Tandy 1000 TL Technical Reference Manual
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The Technical Reference Manual for the Tandy 1000 TL describes the computer hardware components and their relationships to one another, as well as the BIOS (Basic Input Output Services).

The information in this manual is intended for hardware and software designers, engineers, programmers, and anyone who requires an understanding of the design and operation of the computer.

Timing diagrams for devices used in the system architecture, Schematics, specifications, switch settings and jumpers, and a theory of operation are provided for the following hardware sections:

| Main Logic Board | Keyboards |
| :--- | :--- |
| Devices | Disk Drives |
| Power Supplies |  |

The Software section contains the following:
A Quick Reference list of software interrupts (for all device, $1 / O$, and system status services) Keyboard ASCII and scan codes An MS-DOS memory map

The information in this manual is a supplement to and based on a working knowledge of the following literature:

The 1000 TL Installation and Operation Guide (Packaged with the computer)

The Intel 80286 Programing Reference Manual. Intel order number 210498-005

The Intel 80286 Hardware Reference Manual. Intel order number 210760-002

80286 Data Sheet. Intel order number 210253-012
This Intel literature may be ordered directly from Intel at the following number: 1-800-549-4725

Tandy 1000 TL
Page Insertion Guide

Important Customer Note:
A gray stripe has been printed along the right edge of the title page of each of the sections to facilitate your finding the beginning of the section.

Also, a tabbed divider for each section has been provided for insertion at this point.

- Exploded view: Insert at the end of the Assembly/Disassembly section
- Foldout schematic pages: Insert at the end of the Main Logic Board section

```
Schematics
C8000300 - Rev A
Sheets 1 of 10 thru 10 of 10
```

Foldout PCB art: Insert after the Main Logic Board schematics

Silkscreen 1700376 - Rev A
Layer 1 Component Side
Layer 2 GND Plane
Layer $3+5 \mathrm{~V}$ Plane
Layer 4 Solder Side

Foldout schematic page: Insert at the end of the 67 Watt Single Input Power Supply section

Schematic Model No. 8790085

- Foldout schematic page: Insert at the end of the 67 Watt Dual Input Power Supply section

Schematic
Model No. 8790084

- Foldout keyboard art pages: Insert after the Fujitsu Keyboard information in the Keyboard section

| Keyboard Unit Assembly | N860-4703-0001 |
| :--- | :--- |
| Block Diagram | 4700 |
| Circuit Specification | N86C-4700-0001 |
| Circuit Specification | N86C-4700-0101 |

- Foldout schematic page: Insert after the Fujitsu Custom IC Pin Signal sheet 2 of 3 in the Reyboard section
Schematic Fujitsu Custom IC Pin Signals \& Function

Exploded view of Parts Assembly location: Insert after the Section 4 part Replacement portion of the Disk Drive section

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## Introduction

General Description

The Tandy 1000 TL is modular in design to allow maximum flexibility in system configuration. The computer consists of a main unit, and a detachable keyboard with coiled cable. The main unit is supplied with one internal $3 \frac{1}{2}$-inch 720K floppy disk drive. The standard types of monitors used with the Tandy 1000 TL are the monochrome and the color RGB monitor. Since these units are modular, you can place them on top of the main unit or at any convenient location.

The Tandy 1000 TL comes standard with 640 K of system RAM. An optional 128 K RAM can be added on the system board to expand the memory to a full 768k bytes, the maximum RAM allowed by the system memory map.

Other features include a parallel printer port, a serial port, two built-in joystick interfaces, a real-time clock, a speaker for audio feedback, a headphone jack with volume control, and microphone/line audio in jack for sound input.

```
    80286 CPU running at 8 MHz, l wait state, switchable to 4
    MHz
    Socket for 80287 numerical coprocessor
    640K bytes DRAM upgradeable to 768K bytes (l6-bit data bus)
    4 Mbit BIOS ROM with MS-DOS* and Deskmate` (16-bit data bus)
    Tandy 1000 TL video controller that supports:
    - 128K bytes DRAM (used as system and video memory)
    - alphanumeric mode
    - graphics modes including:
        - 160 X 200 16-color
        - 320 X 200 4-color
        - 320 X 200 16-color
        - 640 X 200 2-color
        - 640 X 200 4-color
    8237-5 DMA controller that supports:
    - }3\mathrm{ DMA channels
    - 8-bit transfers
    - 4 MHz clock speed
    8259A interrupt controller for 8 interrupts
8254 interval timer that supports:
    - system interrupt timing
    - sound timing
    - refresh timing
    Custom keyboard interface controller
    10l-key Enhanced keyboard
    Custom parallel printer port
    Serial port (RS-232-C)
    Real-time clock w/battery
    Audio Out interface circuit that supports:
    - internal 8-Ohm speaker
    - headphone jack with user accessible volume control
    Audio In interface circuit that supports:
    - microphone in
    - line audio in
    - sound digitizing and recording
    Joystick interface for two joysticks
    Custom floppy disk controller circuit that supports:
    - 5\frac{1}{-inch 360R floppy disk drives}
    - 3\frac{1}{2}-inch 720K floppy disk drives
    On-board Intelligent Hard Disk interface
    One 3\frac{1}{2}\mathrm{ -inch 720K floppy disk drive}
    Five 8-bit expansion slots
    Reset button and support logic
    67-Watt power supply
```


## Optional Features

- 80287 numerical math coprocessor
- 128K DRAM upgrade (16-bit data bus memory)
- 5t-inch 360K floppy disk drive
- $\quad 3 \frac{1}{2}$-inch 720K floppy disk drive
. Hard disk card (20/40 meg)
. Display adapter boards that support mono, EGA, or other special video modes 300, 1200, or 2400 baud modem boards

```
Physical Specifications
```

(Computer and Keyboard)

```
Processor: Intel* 80286
Dimensions: Computer - 16\frac{1}{2}-inch x 13t-inch x 5\frac{1}{2}-inch
    Keyboard - 16\frac{1}{2}-inch x 8-inch x li-inch
    Computer - }18\mathrm{ lbs. (with 2 floppy disk drives)
    Keyboard - 3 lbs. 4 oz.
```

Power Requirements:
Range: 105 VAC to 135 VAC
Nominal: 120 VAC, $60 \mathrm{~Hz}, 3$ Amp maximum
With 1 Floppy Disk Drive, 640K Memory:
AC Current: $350-400 \mathrm{~mA}$ with floppy doing $R / W$ tests.
Disk Drive:

|  | +5 VDC | +12 VDC |
| :--- | :--- | :--- |
| $\mathrm{R} / \mathrm{W}$ | 560 mA (Min.) | 340 mA (Max.) |
| Main Logic Board: | 1700 mA | 450 mA |

## Ocerating Environment:

Temperature: $\quad 55^{\circ}$ to $85^{\circ} \mathrm{F}\left(13^{\circ}\right.$ to $\left.30^{\circ} \mathrm{C}\right)$ Humidity: $40 \%$ to $80 \%$ non-condensing

Non-Operating Environment:

| Temperature: | $-40^{\circ}$ to $+160^{\circ} \mathrm{F}\left(-40^{\circ}\right.$ to $\left.71^{\circ} \mathrm{C}\right)$ |
| :--- | ---: | :--- |
| Humidity: | $20 \%$ to $90 \%$ non-condensing |

Disk Drive Specifications
Power:
Supply
Voltage +5 VDC Input +12 VDC Input
Ripple
0 to $50 \mathrm{kHz} \quad 100 \mathrm{mV} \quad 100 \mathrm{mV}$
Tolerance
Including Ripple $+/-5$ \%
$+/-58$
Standby Current
Nominal $190 \mathrm{~mA} \quad 160 \mathrm{~mA}$
Worst Case $220 \mathrm{~mA} \quad 190 \mathrm{~mA}$
Operating Current

| Nominal | 260 mA | 600 mA |
| :--- | :--- | :--- |
| Worst Case | 300 mA | 1000 mA |

Assembly/Disassembly

The following instructions explain how the major subassemblies are removed from the Tandy 1000 TL. Re-assembly of major subassemblies is accomplished by reversing the order of the removal procedures.

1. Top Cover Removal
a. Remove the (2) screws from the side of the computer at the rear.
b. Slide the cover forward enough to clear the power button, volume knob, and disk drive eject button and then lift the cover straight up and off.
2. $3 \frac{1}{2}$-inch Floppy Disk Removal
a. Remove the top cover.
b. Unplug the cable from the disk drive.
c. Remove the (3) screws attaching the drive to the drive mounting tower.
d. Slide the drive out of the drive mounting tower.
3. Power Supply Removal
a. Remove the top cover.
b. Remove the rear panel by slightly bending the hooks on each side near the bottom and rotating enough to clear the sheet metal and then lift up.
c. Remove all cables from the main logic board and disk drives.
d. Remove the arm attached to the power supply switch.
e. Remove the (3) screws from the rear of the computer and (1) screw from the side that secure the power supply to the rear of the machine.
f. Slide the power supply up and out.

## 4. Main Logic Board Removal

a. Remove the top cover.
b. Unplug all cables and remove all of the adapter boards from the system.
c. Remove the power supply.
d. Remove the back of the chassis by removing (1) screw at the rear of the computer and pulling the back of the chassis to the rear and down to clear the (3) hooks in the bottom of the chassis. ('also miot reonve if nuts ar back)
e. Remove the (11) screws holding the main logic board in place. Calx must remove drive $\hat{B}$ : hracket) (alsomust remove speaker)
f. Remove the main logic board by carefully pulling straight back from under the drive support and out of the chassis. (also must remove volime control clip \& (snctu)
NOTE: WHEN REPLACING THE MAIN LOGIC BOARD, BE SURE THAT THE VOLUME CONTROL KNOB POST SLIDES INTO THE VOLUME CONTROL POT CORRECTLY.



MECHANICAL BILL OF MATERIAL - TANDY 1000 TL

## TANDY 1000 TL FINAL ASSEMBLY

| QTY. | DESCRIPTION | PART <br> NUMBER |
| :---: | :---: | :---: |
| 1 | CHASSIS - WELDMENT | 8729710 |
| 5 | PANEL - OPTION SLOT Rev D | 8729562 |
| 5 | SCREW - \#4-40 X 3/16 | 8569333 |
|  | OPTION SLOT PANEL |  |
| 4 | FOOT | 8590179 |
| 4 | RIVET - \#1661-0512 | 8565014 |
| 1 | PC BOARD SUBASSY - TANDY 1000 TL MAIN LOGIC A | 8859022 |
| 1 | BATTERY - LITHIUM CR2032 | 8491012 |
| 1 | SHIELD - GROUND, EARPHONE \& MICROPHONE | 8729720 |
| 10 | SCREW - \#6-32 X 1/4 | 8569326 |
|  | MAIN BOARD |  |
| 4 | JACKNUT - \#4-40 X 3/16 | 8569341 |
|  | 9 PIN CONNECTORS |  |
| 4 | SHIELDING STRIP FLITE-WAY ENGR | 8729658 |
| 1 | KNOB - VOLUME CONTROL | 8719624 |
| 1 | CLIP - HAIRPIN | 8559080 |
| 1 | CHASSIS - POWER SUPPLY Rev B | 8729690 |
| 1 | POWER SUPPLY - 67 WATT | 8790084 |
|  | INT'L |  |
|  | DOM. | 8790085 |
|  | DOM. | 8790091 |
| 4 | SCREW - \#6-32 x 5/16 | 8569339 |
|  | POWER SUPPLY |  |
| 1 | DC HARNESS | 8709857 |
| 1 | SWITCH - POWER | 8489111 |
| 1 | SWITCH - POWER | 8489112 |
|  | INT'L |  |
| 2 | SCREW - M3 x 5PPH | 8569293 |
| 3 | SCREW - \#6-32 X 5/16 | 8569339 |
| 1 | ACTUATOR - POWER SWITCH | 8719620 |
| 1 | BUTTON - POWER | 8719625 |
| 1 | RECEPTACLE - AC | 8519246 |
| 1 | harness - AC | 8709868 |
|  | DOMESTIC |  |
| 1 | HARNESS - AC Rev B | 8709873 |
|  | INT'L |  |
| 2 | CAPACITOR - 1000 PFD, 400 V | 8352106 |
| 1 | TORROID - CORE FAIRRITE | 8419030 |
| 2 | NUT - KEPS, \#6-32 | 8579004 |
| 1 | FAN - 80 MM ; 12 VDC | 8790424 |
| 4 | SCREW - \#l0 TAPIT THREAD | 8569301 |
| 1 | ENDPLATE - POWER SUPPLY | 8729691 |
| 1 | CHASSIS - REAR | 8729693 |
| 2 | SCREW - \#6-32 X 5/16 | 8569339 |
| 1 | BRACKET - $31 / 2$ DISK DRIVE | 8729687 |
| 1 | BRACKET - HARD CARD | 8729704 |
| 2 | SCREW - \#6-32 x 5/16 | 8569339 |

## MECHANICAL BILL OF MATERIAL - TANDY 1000 TL

| TANDY 1000 TL FINAL ASSEMBLY |  |  |  |
| :---: | :---: | :---: | :---: |
| QTY. DESCRIPTION |  |  | PART |
|  |  |  | NUMBER |
| 1 | DISK DRIVE - $31 / 2^{\prime \prime}$ 720KB SONY | MP-F11W-70D | 8790144 |
| 1 | BUTTON - DISK EJECT |  | 8719596 |
| 3 | SCREW - M3 X 5 PPH |  | 8569293 |
| 1 | CABLE - SIGNAL |  | 8709854 |
| 2 | RAIL - $51 / 4{ }^{\prime \prime}$ DRIVE |  | 8719603 |
| 2 | CLIP - GROUNDING, DRIVE |  | 8529064 |
| 2 | SCREW - \#6 - $32 \times 1 / 4$ PHILLIPS PAN |  | 8569098 |
| 1 | SUPPORT - 3 1/2" DRIVE BRACKET |  | 8729722 |
| 1 | SPEARER W/CABLE |  | 8490013 |
| 4 | SCREW - \#6-32 X 5/16 |  | 8569339 |
| 1 | PANEL - REAR |  | 8719602 |
| 1 | BEZEL - FRONT |  | 8719604 |
| 2 | LENS |  | 8719560 |
| 2 | PIN - GUIDE |  | 8739038 |
| 1 | CASE - TOP |  | 8729686 |
| 1 | PANEL - DRIVE COVER |  | 8719621 |
| 5 | SCREW - \#6 X 3/8 |  | 8569294 |
| 2 | SCREW - \#l0-24 UNC 3/8" |  | 8569354 |
|  | PHILLIPS OVAL HEAD MACHINE SCREW |  |  |
| 1 | BUTTON - RESET, FRONT |  | 8719440 |
| 1 | BUTTON - RESET, REAR |  | 8719441 |
| 1 | SPRING - RESET BUTTON |  | 8739018 |
| 1 | CORD - POWER 18/3 60/C |  | 8709057 |
| 1 | NAMEPLATE |  | 8719613 |
| 1 | LABEL - SERIAL UL/FCC |  | 87891642 |
| 1 | LABEL - SERIAL CSA |  | 87891640 |
| 1 | LABEL - SERVICE ADVISEMENT ( 6 LANG) |  | 87891571 |
| 1 | LABEL - CAUTION (6 LANG) |  | 87891572 |
| 1 | LABEL - SERIAL INT'L |  | 87891641 |
| 1 | LABEL - BATTERY WARNING |  | 87891570 |
| 1 | LABEL - EARTH GROUND |  | 87891253 |
|  | INT'L - |  |  |
| 1 | LABEL - VIDEO, MONOCHROME COMMAND |  | 87891648 |
| 1 | KEYBOARD ASSEMBLY |  |  |

Main Logic Board

## Main Logic Board

## Introduction

The main unit is the heart of the Tandy 1000 TL. It houses the main logic assembly, system power supply, and floppy disk drive.

The main logic assembly is a large board mounted to the bottom of the main unit and interconnected to the keyboard, power supply, and disk drive by a series of cables.

The power supply is a 67 W switching regulator type, designed to provide adequate power capacity for a fully configured system that has all the option slots in use.

The floppy disk drive uses $3 \frac{1}{2}$-inch double-sided, double-density diskettes to read, write, or store data. These are soft sector diskettes. The disk drive assembly comes installed in the main unit. The floppy diskette stores approximately 720 K bytes (formatted) of data. All system programs, with the exception of the system startup sequence, are stored on diskette.

Switch Settings and Jumper Pin Configurations

Main Logic Board

Jumper
E1-E2

E2-E3

Jumper


Function
Select Video Interrupt
on IRQ5
Normal Video Interrupt

Function
Select Direct Line Audio Input

Select Mic Audio Input

Default
E2-E3

Default
E7-E8

This is inaccurate. See circuit diagram, Sheet 9 of 10 , which agrees with the mother board labelling. Layer Top Silkscreen agrees with the above and is also in error. Owner's manual, page 15 is correct. The correct settings are:

Jumper Function
Default
EG-ET Select MIC Audio Input

$$
E 6-E 7
$$

ET-E8 Select Direct Line Audio Input
Connecting unattenuated direct-line audio with jumper at E6-ET will fry the sound circuitry.
Jumper set to ET-E8, 4/1/a4.

## Theory of Operation

## 80286 Microprocessor

The 80286 (U31) is an advanced, high-performance, l6-bit microprocessor with special capabilities for multi-tasking and multi-user systems. Two modes of operation are available in the 80286, the Real Address mode, and the protected Virtual Address mode. In the Real Address mode, the 80286 is compatible with existing 8086 and 8088 software and allows addressing of one megabyte of memory space. The Tandy 1000 TL does NOT support the Protected Virtual Address mode.

## 80287 Numerical Math Coprocessor

The 80287 (U60) performs high-speed arithmetic and logarithmic functions and trigonometric operations that increase the performance of an 80286 system. Performance increases are obtained by the 80287's ability to perform math calculations faster than the 80286, and also by executing math instructions in parallel with the 80286.

Clock Generation (Night Blue)

All clocks required by the, system are generated by the custom CPU Controller (Ul7). There are two independent clock circuits supplied by a Dual Oscillator Clock (Y2) from which all other clocks are derived.

The 16 MHz Clock is routed into the CPU Controller, which generates the output signals PRCLK, DMACLK, and SCLK. The Clock Switch circuitry required to toggle the 80286 Microprocessor between 8 MHz and 4 MHz mode, as well as the logic to prevent any short cycling during a clock switch cycle, are implemented in the CPU Controller IC. If the signal XD3 is asserted high during an I/O write to port 062 (hex), then the output signal PRCLK is 16 MHz , which operates the 80286 in 8 MHz mode. If the signal XD 3 is asserted low during an $I / 0$ write to port 062 (hex), the output signal PRCLK is at 8 MHz , operating the 80286 in the 4 MHz mode. When Reset is generated, the signal RES- is asserted low and defaults the Tandy 1000 TL to the 8 MHz mode.

The CPU Controller Chip also controls wait states to insert the proper number of wait states required for a two clock mode of operation. When the PRCLK signal is $16 \mathrm{MHz}(8 \mathrm{MHz}$ Mode), then four wait states are inserted in all 8-bit Memory and I/O cycles. When the signal PRCLK is 8 MHz ( 4 MHz mode) then two wait states are inserted during all 8 -bit Memory and $1 / 0$ cycles. During all 16-bit memory cycles, only one wait state is inserted in both the 8 MHz and 4 MHz modes.

PRCLKO is then routed through a damping resistor to produce the signal PRCLK for the 80286 , PRCLRA for the 80287 math coprocessor, and PRCLKB for the DRAM/DMA control logic.

DMACLK and SCLK are output signals for system use. The DMACLK output frequency is $\}$ of the PRCLK signal, and the SCLK output frequency is $\frac{1}{2}$ of the PRCLK signal. Both are synchronized by Reset to the PRCLK output signal. After a Reset, DMACLK and SCLK are held low until the 80286 asserts status $\mathrm{Sl}=0$. SCLK and DMACLK make the first transition on the falling edge of PRCLK, following with a Ts state that synchronizes them to PRCLR.

SCLK is buffered and filtered, then routed to the Expansion Bus for option board use. DMACLK is filtered and then routed to the DMA Controller.

Table 2 shows all the clocks generated from 16 MHz in both modes:

|  | 8 MHz Mode | 4 MHz Mode |
| :--- | :--- | :--- |
| PRCLK | 16 MHz | 8 MHz |
| SCLR | 8 MHz | 4 MHz |
| DMACLK | 4 MHz | 2 MHz |

Table 2. Clocks Generated From 16MHz.

Command and Control Signal Generation

The command and control signals required for the Tandy 1000 TL operation are generated by the CPU Controller (Ul7). The command signals are decoded from the CPU status signals $S 0$ - through Sland $M /(I O-)$ during the $T s$ cycle. The decoded signals indicate the type of cycle that is to be executed (MEMR, MEMW, IOR-, IOW-, INTA-). The control signals (ALE, DT/R, DSDEN0-, DSDEN1-, MEMCYC) control the latching of addresses, determine the direction and enabling of the data bus buffers, and start a memory cycle. Table 3 indicates the decoding of the CPU status signals.

| M/(IO-) | Sl- | SO- | Type of Bus Cycle |
| :---: | :---: | :--- | :--- |
| 0 | 0 | 0 | Interrupt Acknowledge |
| 0 | 0 | 1 | I/O Read |
| 0 | 1 | 0 | I/O Write |
| 0 | 1 | 1 | None: Idle |
| 1 | 0 | 0 | Halt or Shutdown |
| 1 | 0 | 1 | Memory Read |
| 1 | 1 | 0 | Nemory Write |
| 1 | 1 | 1 | None: Idle |

Table 3. CPU Status Signal Decoding.
$A 0$ and BHE- are decoded to determine the data transfer width to and from the CPU. Table 4 shows the data transfer width depending on the state of $A 0$ and BHE-.


Table 4. Data Transfer Width Decode.

## Comand Buffer

Some of the command signals generated by the CPU Controller require buffering to the system. This is accomplished by $\frac{1}{2}$ of an ALS244 (U25) and $\frac{1}{2}$ of another ALS 244 (U9). These ICs buffer the command signals to the system bus for the expansion bus slots, the peripheral devices, and also the video Controller.

DRAM Control

The CPU address decode for the Dynamic Random Access Memory (DRAM) array is generated by the Custom DRAM/DMA Control IC (U22). These signals are latched by ALE internally to the DRAM/DMA Control IC and held for the complete cycle. The address decode signals are RASO-, RASI-, CASL-, and CASH-. Memory configurations supported by the Tandy 1000 TL are 640 K bytes or 768 K bytes (which includes 128 K of video memory). Table 5 shows the different options available on the DRAM/DMA Control IC.

| Memory <br> Option | MCl MC0 | System <br> Memory | Total System <br> Memory* | Control | BankAddress <br> Range |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 0 | 1 | 512 K | 640 K | RASO | 512 K | $000000-07 \mathrm{FFFF}$ |
| 2 | 1 | 1 | 540 K | 768 K | RASO <br> RAS | 512 K <br> 128 K | $000000-07 \mathrm{FFFF}$ <br> $080000-09 \mathrm{FFFF}$ |

Note: Total system memory includes 128 K of video memory.
Table 5. Memory Configurations.

