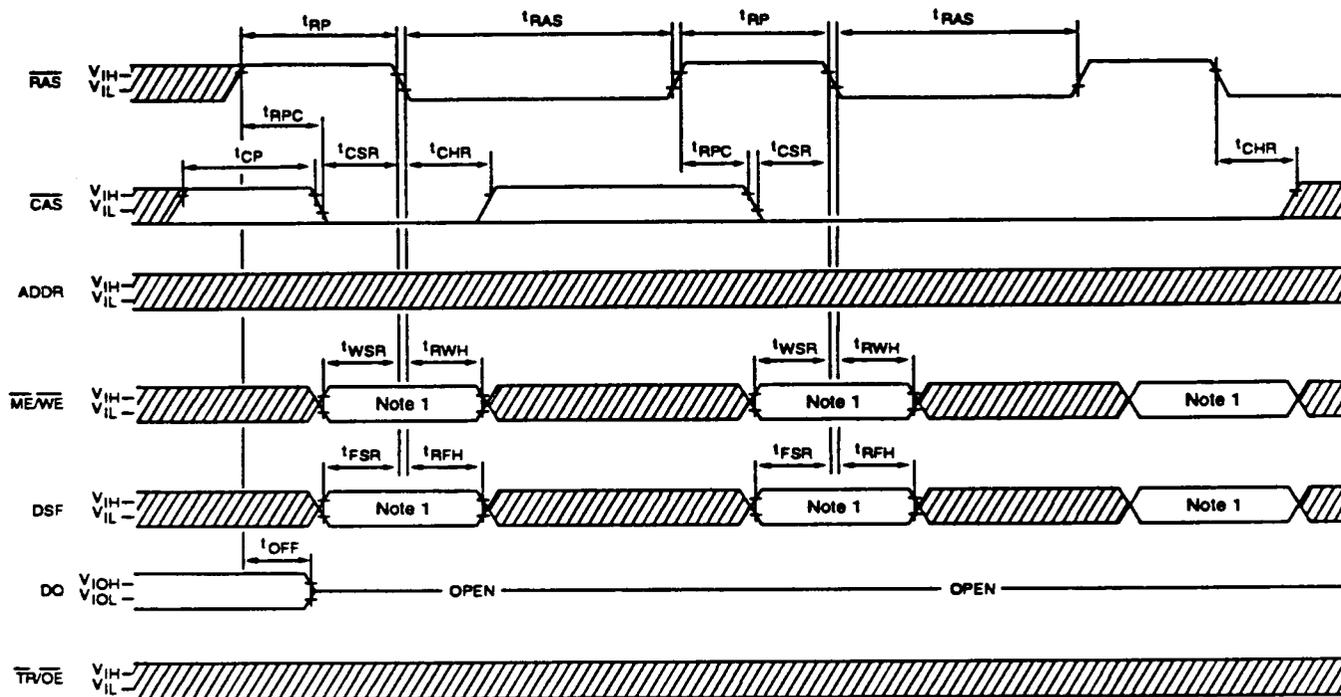


NOTE: 1. The logic states of "A", "B", "C", "D" and "E" determine the type of WRITE operation performed. See the Write Cycle Function Table for a detailed description.

DRAM RAS-ONLY REFRESH CYCLE (ADDR = A0-A8)