Port FFEA hex, Bits $6 \& 7$, control the Memory Configuration Options. See the I/O Map later in this manual for details. MEMCYC triggers the control signals for the DRAM array. The MEMCYC signal (generated by the CPU Controller) indicates that a DRAM bus cycle is in progress. MEMCYC enables RASO-, RASl-, CASH-, and CASL-, depending on the address of the bus cycle. The selected RAS(x)- lines become active at the next falling edge of PRCLK.

After PRCLK cycle at the rising edge of the clock, MUX is generated and switches the DRAM address (MA0-MA8) from Row Address to Column Address. After another PRCLK cycle at the next rising edge of the clock, CASL- and/or CASH- are asserted. Two CAS signals are generated internally to the DRAM/DMA Control IC to provide the ability to access word or byte cycles in the DRAM array. Table 6 shows the state of each control signal during each type of bus cycle.

| Address Range | Bus <br> Width | RAS0- RAS1- CASL- | CASH- |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $000000-07$ FFFFH Even | Byte | 0 | 1 | 0 | 1 |  |
| $000000-07 F F F F H$ | Odd | Byte | 0 | 1 | 1 | 0 |
| $000000-07 F F F F H$ | Even | Word | 0 | 1 | 0 | 0 |
| $080000-09 F F F F H$ | Even | Byte | 1 | 0 | 0 | 1 |
| $080000-09 F F F F H$ | Odd | Byte | 1 | 0 | 1 | 0 |
| $080000-09 F F F F H$ | Even Word | 1 | 0 | 0 | 0 |  |

Table 6. Signal State Control Signals.

The signals WEO- and WEl- provide read and write control. Both are asserted at the same time and are controlled by MEMW- (memory write). If WEO- and WEl- are asserted high, it is a read cycle; if they are asserted low, it is a write cycle.

## Refresh Control

The REFREQ pin of the KFIT custom IC (Ul3) generates an active high pulse every 15 usec. The rising edge of the REFREQ signal clocks the 8237 DMA controller. This input to the DMA controller is actually Data Transfer Request 0 , (DREQO), which requests the DMA to perform a DMA cycle. The DMA controller channel 0 has been programmed to perform a single transfer from memory to an I/O device, such as a floppy drive. This causes a memory read at a certain address to be performed. Each time the REFREQ signal is generated, the DMA controller increments the address and performs another memory read. This causes all memory rows to be read every 4ms to keep data in the Drams stable. Refer to the section on the DMA controller for more information on DMA cycles.

## BIOS ROM Control

The DRAM/DMA IC (U22) provides the CPU address decode used for the ROM select. The signal generated is called ROMCS- (ROM Chip Select) and is used as part of the decode used by PLS173 IFL U44. The PLS173 IFL then generates the ROM Page Selects (RPCS-) and Chip Enable for the BIOS ROMs U54, U55, U56, and U57. This output is asserted whenever any of three addressed ranges is detected, CPHLDA is inactive, and ALE is asserted. The three address ranges are 0E0000-0FFFFFH, EE0000-EFFFFFH, and FE0000-FFFFFFH. The address lines SAl-SAl5 are provided to the BIOS ROMS for lower address control. The data is buffered onto the MDO-MDl5 data bus, controlled by the 82C205 IC.

## Reset Circuit

The CPU Control IC (Ul7) controls the system reset required either to initialize the complete system after power-up or to reboot. Two reset output signals, RESET and RESCPU, are active high and generated when a power-up condition is detected or when the reset button on the front of the computer is pressed.

The RESET signal is used as a general system reset, while the RESCPU signal resets the 80286 Processor. The RES- signal is the input to (Ul7), which signifies a reset condition. The RESsignal is generated either from an RC network during power up or from the reset switch. During a power-up, RES- is held low for the time period generated by the RC time constant of R41 and C62. This is the time it takes C62 to change to an active high. Also, if the reset switch is pressed, it applies a ground to C62 and discharges it. This asserts RES- low until C62 charges again to a logic high.

During power-up, the RES- signal is generated twice to provide a proper reset to the CPU control IC. A second RC time constant is generated by R27 and C83 to the input of a schmidt trigger inverter (U14). This holds the input pin of U 26 ( $\frac{1}{2}$ of a 74LS74) high for approximately 150 ms . When RESCPU is negated after the first reset, the CPU issues the first command to the CPU control IC by driving SI- low. This generates the first rising edge of DMACLK, which latches into U26. The $Q$ output of $U 26$ is then routed to an open collector inverter, (U6), which discharges C62, again asserting RES- low and generating the second reset. CRI provides a reset to U 26 when C 62 is discharged to at least. 7 v . After U26 is reset, RES- is released, and C62 is allowed to charge, negating RES- and finishing the second reset pulse. When the CPU issues the first command again, which starts DMACLK, U26 latches a low. This is because the $D$ input of U 26 has transitioned to a low by the end of the second reset.

The CPU Control IC (U17) also internally controls the RESCPU signal to meet the requirements of the 80286 during a detected shutdown condition.

## Wait State and Ready Logic

Wait state control is implemented internally to the CPU Control IC. The function of the wait state control logic is to match the speed of the various devices in the Tandy 1000 TL to the speed of the 80286 CPU. Two circuits assert wait states within memory and I/o cycles. One method is controlled by the device being accessed, using the IOCHRDY signal input to the CPU Control IC. If a device requires additional wait states within the bus cycle, the device should negate IOCHRDY low until it can service the bus cycle. After the required number of wait states have been inserted, the device should assert IOCHRDY, causing the READYoutput of the CPU Control IC to be asserted low, which tells the CPU to terminate the cycle.

The second method (internal to the CPU Control IC) is several default wait states during the various bus cycles. During a 16bit memory cycle (which is determined by the assertion of AFl6-), one wait state is automatically inserted. The default of an 8-bit memory or $1 / 0$ cycle is four wait states. This can be overridden by driving IOCHRDY low as mentioned above. As long as IOCHRDY is at a logic low level, wait states are inserted indefinitely.

Note: IOCARDY should not be held low for longer than 15 usec because it will stop DRAM refresh cycles.

NMI- Logic

In the Tandy 1000 TL , the Non-Maskable Interrupt (NMI- indicates an I/O error condition and Numerical Math Coprocessor 80287 condition signal (NMI-). Both error conditions are being enabled by the NMI- signal, which is generated from the System data bus bit SD7, PAL CUl, and I/O write (IOW-. The INT287- signal (from the CPU Control IC) becomes active when the ERROR- signal is asserted by the Numerical Math Coprocessor.

## 80287 Control Logic

Incorporated into the CPU Control IC is the logic required to interface the 80287 Math Coprocessor to the 80286 CPU. This logic decodes the signals that select and reset the 80287 and also handles the Busy/(Error-) signals from the 80287 to the CPU.

The input signal 287 CS - is a user I/O address decoded signal used by the CPU Control IC to generate the control signals to the 80287 IC. The 287CS- signal is asserted during I/O address 0FOh 0FFh. Further decoding is provided by the CPU Control IC, which generates RES287, NPCS-, and BUSY287- signals. Table 7 defines the internal decode.

| Hex Address | Description |
| :--- | :--- |
| 0 F0 | Clear Math Coprocessor Busy |
| 0 Fl | Reset Math Coprocessor Busy |
| OF8-0FF |  |

Table 7. 287CS- Decode.

Given the command to perform a task, the 80287 coprocessor issues a BUSY- signal to the CPU Control IC. With the assertion of the BUSY287- output, this signal is passed to the CPU. Normally the BUSY- input is passed through to the BUSY287- output. Deassertion of BUSY- results in deassertion of BUSY287-. The BUSY287- output is latched, and the INT287- output pin is forced HIGH during this busy period if the ERROR- input becomes active (signaling a Numerical processor error). Until cleared by an I/O write cycle to address $0 F O h$ or $0 F 1 h$, both signals then remain active. Both the interrupt latch for INT287- and the busy latch for BUSY287are cleared after a system reset.

The CPU Control IC's RES287 output pin handles resetting of the 80287 coprocessor. This can be activated by a system reset or an I/O write to address 0Flh. This signal is active only for the period of time that the source signal is active, as it is not latched internally.

## CPU Address Buffers

(U30), (U43), and (U27) 74ALS573 implement buffering of the address lines to the system. SAO1-SAl9 are buffered and latched for the expansion bus slots and I/O peripherals. ALE is used to latch SAOl-SA19 and held for the complete bus cycle. SAOl-SAl9 are also used to address the BIOS ROMS and DRAM/DMA Control. Al7A23 are routed directly to DRAM/DMA control to generate memory address decoded signals. The multiplexed address lines MA0-MA7 are also generated and buffered to the DRAM memory by the DRAM/DMA Control IC. To meet address requirements for the DRAMS, the MUX signal multiplexes SAl-SAl6 internally to the DRAM/DMA Control IC.

During a DMA cycle, a 74 LS 245 (U43), is used to buffer S0-S7 directly from the DMA controller. S8-Sl6 are buffered internally to the DRAM/DMA controller. S17-S19 are buffered by AU7 ( $\frac{1}{2}$ of a 74ALS244).

## Data Buffers and Conversion Logic

The 82A205 IC (U61) provides the data buses, buffers, and drivers for D0-D15 to the system. Three data buses are generated, SD0-SD7 for the expansion bus slots, MDO-MD15 for memory access from ROM and DRAM, and D0-D15, which is routed to the 80286 CPU and 80287 Coprocessor data bus. The direction and control of the data buffers are provided by the input signals to the 82A205 IC (DT/R, DSDEN0-, DSDEN1-, SBHE-, and SA0).

DT/R controls the direction of the data path during a read or write. The DSDEN0- and DSDENl- signals control the word and byte data transfers, while SBHE- and SAO determine the buffers to be enabled during a byte access.

Conversion logic is also implemented in the 82A205 IC, controlled by ENHLB- and DIRHLB. This conversion logic allows data to be transferred from the lower to upper or upper to lower data byte to meet the requirements of the CPU or receiving device.

## I/O Decode

Two plsl73 IFLs accomplish the I/O Address decoding. These two ICs provide all the necessary chip select signals to the system. The A, B, and C output signals of the PLSI73 IFL (U42) are encoded device select lines that are fed directly to the PLSl73 IFL (U44), in which I/O address decoding is generated. The second PLS173 (U44) decodes the system address to generate the other I/O address decode select signals. Refer to the Tandy 1000 TL I/O Map for details.

## Floppy Disk Controller

The on-board Floppy Disk Controller (FDC) and KFIT custom IC interface the system to the Floppy Disk Drive (FDD). Up to three floppy disk drives can be supported.

The FDC circuit can be organized into the following subsections:

- uPD765A FDC Chip
- System Interface
- Clock Generation
- Precompensation
- Data Separator
- Disk Drive Interface
uPD765A Chip. The uPD765A FDC chip (U23) integrates most of the control logic necessary to:
. interface the Serial bit stream to or from the FDD to the parallel bus of the system
- implement the commands necessary to operate the FDD
. maintain information about the status of the FDD
During a read or write data operation to the FDD, the FDC chip generates a DMA request for a byte transfer to or from memory. The FDC chip continues to generate DMA requests until the preprogrammed amount of data is transferred as signified by generation of a Termination Count (TC) Signal. After the TC is reached, the FDC chip generates an interrupt to the system through INT so that status and result data can be serviced.

Refer to the device data sheet for complete descriptions of the available commands and the command and status registers.

System Interface. Various ICs, along with the KFIT custom IC, latch and buffer data to and from the system. A DOR Write (Digital Output Register) is generated on an $1 / 0$ write to port 3F2 (hex). This signal latches the data byte that is bit defined as the Drive Select, DSO-, DS1-, and DS2-, Motor On, MTRON-, DMA, (FDCDRQ), Interrupt Request (FDCINT), and a reset signal (FDCRST-) to the FDC controller U23.

Clock Generation. The FDC Support IC (Ul5) generates all clocks required by the Floppy Disk circuit. These clocks are derived from a 16 MHz input signal. FDCCLK, required by the FDC Controller (U23), is derived by dividing the 16 MHz clock by 4. The resulting 4 mHz clock is also used as a delay counter for the DMA request signal DRQ as well as a reference clock for the write precompensation circuit. The 4 MHz clock also generates a 250 nanosecond pulse at a frequency of 500 KHz . The 500 KHz signal is used as a write clock for the FDC Controller.

Precompensation. The precompensation circuit is implemented internally to the FDC Support IC (Ul5). The write data bit can be shifted either early or late in the serial bit stream, depending on the requirements of the Floppy Disk Drive. This function is programmable and controlled by the FDC IC signals PSO and PSI.

Data Separator. The FDC Support IC (Ul5) also contains the data separator circuit. The data separator recovers the clock and data signals from the serial bit stream of the Floppy Disk Drive. The FDC Support IC supports only MFM or Double-density mode.

Disk Drive Interface. All FDC outputs to the FDD are driven by high current open collector buffers inside the KFIT custom IC. All FDC inputs from the FDD are buffered by 74 HCTl 4 SCHMIDT triggered inverters. The inputs are pulled up on-board by lif ohm terminating resistors. All outputs should be terminated on the last FDD by lk ohm resistors.

## Interrupt Controller

The Interrupt Controller is contained in the KFIT custom IC (Ul3) and supplies the maskable interrupt input to the CPU. The KFIT custom IC has eight interrupt inputs controlled through software commands. It can mask (disable) and prioritize (arrange priority) to generate the interrupt input to the CPU. The eight interrupts are assigned as follows:

| \#0 | Timer Channel 0 |
| :--- | :--- |
| \#1 | Keyboard |
| $\# 2$ | Interrupt on the Bus |
| \#3 | Interrupt on the Bus |
| $\# 4$ | Interrupt on the Bus |
| $\# 5$ | Interrupt on the Bus |
| $\# 6$ | Floppy Disk Controller |
| $\# 7$ | Printer |

```
Software Timer
Keyboard Code Received
optional Bus Interrupt
Modem (COM2)
RS-232 (COMI)
Hard Disk Controller/ Vertical Sync
Optional Bus Interrupt optional Bus Interrupt
```

Interrupts 0 and 1 are connected to system board functions as indicated in the chart. Interrupts 2-7 are connected directly to the Expansion Bus, with the normal assigned functions listed in the chart.

## Video Controller

The next major block of the Tandy 1000 TL is the video interface circuitry. This custom part contains all the logic necessary to generate an IBM-compatible color video display. The video interface logic consists of the 100 -pin custom video circuit (U19), four 64K X 4 DRAMS (U32, U33, U34, and U35), a 74LS244 buffer (U8), and associated logic for generating composite and RGBI video.

The Tandy 1000 TL video interface circuitry controls 128 K of memory. This DRAM is shared by the CPU and the video. Normally, the video requires only 16 K or 64 K for the video screen, and the remainder of the 128 K is available for system memory use.

The Tandy 1000 TL video interface custom circuit is composed of a 6845 equivalent design, dynamic RAM address generation/timing, and video attribute controller logic.

Normal function of the video interface custom circuit is as follows. After the 6845 is programmed with a correct set of operating values, a 6:l multiplexer generates the address inputs to the dynamic RAMs. This MUX switches between video (6845) address and CPU address as well as between row and column address. Also, the video interface chip provides the RAM timing signals and generates a wait signal, VIDWT-, to the CPU for proper synchronization with the video RAM access cycles.

The outputs from the RAM chips are connected only to the video interface custom circuit, so all CPU read/write operations are buffered by this part. During a normal display cycle, video data from the RAM chips is first latched in the video Attribute latch and the Video Character latch. The video interface requires a memory organization of 64 K X 16 and latches 16 bits of memory during each access to RAM. From the output of the two latches, the data is supplied to the character ROM for the alpha modes or to the shift registers for the graphics modes. A final 2:1 MUX switches between foreground or background in the alpha modes.

From the 2:1 MUX, the RGBI data is combined with the PC color select data and latched in the Pre-Palette latch. This latch synchronizes the RGBI data before it is used to address the Palette. The palette mask MUX switches between incoming RGBI data and the palette address register. During a CPU write to the palette, this address register selects one of the 16 palette locations. Also, the palette mask MUX allows any of the input RGBI bits to be set to zero.

The palette allows the 16 colors to be remapped in any desired organization. Normally, the Palette is set for a $1: 1$ mapping (red = red, blue = blue, and so on) for PC compatibility. However, instantly changing the on-screen colors is a powerful tool for animation or graphics programs.

After the palette, the RGBI data is resynchronized in the post Palette register. The final logic before the RGBI data is buffered off the chip is the Border mUX. This MUX allows the Border to be replaced with any color selected by the border color latch. This latch is normally disabled in PC modes, but it is used in all PC jr modes.

## Timer

The final Tandy 1000 TL function, other than $I / O$, is the timer found in the KFIT custom IC (Ul3). This part is composed of three independent programmable counters. The clock for all three counters is 1.1925 MHz , which is derived from $14 \mathrm{MHz} / 12$. Counters 0 and 1 are permanently enabled. Counter 2 is controlled by port Hex 0061 , Bit 0 . Counter 0 is connected to system interrupt 0 and is used for software timing functions. Counter 1 is used for refresh function timing. Counter 2 is connected to the sound circuit and also to port Hex 0062, Bit 5.

## Joystick Interface

The joystick interface contained in the PSSJ custom IC (Ull) converts positional information from hand-held joysticks (l or 2) into CPU data. Each joystick provides one or two push-buttons and $X, Y$ position for a total of four bits each. You can use two joysticks.

The joystick handle is connected to two potentiometers mounted perpendicular to each other; one for $X$ position, one for $Y$ position. Through the cable, the main logic board applies +5 VDC to one side and ground to the other of the pots. The pot wiper is the position signal: a voltage between 0 and +5 VDC. This signal is applied to one input of a comparator HU2. The other comparator input is the reference signal (a ramp between 0.0 to +5.0 volts).

When the position signal is equal to or less than the reference signal, the comparator output goes true. This comparator output is the $X$ or $Y$ position data bit. The ramp is reset to 0.0 VDC whenever an I/O Write is made at port 200/201 Hex. The joystick information is "read" by the CPU at Port 200/201 Hex through Ull. See the Joystick Block Diagram.

## Keyboard Interface

The next $1 / 0$ function of the Tandy 1000 TL is the Keyboard interface custom circuit, part of the KFIT custom IC. The heart of this custom part is several read/write registers that are used to control the keyboard interface logic. the keyboard connector, a 164- type shift register is used to load the serial data and allow the CPU to read it as 8 parallel bits.

## Sound Out Circuit

The sound circuit is one of the five I/O functions of the Tandy 1000 TL. The circuit provides sound output for the internal speaker as well as for an external sound circuit.

The main source of sound in the Tandy 1000 TL is the PSSJ custom IC (Ull). It contains the equivalent of a 76496 complex sound generator. This device has three tone generators and one white noise generator. Each tone generator can be programmed for frequency and attenuation. Also, this device has an audio input pin connected to the gated output of timer channel 2. This audio input signal is mixed with the sound generator signal and supplied to the audio output pin.

From the output of the 76496 , the sound signal is connected to a dual analog multiplexer. The multiplexer is switched by Port 61, Bit 4, and turns off the audio signal to the speaker, headphone jacks, and external audio output. The output of the multiplexer is routed to audio amplifiers U71 for the internal speaker and headphone jacks. The volume of the internal speaker can be adjusted by a user-accessible volume control (R59). When the headphone jack is used, the internal speaker is disabled.

Sound In Circuit

An additional feature of the Tandy 1000 TL sound circuitry is a Digital to Analog Converter (DAC). The DAC is controlled by read/write Ports C4-C7. The DAC can be used to convert prerecorded digital sound, voice, or music into analog audio output. A microphone jack and audio input circuitry are provided for recording analog sound, voice or music, and converting it to digital data. A set of jumpers is provided with the sound circuit to allow selection between microphone and direct line input. Jumpers Ell to El2 select microphone input, and jumpers E12 to El3 select direct line input. The audio input signal is fed through an amplifier, U72, and increased by a multiple of 100, before being sent to the AUDIOIN pin of the PSSJ custom IC. When direct line input is selected, the audio input signal is reduced in amplitude by resistors R57A and R56A before being sent to amplifier U72. Bit programming data for these ports is available in the data sheets on the 8079021 Custom IC located in the "Devices" section of this manual.

## Real-time Clock

The Real-time Clock in the Tandy 1000 TL system is the Dallas Semiconductor DSl215. Its input is derived from a 32.768 KHz crystal CYl. When system power is removed, operation of the DSl215 is maintained by a $3 v$ battery.

The DS1215 is capable of operating in both 12 -Hour mode, with an AM/PM indicator, or $24-$ Hour mode. The real-time clock is initialized by reading a consecutive 64 -bit stream with SA2 high and the ROMCS- signal asserted. Then another 64 -bit read or write may be done to set, read, or update the clock. The 64 -bit pattern is programmed into 88 -bit registers in the DS1215. All 8 registers must be programmed at the same time, and must be sent from Register 0 to Register 7. The DSl2l5 is automatically disabled after the 64 -bit stream is read. Register and bit definition for the DSl2l5 can be found in the data sheets on the DS1215 in the "Devices" section of this manual.

## DMA Controller

The major components of the Direct Memory Access (DMA) circuit consists of an 8237A-5 DMA controller (Ull), the DRAM/DMA Control (U53), and a bi-directional address buffer 74ALS245 (U3).

Initialization--A DMA Operation. When a DMA operation is requested by software or by a peripheral through a DREQ line, the 8237A-5 DMA controller initiates a Bus Hold Request to the 80286 CPU through the CPU Controller IC. The CPU Controller arbitrates the CPU Hold Request from the DMA controller to the CPU.

When the CPU acknowledges the Hold request, the CPU control, address, and data lines are tri-stated. The CPU Controller controls the direction and enables the memory or peripheral address and data buses that correspond to the requested DMA operation.

During the DMA operation, the 8237A-5 acts as the bus master and, along with the CPU Controller IC, generates all bus control signals and address and data signals. The DMA transfers continue for the number of counts and to the destination address that was previously programmed into the DMA registers. See the device data sheet and the IO map for complete descriptions of the registers, their locations, and their functions.

DMA Bus cycles. During the data bus cycle, the 8237A-5 first outputs the upper address (A8-A15) on its data outputs (XD0-XD7), to be latched in the buffer internally to the DRAM/DMA Control by the Address Strobe signal (AS) from the 8237A-5. Next, the lower address (A0-A7) is put directly on the $S$ address bus by the 8237A-5.

The DMA request acknowledge signals, DACK2 and DACK3-, are used along with RFRSH- and ACK* to enable the page register to be output as the upper address (SAl6 and Al7 through A23), which are buffered by U 22 to the system expansion slots.

A DMA bus cycle can be extended by the RDY input of the DMA controller. The DMA memory read DMAMR is routed to the CPU Controller IC for extending the DMA bus cycle by inserting on DMA clock period as a wait state. The CPU Controller inserts the wait state by controlling the DMARDY input of the DMA controller.