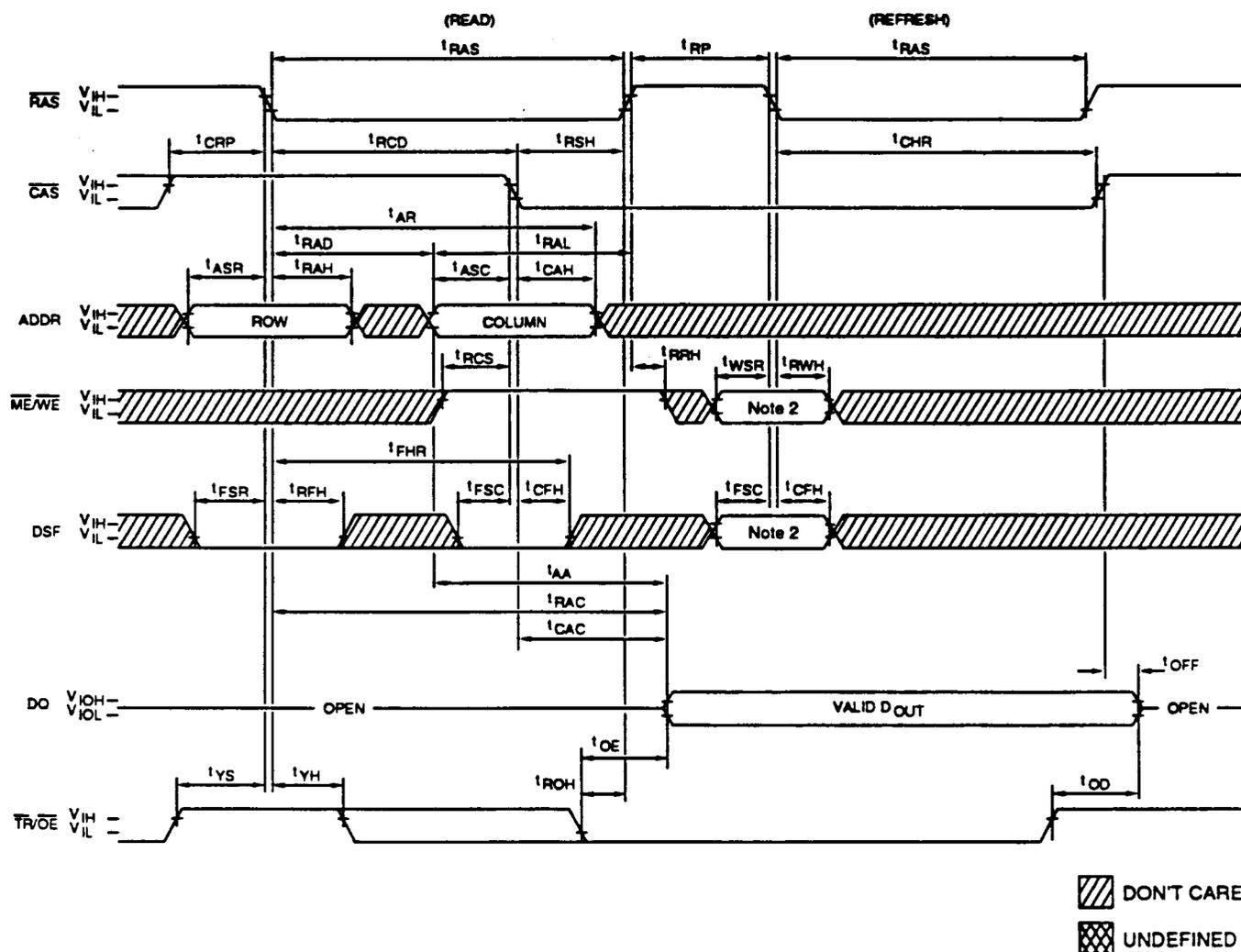
CAS-BEFORE-RAS REFRESH CYCLE 2



 DON'T CARE
 UNDEFINED

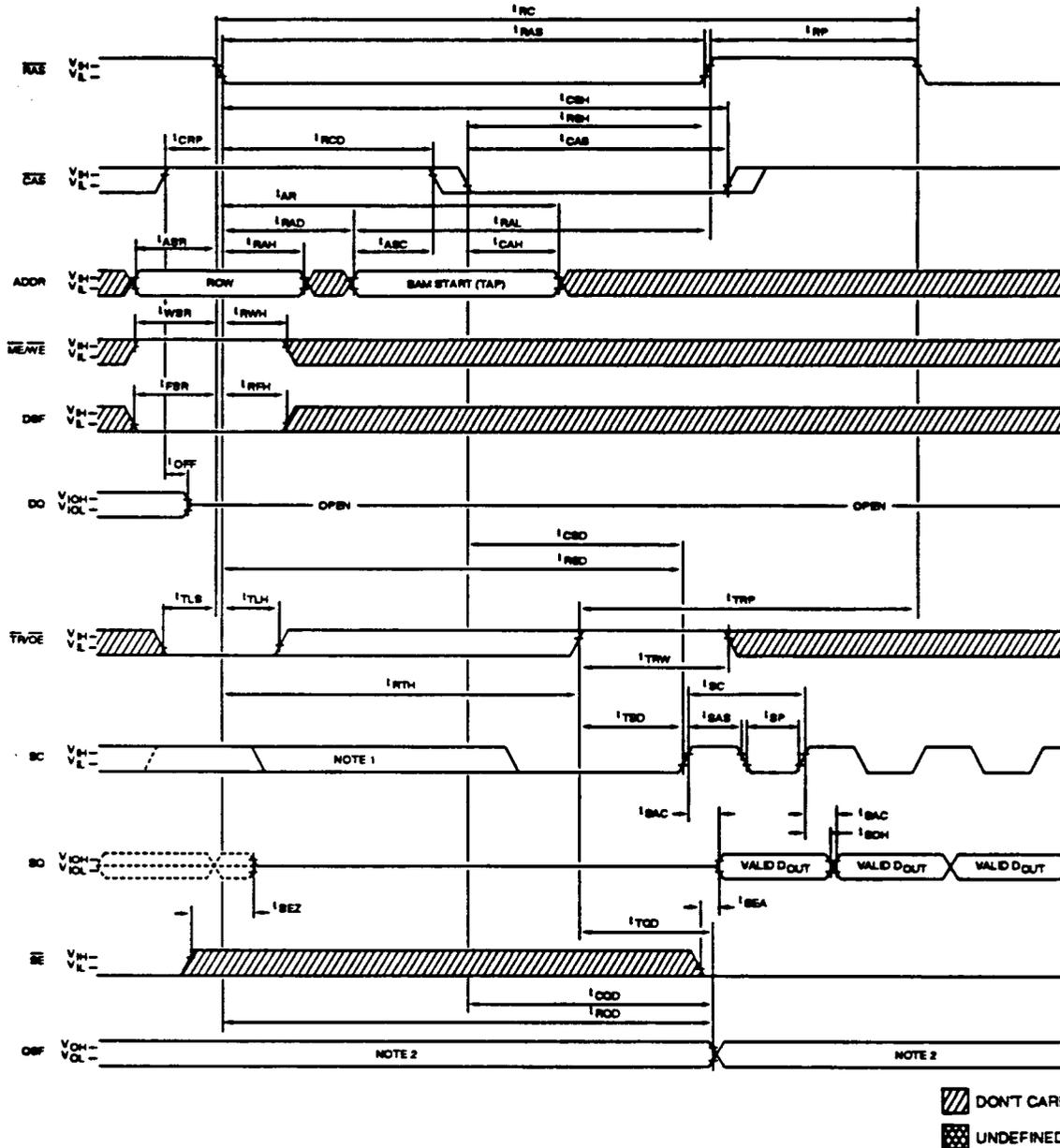
NOTE: 1. The MT42C8255 operates with $\overline{ME/WE}$ and DSF = "don't care," but to ensure compatibility with all 2 Meg VRAM feature sets, it is recommended that they be HIGH ("1").