I/O devices can extend the DMA bus cycle by controlling the IOCHRDY signal of the expansion bus. Setup times must be observed for IOCHRDY to be recognized.

## RS-232 Serial Port Interface

The RS-232 Port is a single-channel, asynchronous communications port. The heart of the serial port is the PSSJ custom IC that functions as a serial data input/output interface. It performs serial-to-parallel conversion on data characters received from a peripheral device or modem and parallel-to-serial conversion on data characters received from the CPU.

Status information reported includes the type and condition of the ACE's transfer operations as well as any error conditions detected during serial data operations. The pSSJ custom IC includes a programmable Baud Rate Generator that allows operation from 50 to 9600 Baud. The PSSJ custom IC is supplied with a clock of 14 MHz from oscillator (Y2). The PSSJ can be tailored to the user's requirements by being able to remove start bits, stop bits, and parity bits. It supports $5,6,7$, or 8 data bit characters with 1 , $1 \frac{1}{2}$, or 2 stop bits. Diagnostic capabilities provide loopback functions of transmit/receive and input/output signals.

The PSSJ serial port is programmed by selecting the I/O address $3 F 8$ - 3FE hex for primary and 2FB-2FE hex for secondary and writing data out to the port. Address bits AO, Al, and A2 are used to define the modes of operation by selecting the different registers to be programmed or read.

One interrupt is provided to the system from IRQ4 for primary operation and IRQ3 for secondary operation. This interrupt is active high. Bit 3 of the modem control register must be set high in order to send interrupts to the system. When this bit is high, any interrupts allowed by the interrupt enable register cause an interrupt.

## Parallel Printer Port Interface

The final I/O interface of the Tandy 1000 TL is the printer Interface contained in the PSSJ custom IC (Ull). This part supplies all the signals required to interface to a typical parallel printer. These signals are 8 data out lines, plus various handshake control signals. Also, the printer interface generates an interrupt to the CPU if enabled.

## Expansion Ports

## System Expansion Bus

This section identifies the $1 / 0$ interface requirements for the 8bit, PC-compatible option cards. Each of the five slots has a 62pin connector socket.

The following connector pin assignment is used on the PC option slots; this connector socket has 62 pins.

| Pin | Signal Name | I/0 | Pin | Signal Name | I/0 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Al | NMI- |  |  |  |  |
| A2 | SD7 | 1/0 | B2 | BRESET | 0 |
| A3 | SD6 | 1/0 | B3 | +5V | POWER |
| A4 | SD5 | I/0 | B4 | IRQ2 | I |
| A5 | SD4 | 1/0 | B5 | -5v | POWER |
| A6 | SD3 | I/O | B6 | FDCDRQ | I |
| A7 | SD2 | 1/0 | B7 | -12V | POWER |
| A8 | SD1 | 1/0 | B8 | N/C |  |
| A9 | SD0 | I/O | B9 | +12V | POWER |
| Al0 | IOCHRDY | I | Bl0 | GND | GROUND |
| All | AEN | 0 | B11 | SMEMW- | 0 |
| A12 | SA19 | 0 | B12 | SMEMR- | 0 |
| Al3 | SAl8 | 0 | B13 | IOW- | 0 |
| Al4 | SA17 | 0 | Bl4 | IOR- | 0 |
| Al5 | SA16 | 0 | Bl5 | DACK3- | 0 |
| 416 | SAl5 | 0 | B16 | DRQ3 | 1 |
| 117 | SAl4 | 0 | B17 | DACKI- | 0 |
| 118 | SAl3 | 0 | B18 | DRQ1 | 1 |
| 419 | SAl2 | 0 | B19 | RFRSH- | 0 |
| A20 | SAll | 0 | B20 | SCLK | 0 |
| 121 | SAl0 | 0 | B21 | IRQ7 | I |
| A22 | SA9 | 0 | B22 | IRQ6 | I |
| A23 | SA8 | 0 | B23 | IRQ5 | I |
| A24 | SA7 | 0 | B24 | IRQ4 | I |
| 425 | SA6 | 0 | B25 | IRQ3 | I |
| A26 | SA5 | 0 | B26 | FDCDACK- | 0 |
| A27 | SA4 | 0 | B27 | T/C | 0 |
| A28 | SA3 | 0 | B28 | BALE | 0 |
| A29 | SA2 | 0 | B29 | +5v | POWER |
| A30 | SAl | 0 | B30 | OSC | 0 |
| A31 | SAO | 0 | B31 | GND | GROUND |

## Expansion Bus Signal Description

The following signal descriptions for the System I/O Bus are for PC bus-compatible option cards. Note that all signal lines are TTL compatible levels and that I/O adapters should be designed with a maximum of two low power Shottky (LS) loads per line.

SCLK (B20). SCLK is the System clock and has a period of 125 ns in 8 MHz mode, or 250 ns in 4 MHz mode. It has a $50 \%$ duty cycle and is used only for synchronization with the CPU. It is not intended for uses requiring a fixed frequency.

SAO through SAl9 (Al2-A31). These lines are 20 address bits used to address memory and I/O devices within the Tandy 1000 TL. They are gated on the system bus when the BALE signal is high and are latched on the falling edge of the BALE signal. Generation of these signals is accomplished by the CPU or a DMA controller. SAO-SAl9 are active high.

BALE (B28). BALE is a Buffered Address Latch Enable generated by the CPU Control IC. It is used to latch valid addresses from the CPU, and can be used by an I/O board to indicate a valid CPU address, in conjunction with AEN. BALE is pulled to a high state during DMA cycles, which include Refresh cycles. BALE is active high.

AEN (All). AEN is an Address Enable signal used to remove the CPU and other devices from the bus to allow DMA transfers to take place. During AEN active, the DMA controller has control of the address bus, the data bus, the READ command lines, and the WRITE command lines. AEN is active high.

SDO through SD7 (A2-A9). These signals are the data bus Bits 0 through 7 from the CPU to memory and I/O devices on the bus. SDO is the least significant bit (lsb), and SD7 is the most significant bit (msb).

BRESET (B2). BRESET is used to reset or initialize the expansion logic during power-up time, line voltage outage, or when the Reset switch on the front panel is pressed. BRESET is active high.

NMI- (Al). This signal indicates an uncorrectable system error when active. The NMI- signal provides the system board with parity information about memory or devices on the bus. NMI- is active low.

IOCHRDY (AIO). This signal is used to lengthen $I / O$ or memory cycles when driven low by the active device. (This signal should not be held low more than 15 microseconds.) Any slow device using this line should drive it low immediately upon detecting its valid address and a READ or WRITE command. See the timing diagram for setup times. IOCHRDY is active high (Ready condition).

IRQ2 through IRQ7 (B4, B21-B25). These signals are used to tell the CPU that an $I / O$ device needs attention. The Interrupt Requests are prioritized with IRQ2 having the highest priority and IRQ7 the lowest. An Interrupt Request is generated when any IRQ signal is driven high and held high until the CPU acknowledges the interrupt.

IOR- (Bl4). IOR- is a read signal that instructs an I/O device to drive its data onto the data bus (SDO-SD7). This line can be driven by the CPU Control IC or by the DMA controller. IOR- is active low.

IOW- (B13). IOW- is a write signal that instructs an $I / O$ device to read, or latch, the data from the data bus (SD0-SD7). This line can be driven by the CPU Control IC or by the DMA controller. IOW- is active low.

SMEMR- (B12). SMEMR- is a read signal that instructs a memory device to drive its data onto the data bus (SD0-SD7). This line can be driven by the CPU Control IC or by the DMA controller through the CPU Control IC. SMEMR- is active only when the memory address is within the first 1 megabyte range (000000-0FFFFFH). SMEMR- is active low.

SMEMW- (BII). SMEMW- is a write signal that instructs a memory device to read, or latch, the data from the data bus (SD0-SD7). This line can be driven by the CPU Control IC or by the DMA controller through the CPU Control IC. SMEMW- is active only when the memory address is within the first 1 megabyte range (000000OFFFFFH). SMEMW- is active low.

DRQ1, FDCDRQ, and DRQ3 (B18, B6, B16). These lines are asynchronous DMA requests by peripheral devices to gain DMA service. They are prioritized with DRQ1 having the highest priority, FDCDRQ next, and DRQ3 lowest. A DMA request is generated by driving a DRQ line active high and holding it until the corresponding DACK (DMA acknowledge) signal goes active. DROL FDCDRQ, and DRQ3 perform only 8-bit transfers. All DRQ lines are active high.

DACK1-, FDCACK-, and DACK3-, B26, B15). These lines are DMA acknowledge signals used to acknowledge DMA requests DRQ1, FDCDRQ, and DRQ3. All DACK signals are active low.

RFRSH- (B19). This signal is used to indicate a refresh cycle that can be used by a memory board to refresh Dynamic memory. RFRSH- is active low and generated every 15 usec.

T/C (B27). T/C is a signal that provides a pulse when the terminal count for any DMA channel is reached. T/C is active high.

OSC (B30). OSC is an oscillator signal that is a high-speed clock with a 70 nanosecond period (14.31818 megahertz). It has a 50\% duty cycle.

Memory Map

## Memory Map

| Address | Name | Allocated Function |
| :---: | :---: | :---: |
| 000000-07FPFP | 512K System RAM | System Memory |
| 080000-09FFFF | 128K System/Video RAM or 128 K Expansion Memory | System Memory and Video Display Memory or System Memory |
| OA0000-0BFFFF | 128K Video RAM | Reserved for Graphics Display Memory |
| $\begin{gathered} \text { 0E0000-0FFFFF, } \\ \text { EE0000-EFFFFF, } \\ \text { or } \\ \text { FE } 0000-\mathrm{FFFFFF} \end{gathered}$ | 16K BIOS ROM Memory | Reserved for BIOS ROM Memory |

I/O Port Map of System

## I/O Port Map Summary

| Block | Usage | Function |
| :---: | :---: | :---: |
| 0000-001F | 0000-000F | DMA Function |
| 0020-003F | 0020-0027 | Interrupt Controller |
| 0040-005F | 0040-0047 | Timer |
| 0060-007F | 0060-0067 | PIO Function |
| 0080-009F | 0080-008F | DMA Page Register |
| 00A0-00BF | 00A0-00A7 | NMI- Mask Register |
| 00C0-00DF | 00C0-00c7 | Sound Generator |
| 00E0-00FF | 00F0-00FF | Numerical Coprocessor |
| 0100-01FF |  | Reserved |
| 0200-020F | 0200-0207 | Joystick Interface |
| 0210-02F7 |  | Reserved |
| 02F8-02FF | 02F8-02FF | Serial Port Secondary (COM2 optional) |
| 0300-031F |  | Reserved |
| 0320-032F |  | Hard Disk Controller (optional) |
| 0330-036F |  | Reserved |
| 0370-0377 | 0370-0377 | Floppy Disk Controller 2 (optional) |
| 0378-037F | 0378-037F | Printer |
| 0380-03CF |  | Reserved |
| 03D0-03DF | 03D0-03DF | System Video |
| 03E0-03EF |  | Reserved |
| $03 \mathrm{FO}-03 \mathrm{F7}$ | 03F0-03F7 | Floppy Disk Controller 1 |
| 03F8-03FF | 03F8-03FF | Serial Port primary (COMl) |
| 0400-FFE7 |  | Not Used |
| FFE8-FFEF | FFE8-FFEF | System Control Registers |
| FFF0-FFFF |  | Reserved |


| Address | must do JMP $\$+2$ when programming DMA contraller |
| :---: | :---: |
| 0000 | DMA Controller |
|  | IOW- = 0: Channel 0 Base and Current Address |
|  | Internal Flip/Flop $=0$ : Write A0-A7 |
|  | Internal Flip/Flop = 1: Write A8-Al5 |
|  | IOR- = 0: Channel 0 Current Address |
|  | Internal Flip/Flop $=0$ : Read A0-A7 |
|  | Internal Flip/Flop $=1$ : Read A8-Al5 |
| 0001 | DMA Controller |
|  | IOW- = 0: Channel 0 Base and Current Word Count |
|  | Internal Flip/Flop $=0$ : Write W0-W7 |
|  | Internal Flip/Flop = 1: Write AW-Wl5 |
|  | IOR- = 0: Channel 0 Current Word Count |
|  | Internal Flip/Flop $=0$ : Read w0-w7 |
|  | Internal Flip/Flop = 1: Read w8-W15 |
| 0002 | DMA Controller |
|  | IOW- = 0: Channel 1 Base and Current Address |
|  | Internal Flip/Flop $=0$ : Write A0-A7 |
|  | Internal Flip/Flop = l: Write A8-Al5 |
|  | IOR- = 0: Channel 1 Current Address |
|  | Internal Flip/Flop $=0$ : Read A0-A7 |
|  | Internal Flip/Flop = 1: Read A8-Al5 |
| 0003 | DMA Controller |
|  | IOW- = 0: Channel 1 Base and Current Word Count |
|  | Internal Flip/Flop $=0$ : Write w0-w7 |
|  | Internal Flip/Flop $=1:$ Write AW-Wl5 |
|  | IOR- = 0: Channel 1 Current Word Count |
|  | Internal Flip/Flop $=0$ : Read W0-W7 |
|  | Internal Flip/Flop = 1: Read w8-Wl5 |
| 0004 | DMA Controller |
|  | IOW- = 0: Channel 2 Base and Current Address |
|  | Internal Flip/Flop $=0$ : Write A0-A7 |
|  | Internal Flip/Flop = 1: Write A8-Al5 |
|  | IOR- = 0: Channel 2 Current Address |
|  | Internal Flip/Flop $=0$ : Read A0-A7 |
|  | Internal Flip/Flop $=1$ : Read A8-Al5 |
| 0005 | DMA Controller |
|  | IOW- = 0: Channel 2 Base and Current word Count |
|  | Internal Flip/Flop $=0$ : Write W0-W7 |
|  | Internal Flip/Flop = 1: Write AW-Wl5 |
|  | IOR- = 0: Channel 2 Current Word Count |
|  | Internal Flip/Flop $=0$ : Read W0-W7 |
|  | Internal Flip/Flop = 1: Read W8-Wl5 |

## Description

| 0006 | DMA Controller <br> IOW- $=0$ : Channel 3 Base and Current Address <br> Internal Flip/Flop $=0$ : Write A0-A7 <br> Internal Flip/Flop $=1$ : Write A8-Al5 <br> IOR- = 0: Channel 3 Current Address <br> Internal Flip/Flop $=0$ : Read A0-A7 <br> Internal Flip/Flop $=1$ : Read A8-Al5 |
| :---: | :---: |
| 0007 | DMA Contrcller <br> IOW- = 0: Channel 3 Base and Current Word Count <br> Internal Flip/Flop $=0$ : Write W0-W7 <br> Internal Flip/Flop = 1: Write AW-W15 <br> IOR- = 0: Channel 3 Current Word Count <br> Internal Flip/Flop $=0$ : Read w0-w7 <br> Internal Flip/Flop = 1: Read W8-Wl5 |
| 0008 | DMA Controller <br> IOW- $=0$, Write Command Register |

Bit Description
$0 \quad 0=$ Memory to Memory Disable $1=$ Memory to Memory Enable

I $0=$ Channel 0 Address Hold Disable 1 = Channel 0 Address Hold Enable
$X$ If Bit $0=0$
$20=$ Controller Enable
1 = Controller Disable
$3 \quad 0=$ Normal Timing
$1=$ Compressed Timing
$\mathrm{X} \quad$ If Bit $0=1$
$4 \quad 0=$ Fixed Priority $1=$ Rotating Priority
$5 \quad 0=$ Late Write Selection $1=$ Extended Write Selection x If Bit $3=1$
$6 \quad 0=$ DREQ Sense Active High 1 = DREQ Sense Active Low
$70=$ DACK Sense Active Low $1=$ DACK Sense Active High



Address Description

```
000F DMA Controller
    IOW- = 0, Write All Mask Register Bits
    Bit Description
    0 0 = Clear Channel 0 Mask Bit (Enable)
    l = Set Channel O Mask Bit (Disable
    l 0 = Clear Channel 1 Mask Bit (Enable)
    l = Set Channel l Mask Bit (Disable
    2 0 = Clear Channel 2 Mask Bit (Enable)
        l = Set Channel 2 Mask Bit (Disable
        3 O Clear Channel 3 Mask Bit (Enable)
        l = Set Channel 3 Mask Bit (Disable
    4-7 Don't Care
    IOR- = 0, Illegal
```

0010-001F Reserved
0020 8259A Interrupt Controller
Note: Initialization Words are set up by the operating
system and are generally not to be changed.
Writing an initialization word might cancel
pending interrupts.
Bit $4=1$ Initialization Command Word 1
Bit $0=0$ ICW4 Needed
$=1$ ICW 4 Not Needed
Bit $1=0$ Cascade Mode
$=1$ Single Mode
Bit 2 Not Used
Bit $3=0$ Edge Triggered Mode
= 1 Level Triggered Mode
Bit 5-7 Not Used
Bit $4=0 \&$ Operation Control Word 2

Description
0020 8259A Interrupt Controller
Bit 3 = 0 Bits 0-2: Determine the Interrupt Level Acted on when the SL Bit is Active

```
Interrupt Level = 0 1 2 3 4 5 6 7
Bit 0 (LO): 0 l 0 1 0 1 0 1
Bit l (Ll): 0 0 1 l 0 0 l 1
Bit 2 (L2): 0 0 0 0 1 1 1 1
```

Bits 5-7: Control Rotate and End of Interrupt Modes
B7 B6 B5
$0 \quad 1$ Non-Specific EOI Command End of Interrupt
011 Specific EOI Command End of Interrupt
101 Rotate on Non-Specific EOI Auto Rotation
100 Rotate in Automatic EOI Mode (Set) Auto Rotation
000 Rotate in Automatic EOI Mode (Clear) Auto Rotation
111 *Rotate on Specific EOI Command Specific Rotation
110 *Set Priority Command Specific Rotation
010 No Operation
(*LO - L2 Are Used)
Bit $4=0$ \& Operation Control Word 3
Bit $3=1$ Bits $0-1$ :
Bit 1 Bit 0 - Read Register Command
$0 \quad 0$ No Action

01 No Action
$1 \quad 0 \quad$ Read IR Register on next IOR- Pulse
$1 \quad 1$ Read IS Register on next IOR- Pulse
Bit $2=0$ : No Poll Command = 1: Poll Command

Bits 5-6

| Bit 5 | Bit 6 | Special Mask Mode |
| :---: | :---: | :---: |
| 0 | 0 | No Action |
| 0 | 1 | No Action |
| 1 | 0 | Reset Special Mask |
| 1 | 1 | Set Special Mask |

Bit $7=0$

| Address | Description |
| :---: | :---: |
| 0021 | 8259A Interrupt Controller |
|  | Initialization Control Word 2 |
|  | Bits 0-7: Not Used |
|  | Bits 3-7: T3-T7 of Interrupt Vector Address (8086/8088/80286 Mode) |
|  | Initialization Control Word 3 (Master Device) |
|  | $\begin{aligned} \text { Bits } \begin{aligned} 0-7: & =1 \\ & =0 \quad \text { Indicated IR Input has a Slave } \\ & \text { Indicated IR Input does not have a } \\ & \text { Slave } \end{aligned} \quad . \quad l \end{aligned}$ |
|  | Initialization Control Word 3 (Slave Device) |
|  | Bits 0-2: $=$ ID0-2 |
|  | Bit 0 Bit 1 Bit 2 - Slave ID \# |
|  | 000 |
|  | $\begin{array}{llll}0 & 0 & 1\end{array}$ |
|  | $\begin{array}{llll}0 & 1 & 0 & 2\end{array}$ |
|  | $\begin{array}{llll}0 & 1 & 1\end{array}$ |
|  | 1000 |
|  | 1005 |
|  | $1 \begin{array}{lll}1 & 1 & 0\end{array}$ |
|  | $1 \begin{array}{llll}1 & 1 & 1 & 7\end{array}$ |
|  | Bits 3-7: $=0$ (Not Used) |
|  | Initialization control word 4 |
|  | Bit 0: Type of processor |
|  | $\begin{aligned} & =0 \text { MCS }-80 / 85 \text { Mode } \\ & =18086 / 8088 / 80286 \text { Mode } \end{aligned}$ |
|  | Bit 1: Type of End of Interrupt |
|  | $=0 \text { Normal EOI }$ |
|  | $=1 \text { Auto EOI }$ |
|  | Bits 2-3: Buffering Mode |
|  | Bit 3 Bit 2 |
|  | 0 X Non-Buffered Mode |
|  | 10 Buffered Mode/Slave |
|  | 111 Buffered Mode/Master |

```
Address Description
    Initialization Control Word 4
    Bit 4: Nesting Mode
    =0 Not Special Fully Nested Mode
    =1 Special Fully Nested Mode
    Bits 5-7:
    =0 (Not Used)
    Operation Control Word l (IOR-)
    Bits 0-7: Interrupt Mask for IRQ0-IRQ7
    =0 Mask Reset (Enable)
    =1 Mask Set (Disable)
    Note: Peripherals requesting an interrupt service must
        generate a low to high edge and then remain at a
        logic high level until service is acknowledged.
        Failure to do so results in a Default Service for
        IRQ7.
0022-0027 Same as 0020-0021
0028-003F Not Used
0040 8254-2 Timer
    IOW- = 0: Load Counter No. 0
    IOR- = 0: Read Counter No. 0
0041 8254-2 Timer
    IOW- = 0: Load Counter No. l
    IOR- = 0: Read Counter No. l
0042 8254-2 Timer
    IOW- = 0: Load Counter No. 2
    IOR- = 0: Read Counter No. 2
```

```
    Address Description
    0043 8254-2 Timer
    IOW- = 0: Write Mode Word
    Control Word Format
    Bit 0: BCD
    =0: BCD Counter (4 Decades)
    =1: Binary Counter 16 Bits
    Bits 1-3: Mode Selection
    Bit 3 Bit 2 Bit 1
\begin{tabular}{lllll}
0 & 0 & 0 & Mode 0 \\
0 & 0 & 1 & Mode 1 \\
\(X\) & 1 & 0 & Mode 2 \\
\(X\) & 1 & 1 & Mode 3 \\
1 & 0 & 0 & Mode 4 \\
1 & 0 & 1 & Mode 5
\end{tabular}
    Bits 4-5: Read/Load
    Bit 5 Bit 4
        0 0 Counter Latching Operation
        0 1 Read/Load LSB Only
        1 0 Read/Load MSB Only
        1 1 Read/Load LSB First, Then MSB
    Bits 6-7: Select Counter
    Bit 7 Bit 6
        0 Select Counter 0
        0 1 Select Counter 1
        1 0 Select Counter 2
        1 I Illegal
    IOR- = 0: No-Operation 3-State
0044-0047 Same as 0040-0043
0048-005F Not Used
```

```
Address Description
0060 Port A / Keyboard Interface Control Ports
    (Read Only)
    Bit Description
        0 Keyboard Bit 0-LSB
        1 Keyboard Bit 1
        2 Keyboard Bit 2
        3 Keyboard Bit 3
        4 Keyboard Bit 4
        5 Keyboard Bit 5
        6 Keyboard Bit }
        7 Keyboard Bit 7-MSB
0061 Port B Read or Write
    Bit Description
    0 8253 Gate 2 Enable
        1 = Enable
        0 = Disable
    1 Speaker Data Out Enable
        l = Enable
        0 = Disable
    2 Not Used
    N Not Used
    4 Internal Speaker Enable
        l = Disable
        0 = Enable
    N Not Used
    6 HOLDCK (if IBM PC Keyboard Mode)
        l = Tristate Keyboard Clock Line
        O = Pull Keyboard Clock Line Low
    7 Keyboard Clear
    l = Clear Buffer and Reset Keyboard Interrupt
    0 = Release Clear and Reset
```


## Address Description

```
0062 Port C Read/Write: Bits 0-3; Read Only Bits: 4-7
Bit Description
    0 Not Used (Read/Write)
    l Not Used (Read/Write)
    2 Not Used (Read/Write)
    3 (Output) CPU Clock Rate (Read/Write)
        0=4.00 MHz (PC Compatible Rate)
        l = 8.00 MHz (Default By Boot ROM)
    4 EEPROM Data Input (Read Only)
    5 8253 Out #2 (Read Only)
    6 Monochrome Mode
        0 = Color Monitor
        l = 350 Line Monitor, Mono
    Reserved
```

0063-0064 Reserved
0065 Planar Control Register (Read/Write)
1 = Enable
$0=$ Disable
Bit Description
not really. Always reads osinor
dion, retests bit 3
0 Hard Disk Select Enable
1 Parallel Port Select Enable
only.
2 Video Port Select Enable
3 Floppy Disk Port Select Enable
4 Serial Port Select Enable
5 Reserved
6 Reserved
7 Parallel Port Output Enable
0066-0067 Reserved
0068-007F Reserved

0080 DMA Page Register (Reserved for Diagnostics) Write only

```
0081 DMA Channel 2 Page Register (Write Only)
    Address Description
    Bit 0 Address Al6
    Bit 1 Address Al7
    Bit 2 Address Al8
    Bit 3 Address Al9
    Bit 4 Address A20
    Bit 5 Address A21
    Bit 6 Address A22
    Bit 7 Address A23
0082 DMA Channel 3 Page Register (Write Only)
    Address Description
    Bit 0 Address Al6
    Bit 1 Address Al7
    Bit 2 Address Al8
    Bit 3 Address Al9
    Bit 4 Address A20
    Bit 5 Address A21
    Bit 6 Address A22
    Bit 7 Address A23
0 0 8 3 ~ D M A ~ C h a n n e l ~ 0 - 1 ~ P a g e ~ R e g i s t e r ~ ( W r i t e ~ O n l y )
    Address Description
    Bit 0 Address Al6
    Bit 1 Address Al7
    Bit 2 Address Al8
    Bit 3 Address Al9
    Bit 4 Address A20
    Bit 5 Address A2l
    Bit 6 Address A22
    Bit 7 Address A23
0084-008F Same as 0080-0083
```



```
Address Description
00C4-00C7 DAC Function
00C4 DAC Mode Register (Write Commands)
    Bits 0-1
    Bit 0 Bit 1 
    Bit 2 DMA Enable (for SA, Direct R/W) OM PSSTFor
        O DMA Disabled
            1 DMA Enabled for SA, DA
    Bit 3 DMA Interrupt Clear
        0 DMA Interrupt Held Clear
            I DMA Interrupt Allowed
Bit 4 DMA Interrupt Enable
    O DMA EOP Interrupt Disabled
    l DMA EOP Interrupt Enabled
Bit 5 Sound Divider Sync Enable
    0 Synchronization Disabled
    l Synchronization Enabled (Write to, 00C6 or
        00C7 reloads all dividers)
    Bit 6 Sound Chip Extra Divide Enable
    O Extra Divide Disabled
    1 Extra Divide Enabled
Bit 7 Reserved
00C4 DAC MOde Register (Read Commands)
Bit 3 DMA Interrupt Flag. A DMA Interrupt has
        occurred. To clear the interrupt flag, Bit 3
        must be brought low and then high again.
Bit }7\mathrm{ Successive Approximation Done. Useful when
        polling instead of using DMA.
```


## Address Description



## Address Description

$00 \mathrm{C} 7 \mathrm{R} / \mathrm{W}$ Amplitude/frequency msb for DAC sound channel

| Bit | 0 | F8 |
| :--- | :--- | :--- |
| Bit | 1 | F9 |
| Bit | 2 | Fl0 |
| Bit | 3 | Fll |
| Bit | 4 | Reserved |
| Bit | 5 | AMP |
| Bit | 6 | AMP |
| Bit | $\mathbf{2}$ | AMP |
| Bit | 7 | AMP |

00C8-00CF Reserved
00E0-00EF Reserved

| 00F0 | Clear Numerical Coprocessor Busy |
| :--- | :--- |
| 00F1 | Reset Numerical Coprocessor to Real Mode |
| 00F2 | Same as 00F0 |
| 00F3 | Same as 00F1 |
| $00 \mathrm{F4}$ | Same as 00F0 |
| 00F5 | Same as 00F1 |
| 00F6 | Same as 00F0 |
| 00F7 | Same as 00F1 |
| 00 FB -00FF | Math Coprocessor Chip Select |

0100-01FF Reserved
0200-0207 Joystick
Clear (Resets Integrator to 0)
0201 Read $\mathrm{R}=$ Right Joystick, $\mathrm{L}=$ Left Joystick
Bit Description
$0 \quad \mathrm{R}$ - X Horizontal Position
$1 \quad \mathrm{R}$ - Y Vertical Position
2 L - X Horizontal position
3 L - Y Vertical position
4 R Button \#l (Logic $0=$ Button Pressed)
5 R Button $\# 2$ (Logic $0=$ Button Pressed)
6 L Button \#l (Logic $0=$ Button Pressed)
7 L Button ${ }^{\#} 2$ (Logic $0=$ Button Pressed)
0208-020F Not Used
0210-02F7 Reserved

```
Address Description
2F8-2FF Serial Port Secondary (COM2 Optional)
0300-036F Reserved
0370-0377 Floppy Disk Controller 2 (optional)
0 3 7 8 ~ P r i n t e r ~ - ~ D a t a ~ L a t c h ~
    Bit Description
    0 Bit 0 - LSB
    1 Bit 1
    2 Bit 2
    3 Bit 3
    4 Bit 4
    5 Bit 5
    Bit 6
    7 Bit 7 - MSB
0379 Printer - Read Status
    Bit Description
        O Not Used
        1 Not Used
        2 Not Used
        3 "O" = Error (Fault)
    4 "l" = printer Select In
    5 "0" = Out of Paper (Paper Empty)
    6 "0" = Acknowledge
    7 "0" = Busy
037A Printer - Control Latch
    Bit Description
        0 "0" = Strobe
        1 "O" = Auto FD XT
        2 "0" = Initialize
        3 "O" = Select Printer Out
        4 "1" = Enable Interrupt
        5 "O" = Enable Output Data
        6 ~ N o t ~ U s e d ~
        7 \text { Not Used}
037B Not Used
```


## Address Description

037C EEPROM Control Register (Write Only)
Bit Description
0 NOVDO - EEPROM Data Output
1 NOVCE - EEPROM Chip Enable
$0=$ Enabled or Selected
$l=$ Disabled or Deselected
2 NOVCLK - EEPROM Clock
$0=$ Toggle Clock Low
$1=$ Toggle Clock High
037D-037F Reserved
0378-03CF not used $\$$ some $3 B x$ used for mono video (q.v.l
03D0-03D3 Reserved
03D4 6845 Address Register
03D5 6845 Data Register
03D6 Not Used
03D7 Not Used
03D8 Mode Select Register
Bito High Resolution Clock
$=0$ Selects $40 \times 25$ Alphanumeric Mode
$=1$ Selects $80 \times 25$ Alphanumeric Mode
Bitl Graphics Select
$=0$ Selects Alphanumeric Mode
$=1$ Selects $320 \times 200$ Graphics Mode
Bit2 Black and white
= 0 Selects Color Mode
= l Selects Black and White Mode
Bit3 video Enable
= 0 Disables Video Signal
$=1$ Enables Video Signal
Bit4 640 Dot Graphics
= 0 Disables $640 \times 200$ B\&W Graphics Mode
$=1$ Enables $640 \times 200$ B\&W Graphics Mode
Bit5 Blink Enable
= O Disables Blinking
$=1$ Enables Blinking



Address Description

| $03 \mathrm{F6}$ | Reserved |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 03F7 | FDC Data Rate Selection |  |  |  |  |  |  |
|  | Bit Description |  |  |  |  |  |  |
|  | 0 Not Used |  |  |  |  |  |  |
|  | 1 |  | Write - Data Rate |  |  |  |  |
|  |  |  | $0=500 \mathrm{~K}$ bits per second |  |  |  |  |
|  |  |  | $1=250 \mathrm{~K}$ bits per second |  |  |  |  |
|  | 2 |  | Not Used |  |  |  |  |
|  | 3 |  | Not Used |  |  |  |  |
|  | 4 |  | Not Used |  |  |  |  |
|  | 5 |  | Not Used |  |  |  |  |
|  | 6 |  | Not Used |  |  |  |  |
|  | 7 |  | $0=$ Disk Change |  |  |  |  |
| 03F8-03FF | Serial Port Primary ( $C O M 1$ ) |  |  |  |  |  |  |
| 03 F 8 | Write Transmitter Holding Register (Character to Send) |  |  |  |  |  |  |
|  | Bit Description |  |  |  |  |  |  |
|  | 0 | Bit 0 - LSB (First Bit sent Serially) |  |  |  |  |  |
|  | 1 |  |  |  |  |  |  |
|  | 2 | Bit 2 |  |  |  |  |  |
|  | 3 | Bit 3 |  |  |  |  |  |
|  | 4 | Bit 4 |  |  |  |  |  |
|  | 5 | Bit 5 |  |  |  |  |  |
|  | 6 | Bit 6 |  |  |  |  |  |
|  | 7 | Bit 7 - MSB |  |  |  |  |  |
| $03 F 8$ | Read | Receiver Buffer Register (Character Received) |  |  |  |  |  |
|  | Bit | Description |  |  |  |  |  |
|  | 0 | Bit 0 - LSB (First Bit Received Serially) |  |  |  |  |  |
|  | 1 | Bit 1 |  |  |  |  |  |
|  | 2 | Bit 2 |  |  |  |  |  |
|  | 3 | Bit 3 |  |  |  |  |  |
|  | 4 | Bit 4 |  |  |  |  |  |
|  | 5 | Bit 5 |  |  |  |  |  |
|  | 6 | Bit ${ }^{\text {bit }} 7$ - MSB |  |  |  |  |  |
|  | 7 |  |  |  |  |  |  |




```
Address Description
03FE Modem Status Register
    Bit Description
    0 Delta clear to Send (DCTS)
    1 DeIta Data Set Ready (DDSR)
    2 Trailing Edge Ring Indicator
        "1" = On
        "0" = Off
    3 Delta Received Line Signal Detect (If Bit 0, 1, 2,
        or 3 is set to a "l" modem status interrupt is
        generated
    4 "0" = Clear to Send (CTS)
    5 "0n = Data Set Ready (DSR)
    6 "0" = Ring Indicator (RI)
    7 "0" = Received Line Signal Detect (RLSD)
03FF Reserved
0400-FFE7 Not Used
FFE8-FFEB System Control Registers
FFE8 Video Configuration Register (urite-only)
    Bit Description
```



```
    2 Memory Configuration 2 M Mrónably 1006 w/o
4 Reserved
5 llor-Bit CPU Memory = 1 (aust be X)
6 Reserved
R Reserved
FFE9 Reserved
```

Address Description
FFEA WRITE/READ
Bit Description
Bit 0: ROM PAGING 0
Bit 1: ROM PAGING 1
Bit 2: ROM PAGING 2
Bit 3: ROM PAGING 3
Bit 4: ROM PAGING 4
Bit 5: System Type (Reserved - See Note)
6-7 Bit 7 Bit 6
$0 \quad 0 \quad 512 \mathrm{~K}$ System Memory
$0 \quad 1 \quad 512 \mathrm{~K}$ System Memory
$1 \quad 0 \quad 512 \mathrm{~K}$ System Memory

NOTE: When reading Port FFEA, Bit 4 will be inverted from what was written, (ie. when a 0 is written, a 1 will be read; when a 1 is written, a 0 will be read.) FFEA Bit 4 can be used to determine the system type. If Bit 4 is read back inverted, the system is identified as a mandy 1000 SL. If Bit 4 is NOT inverted, the system is identified as a mandy 1000 TL.

Writing 1 to bit 4 disables access to the RON segment at E0000.
Writing $A$ to bit 6 or 7 , with the 1286 vide
RAll upgrade installed, disables access to memory between SD000-9FFFF.

ROM Paging Definition:




$$
\text { have this: } 512 \mathrm{kPam} \text { on } 2256 \mathrm{k} \text { chips }=864 \mathrm{k} \text { pages. }
$$

| FFEB | UART Clo | , Joystick, and Sound Enable |
| :---: | :---: | :---: |
|  | Bit | Description |
| $\cdots$ | 0 | $0=$ Clock Divided by 13 |
|  |  | 1 = Clock Divided by 1 |
|  | 1 | $0=$ Disable Joystick |
|  |  | 1 = Enable Joystick |
|  | 2 | $0=$ Disable Sound Chip |
|  |  | 1 = Enable Sound Chip |
|  | 3 | Reserved |
|  | 4 | Reserved |
| - | 5 | Keyboard Interrupt (Read only) |
|  |  | $1=$ Keyboard Interrupt Active |
|  |  | $0=$ Keyboard Interrupt Inactive |
|  | 6 | Reserved |
|  | 7 | Keyboard Type (Read Only) |
|  |  | $1=101$ Key Enhanced Keyboard |
|  |  | $0=$ Original Tandy 1000 Keyboard |
| FFEC-FFEF | Reserved |  |
| FFPO-FFFF | Reserved |  |

ELECTRICAL BILL OF MATERIAL - TANDY 1000 TL

MAIN LOGIC ASSY.

| QTY. | DESCRIPTION | DESIGNATOR | VENDOR | PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| 1 | CPU PCB TANDY 1000 | TL REV. A |  | 8709842 |
| 8 | STAKING PINS | E1-8 | AMP\#1-87022-0 | 8529014 |
| 1 | SOCKET 8-PIN DIP | U12 | AMP\#640460-3 | 8509011 |
| 17 | SOCKET 16-PIN DIP | U3,45-52,62-69 | AMP\# 2-644100-3 | 8509036 |
| 8 | SOCKET 18-PIN DIP | U32-39 | AMP \# 2-383060-3 | 8509037 |
| 3 | SOCKET 24-PIN DIP | U15,42,44 | AMP\#640962-3 | 8509029 |
| 1 | SOCKET 28-PIN DIP | U24 | AMP\#2-641605-3 | 8509007 |
| 2 | SOCKET 32-PIN DIP | U54,56 OR 55,57 | AMP * 2-644018-3 | 8505048 |
| 2 | SOCKET 32-PIN DIP | U58,59 | AMP\#2-644018-3 | 8505048 |
| 3 | SOCKET 40-PIN DIP | U23,28,60 | AMP\# 2-641606-3 | 8509002 |
| 6 | SOCKET 68-PIN, PLCC | $\begin{aligned} & \text { U11,13,17,22, } \\ & 31,61 \end{aligned}$ | AMP\#821574-1 <br> BURNDY\#68P410T | 8509020 |
| 1 | SWITCH, RESET | Sl | ALPS\#KHC15901 | 8489065 |
| 1 | STEREO HEAD. JACK | J14 | HOSIDEN\# HSJ094 2-01-1020 | 8519322 |
| 1 | MICROPHONE JACK | J16 | HOSIDEN\#HSJ0942-01-1060 | 8519355 |
| 1 | BATTERY HOLDER | J13 | SPECIALTY ELEC. 2S2032-0. | 8491013 |
| 1 | CONNECTOR, 2-PIN | J15 (SPKR) | MOLEX\#22-29-2021 | 8519193 |
| 2 | CONNECTOR, 6-PIN | J3,4 (JOYSTICK) | HOSIDEN\#TCS5040-16-1911 | 8519318 |
| 1 | CONNECTOR, 7-PIN | J2 (KEYBOARD) | HOSIDEN\#TCS5040-17-4071 | 8519358 |
| 1 | CONNECTOR, 9-PIN | J12 (POWER) | MOLEX\#26-48-1095 | 8519191 |
| 1 | CONNECTOR, 9-PIN <br> FEMALE "D" SUB | Ul (VIDEO) | HOLMBERG\#H4S09RA28CM42 <br> AMP\#745988-3 <br> MOLEX\#82009-2052 | 8519279 |
| 1 | $\begin{aligned} & \text { CONNECTOR, } 9-\text { PIN } \\ & \text { "D" SUB } \end{aligned}$ | J5 (SERIAL) | AMP\#747840-3 | 8519269 |
| 1 | CONNECTOR, 17-PIN HEADER (SHROUDED) | Jll (FLOPPY) | MOLEX\# 70246-3402 | 8519324 |
| 5 | CONNECTOR, 31-PIN CARD EDGE | J6-10 | TEKA*021-31014-200 <br> VIKING\#3KT31/2JFFl2 <br> JAE\#PB21-62T1-Cl-S <br> AMP ${ }^{*}$ 6-530843-5 <br> HOLMBERG\#A8D31DS27C2 <br> BURNDY*PWBH31DDSIB | 8519236 |
| 1 | RESISTOR, VAR. 10K | R63 | PIHER PTI5NH510KA | 8270510 |
| 1 | RESISTOR, VAR. TBD | R27 |  |  |
| 1 | RESISTOR 10 OHM | R48 | GENERIC | 8207010 |
|  | 1/4 WATT 5\% |  |  |  |
| 15 | RESISTOR 33 OHM | R6-9,17,21,22, | " | 8207033 |
|  | 1/4 WATT 5\% | 31,32,34-39 |  |  |
| 1 | RESISTOR 300 OHM | R50,49 | " | 8207130 |


| MAIN LOGIC ASSY. |  |  |  | TANDY 1000 TL |
| :---: | :---: | :---: | :---: | :---: |
| QTY. | DESCRIPTION |  |  | PART |
|  |  | DESIGNATOR | VENDOR | NUMBER |
| 1 | RESISTOR 510 OHM | R24 | " | 8207151 |
|  | 1/4 WATT 5\% |  |  |  |
| 2 | RESISTOR 1K OHM | R5,29 | GENERIC | 8207210 |
|  | 1/4 WATT 5\% |  |  |  |
| 3 | RESISTOR 1.2K OHM | R54,55,57 | " | 8207212 |
|  | 1/4 WATT 1\% |  |  |  |
| 1 | RESISTOR 1.2K OHM | R26 |  |  |
|  | 1/4 WATT 18 |  |  |  |
| 1 | RESISTOR 1.3K OHM | R58 | " | 8207213 |
|  | 1/4 WATT 5\% |  |  |  |
| 1 | RESISTOR 2.2 K OHM | R51 | " | 8207222 |
|  | 1/4 WATT 5\% |  |  |  |
| 1 | RESISTOR 2.4K OHM | R60 | " | 8207244 |
|  | 1/4 WATT 5\% |  |  |  |
| 1 | RESISTOR 2.7K OHM | R62 | " | 8207227 |
|  | 1/4 WATT 5\% |  |  |  |
| 1 | RESISTOR 3.3K OHM | R52 | " | 8207233 |
|  | 1/4 WATT 5\% |  |  |  |
| 4 | RESISTOR 4.7K OHM | R33,41,42,61 | " | 8207247 |
|  | 1/4 WATT 5\% |  |  |  |
| 9 | RESISTOR 10 K OHM | R1-4, 30, 40, 45-47 | " | 8207310 |
|  | 1/4 WATT 5\% |  |  |  |
| 2 | RESISTOR 13K OHM | R53,56 | " | 8207313 |
|  | 1/4 WATT 5\% |  |  |  |
| 5 | RESISTOR 27 K OHM | R12,13,20,23,25 | " | 8207327 |
|  | 1/4 WATT 5\% |  |  |  |
| 3 | RESISTOR 47K OHM | R14,19,43 | " | 8207347 |
|  | 1/4 WATT 5\% |  |  |  |
| 1 | RESISTOR 68K OHM | R28 | n | 8207368 |
|  | 1/4 WATT 5\% |  |  |  |
| 1 | RESISTOR 91K OHM | R59 |  | 8207391 |
|  | $1 / 4$ WATT 5\% |  |  |  |
| 2 | RESISTOR 100K OHM | R15,44 | " | 8207410 |
|  | 1/4 WATT 5\% |  |  |  |
| 4 | RESISTOR 1 M MEG OHM | R10,11,16,18 | " | 8207510 |
|  | 1/4 WATT 5\% ${ }^{\text {R }}$ ( 33 OHM |  | * |  |
| 6 | $16-\mathrm{PIN} \text { DIP }$ | RP1-4,12,13 |  | 8290044 |
| 5 | RESISTOR PAK 33 OHM | RP5,14,15,18,19 | $\cdots$ | 8295033 |
|  | 8-PIN SIP |  |  |  |
| 2 | RESISTOR PAK 1K OHM | RP7,10 | " | 8290210 |
|  | 6-PIN SIP |  |  |  |

ELECTRICAL BILL OF MATERIAL - TANDY 1000 TL
 MAIN LOGIC ASSY.

| QTY. | DESCRIPTION | DESIGNATOR | VENDOR | PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| 2 | RES. PAK 4.7K OHM 10-PIN SIP | RP6, 17 | GENERIC | 8294247 |
| 3 | RES.PAK 10 K OHM 6-PIN SIP | RP9,11,20 | " | 8290032 |
| 2 | $\begin{aligned} & \text { RES. PAK } 10 K \text { OHM } \\ & \text { 8-PIN SIP } \end{aligned}$ | RP8,16 | " | 8292310 |
| 2 | RES. PAK 10K OHM 10-PIN SIP | RP21, 22 | " | 8290010 |
| 45 | CAPACITOR 220 PFD 25V 10\% SMD | $\begin{aligned} & C 3,4,5,8,9, \\ & 10,11,12,13, \\ & 14,15,17,19, \\ & 20,21,22,23, \\ & 24,25,26,27, \\ & 28,29,30,31, \\ & 32,33,34,38, \\ & 39,40,41,42, \\ & 43,44,45,46, \\ & 47,48,49,50, \\ & 51,52,176,180 \end{aligned}$ | " | X30122243 |
| 7 | CAPACITOR 330 PFD 25v 10\% SMD | $\begin{aligned} & C 6,7,18,56, \\ & 57,58,59 \end{aligned}$ | " | X30133244 |
| 9 | CAPACITOR 47 PFD 50V 5\% NPO CD | $\begin{aligned} & C 70,75,76,78, \\ & 83,90,94,96, \\ & 98 \end{aligned}$ | n | 8300475 |
| 2 | $\begin{aligned} & \text { CAP } 1000 \mathrm{PFD} \\ & 50 \mathrm{~V} 10 \% \mathrm{z} 5 \mathrm{P} \text { CD } \end{aligned}$ | C169,171 | " | 8302104 |
| 5 | $\begin{aligned} & \text { CAP } .022 \mathrm{MFD} \\ & 50 \mathrm{~V} \text { 25C } \\ & +80 /-20 \% \text { CD } \end{aligned}$ | C65,69,95,77,111 | " | 8303224 |
| 1 | $\begin{aligned} & \text { CAP } 047 \mathrm{MFD} \\ & 50 \mathrm{~V} \text { 10\% } 25 \mathrm{P} \end{aligned}$ | Cl63 | " | 8373474 |
| 1 | CD OR EQUIV. <br> CAP . 047 MFD <br> 50V 10\% X7R <br> MONO. AX. | Cl31 | " | 8373484 |