DRAM HIDDEN-REFRESH CYCLE



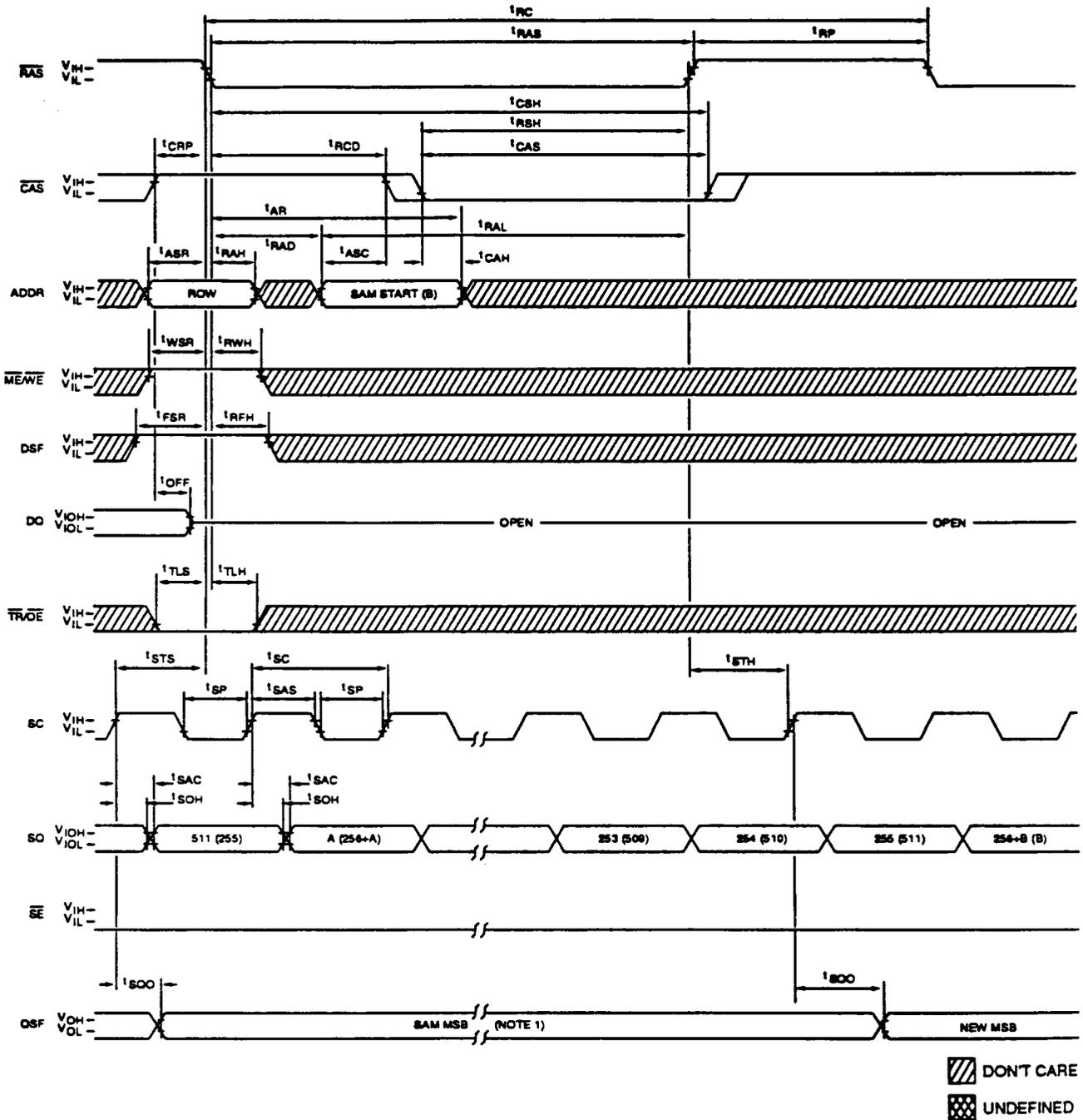
- NOTE:**
1. A HIDDEN REFRESH may also be performed after a WRITE or TRANSFER cycle. In this case, $\overline{ME/WE}$ = LOW (when \overline{CAS} goes LOW) and $\overline{TR/OE}$ = HIGH. In the TRANSFER case, $\overline{TR/OE}$ = LOW (when \overline{RAS} goes LOW) and the DQ pins stay High-Z during the refresh period, regardless of $\overline{TR/OE}$.
 2. The MT42C8255 operates with $\overline{ME/WE}$ and DSF = "don't care", but to ensure compatibility with all 2 Meg VRAM feature sets, it is recommended that they be HIGH ("1").