ELECTRICAL BILL OF MATERIAL - TANDY 1000 TL

| MAIN | LOGIC ASSY. |  | TANDY 1000 TL |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| QTY. DESCRIPTION |  | DESIGNATOR | VENDOR |  |  | PART NUMBER |
| 50 | CAP 0.1 MFD 50V | Cl, 16,35,53, | GENERIC |  |  | 8374104 |
|  | +80/-20\% Z5U | 62,66,72,73, |  |  |  |  |
|  | MONO. AX. | 74,82,87,88, |  |  |  |  |
|  |  | 89,91,93,97, |  |  |  |  |
|  |  | 99,101,103, |  |  |  |  |
|  |  | 107,108,109, |  |  |  |  |
|  |  | 110,111,113, |  |  |  |  |
|  |  | 114,116,117, |  |  |  |  |
|  |  | 126,127,128, |  |  |  |  |
|  |  | 129,130,140, |  |  |  |  |
|  |  | 141,144,145, |  |  |  |  |
|  |  | 154,155,157, |  |  |  |  |
|  |  | 158,159,160, |  |  |  |  |
|  |  | 161,165,166, |  |  |  |  |
|  |  | 175,177,178, |  |  |  |  |
|  |  | 179 |  |  |  |  |
| 24 | CAP . 33 MFD | Cl18,119,120, | " |  |  | 8374334 |
|  | $50 \mathrm{~V}+80 /-20 \% \mathrm{Z} 5 \mathrm{U}$ | 121,122,123, |  |  |  |  |
|  | MONO. AX. | 124,125,132, |  |  |  |  |
|  |  | 133,134,135, |  |  |  |  |
|  |  | 136,137,138, |  |  |  |  |
|  |  | 139,146,147. |  |  |  |  |
|  |  | 148,149,150, |  |  |  |  |
|  |  | 151,152,153 |  |  |  |  |
| 3 | CAP . 47 MFD 50V CD | C64,182,183 | " |  |  | 8384475 |
| 13 | CAP RFI TBD | C2,36,37,81, | " |  |  |  |
|  |  | 92,100,102, |  |  |  |  |
|  |  | 104,105,142, |  |  |  |  |
|  |  | 143,164,112 |  |  |  |  |
| 3 | CAP 3.3 MFD | C168,170,172 | " |  |  | 8335322 |
| 9 | CAP 10 MFD | C60,61,63,68, | " |  |  | 8326101 |
|  | 16V ELEC.RAD. 20\% | 79,80,173,174, |  |  |  |  |
|  |  | 181 |  |  |  |  |
| 8 | CAP 22 MFD | C54,55,84,85, | " |  |  | 8326221 |
|  | 16V ELEC. RAD. $20 \%$ | 86,115,156,162 |  |  |  |  |
| 1 | CAP 47 MFD | C71 | " |  |  | 8326474 |
|  | 16V ELEC.RAD. 20\% |  |  |  |  |  |
| 1 | CAP 100 MFD | C167 | " |  |  | 8327101 |
|  | 16V ELE.RAD. |  |  |  |  |  |
| 5 | EMI FILTER . 022 UF | CFl, 3, 4,5,6 | MURATA | \#DST310 | 55D-223S | 8418013 |
|  | W/FER. BEAD |  |  |  |  |  |
| 1 | EMI CAP. . 022 | CF2 | MURATA | \#DS310 | 55D-223S | 8418014 |
|  | (NO BEAD) |  |  |  |  |  |
| 8 | FERRITE BEAD | FB2-5,7-10 | FARRITE | \#27430 | 02121 | 8419013 |
| 4 | FERRITE BEAD (801) | FBl, 6,11,12 | FARRITE | \#26430 | 00801 | 8419098 |

ELECTRICAL BILL OF MATERIAL - TANDY 1000 TL


## MAIN LOGIC ASSY.

| $\frac{\text { QTY. }}{1}$ | DESCRIPTION | DESIGNATOR | VENDOR | PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
|  | FUSE - 1 AMP <br> (CANADA ONLY) | F1 |  | 8749045 |
| 2 | DIODE 1N4148 | CR1, 2 |  | 8150148 |
| 2 | REGULATOR 78L05 | VR2,3 | FAIRCHILD UA78L05 |  |
|  |  |  |  |  |
|  |  |  |  |  |
| 1 | REGULATOR 79M0A5CT | VR1 | MOTOROLA MC79M05CT <br> FAIRCHILD UA7905 <br> TEXAS INST. UA79M05CKC | 8190005 |
|  |  |  |  |  |
|  |  |  |  |  |
| 1 | DUAL OSC. <br> 28.63636/32.514 MHZ | Y3 | DIAWA, MF | 8409076 |
|  |  |  |  |  |
| 1 | DUAL OSC. 24/16 MHZ | Y2 | DIAWA, MF | 8409075 |
|  |  |  |  |  |
| 1 | CRYSTAL 32.768 KHZ | Y1 |  | 8404033 |
| 1 | IC 7416 | U6 | GENERIC | 8000016 |
| 1 | IC 74LS04 | U16 | " | 8020004 |
| 1 | IC 74LS32 | U29 | " | 8020032 |
| 1 | IC 74LS74 | U26 | $\square$ | 8020074 |
| 1 | IC 74LS244 | U8 | MOTOROLA | 8020244 |
| 2 | IC 74ALS244 | U9, U25 | GENERIC | 8025244 |
| 1 | IC 74ALS245 | U40 | , | 8025245 |
| 3 | IC 74ALS573 | U27,30,43 | - | 8025573 |
| 1 | IC $74 \mathrm{FO8}$ | U20 | " | 8015008 |
| 1 | IC $74 \mathrm{HCTOO*}$ | U18 | " | 8026000 |
| 1 | IC 74HCT14* | U14 | " | 8026014 |
| 1 | IC 74HCT32* | U41 | " | 8026032 |
| 1 | IC 74HCT244* | U53 | $\cdots$ | 8026244 |
| 2 | IC 74HCT273* | U21,70 | " | 8026273 |
| 1 | IC LM339 | U7 | " | 8050339 |
| 1 | IC LM386 | U71 | " | 8050386 |
| 1 | IC MCl458 | U72 |  | 8051458 |
| 1 | IC MCl458S | U10 |  | 8052458 |
| 1 | IC MC1488 | U2 | MOTOROLA | 8050188 |
| 2 | IC MCl489 | U4,5 | MOTOROLA | 8050189 |
| 1 | IC TANVID 2100 PIN | U19 | NCR | $\times 07900100$ |

[^0]| QTY. | DESCRIPTION | DESIGNATOR | VENDOR | PART NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| 1 | CPU TANDY 1000 TL SUB A | ASSY. (REV. A) |  | 8859022 |
| 2 | JUMPER PLUG | E2-E3,E6-E7 |  | 8519098 |
| 1 | IC 802868 MHZ CPU | U31 | INTEL, AMD | 8041286 |
|  | C,E STEP |  |  |  |
| 1 | IC DATA BUFFER * | U61 | MOTOROLA, FUJITSU | 8079022 |
| 1 | IC CPU CONTROLLER <br> (NT. BLUE) | U17 | VLSI, RCA | 8040810 |
| 1 | IC DRAM/DMA CONTROLLER | U22 | MOTOROLA, RCA | 8040142 |
| 1 | IC PSSJ | Ull | NCR | 8079021 |
| 1 | IC KFIT | Ul3 | NCR | 8079019 |
| 1 | IC 8237A-5 | U28 | INTEL, AMD, NEC | 8040237 |
| 1 | IC 8272A FDC CONT. * | U23 | INTEL, ROCKWELL, <br> ZILOG | 8040272 |
| 1 | IC FDC SUPPORT | 015 | MOTOROLA, NCR | 8041401 |
| 1 | IC DSl215 REAL TIME CLK | U3 | DALLAS SEMI. | 8079023 |
| 1 | IC EEPROM $64 \times 169346$ | U12 | NATIONAL, HYUNDAI, AMI/GOULD | 8040346 |
| 16 | IC $256 \mathrm{~K} \times 1 . \mathrm{DRAM} 150 \mathrm{NS}$ | U45-52,62-69 | FUJITSU, HITACHI, NEC, TI, SAMSUNG, MICRON | 8049008 |
| 4 | IC $64 \mathrm{~K} \times 4$ DRAM 120 NS | U32-35 | FUJITSU, HITACHI, NEC, TI, SAMSUNG, SHARP, MOTOROLA | 8040464 |
| 1 | IC $256 \mathrm{~K} \times 8$ <br> BIOS ROM EVEN 200NS (1) | U55 | SHARP | 8076323 |
| 1 | IC $256 \mathrm{~K} \times 8$ <br> BIOS ROM ODD 200NS (1) | U57 | SHARP | 8075323 |
| 1 | IC 256K X 8 <br> BIOS ROM EVEN 200NS (1) | U54 | HITACHI | 8079025 |
| 1 | IC $256 \mathrm{~K} \times 8$ <br> BIOS ROM ODD 200NS (1) | U56 | HITACHI | 8079026 |
| 1 | IC $16 \mathrm{~K} \times 8$ CHARA.GEN. ROM 200NS | U24 | HITACHI, <br> SHARP, NEC | 8079027 |
| 1 | IC PLSI73 I/O DECODE | U42 | SIGNETICS | 8077173 |
| 1 | IC PLSI73 ROM CNTL. | U44 | SIGNETICS | 8076173 |

ELECTRICAL BILL OF MATERIAL - PROJECT 878
MAIN LOGIC ASSY.DESIGNATOR VENDORNUMBER
*THE FOLLOWING PARTS MAY BE SUBSTITUTED IF NECESSARY:
1 IC UPD765 U23 NEC 8041765
1 IC 82A205 U61 CHIPS ..... 8075205

MAIN LOGIC TANDY 1000 TL SUB ASSY.

| TTY. | DESCRIPTION | DESIGNATOR | VENDOR | PART <br> NUMBER |
| :---: | :---: | :---: | :---: | :---: |
| JPTION KITS: |  |  |  |  |
| 1 | IC 80287-6 OR 8 | U60 | INTEL |  |
| 4 | IC $64 \mathrm{~K} \times 4$ DRAM | U36-39 | FUJITSU, | 8040464 |
|  | 150 OR 120NS |  | HITACHI, NEC, |  |
|  |  |  | TI, SAMSUNG, |  |
|  |  |  | MICRON, |  |
|  |  |  | SHARP, |  |

















Devices

# High Performance Microprocessor with Memory Management and Protection 

 (80286-12, 80286-10, 80286-8)High Performance Processor (Up to six times 8086)

- Large Address Space: - 16 Megabytes Physical - 1 Gigabyte Virtual per Task
- Integrated Memory Management, FourLevel Memory Protection and Support for Virtual Memory and Operating Systems
- High Bandwidth Bus Interface (12.5 Megabyte/Sec)
- Industry Standard O.S. Support: - IRMX ${ }^{\text {© }}$
- XENIX*
- UNIX*
- MS-DOS*
- Optional Processor Extension: - 80287 High Performance 80-bit Numeric Data Processor
- Two 8086 Upward Compatible Operating Modes:
- 8086 Real Address Mode
- Protected Virtual Address Mode
- Range of Clock Rates
- 12.5 MHz for 80286-12
- 10 MHz for 80286-10
- 8 MHz for $\mathbf{8 0 8 2 6 - 8}$
-6 MHz for $80286-6$
- Complete System Development Support:
- Development Software: Assembler, PL/M, Pascal, FORTRAN, and System Utilities
- In-Circuit-Emulator (ICETM-286)
- Available in 68 Pin Ceramic LCC (Leadless Chip Carrier), PGA (Pin Grid Array), and PLCC (Plastic Leaded Chip Carrier) Packages
(See Packaging Spec., Order *231369)

The 80286 is an advanced, high-performance microprocessor with specially optimized capabilities for multiple user and multi-tasking systems. The 80286 has built-in memory protection that supports operating system and task isolation as well as program and data privacy within tasks. A 12 MHz 80286 provides six times or more throughput than the standard 5 MHz 8086 . The 80286 includes memory management capabilities that map $2^{30}$ (one gigabyte) of virtual address space per task into $2^{24}$ bytes ( 16 megabytes) of physical memory.
The 80286 is upward compatible with 8086 and 88 software. Using 8086 real address mode, the 80286 is object code compatible with existing 8086, 88 software. In protected virtual address mode, the 80286 is source code compatible with 8086, 88 software and may require upgrading to use virtual addresses supported by the 80286's integrated memory management and protection mechanism. Both modes operate at full 80286 performance and execute a superset of the 8086 and 88 instructions.
The 80286 provides special operations to support the efficient implementation and execution of operating systems. For example, one instruction can end execution of one task, save its state, switch to a new task, load its state, and start execution of the new task. The 80286 also supports virtual memory systems by providing a segment-not-present exception and restartable instructions.
*XENIX and MS-DOS are trademarks of Microsoft Corp.
-UNIX is a trademark of Bell Labs or AT\&T


Figure 1. 80286 Internal Block Diagram

Component Pad Views-As viewed from underside of component when mounted on the board.
P.C. Board Vlews-As viewed from the component side of the P.C. board.

N.C. signals must not be connected


Figure 2. 80286 Pin Configuration

Table 1. Pin Description
The following pin function descriptions are for the 80286 microprocessor:

| Symbol | Type | Name and Function |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CLK | 1. | SYSTEM CLOCK provides the fundamental timing for 80286 systems. It is divided by two inside the 80286 to generate the processor clock. The internal divide-by-two circuitry can be synchronized to an external clock generator by a LOW to HIGH transition on the RESET input. |  |  |  |  |
| $\mathrm{D}_{15}-\mathrm{D}_{0}$ | 1/0 | DATA BUS inputs data during memory, 1/0, and interrupt acknowledge read cycles; outputs data during memory and I/O write cycles. The data bus is active HIGH and floats to 3-state OFF during bus hold acknowledge. |  |  |  |  |
| $A_{23}-A_{0}$ | 0 | ADDRESS BUS outputs physical memory and I/O port addresses. AO is LOW when data is to be transferred on pins $\mathrm{D}_{7-0} . \mathrm{A}_{23}-\mathrm{A}_{16}$ are LOW during 1/O transfers. The address bus is active HIGH and floats to 3 -state OFF during bus hold acknowledge. |  |  |  |  |
| BHE | 0 | BUS HIGH ENABLE indicates transfer or data on the upper byte of the data bus. $D_{15-8}$ -Eight-bit oriented devices assigned to the upper byte of the data bus would normally use $\overline{\mathrm{BHE}}$ to condition chip select functions. $\overline{\mathrm{BHE}}$ is active LOW and floats to 3 -state OFF during bus hold acknowledge. |  |  |  |  |
|  |  | BHE and AO Encodings |  |  |  |  |
|  |  | BHE Value | AO Value |  |  | Function |
|  |  | $\begin{aligned} & 0 \\ & 0 \\ & 1 \\ & 1 \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline 0 \\ & 1 \\ & 0 \\ & 1 \\ & \hline \end{aligned}$ | Word transfer <br> Byte transfer on upper half of data bus ( $\mathrm{D}_{15}-\mathrm{D}_{8}$ ) <br> Byte transfer on lower half of data bus ( $\mathrm{D}_{7-0}$ ) <br> Will never occur |  |  |
| $\overline{\mathrm{S} 1}, \overline{\mathrm{~s} 0}$ | 0 | BUS CYCLE STATUS indicates initiation of a bus cycle and, along with M/ $\overline{\mathrm{O}}$ and COD/ $\mathbb{I N T A}$, defines the type of bus cycle. The bus is in a $T_{s}$ state whenever one or both are LOW, $\overline{\mathrm{S} 1}$ and $\overline{\mathrm{S} 0}$ are active LOW and float to 3 -state OFF during bus hold acknowledge. |  |  |  |  |
|  |  | 80286 Bus Cycle Status Definition |  |  |  |  |
|  |  | COD/INTA | M/ $\overline{\mathbf{O}}$ | $\overline{51}$ | $\overline{\mathbf{s} 0}$ | Bus Cycle Initiated |
|  |  | $\begin{array}{\|l\|} \hline 0 \text { (LOW) } \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \\ 1 \text { (HIGH) } \\ 1 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1 \end{array}$ | 0 0 0 0 1 1 1 1 0 0 0 0 1 1 1 1 | $\begin{aligned} & 0 \\ & 0 \\ & 1 \\ & 1 \\ & 1 \\ & 0 \\ & 0 \\ & 1 \\ & 1 \\ & 0 \\ & 0 \\ & 1 \\ & 1 \\ & 1 \\ & 0 \\ & 0 \\ & 1 \end{aligned}$ | $\begin{aligned} & 0 \\ & 1 \\ & 1 \\ & 0 \\ & 1 \\ & 0 \\ & 1 \\ & 0 \\ & 0 \\ & 1 \\ & 0 \\ & 1 \\ & 0 \\ & 1 \end{aligned}$ | Interrupt acknowledge <br> Will not occur <br> Will not occur <br> None; not a status cycle <br> IF A1 $=1$ then halt; else shutdown <br> Memory data read <br> Memory data write <br> None; not a status cycle <br> Will not occur <br> 1/O read <br> 1/O write <br> None; not a status cycle <br> Will not occur <br> Memory instruction read <br> Will not occur <br> None; not a status cycle |
| M/ $\overline{\mathrm{O}}$ | 0 | MEMORY I/O SELECT distinguishes memory access from I/O access. If HIGH during $\mathrm{T}_{\mathrm{s}}$, a memory cycle or a halt/shutdown cycle is in progress. If LOW, an I/O cycle or an interrupt acknowledge cycle is in progress. M/IO floats to 3 -state OFF during bus hold acknowledge. |  |  |  |  |
| COD/ $\overline{N T T A}$ | 0 | CODE/INTERRUPT ACKNOWLEDGE distinguishes instruction fetch cycles from memory data read cycles. Also distinguishes interrupt acknowledge cycles from I/O cycles. COD/ $\overline{\mathbb{N T A}}$ floats to 3 -state OFF during bus hold acknowledge. Its timing is the same as $\mathrm{M} / \overline{\mathrm{IO}}$. |  |  |  |  |
| $\overline{\text { LOCK }}$ | 0 | BUS LOCK indicates that other system bus masters are not to gain control of the system bus for the current and the following bus cycle. The LOCK signal may be activated explicitly by the "LOCK" instruction prefix or automatically by 80286 hardware during memory XCHG instructions, interrupt acknowledge, or descriptor table access. LOCK is active LOW and floats to 3 -state OFF during bus hold acknowledge. |  |  |  |  |
| READY | 1 | BUS READY terminates a bus cycle. Bus cycles are extended without limit until terminated by AEADY LOW. READY is an active LOW synchronous input requiring setup and hold times relative to the system clock be met for correct operation. READY is ignored during bus hold acknowledge. |  |  |  |  |

Table 1. Pin Description (Continued)

| Symbol | Type | Name and Function |
| :---: | :---: | :---: |
| $\begin{aligned} & \text { HOLD } \\ & \text { HLDA } \end{aligned}$ | $\begin{aligned} & 1 \\ & 0 \end{aligned}$ | BUS HOLD REQUEST AND HOLD ACKNOWLEDGE conirol ownership of the 80286 local bus. The HOLD input allows another local bus master to request control of the local bus. When control is granted, the 80286 will float its bus drivers to 3 -state OFF and then activate HLDA, thus entering the bus hold acknowledge condition. The local bus will remain granted to the requesting master until HOLD becomes inactive which results in the 80286 deactivating HLDA and regaining control of the local bus. This terminates the bus hold acknowledge condition. HOLD may be asynchronous to the system clock. These signals are active HIGH. |
| INTR | 1 | INTERRUPT REQUEST requests the 80286 to suspend its current program execution and service a pending external request. Interrupt requests are masked whenever the interrupt enable bit in the flag word is cleared. When the 80286 responds to an interrupt request, it performs two interrupt acknowledge bus cycles to read an 8 -bit interrupt vector that identifies the source of the interrupt. To assure program interruption, INTR must remain active until the first interrupt acknowledge cycle is completed. INTR is sampled at the beginning of each processor cycle and must be active HIGH at least two processor cycles before the current instruction ends in order to interrupt before the next instruction. INTR is level sensitive, active HIGH, and may be asynchronous to the system clock. |
| NMI | 1 | NON-MASKABLE INTERRUPT REQUEST interrupts the 80286 with an internally supplied vector value of 2 . No interrupt acknowledge cycles are performed. The interrupt enable bit in the 80286 flag word does not affect this input. The NMI input is active HIGH, may be asynchronous to the system clock, and is edge triggered after internal synchronization. For proper recognition, the input must have been previously LOW for at least four system clock cycles and remain HIGH for at least four system clock cycles. |
| $\begin{aligned} & \overline{\text { PEREQ }} \\ & \hline \text { PEACK } \end{aligned}$ | $\begin{aligned} & 1 \\ & 0 \end{aligned}$ | PROCESSOR EXTENSION OPERAND REQUEST AND ACKNOWLEDGE extend the memory management and protection capabilities of the 80286 to processor extensions. The PEREQ input requests the 80286 to periorm a data operand transfer for a processor extension. The PEACK output signals the processor extension when the requested operand is being transferred. PEREQ is active HIGH and floats to 3 -state OFF during bus hold acknowledge. PEACK may be asynchronous to the system clock. PEACK is active LOW. |
| $\begin{aligned} & \overline{\overline{B U S Y}} \\ & \overline{\text { ERROR }} \end{aligned}$ | $T$ | PROCESSOR EXTENSION BUSY AND ERROR indicate the operating condition of a processor extension to the 80286. An active BUSY input stops 80286 program execution on WAIT and some ESC instructions until BUSY becomes inactive (HIGH). The 80286 may be interrupted while waiting for BUSY to become inactive. An active ERROR input causes the 80286 to perform a processor extension interrupt when executing WAIT or some ESC instructions. These inputs are active LOW and may be asynchronous to the system clock. These inputs have internal pull-up resistors. |

Table 1. Pin Description (Continued)

| Symbol | Type | Name and Function |  |
| :---: | :---: | :---: | :---: |
| RESET | 1 | SYSTEM RESET clears the internal logic of the 80286 and is active HIGH. The 80286 may be reinitialized at any time with a LOW to HIGH transition on RESET which remains active for more than 16 system clock cycles. During RESET active, the output pins of the 80286 enter the state shown below: |  |
|  |  | 80286 Pin State During Reset |  |
|  |  | Pin Value | Pin Names |
|  |  | $\begin{aligned} & 1 \text { (HIGH) } \\ & 0 \text { (LOW) } \\ & 3 \text {-state OFF } \end{aligned}$ | S̄, S1, PEACK, A23-AO, BHE, LOCK M/IO, COD/INTA, HLDA (Note 1) $D_{15}-D_{0}$ |
|  |  | Operation of the 80286 begins after a HIGH to LOW transition on RESET. The HIGH to LOW transition of RESET must be synchronous to the system clock. Approximately 38 CLK cycles from the trailing edge of RESET are required by the 80286 for internal initialization before the first bus cycle, to fetch code from the power-on execution address, occurs. <br> A LOW to HIGH transition of RESET synchronous to the system clock will end a processor cycle at the second HIGH to LOW transition of the system clock. The LOW to HIGH transition of RESET may be asynchronous to the system clock; however, in this case it cannot be predetermined which phase of the processor clock will occur during the next system clock period. Synchronous LOW to HIGH transitions of RESET are required only for systems where the processor clock must be phase synchronous to another clock. |  |
| $V_{S S}$ | 1 | SYSTEM GROUND: 0 Volts. |  |
| $V_{\text {CC }}$ | 1 | SYSTEM POWER: +5 Volt Power Supply. |  |
| CAP | 1 | SUBSTRATE FILTER CAPACITOR: a $0.047 \mu \mathrm{~F} \pm 20 \% 12 \mathrm{~V}$ capacitor must be connected between this pin and ground. This capacitor filters the output of the internal substrate bias generator. A maximum DC leakage current of $1 \mu \mathrm{~A}$ is allowed through the capacitor. <br> For correct operation of the 80286, the substrate bias generator must charge this capacitor to its operating voltage. The capacitor chargeup time is 5 milliseconds (max.) after $V_{C C}$ and CLK reach their specified AC and DC parameters. RESET may be applied to prevent spurious activity by the CPU during this time. After this time, the 80286 processor clock can be synchronized to another clock by pulsing RESET LOW synchronous to the system clock. |  |

## NOTE:

1. HLDA is only Low if HOLD is inactive (Low).

## FUNCTIONAL DESCRIPTION

## Introduction

The 80286 is an advanced, high-performance microprocessor with specially optimized capabilities for multiple user and multi-tasking systems. Depending on the application, a 12 MHz 80286 's performance is up to six times faster than the standard 5 MHz 8086's, while providing complete upward software compatibility with Intel's 8086, 88, and 186 family of CPU's.