**READ TRANSFER
(DRAM-TO-SAM TRANSFER)**
(When serial part was previously High-Z or SC idle)



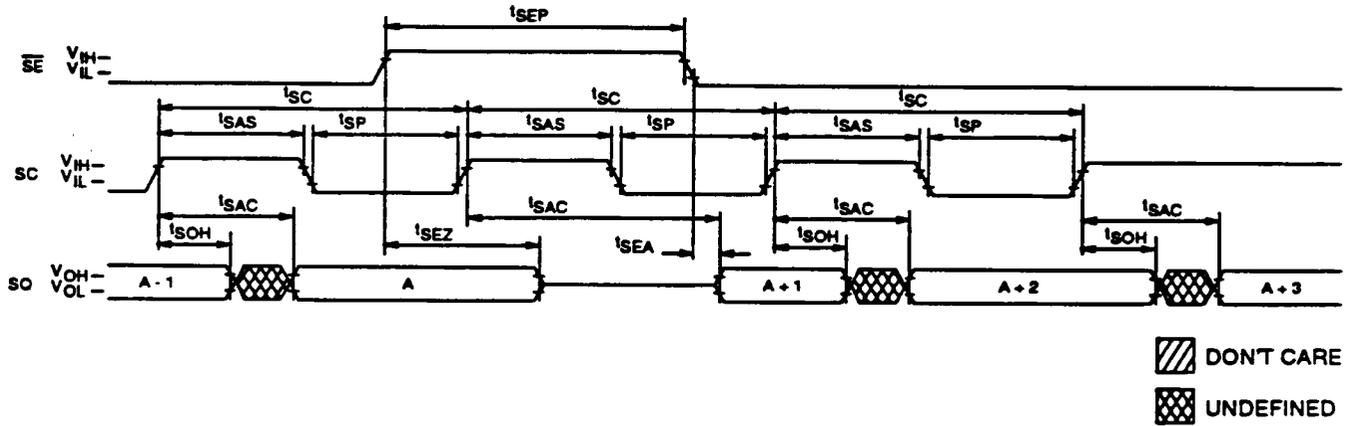
- NOTE:**
1. There must be no rising edges on the SC input during this time period.
 2. QSF = 0 when the Lower SAM (bits 0–255) is being accessed.
QSF = 1 when the Upper SAM (bits 256–511) is being accessed.
 3. If t_{TLH} is timing for the $\overline{TR}/\overline{OE}$ rising edge, the transfer is self-timed and the t_{CSD} and t_{RSD} times must be met. If t_{RTH} is timing for the $\overline{TR}/\overline{OE}$ rising edge, the transfer is done off of the $\overline{TR}/\overline{OE}$ rising edge and t_{TSD} must be met.

**SPLIT READ TRANSFER
(SPLIT DRAM-TO-SAM TRANSFER)**



NOTE: 1. QSF = 0 when the Lower SAM (bits 0-255) is being accessed.
QSF = 1 when the Upper SAM (bits 256-511) is being accessed.

SAM SERIAL OUTPUT



028650 ✓