The 80286 operates in two modes: 8086 real address mode and protected virtual address mode. Both modes execute a superset of the 8086 and 88 instruction set.

In 8086 real address mode programs use real addresses with up to one megabyte of address space. Programs use virtual addresses in protected virtual address mode, also called protected mode. In protected mode, the 80286 CPU automatically maps 1 gigabyte of virtual addresses per task into a 16 megabyte real address space. This mode also provides memory protection to isolate the operating system and ensure privacy of each tasks' programs and data. Both modes provide the same base instruction set, registers, and addressing modes.

The following Functional Description describes first, the base 80286 architecture common to both modes, second, 8086 real address mode, and third, protected mode.

## 80286 BASE ARCHITECTURE

The 8086, 88, 186, and 286 CPU family all contain the same basic set of registers, instructions, and
addressing modes. The 80286 processor is upward compatible with the 8086, 8088, and 80186 CPU's.

## Register Set

The 80286 base architecture has fifteen registers as shown in Figure 3. These registers are grouped into the following four categories:

General Registers: Eight 16-bit general purpose registers used to contain arithmetic and logical operands. Four of these (AX, BX, CX, and DX) can be used either in their entirety as 16 -bit words or split into pairs of separate 8 -bit registers.

Segment Registers: Four 16-bit special purpose registers select, at any given time, the segments of memory that are immediately addressable for code, stack, and data. (For usage, refer to Memory Organization.)

Base and Index Registers: Four of the general purpose registers may also be used to determine offset addresses of operands in memory. These registers may contain base addresses or indexes to particular locations within a segment. The addressing mode determines the specific registers used for operand address calculations.

Status and Control Registers: The 3 16-bit special purpose registers in figure 3A record or control certain aspects of the 80286 processor state including the Instruction Pointer, which contains the offset address of the next sequential instruction to be executed.


Figure 3. Register Set


Figure 3a. Status and Control Register Bit Functions

## Flags Word Description

The Flags word (Flags) records specific characteristics of the result of logical and arithmetic instructions (bits $0,2,4,6,7$, and 11) and controls the operation of the 80286 within a given operating mode (bits 8 and 9). Flags is a 16 -bit register. The function of the flag bits is given in Table 2.

## Instruction Set

The instruction set is divided into seven categories: data transfer, arithmetic, shift/rotate/logical, string manipulation, control transfer, high level instructions, and processor control. These categories are summarized in Figure 4.

An 80286 instruction can reference zero, one, or two operands; where an operand resides in a register, in the instruction itself, or in memory. Zero-operand instructions (e.g. NOP and HLT) are usually one byte long. One-operand instructions (e.g. INC and DEC) are usually two bytes long but some are encoded in only one byte. One-operand instructions may reference a register or memory location. Two-operand instructions permit the following six types of instruction operations:
-Register to Register
-Memory to Register
-Immediate to Register
-Memory to Memory
-Register to Memory
-Immediate to Memory

Table 2. Flags Word Bit Functions

| Bit <br> Position | Name | Function |
| :---: | :---: | :--- |
| 0 | CF | Carry Flag-Set on high-order bit <br> carry or borrow; cleared otherwise |
| 2 | PF | Parity Flag-Set if low-order 8 bits <br> of result contain an even number of <br> 1-bits; cleared otherwise |
| 4 | AF | Set on carry from or borrow to the <br> low order four bits of AL; cleared <br> otherwise |
| 6 | ZF | Zero Flag-Set if result is zero; <br> cleared otherwise |
| 7 | SF | Sign Flag-Set equal to high-order <br> bit of result (0 if positive, 1 if negative) |
| 11 | OF | Overflow Flag-Set if result is a too- <br> large positive number or a too-small <br> negative number (excluding sign--bit) <br> to fit in destination operand; cleared <br> otherwise |
| 8 | TF | Single Step Flag-Once set, a sin- <br> gle step interrupt occurs after the <br> next instruction executes. TF is <br> cleared by the single step interrupt. |
| 9 | IF | Interrupt-enable Flag-When set, <br> maskable interupts will cause the <br> CPU to transfer control to an inter-- <br> rupt vector specified location. |
| 10 | DF | Direction Flag-Causes string <br> instructions to auto decrement <br> the appropriate index registers <br> when set. Clearing DF causes <br> auto increment. |

Two-operand instructions (e.g. MOV and ADD) are usually three to six bytes long. Memory to memory operations are provided by a special class of string instructions requiring one to three bytes. For detailed instruction formats and encodings refer to the instruction set summary at the end of this document.

For detailed operation and usage of each instruction, see Appendix of 80286 Programmer's Reference Manual (Order No. 210498)

| GENERAL PURPOSE |  |
| :--- | :---: |
| MOV | Move byte or word |
| PUSH | Push word onto stack |
| POP | Pop word off stack |
| PUSHA | Push all registers on stack |
| POPA | Pop all registers from stack |
| XCHG | Exchange byte or word |
| XLAT | Translate byte |
|  | INPUT/OUTPUT |
| IN | Input byte or word |
| OUT | Output byte or word |
| LEA | Load effective address |
| LDS | Load pointer using DS |
| LES | Load pointer using ES |
|  | FLAG TRANSFER |
| LAHF | Load AH register from flags |
| SAHF | Store AH register in flags |
| PUSHF | Push flags onto stack |
| POPF | Pop flags off stack |

Figure 4a. Data Transfer Instructions

| MOVS | Move byte or word string |
| :--- | :--- |
| INS | Input bytes or word string |
| OUTS | Output bytes or word string |
| CMPS | Compare byte or word string |
| SCAS | Scan byte or word string |
| LODS | Load byte or word string |
| STOS | Store byte or word string |
| REP | Repeat |
| REPE/REPZ | Repeat while equal/zero |
| REPNE/REPNZ | Repeat while not equal/not zero |

Figure 4 c . String Instructions

| ADDITION |  |  |  |
| :--- | :--- | :---: | :---: |
| ADD | Add byte or word |  |  |
| ADC | Add byte or word with carry |  |  |
| INC | Increment byte or word by 1 |  |  |
| AAA | ASCII adjust for addition |  |  |
| DAA | Decimal adjust for addition |  |  |
|  | SUBTRACTION |  |  |
| SUB | Subtract byte or word |  |  |
| SBB | Subtract byte or word with borrow |  |  |
| DEC | Decrement byte or word by 1 |  |  |
| NEG | Negate byte or word |  |  |
| CMP. | Compare byte or word |  |  |
| AAS | ASCII adjust for subtraction |  |  |
| DAS | Decimal adjust for subtraction |  |  |
| MUL | MULTIPLICATION |  |  |
| IMUL | Multiple byte or word unsigned |  |  |
| AAM | ASCII adjust for multiply |  |  |
| DIVISION |  |  |  |
| DIV | Divide byte or word unsigned |  |  |
| IDIV | Integer divide byte or word |  |  |
| AAD | ASCII adjust for division |  |  |
| CBW | Convert byte to word |  |  |
| CWD | Convert word to doubleword |  |  |

Figure 4b. Arithmetic Instructions

| LOGICALS |  |  |
| :--- | :--- | :---: |
| NOT | "Not" byte or word |  |
| AND | "And" byte or word |  |
| OR | "Inclusive or" byte or word |  |
| XOR | "Exclusive or"' byte or word |  |
| TEST | "Test" byte or word |  |
| SHIFTS |  |  |
| SHL/SAL | Shift logical/arithmetic left byte or word |  |
| SHR | Shift logical right byte or word |  |
| SAR | Shift arithmetic right byte or word |  |
|  | ROTATES |  |
| ROL | Rotate left byte or word |  |
| ROR | Rotate right byte or word |  |
| RCL | Rotate through carry left byte or word |  |
| RCR | Rotate through carry right byte or word |  |

Figure 4d. Shift/Rotate Logical Instructions

| CONDITIONAL TRANSFERS |  | UNCONDITIONAL TRANSFERS |  |
| :---: | :---: | :---: | :---: |
| JA/JNBE | Jump if above/not below nor equal | CALL | Call procedure |
| JAE/JNB | Jump if above or equal/not below | RET | Return from procedure |
| JB/JNAE | Jump if below/not above nor equal | JMP | Jump |
| JBE/JNA | Jump if below or equal/not above |  |  |
| JC | Jump if carry | ITERATION CONTROLS |  |
| JE/JZ | Jump if equal/zero | LOOP | Loop |
| JG/JNLE | Jump if greater/not less nor equal |  |  |
| JGE/JNL | Jump if greater or equal/not less | LOOPE/LOOPZ | Loop if equal/zero |
| JL/JNGE | Jump if less/not greater nor equal | LOOPNE/LOOPNZ | Loop if not equal/not zero |
| JLE/JNG | Jump if less or equal/not greater | JCXZ | Jump if register CX $=0$ |
| JNC | Jump if not carry |  |  |
| JNE/JNZ | Jump if not equal/not zero | INTERRUPTS |  |
| JNO | Jump if not overflow | INT | Interrupt |
| JNP/JPO | Jump if not parity/parity odd |  |  |
| JNS | Jump if not sign | INTO | Interrupt if overflow |
| JO | Jump if overtiow | IRET | Interrupt return |
| JP/JPE | Jump if parity/parity even |  |  |
| JS | Jump if sign |  |  |

Figure 4e. Program Transfer Instructions

| FLAG OPERATIONS |  |  |
| :--- | :--- | :---: |
| STC | Set carry flag |  |
| CLC | Clear carry fiag |  |
| CMC | Complement carry flag |  |
| STD | Set direction flag |  |
| CLD | Clear direction flag |  |
| STI | Set interrupt enable flag |  |
| CLI | Clear interrupt enable flag |  |
| EXTERNAL SYNCHRONIZATION |  |  |
| HLT | Halt until interrupt or reset |  |
| WAIT | Wait for BUSY not active |  |
| ESC | Escape to extension processor |  |
| LOCK | Lock bus during next instruction |  |
| NO OPERATION |  |  |
| NOP | No operation |  |
|  |  |  |
| EXECUTION ENVIRONMENT CONTROL |  |  |
| LMSW |  |  |
| SMSW | LOad machine status word |  |
| Store machine status word |  |  |

Figure 4f. Processor Control Instructions

| ENTER | Format stack for procedure entry |
| :--- | :--- |
| LEAVE | Restore stack for procedure exit |
| BOUND | Detects values outside prescribed range |

Figure 4g. High Level Instructions

## Memory Organization

Memory is organized as sets of variable length segments. Each segment is a linear contiguous sequence of up to $64 \mathrm{~K}\left(2^{16}\right) 8$-bit bytes. Memory is addressed using a two component address (a pointer) that consists of a 16 -bit segment selector, and a 16 -bit offset. The segment selector indicates the desired segment in memory. The offset component indicates the desired byte address within the segment.


Figure 5. Two Component Address

Table 3. Segment Register Selection Rules

| Memory <br> Reference Needed | Segment Register <br> Used | Implicit Segment <br> Selection Rule |
| :--- | :--- | :--- |
| Instructions | Code (CS) | Automatic with instruction prefetch |
| Stack | Stack (SS) | All stack pushes and pops. Any memory reference which uses BP <br> as a base register. |
| Local Data | Data (DS) | All data references except when relative to stack or <br> string destination |
| External (Global) Data | Extra (ES) | Alternate data segment and destination of string operation |

All instructions that address operands in memory must specify the segment and the offset. For speed and compact instruction encoding, segment selectors are usually stored in the high speed segment registers. An instruction need specify only the desired segment register and an offset in order to address a memory operand.

Most instructions need not explicitly specify which segment register is used. The correct segment register is automatically chosen according to the rules of Table 3. These rules follow the way programs are written (see Figure 6) as independent modules that require areas for code and data, a stack, and access to external data areas.

Special segment override instruction prefixes allow the implicit segment register selection rules to be overridden for special cases. The stack, data, and extra segments may coincide for simple programs. To access operands not residing in one of the four immediately available segments, a full 32-bit pointer or a new segment selector must be loaded.

## Addressing Modes

The 80286 provides a total of eight addressing modes for instructions to specify operands. Two addressing modes are provided for instructions that operate on register or immediate operands:

Register Operand Mode: The operand is located in one of the 8 or 16-bit general registers.
Immediate Operand Mode: The operand is included in the instruction.

Six modes are provided to specify the location of an operand in a memory segment. A memory operand address consists of two 16 -bit components: segment selector and offset. The segment selector is supplied by a segment register either implicitly chosen by the addressing mode or explicitly chosen by a segment override prefix. The offset is calculated by summing any combination of the following three address elements:
the displacement (an 8 or 16-bit immediate value contained in the instruction)
the base (contents of either the BX or BP base registers)


Figure 6. Segmented Memory Helps Structure Software
the index (contents of either the SI or DI index registers)

Any carry out from the 16 -bit addition is ignored. Eight-bit displacements are sign extended to 16-bit values.

Combinations of these three address elements define the six memory addressing modes, described below.

Direct Mode: The operand's offset is contained in the instruction as an 8 or 16 -bit displacement element.

Register Indirect Mode: The operand's offset is in one of the registers SI, DI, BX, or BP.

Based Mode: The operand's offset is the sum of an 8 or 16-bit displacement and the contents of a base register ( $B X$ or $B P$ ).

Indexed Mode: The operand's offset is the sum of an 8 or 16 -bit displacement and the contents of an index register (SI or DI).

Based Indexed Mode: The operand's offset is the sum of the contents of a base register and an index register.

Based Indexed Mode with Displacement: The operand's offset is the sum of a base register's contents, an index register's contents, and an 8 or 16-bit displacement.

## Data Types

The 80286 directly supports the following data types:
Integer: A signed binary numeric value contained in an 8 -bit byte or a 16 -bit word. All operations assume a 2's complement representation. Signed 32 and 64-bit integers are supported using the Numeric Data Processor, the 80287.
Ordinal: An unsigned binary numeric value contained in an 8 -bit byte or 16 -bit word.
Pointer: A 32-bit quantity, composed of a segment selector component and an offset component. Each component is a 16 -bit word.
String: $\quad$ A contiguous sequence of bytes or words. A string may contain from 1 byte to 64 K bytes.
ASCII: A byte representation of alphanumeric and control characters using the ASCll standard of character representation.
BCD: $\quad$ A byte (unpacked) representation of the decimal digits 0-9.
Packed BCD: A byte (packed) representation of two decimal digits $0-9$ storing one digit in each nibble of the byte.
Floating Point: A signed 32,64 , or 80 -bit real number representation. (Floating point operands are supported using the 80287 Numeric Processor).

Figure 7 graphically represents the data types supported by the 80286.

## I/O Space

The I/O space consists of 64 K 8 -bit or 32 K 16 -bit ports. I/O instructions address the I/O space with
either an 8-bit port address, specified in the instruction, or a 16-bit port address in the DX register. 8-bit port addresses are zero extended such that $A_{15}-A_{8}$ are LOW. I/O port addresses $00 F 8(\mathrm{H})$ through $00 F F(H)$ are reserved.


Figure 7. 80286 Supported Data Types

Table 4. Interrupt Vector Assignments

| Function | Interrupt <br> Number | Related <br> Instructions | Does Return Address <br> Polnt to Instruction <br> Causing Exception? |
| :--- | :---: | :--- | :---: |
| Divide error exception | 0 | DIV, IDIV | Yes |
| Single step interrupt | 1 | All |  |
| NMI interrupt | 2 | INT 2 or NMI pin |  |
| Breakpoint interrupt | 3 | INT 3 |  |
| INTO detected overflow exception | 4 | INTO | No |
| BOUND range exceeded exception | 6 | BOUND | Yes |
| Invalid opcode exception | Any undefined opcode | Yes |  |
| Processor extension not available exception | 7 | ESC or WAIT | Yes |
| Intel reserved-do not use | $8-15$ |  |  |
| Processor extension error interrupt | 16 | ESC or WAIT |  |
| Intel reserved-do not use | $17-31$ |  |  |
| User defined | $32-255$ |  |  |

## Interrupts

An interrupt transfers execution to a new program location. The old program address (CS:IP) and machine state (Flags) are saved on the stack to allow resumption of the interrupted program. Interrupts fall into three classes: hardware initiated, INT instructions, and instruction exceptions. Hardware initiated interrupts occur in response to an external input and are classified as non-maskable or maskable. Programs may cause an interrupt with an INT instruction. Instruction exceptions occur when an unusual condition, which prevents further instruction processing, is detected while attempting to execute an instruction. The return address from an exception will always point at the instruction causing the exception and include any leading instruction prefixes.

A table containing up to 256 pointers defines the proper interrupt service routine for each interrupt. Interrupts $0-31$, some of which are used for instruction exceptions, are reserved. For each interrupt, an 8 -bit vector must be supplied to the 80286 which identifies the appropriate table entry. Exceptions supply the interrupt vector internally. INT instructions contain or imply the vector and allow access to all 256 interrupts. Maskable hardware initiated interrupts supply the 8 -bit vector to the CPU during an interrupt acknowiedge bus sequence. Non-maskable hardware interrupts use a predefined internally supplied vector.

## MASKABLE INTERRUPT (INTR)

The 80286 provides a maskable hardware interrupt request pin, INTR. Software enables this input by
setting the interrupt flag bit (IF) in the flag word. All 224 user-defined interrupt sources can share this input, yet they can retain separate interrupt handlers. An 8 -bit vector read by the CPU during the interrupt acknowledge sequence (discussed in System Interface section) identifies the source of the interrupt.

Further maskable interrupts are disabled while servicing an interrupt by resetting the IF bit as part of the response to an interrupt or exception. The saved flag word will reflect the enable status of the processor prior to the interrupt. Until the flag word is restored to the flag register, the interrupt flag will be zero unless specifically set. The interrupt return instruction includes restoring the flag word, thereby restoring the original status of IF.

## NON-MASKABLE INTERRUPT REQUEST (NMI)

A non-maskable interrupt input (NMI) is also provided. NMI has higher priority than INTR. A typical use of NMI would be to activate a power failure routine. The activation of this input causes an interrupt with an internally supplied vector value of 2 . No external interrupt acknowledge sequence is 'performed.

While executing the NMI servicing procedure, the 80286 will service neither further NMI requests, INTR requests, nor the processor extension segment overrun interrupt until an interrupt return (IRET) instruction is executed or the CPU is reset. If NMI occurs while currently servicing an NMI, its presence will be saved for servicing after executing the first IRET instruction. IF is cleared at the beginning of an NMI interrupt to inhibit INTR interrupts.

## SINGLE STEP INTERRUPT

The 80286 has an internal interrupt that allows programs to execute one instruction at a time. It is called the single step interrupt and is controlled by the single step flag bit (TF) in the flag word. Once this bit is set, an internal single step interrupt will occur after the next instruction has been executed. The interrupt clears the TF bit and uses an internally supplied vector of 1 . The IRET instruction is used to set the TF bit and transfer control to the next instruction to be single stepped.

## Interrupt Priorities

When simultaneous interrupt requests occur, they are processed in a fixed order as shown in Table 5. Interrupt processing involves saving the flags, return address, and setting CS:IP to point at the first instruction of the interrupt handler. If other interrupts remain enabled they are processed before the first instruction of the current interrupt handler is executed. The last interrupt processed is therefore the first one serviced.

Table 5. Interrupt Processing Order

| Order | Interrupt |
| :--- | :--- |
| 1 | Instruction exception |
| 2 | Single step |
| 3 | NMI |
| 4 | Processor extension segment overrun |
| 5 | INTR |
| 6 | INT instruction |

## Initialization and Processor Reset

Processor initialization or start up is accomplished by driving the RESET input pin HIGH. RESET forces the 80286 to terminate all execution and local bus activity. No instruction or bus activity will occur as long as RESET is active. After RESET becomes inactive and an internal processing interval elapses, the 80286 begins execution in real address mode with the instruction at physical location FFFFFO(H). RESET also sets some registers to predefined values as shown in Table 6.

Table 6. 80286 Initial Register State after RESET

| Flag word | $0002(\mathrm{H})$ |
| :--- | :---: |
| Machine Status Word | FFFo(H) |
| Instruction pointer | FFFO(H) |
| Code segment | FO00(H) |
| Data segment | $000(\mathrm{H})$ |
| Extra segment | $000(\mathrm{H})$ |
| Stack segment | $0000(\mathrm{H})$ |

HOLD must not be active during the time from the leading edge of RESET to 34 CLKs after the trailing edge of RESET.

## Machine Status Word Description

The machine status word (MSW) records when a task switch takes place and controls the operating mode of the 80286 . It is a 16 -bit register of which the lower four bits are used. One bit places the CPU into protected mode, while the other three bits, as shown in Table 7, control the processor extension interface. After RESET, this register contains FFFO(H) which places the 80286 in 8086 real address mode.

Table 7. MSW Bit Functions

| BIt <br> Position | Name | Function |
| :---: | :---: | :--- |
| 0 | PE | Protected mode enable places the <br> 80286 into protected mode and cannot <br> be cleared except by RESET. |
| 1 | MP | Monitor processor extension allows <br> WAIT instructions to cause a processor <br> extension not present exception <br> (number 7). |
| 2 | EM | Emulate processor extension causes a <br> processor extension not present <br> exception (number 7) on ESC <br> instructions to allow emulating a <br> processor extension. |
| 3 | TS | Task switched indicates the next <br> instruction using a processor extension <br> will cause exception 7, allowing software <br> to test whether the current processor <br> extension context belongs to the current <br> task. |

The LMSW and SMSW instructions can load and store the MSW in real address mode. The recommended use of TS, EM, and MP is shown in Table 8.

Table 8. Recommended MSW Encodings For Processor Extension Control

| TS | MP | EM |  | Instructions <br> Causing <br> Exception 7 |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Recommended Use | Initial encoding after RESET. 80286 operation is identical to 8086, 88. | None |
| 0 | 0 | 1 | No processor extension is available. Software will emulate its function. | ESC |  |
| 1 | 0 | 1 | No processor extension is available. Software will emulate its function. The current <br> processor extension context may belong to another task. | ESC |  |
| 0 | 1 | 0 | A processor extension exists. | None |  |
| 1 | 1 | 0 | A processor extension exists. The current processor extension context may belong to <br> another task. The Exception 7 on WAIT allows software to test for an error pending <br> from a previous processor extension operation. | ESC or <br> WAIT |  |

## Halt

The HLT instruction stops program execution and prevents the CPU from using the local bus until restarted. Either NMI, INTR with IF $=1$, or RESET will force the 80286 out of halt. If interrupted, the saved CS:IP will point to the next instruction after the HLT.

## 8086 REAL ADDRESS MODE

The 80286 executes a fully upward-compatible superset of the 8086 instruction set in real address mode. In real address mode the 80286 is object code compatible with 8086 and 8088 software. The real address mode architecture (registers and addressing modes) is exactly as described in the 80286 Base Architecture section of this Functional Description.

## Memory Size

Physical memory is a contiguous array of up to 1,048,576 bytes (one megabyte) addressed by pins $\mathrm{A}_{0}$ through $\mathrm{A}_{19}$ and $\overline{\mathrm{BHE}}$. $\mathrm{A}_{20}$ through $\mathrm{A}_{23}$ should be ignored.

## Memory Addressing

In real address mode physical memory is a contiguous array of up to $1,048,576$ bytes (one megabyte) addressed by pins $A_{0}$ through $\mathrm{A}_{19}$ and BHE. Address bits $A_{20}-A_{23}$ may not always be zero in real mode. $\mathrm{A}_{20}-\mathrm{A}_{23}$ should not be used by the system while the 80286 is operating in Real Mode.

The selector portion of a pointer is interpreted as the upper 16 bits of a 20 -bit segment address. The lower four bits of the 20 -bit segment address are always zero. Segment addresses, therefore, begin on multiples of 16 bytes. See Figure 8 for a graphic representation of address information.

All segments in real address mode are 64 K bytes in size and may be read, written, or executed. An exception or interrupt can occur if data operands or instructions attempt to wrap around the end of a segment (e.g. a word with its low order byte at offset FFFF(H) and its high order byte at offset $0000(\mathrm{H})$. If, in real address mode, the information contained in a segment does not use the full 64 K bytes, the unused end of the segment may be overlayed by another segment to reduce physical memory requirements.
tion area and interrupt table area. Locations from addresses $\operatorname{FFFFO}(\mathrm{H})$ through $\operatorname{FFFFF}(\mathrm{H})$ are reserved for system initialization. Initial execution begins at location FFFFO(H). Locations $00000(\mathrm{H})$ through 003FF(H) are reserved for interrupt vectors.


210253-8
Figure 8. 8086 Real Address Mode Address Calculation


INITIAL CS:IP VALUE IS F000:FFFO.
210253-9
Figure 9. 8086 Real Address Mode Initlally Reserved Memory Locations

## Reserved Memory Locations

The 80286 reserves two fixed areas of memory in real address mode (see Figure 9); system initializa-

Table 9. Real Address Mode Addressing Interrupts

| Function | Interrupt <br> Number | Related <br> Instructions | Return Address <br> Before Instruction? |
| :--- | :---: | :--- | :---: |
| Interrupt table limit too small exception | 8 | INT vector is not within table limit | Yes |
| Processor extension segment overrun <br> interrupt | 9 | ESC with memory operand extend- <br> ing beyond offset FFFF(H) | No |
| Segment overrun exception | 13 | Word memory reference with offset <br> $=$ FFFF(H) or an attempt to exe- <br> cute past the end of a segment | Yes |

## Interrupts

Table 9 shows the interrupt vectors reserved for exceptions and interrupts which indicate an addressing error. The exceptions leave the CPU in the state existing before attempting to execute the failing instruction (except for PUSH, POP, PUSHA, or POPA). Refer to the next section on protected mode initialization for a discussion on exception 8.

## Protected Mode Initialization

To prepare the 80286 for protected mode, the LIDT instruction is used to load the 24-bit interrupt table base and 16 -bit limit for the protected mode interrupt table. This instruction can also set a base and limit for the interrupt vector table in real address mode. After reset, the interrupt table base is initialized to $000000(\mathrm{H})$ and its size set to 03FF(H). These values are compatible with 8086,88 software. LIDT should only be executed in preparation for protected mode.

## Shutdown

Shutdown occurs when a severe error is detected that prevents further instruction processing by the CPU. Shutdown and halt are externally signalled via a halt bus operation. They can be distinguished by $A_{1}$ HIGH for halt and $A_{1}$ LOW for shutdown. In real address mode, shutdown can occur under two conditions:

- Exceptions 8 or 13 happen and the IDT limit does not include the interrupt vector.
- A CALL INT or PUSH instruction attempts to wrap around the stack segment when SP is not even.

An NMI input can bring the CPU out of shutdown if the IDT limit is at least $000 \mathrm{~F}(\mathrm{H})$ and SP is greater than $0005(\mathrm{H})$, otherwise shutdown can only be exited via the RESET input.

## - PROTECTED VIRTUAL ADDRESS MODE

The 80286 executes a fully upward-compatible superset of the 8086 instruction set in protected virtual address mode (protected mode). Protected mode also provides memory management and protection mechanisms and associated instructions.

The 80286 enters protected virtual address mode from real address mode by setting the PE (Protection Enable) bit of the machine status word with the Load Machine Status Word (LMSW) instruction. Protected mode offers extended physical and virtual memory address space, memory protection mechanisms, and new operations to support operating systems and virtual memory.

All registers, instructions, and addressing modes described in the 80286 Base Architecture section of this Functional Description remain the same. Programs for the 8086, 88, 186, and real address mode 80286 can be run in protected mode; however, embedded constants for segment selectors are different.

## Memory Size

The protected mode 80286 provides a 1 gigabyte virtual address space per task mapped into a 16 megabyte physical address space defined by the address pin $\mathrm{A}_{23}-\mathrm{A}_{0}$ and BHE . The virtual address space may be larger than the physical address space since any use of an address that does not map to a physical memory location will cause a restartable exception.

## Memory Addressing

As in real address mode, protected mode uses 32bit pointers, consisting of 16 -bit selector and offset components. The selector, however, specifies an index into a memory resident table rather than the upper 16 -bits of a real memory address. The 24 -bit
base address of the desired segment is obtained from the tables in memory. The 16-bit offset is added to the segment base address to form the physical address as shown in Figure 10. The tables are automatically referenced by the CPU whenever a segment register is loaded with a selector. All 80286 instructions which load a segment register will reference the memory based tables without additional software. The memory based tables contain 8 byte values called descriptors.


Figure 10. Protected Mode Memory Addressing

## DESCRIPTORS

Descriptors define the use of memory. Special types of descriptors also define new functions for transfer of control and task switching. The 80286 has segment descriptors for code, stack and data segments, and system control descriptors for special system data segments and control transfer operations. Descriptor accesses are performed as locked bus operations to assure descriptor integrity in multi-processor systems.

## CODE AND DATA SEGMENT DESCRIPTORS ( $\mathrm{S}=1$ )

Besides segment base addresses, code and data descriptors contain other segment attributes including segment size ( 1 to 64 K bytes), access rights (read only, read/write, execute only, and execute/ read), and presence in memory (for virtual memory systems) (See Figure 11). Any segment usage violating a segment attribute indicated by the segment descriptor will prevent the memory cycle and cause an exception or interrupt.

Code or Data Segment Descriptor


Access Rights Byte Definition

\begin{tabular}{|c|c|c|c|c|c|}
\hline \multirow[t]{2}{*}{} \& Bit Position \& Name \& \& Function \& \\
\hline \& 7
\(6-5\)
4 \& \begin{tabular}{l}
Present (P) \\
Descriptor Privilege Level (DPL) Segment Descriptor (S)
\end{tabular} \& \[
\begin{gathered}
P=1 \\
P=0 \\
\\
S=1 \\
S=0
\end{gathered}
\] \& \begin{tabular}{l}
Segment is mapped into physical memory. \\
No mapping to physical memory exits, base and limit are not used. \\
Segment privilege attribute used in privilege tests. \\
Code or Data (includes stacks) segment descriptor \\
System Segment Descriptor or Gate Descriptor
\end{tabular} \& \\
\hline \multirow{3}{*}{\begin{tabular}{l}
Type \\
Field \\
Definition
\end{tabular}} \& 3
2
1 \& \begin{tabular}{l}
Executable (E) Expansion Direction (ED) \\
Writeable (W)
\end{tabular} \& \[
\begin{aligned}
\& E=0 \\
\& E D=0 \\
\& E D=1 \\
\& W=0 \\
\& W=1
\end{aligned}
\] \& \begin{tabular}{l}
Data segment descriptor type is: \\
Expand up segment, offsets must be \(\leq\) limit. Expand down segment, offsets must be > limit. Data segment may not be written into. \\
Data segment may be written into.
\end{tabular} \& \begin{tabular}{l}
If \\
Data \\
Segment
\[
(S=1
\]
\[
E=0)
\]
\end{tabular} \\
\hline \& 3
2

1 \& | Executable (E) |
| :--- |
| Conforming (C) |
| Readable (R) | \& \[

$$
\begin{aligned}
& E=1 \\
& C=1 \\
& R=0 \\
& R=1
\end{aligned}
$$

\] \& | Code Segment Descriptor type is: |
| :--- |
| Code segment may only be executed |
| when CPL $\geq$ DPL and CPL |
| remains unchanged. |
| Code segment may not be read |
| Code segment may be read. | \& | If |
| :--- |
| Code |
| Segment $\begin{aligned} & (S=1, \\ & E=1) \end{aligned}$ | <br>

\hline \& 0 \& Accessed (A) \& $$
\begin{aligned}
& A=0 \\
& A=1
\end{aligned}
$$ \& Segment has not been accessed. Segment selector has been loaded into segment register or used by selector test instructions. \& <br>

\hline
\end{tabular}

Figure 11. Code and Data Segment Descriptor Formats

Code and data (including stack data) are stored in two types of segments: code segments and data segments. Both types are identified and defined by segment descriptors $(S=1)$. Code segments are identified by the executable ( E ) bit set to 1 in the descriptor access rights byte. The access rights byte of both code and data segment descriptor types have three fields in common: present ( P ) bit, De scriptor Privilege Level (DPL), and accessed (A) bit. If $\mathrm{P}=0$, any attempted use of this segment will cause a not-present exception. DPL specifies the privilege level of the segment descriptor. DPL controls when the descriptor may be used by a task (refer to privilege discussion below). The A bit shows whether the segment has been previously accessed for usage profiling, a necessity for virtual memory systems. The CPU will always set this bit when accessing the descriptor.

Data segments ( $S=1, E=0$ ) may be either readonly or read-write as controlled by the $W$ bit of the access rights byte. Read-only ( $W=0$ ) data segments may not be written into. Data segments may grow in two directions, as determined by the Expansion Direction (ED) bit: upwards ( $E D=0$ ) for data segments, and downwards ( $E D=1$ ) for a segment containing a stack. The limit field for a data segment descriptor is interpreted differently depending on the ED bit (see Figure 11).
A code segment ( $S=1, E=1$ ) may be executeonly or execute/read as determined by the Readable ( $R$ ) bit. Code segments may never be written into and execute-only code segments ( $R=0$ ) may not be read. A code segment may also have an attribute called conforming (C). A conforming code segment may be shared by programs that execute at different privilege levels. The DPL of a conforming code segment defines the range of privilege levels at which the segment may be executed (refer to privilege discussion below). The limit field identifies the last byte of a code segment.

## SYSTEM SEGMENT DESCRIPTORS (S = 0, TYPE $=1-3$ )

In addition to code and data segment descriptors, the protected mode 80286 defines System Segment Descriptors. These descriptors define special system data segments which contain a table of descriptors (Local Descriptor Table Descriptor) or segments which contain the execution state of a task (Task State Segment Descriptor).

Figure 12 gives the formats for the special system data segment descriptors. The descriptors contain a 24 -bit base address of the segment and a 16 -bit limit. The access byte defines the type of descriptor, its state and privilege level. The descriptor contents are valid and the segment is in physical memory if $\mathrm{P}=1$. If $P=0$, the segment is not valid. The DPL field is only used in Task State Segment descriptors and indicates the privilege level at which the descrip-
tor may be used (see Privilege). Since the Local Descriptor Table descriptor may only be used by a special privileged instruction, the DPL field is not used. Bit 4 of the access byte is 0 to indicate that it is a system control descriptor. The type field specifies the descriptor type as indicated in Figure 12.

System Segment Descriptor


System Segment Descriptor Fields

| Name | Value | Description |
| :---: | :---: | :--- |
| TYPE | 1 | Available Task State Segment (TSS) <br>  <br>  $2^{2}$ |
| Local Descriptor Table |  |  |
| Busy Task State Segment (TSS) |  |  |
|  | 0 | Descriptor contents are not valid |
|  | 1 | Descriptor contents are valid |
| DPL | $0-3$ | Descriptor Privilege Level |
| BASE | $24-$ bit <br> number | Base Address of special system data <br> segment in real memory |
| LIMIT | $16-$ bit <br> number | Offset of last byte in segment |

Figure 12. System Segment Descriptor Format

## GATE DESCRIPTORS ( $\mathrm{S}=0$, TYPE $=4-7$ )

Gates are used to control access to entry points within the target code segment. The gate descriptors are call gates, task gates, interrupt gates and trap gates. Gates provide a level of indirection between the source and destination of the control transfer. This indirection allows the CPU to automatically perform protection checks and control entry point of the destination. Call gates are used to change privilege levels (see Privilege), task gates are used to perform a task switch, and interrupt and trap gates are used to specify interrupt service routines. The interrupt gate disables interrupts (resets IF) while the trap gate does not.

Gate Descriptor


210253-13
*Must be set to 0 for compatibility with 80386 ( $X$ is don't care)

Gate Descriptor Fields

| Name | Value | Description |
| :---: | :---: | :--- |
| TYPE | 4 | -Call Gate <br> -Task Gate <br> - Interrupt Gate <br> -Trap Gate |
| P | 0 | - Descriptor Contents are not <br> valid <br> -Descriptor Contents are <br> valid |
| DPL | $0-3$ | Descriptor Privilege Level |
| WORD | $0-31$ | Number of words to copy <br> from callers stack to called <br> procedures stack. Only used <br> with call gate. |
| DESTINATION | $16-$ bit |  |
| SELECTOR | selector | Selector to the target code <br> segment (Call, Interrupt or <br> Trap Gate) <br> Selector to the target task <br> state segment (Task Gate) |
| DESTINATION | 16-bit <br> Offset | Entry point within the target <br> Code segment |

## Figure 13. Gate Descriptor Format

Figure 13 shows the format of the gate descriptors. The descriptor contains a destination pointer that points to the descriptor of the target segment and the entry point offset. The destination selector in an interrupt gate, trap gate, and call gate must refer to a code segment descriptor. These gate descriptors contain the entry point to prevent a program from constructing and using an illegal entry point. Task gates may only refer to a task state segment. Since task gates invoke a task switch, the destination offset is not used in the task gate.
Exception 13 is generated when the gate is used if a destination selector does not refer to the correct descriptor type. The word count field is used in the call gate descriptor to indicate the number of parameters ( $0-31$ words) to be automatically copied from the caller's stack to the stack of the called routine when a control transfer changes privilege levels. The word count field is not used by any other gate descriptor.
The access byte format is the same for all gate descriptors. $P=1$ indicates that the gate contents are valid. $P=0$ indicates the contents are not valid and causes exception 11 if referenced. DPL is the de-
scriptor privilege level and specifies when this descriptor may be used by a task (refer to privilege discussion below). Bit 4 must equal 0 to indicate a system control descriptor. The type field specifies the descriptor type as indicated in Figure 13.

## SEGMENT DESCRIPTOR CACHE REGISTERS

A segment descriptor cache register is assigned to each of the four segment registers (CS, SS, DS, ES). Segment descriptors are automatically loaded (cached) into a segment descriptor cache register (Figure 14) whenever the associated segment register is loaded with a selector. Only segment descriptors may be loaded into segment descriptor cache registers. Once loaded, all references to that segment of memory use the cached descriptor information instead of reaccessing the descriptor. The descriptor cache registers are not visible to programs. No instructions exist to store their contents. They only change when a segment register is loaded.

## SELECTOR FIELDS

A protected mode selector has three fields: descriptor entry index, local or global descriptor table indicator (TI), and selector privilege (RPL) as shown in Figure 15. These fields select one of two memory based tables of descriptors, select the appropriate table entry and allow highspeed testing of the selector's privilege attribute (refer to privilege discussion below).


Figure 15. Selector Fields


Figure 14. Descriptor Cache Registers

## LOCAL AND GLOBAL DESCRIPTOR TABLES

Two tables of descriptors, called descriptor tables, contain all descriptors accessible by a task at any given time. A descriptor table is a linear array of up to 8192 descriptors. The upper 13 bits of the selector value are an index into a descriptor table. Each table has a 24-bit base register to locate the descriptor table in physical memory and a 16 -bit limit register that confine descriptor access to the defined limits of the table as shown in Figure 16. A restartable exception (13) will occur if an attempt is made to reference a descriptor outside the table limits.

One table, called the Global Descriptor table (GDT), contains descriptors available to all tasks. The other table, called the Local Descriptor Table (LDT), contains descriptors that can be private to a task. Each task may have its own private LDT. The GDT may contain all descriptor types except interrupt and trap descriptors. The LDT may contain only segment, task gate, and call gate descriptors. A segment cannot be accessed by a task if its segment descriptor does not exist in either descriptor table at the time of access.


Figure 16. Local and Global Descriptor Table Definition
The LGDT and LLDT instructions load the base and limit of the global and local descriptor tables. LGDT and LLDT are privileged, i.e. they may only be executed by trusted programs operating at level 0 . The LGDT instruction loads a six byte field containing the 16 -bit table limit and 24-bit physical base address of the Global Descriptor Table as shown in Figure 17. The LLDT instruction loads a selector which refers to a Local Descriptor Table descriptor containing the
base address and limit for an LDT, as shown in Figure 12.


Figure 17. Global Descriptor Table and Interrupt Descriptor Table Data Type

## INTERRUPT DESCRIPTOR TABLE

The protected mode 80286 has a third descriptor table, called the Interrupt Descriptor Table (IDT) (see Figure 18), used to define up to 256 interrupts. It may contain only task gates, interrupt gates and trap gates. The IDT (Interrupt Descriptor Table) has a 24 -bit physical base and 16 -bit limit register in the CPU. The privileged LIDT instruction loads these registers with a six byte value of identical form to that of the LGDT instruction (see Figure 17 and Protected Mode Initialization).


Figure 18. Interrupt Descriptor Table Definition
References to IDT entries are made via INT instructions, external interrupt vectors, or exceptions. The IDT must be at least 256 bytes in size to allocate space for all reserved interrupts.

## Privilege

The 80286 has a four-level hierarchical privilege system which controls the use of privileged instructions and access to descriptors (and their associated segments) within a task. Four-level privilege, as shown in Figure 19, is an extension of the user/supervisor mode commonly found in minicomputers. The privilege levels are numbered 0 through 3. Level 0 is the

most privileged level. Privilege levels provide protection within a task. (Tasks are isolated by providing private LDT's for each task.) Operating system routines, interrupt handlers, and other system software can be included and protected within the virtual address space of each task using the four levels of privilege. Each task in the system has a separate stack for each of its privilege levels.
Tasks, descriptors, and selectors have a privilege level attribute that determines whether the descriptor may be used. Task privilege effects the use of instructions and descriptors. Descriptor and selector privilege only effect access to the descriptor.

## TASK PRIVILEGE

A task always executes at one of the four privilege levels. The task privilege level at any specific instant is called the Current Privilege Level (CPL) and is defined by the lower two bits of the CS register. CPL cannot change during execution in a single code segment. A task's CPL may only be changed by control transfers through gate descriptors to a new code segment (See Control Transfer). Tasks begin executing at the CPL value specified by the code segment selector within TSS when the task is initiated via a task switch operation (See Figure 20). A task executing at Level 0 can access all data segments defined in the GDT and the task's LDT and is considered the most trusted level. A task executing a Level 3 has the most restricted access to data and is considered the least trusted level.

## DESCRIPTOR PRIVILEGE

Descriptor privilege is specified by the Descriptor Privilege Level (DPL) field of the descriptor access byte. DPL specifies the least trusted task privilege level (CPL) at which a task may access the descrip-
tor. Descriptors with DPL $=0$ are the most protected. Only tasks executing at privilege level 0 (CPL $=0$ ) may access them. Descriptors with DPL $=3$ are the least protected (i.e. have the least restricted access) since tasks can access them when CPL $=0,1,2$, or 3 . This rule applies to all descriptors, except LDT descriptors.

## SELECTOR PRIVILEGE

Selector privilege is specified by the Requested Privilege Level (RPL) field in the least significant two bits of a selector. Selector RPL may establish a less trusted privilege level than the current privilege level for the use of a selector. This level is called the task's effective privilege level (EPL). RPL can only reduce the scope of a task's access to data with this selector. A task's effective privilege is the numeric maximum of RPL and CPL. A selector with RPL $=0$ imposes no additional restriction on its use while a selector with RPL $=3$ can only refer to segments at privilege Level 3 regardless of the task's CPL. RPL is generally used to verify that pointer parameters passed to a more trusted procedure are not allowed to use data at a more privileged level than the caller (refer to pointer testing instructions).

## Descriptor Access and Privilege Validation

Determining the ability of a task to access a segment involves the type of segment to be accessed, the instruction used, the type of descriptor used and CPL, RPL, and DPL. The two basic types of segment accesses are control transfer (selectors loaded into CS) and data (selectors loaded into DS, ES or SS).

## DATA SEGMENT ACCESS

Instructions that load selectors into DS and ES must refer to a data segment descriptor or readable code segment descriptor. The CPL of the task and the RPL of the selector must be the same as or more privileged (numerically equal to or lower than) than the descriptor DPL. In general, a task can only access data segments at the same or less privileged levels than the CPL or RPL (whichever is numerically higher) to prevent a program from accessing data it cannot be trusted to use.
An exception to the rule is a readable conforming code segment. This type of code segment can be read from any privilege level.
If the privilege checks fail (e.g. DPL is numerically less than the maximum of CPL and RPL) or an incorrect type of descriptor is referenced (e.g. gate de-
scriptor or execute only code segment) exception 13 occurs. If the segment is not present, exception 11 is generated.
Instructions that load selectors into SS must refer to data segment descriptors for writable data segments. The descriptor privilege (DPL) and RPL must equal CPL. All other descriptor types or a privilege level violation will cause exception 13. A not present fault causes exception 12.

## CONTROL TRANSFER

Four types of control transfer can occur when a selector is loaded into CS by a control transfer operation (see Table 10). Each transfer type can only occur if the operation which loaded the selector references the correct descriptor type. Any violation of these descriptor usage rules (e.g. JMP through a call gate or RET to a Task State Segment) will cause exception 13.

The ability to reference a descriptor for control transfer is also subject to rules of privilege. A CALL or JUMP instruction may only reference a code segment descriptor with DPL equal to the task CPL or a conforming segment with DPL of equal or greater privilege than CPL. The RPL of the selector used to reference the code descriptor must have as much privilege as CPL.
RET and IRET instructions may only reference code segment descriptors with descriptor privilege equal to or less privileged than the task CPL. The selector loaded into CS is the return address from the stack. After the return, the selector RPL is the task's new CPL. If CPL changes, the old stack pointer is popped after the return address.
When a JMP or CALL references a Task State Segment descriptor, the descriptor DPL must be the same or less privileged than the task's CPL. Refer-
ence to a valid Task State Segment descriptor causes a task switch (see Task Switch Operation). Reference to a Task State Segment descriptor at a more privileged level than the task's CPL generates exception 13.
When an instruction or interrupt references a gate descriptor, the gate DPL must have the same or less privilege than the task CPL. If DPL is at a more privileged level than CPL, exeception 13 occurs. If the destination selector contained in the gate references a code segment descriptor, the code segment descriptor DPL must be the same or more privileged than the task CPL. If not, Exception 13 is issued. After the control transfer, the code segment descriptors DPL is the task's new CPL. If the destination selector in the gate references a task state segment, a task switch is automatically performed (see Task Switch Operation).

The privilege rules on control transfer require:

- JMP or CALL direct to a code segment (code segment descriptor) can only be to a conforming segment with DPL of equal or greater privilege than CPL or a non-conforming segment at the same privilege level.
- interrupts within the task or calls that may change privilege levels, can only transfer control through a gate at the same or a less privileged level than CPL to a code segment at the same or more privileged level than CPL.
- return instructions that don't switch tasks can only return control to a code segment at the same or less privileged level.
- task switch can be performed by a call, jump or interrupt which references either a task gate or task state segment at the same or less privileged level.

Table 10. Descriptor Types Used for Control Transfer

| Control Transfer Types | Operation Types | Descriptor Referenced | Descriptor Table |
| :---: | :---: | :---: | :---: |
| Intersegment within the same privilege level | JMP, CALL, RET, IRET* | Code Segment | GDT/LDT |
| Intersegment to the same or higher privilege level Interrupt within task may change CPL. | CALL | Call Gate | GDT/LDT |
|  | Interrupt Instruction, Exception, External Interrupt | Trap or Interrupt Gate | IDT |
| Intersegment to a lower privilege level (changes task CPL) | RET, IRET* | Code Segment | GDT/LDT |
| $\cdots$. | CALL, JMP | Task State Segment | GDT |
| Task Switch | CALL, JMP | Task Gate | GDT/LDT |
|  | IRET** <br> Interrupt Instruction, Exception, External Interrupt | Task Gate | IDT |

*NT (Nested Task bit of flag word) $=0$
**NT (Nested Task bit of flag word) $=1$

## PRIVILEGE LEVEL CHANGES

Any control transfer that changes CPL within the task, causes a change of stacks as part of the operation. Initial values of SS:SP for privilege levels 0,1 , and 2 are kept in the task state segment (refer to Task Switch Operation). During a JMP or CALL control transfer, the new stack pointer is loaded into the SS and SP registers and the previous stack pointer is pushed onto the new stack.

When returning to the original privilege level, its stack is restored as part of the RET or IRET instruction operation. For subroutine calls that pass parameters on the stack and cross privilege leveis, a fixed number of words, as specified in the gate, are copied from the previous stack to the current stack. The inter-segment RET instruction with a stack adjustment value will correctly restore the previous stack pointer upon return.

## Protection

The 80286 includes mechanisms to protect critical instructions that affect the CPU execution state (e.g. HLT) and code or data segments from improper usage. These protection mechanisms are grouped into three forms:

Restricted usage of segments (e.g. no write allowed to read-only data segments). The only segments available for use are defined by descriptors in the Local Descriptor Table (LDT) and Global Descriptor Table (GDT).
Restricted access to segments via the rules of privilege and descriptor usage.
Privileged instructions or operations that may only be executed at certain privilege levels as determined by the CPL and I/O Privilege Level (IOPL). The IOPL is defined by bits 14 and 13 of the flag word.

These checks are performed for all instructions and can be split into three categories: segment load checks (Table 11), operand reference checks (Table 12), and privileged instruction checks (Table 13). Any violation of the rules shown will result in an exception. A not-present exception related to the stack segment causes exception 12.

The IRET and POPF instructions do not perform some of their defined functions if CPL is not of sufficient privilege (numerically small enough). Precisely these are:

- The IF bit is not changed if CPL > IOPL.
- The IOPL field of the flag word is not changed if CPL > 0 .
No exceptions or other indication are given when these conditions occur.

Table 11
Segment Register Load Checks

| Error Description | Exception <br> Number |
| :--- | :---: |
| Descriptor table limit exceeded | 13 |
| Segment descriptor not-present | 11 or 12 |
| Privilege rules violated | 13 |
| Invalid descriptor/segment type seg- |  |
| ment register load: |  |
| -Read only data segment load to |  |
| SS |  |
| -Special Control descriptor load to | 13 |
| DS, ES, SS |  |
| - Execute only segment load to |  |
| DS, ES, SS |  |
| -Data segment load to CS |  |
| -Read/Execute code segment |  |
| load to SS |  |

Table 12. Operand Reference Checks

| Error Description | Exception <br> Number |
| :--- | :--- |
| Write into code segment | 13 |
| Read from execute-only code | 13 |
| segment | 13 |
| Write to read-only data segment | 12 or 13 |
| Segment limit exceeded 1 |  |

NOTE:
Carry out in offset calculations is ignored.
Table 13. Privileged Instruction Checks

| Error Description | Exception <br> Number |
| :---: | :---: |
| CPL $\neq 0$ when executing the following <br> instructions: <br> LID, LLD, LGDT, LTR, LMSW, | 13 |
| CTS, HLT |  |

## EXCEPTIONS

The 80286 detects several types of exceptions and interrupts, in protected mode (see Table 14). Most are restartable after the exceptional condition is removed. Interrupt handlers for most exceptions can read an error code, pushed on the stack after the return address, that identifies the selector involved ( 0 if none). The return address normally points to the failing instruction, including all leading prefixes. For a processor extension segment overrun exception, the return address will not point at the ESC instruction that caused the exception; however, the processor extension registers may contain the address of the failing instruction.

Table 14. Protected Mode Exceptions

| Interrupt <br> Vector | Function | Return <br> Address <br> At Falling <br> Instruction? | Always <br> Restart- <br> able? | Error <br> Code <br> on Stack? |
| :---: | :--- | :---: | :---: | :---: |
| 8 | Double exception detected | Yes | $\mathrm{No}^{2}$ | Yes |
| 9 | Processor extension segment overrun | No | No | No |
| 10 | Invalid task state segment | Yes | Yes | Yes |
| 11 | Segment not present | Yes | Yes | Yes |
| 12 | Stack segment overrun or stack segment not present | Yes | Yes | Yes |
| 13 | General protection | Yes | No $^{2}$ | Yes |

NOTE:

1. When a PUSHA or POPA instruction attempts to wrap around the stack segment, the machine state after the exception will not be restartable because stack segment wrap around is not permitted. This condition is identified by the value of the saved SP being either $0000(\mathrm{H}), 0001(\mathrm{H}), \operatorname{FFFE}(\mathrm{H})$, or $\operatorname{FFFF}(\mathrm{H})$.
2. These exceptions indicate a violation to privilege rules or usage rules has occurred. Restart is generally not attempted under those conditions.

These exceptions indicate a violation to privilege rules or usage rules has occurred. Restart is generally not attempted under those conditions.

All these checks are performed for all instructions and can be split into three categories: segment load checks (Table 11), operand reference checks (Table 12), and privileged instruction checks (Table 13). Any violation of the rules shown will result in an exception. A not-present exception causes exception 11 or 12 and is restartable.

## Special Operations

## TASK SWITCH OPERATION

The 80286 provides a built-in task switch operation which saves the entire 80286 execution state (registers, address space, and a link to the previous task), loads a new execution state, and commences execution in the new task. Like gates, the task switch operation is invoked by executing an inter-segment JMP or CALL instruction which refers to a Task State Segment (TSS) or task gate descriptor in the GDT or LDT. An INT n instruction, exception, or external interrupt may also invoke the task switch operation by selecting a task gate descriptor in the associated IDT descriptor entry.

The TSS descriptor points at a segment (see Figure 20) containing the entire 80286 execution state while a task gate descriptor contains a TSS selector. The limit field of the descriptor must be $>002 \mathrm{~B}(\mathrm{H})$.

Each task must have a TSS associated with it. The current TSS is identified by a special register in the 80286 called the Task Register (TR). This register contains a selector referring to the task state segment descriptor that defines the current TSS. A hidden base and limit register associated with TR are loaded whenever TR is loaded with a new selector.

The IRET instruction is used to return control to the task that called the current task or was interrupted. Bit 14 in the flag register is called the Nested Task (NT) bit. It controls the function of the IRET instruction. If $N T=0$, the IRET instruction performs the regular current task by popping values off the stack; when NT $=1$, IRET performs a task switch operation back to the previous task.
When a CALL, JMP, or INT instruction initiates a task switch, the old (except for case of JMP) and new TSS will be marked busy and the back link field of the new TSS set to the old TSS selector. The NT bit of the new task is set by CALL or INT initiated task switches. An interrupt that does not cause a task switch will clear NT. NT may also be set or cleared by POPF or IRET instructions.
The task state segment is marked busy by changing the descriptor type field from Type 1 to Type 3. Use of a selector that references a busy task state segment causes Exception 13.

## PROCESSOR EXTENSION CONTEXT SWITCHING

The context of a processor extension (such as the 80287 numerics processor) is not changed by the task switch operation. A processor extension context need only be changed when a different task attempts to use the processor extension (which still contains the context of a previous task). The 80286 detects the first use of a processor extension after a task switch by causing the processor extension not present exception (7). The interrupt handler may then decide whether a context change is necessary.

Whenever the 80286 switches tasks, it sets the Task Switched (TS) bit of the MSW. TS indicates that a processor extension context may belong to a different task than the current one. The processor extension not present exception (7) will occur when attempting to execute an ESC or WAIT instruction if TS =1 and a processor extension is present (MP = 1 in MSW).

## POINTER TESTING INSTRUCTIONS

The 80286 provides several instructions to speed pointer testing and consistency checks for maintaining system integrity (see Table 15). These instruc-
tions use the memory management hardware to verify that a selector value refers to an appropriate segment without risking an exception. A condition flag (ZF) indicates whether use of the selector or segment will cause an exception.


Figure 20. Task State Segment and TSS Registers

Table 15. 80286 Pointer Test Instructions

| Instruction | Operands | Function |
| :--- | :--- | :--- |
| ARPL | Selector, <br> Register | Adjust Requested Privilege <br> Level: adjusts the RPL of <br> the selector to the numeric <br> maximum of current selec- <br> tor RPL value and the RPL <br> value in the register. Set <br> zero flag if selector RPL <br> was changed by ARPL. |
| VERR | Selector | VERify for Read: sets the <br> zero flag if the segment re- <br> ferred to by the selector <br> can be read. |
| VERW | Selector | VERify for Write: sets the <br> zero flag if the segment re- <br> ferred to by the selector <br> can be written. |
| LSL | Register, <br> Selector | Load Segment Limit: reads <br> the segment limit into the <br> register if privitege rules <br> and descriptor type allow. <br> Set zero flag if successful. |
| LAR | Register, <br> Selector | Load Access Rights: reads <br> the descriptor access <br> rights byte into the register <br> if privilege rules allow. Set <br> zero flag if successful. |

## DOUBLE FAULT AND SHUTDOWN

If two separate exceptions are detected during a single instruction execution, the 80286 performs the double fault exception (8). If an execution occurs during processing of the double fault exception, the 80286 will enter shutdown. During shutdown no further instructions or exceptions are processed. Either NMI (CPU remains in protected mode) or RESET (CPU exits protected mode) can force the 80286 out of shutdown. Shutdown is externally signalled via a HALT bus operation with $\mathrm{A}_{1}$ LOW.

## PROTECTED MODE INITIALIZATION

The 80286 initially executes in real address mode after RESET. To allow initialization code to be placed at the top of physical memory, $A_{23}-A_{20}$ will be HIGH when the 80286 performs memory references relative to the CS register until CS is changed. $A_{23}-A_{20}$ will be zero for references to the DS, ES, or SS segments. Changing CS in real address mode will force $A_{23}-A_{20}$ LOW whenever CS is used again. The initial CS:IP value of F000:FFFO provides 64 K bytes of code space for initialization code without changing CS.
Protected mode operation requires several registers to be initialized. The GDT and IDT base registers must refer to a valid GDT and IDT. After executing the LMSW instruction to set PE, the 80286 must im-
mediately execute an intra-segment JMP instruction to clear the instruction queue of instructions decoded in real address mode.

To force the 80286 CPU registers to match the initial protected mode state assumed by software, execute a JMP instruction with a selector referring to the initial TSS used in the system. This will load the task register, local descriptor table register, segment registers and initial general register state. The TR should point at a valid TSS since any task switch operation involves saving the current task state.

## SYSTEM INTERFACE

The 80286 system interface appears in two forms: a local bus and a system bus. The local bus consists of address, data, status, and control signals at the pins of the CPU. A system bus is any buffered version of the local bus. A system bus may also differ from the local bus in terms of coding of status and control lines and/or timing and loading of signals. The 80286 family includes several devices to generate standard system buses such as the IEEE 796 standard MULTIBUS.

## Bus Interface Signals and Timing

The 80286 microsystem local bus interfaces the 80286 to local memory and 1/O components. The interface has 24 address lines, 16 data lines, and 8 status and control signals.
The 80286 CPU, 82C284 clock generator, 82288 bus controller, 82289 bus arbiter, tranceivers, and latches provide a buffered and decoded system bus interface. The 82C284 generates the system clock and synchronizes READY and RESET. The 82288 converts bus operation status encoded by the 80286 into command and bus control signals. The 82289 bus arbiter generates Multibus bus arbitration signals. These components can provide the timing and electrical power drive levels required for most system bus interfaces including the Multibus.

## Physical Memory and I/O Interface

A maximum of 16 megabytes of physical memory can be addressed in protected mode. One megabyte can be addressed in real address mode. Memory is accessible as bytes or words. Words consist of any two consecutive bytes addressed with the least significant byte stored in the lowest address.
Byte transfers occur on either half of the 16-bit local data bus. Even bytes are accessed over $\mathrm{D}_{7-0}$ while odd bytes are transferred over $\mathrm{D}_{15-8}$. Even-addressed words are transferred over $D_{15-0}$ in one bus cycle, while odd-addressed word require two bus operations. The first transfers data on $D_{15-8}$, and the second transfers data on $D_{7-0}$. Both byte data transfers occur automatically, transparent to software.

Two bus signals, $A_{0}$ and $\overline{\mathrm{BHE}}$, control transfers over the lower and upper halves of the data bus. Even address byte transfers are indicated by $A_{0}$ LOW and $\overline{\text { BHE HIGH. Odd address byte transfers are indicat- }}$ ed by $\mathrm{A}_{0}$ HIGH and BHE LOW. Both $\mathrm{A}_{0}$ and $\overline{\mathrm{BHE}}$ are LOW for even address word transfers.

The I/O address space contains 64 K addresses in both modes. The 1/O space is accessible as either bytes or words, as is memory. Byte wide peripheral devices may be attached to either the upper or lower byte of the data bus. Byte-wide I/O devices attached to the upper data byte ( $\mathrm{D}_{15-8}$ ) are accessed with odd I/O addresses. Devices on the lower data byte are accessed with even 1/O addresses. An interrupt controller such as Intel's 8259A must be connected to the lower data byte ( $D_{7-0}$ ) for proper return of the interrupt vector.

## Bus Operation

The 80286 uses a double frequency system clock (CLK input) to control bus timing. All signals on the local bus are measured relative to the system CLK input. The CPU divides the system clock by 2 to produce the internal processor clock, which determines bus state. Each processor clock is composed of two. system clock cycles named phase 1 and phase 2. The 82C284 clock generator output (PCLK) identifies the next phase of the processor clock. (See Figure 21.)


Figure 21. System and Processor Clock Relationships
Six types of bus operations are supported; memory read, memory write, I/O read, I/O write, interrupt acknowledge, and halt/shutdown. Data can be transferred at a maximum rate of one word per two processor clock cycles.
The 80286 bus has three basic states: idle ( $\mathrm{T}_{\mathrm{i}}$ ), send status ( $T_{s}$ ), and perform command ( $T_{c}$ ). The 80286 CPU also has a fourth local bus state called hold ( $T_{h}$ ). $T_{h}$ indicates that the 80286 has surrendered control of the local bus to another bus master in response to a HOLD request.
Each bus state is one processor clock long. Figure 22 shows the four 80286 local bus states and allowed transitions.


Figure 22. 80286 Bus States

## Bus States

The idle ( $\mathrm{T}_{\mathrm{i}}$ ) state indicates that no data transfers are in progress or requested. The first active state $\mathrm{T}_{\mathrm{S}}$ is signaled by status line $\overline{\mathrm{S} 1}$ or $\overline{\mathrm{SO}}$ going LOW and identifying phase 1 of the processor clock. During $T_{s}$, the command encoding, the address, and data (for a write operation) are available on the 80286 output pins. The 82288 bus controller decodes the status signals and generates Multibus compatible read/write command and local transceiver control signals.
After $T_{S}$, the perform command ( $T_{C}$ ) state is entered. Memory or $1 / O$ devices respond to the bus operation during $T_{C}$, either transferring read data to the CPU or accepting write data. $T_{C}$ states may be repeated as often as necessary to assure sufficient time for the memory or 1/O device to respond. The $\overline{R E A D Y}$ signal determines whether $T_{C}$ is repeated. A repeated $T_{C}$ state is called a wait state.
During hold ( $T_{h}$ ), the 80286 will float all address, data, and status output pins enabling another bus master to use the local bus. The 80286 HOLD input signal is used to place the 80286 into the $T_{h}$ state. The 80286 HLDA output signal indicates that the CPU has entered $T_{h}$.

## Pipelined Addressing

The 80286 uses a local bus interface with pipelined timing to allow as much time as possible for data access. Pipelined timing allows a new bus operation to be initiated every two processor cycles, while allowing each individual bus operation to last for three processor cycles.
The timing of the address outputs is pipelined such that the address of the next bus operation becomes available during the current bus operation. Or in other words, the first clock of the next bus operation is overlapped with the last clock of the current bus opèration. Therefore, address decode and routing logic can operate in advance of the next bus operation.


Figure 23. Basic Bus Cycle

External address latches may hold the address stable for the entire bus operation, and provide additional $A C$ and $D C$ buffering.

The 80286 does not maintain the address of the current bus operation during all $T_{c}$ states. Instead, the address for the next bus operation may be emitted during phase 2 of any $T_{c}$. The address remains valid during phase 1 of the first $T_{c}$ to guarantee hold time, relative to $A L E$, for the address latch inputs.

## Bus Control Signals

The 82288 bus controller provides control signals; address latch enable (ALE), Read/Write commands, data transmit/receive ( $\mathrm{DT} / \overline{\mathrm{R}}$ ), and data enable (DEN) that control the address latches, data transceivers, write enable, and output enable for memory and I/O systems.
The Address Latch Enable (ALE) output determines when the address may be latched. ALE provides at least one system CLK period of address hold time from the end of the previous bus operation until the address for the next bus operation appears at the latch outputs. This address hold time is required to support MULTIBUS and common memory systems.
The data bus transceivers are controlled by 82288 outputs Data Enable (DEN) and Data Transmit/Receive ( $\mathrm{DT} / \overline{\mathrm{R}}$ ). DEN enables the data transceivers; while DT/ $\overline{\mathrm{R}}$ controls tranceiver direction. DEN and $D T / \bar{R}$ are timed to prevent bus contention between the bus master, data bus transceivers, and system data bus transceivers.

## Command Timing Controls

Two system timing customization options, command extension and command delay, are provided on the 80286 local bus.

Command extension allows additional time for external devices to respond to a command and is analogous to inserting wait states on the 8086. External logic can control the duration of any bus operation such that the operation is only as long as necessary. The READY input signal can extend any bus operation for as long as necessary.

Command delay allows an increase of address or write data setup time to system bus command active for any bus operation by delaying when the system bus command becomes active. Command delay is controlled by the 82288 CMDLY input. After Ts, the bus controller samples CMDLY at each failing edge of CLK. If CMDLY is HIGH, the 82288 will not activate the command signal. When CMDLY is LOW, the 82288 will activate the command signal. After the command becomes active, the CMDLY input is not sampled.

When a command is delayed, the available response time from command active to return read data or accept write data is less. To customize system bus timing, an address decoder can determine which bus operations require delaying the command. The CMDLY input does not affect the timing of ALE, DEN, or DT/R.


Figure 24. CMDLY Controls the Leading Edge of Command Signal

Figure 24 illustrates four uses of CMDLY. Example 1 shows delaying the read command two system CLKs for cycle N-1 and no delay for cycle N, and example 2 shows delaying the read command one system CLK for cycle $\mathrm{N}-1$ and one system CLK delay for cycle N .

## Bus Cycle Termination

At maximum transfer rates, the 80286 bus alternates between the status and command states. The bus status signals become inactive after $T_{s}$ so that they may correctly signal the start of the next bus operation after the completion of the current cycle. No external indication of $T_{c}$ exists on the 80286 local bus. The bus master and bus controller enter $T_{c}$ directly after $T_{s}$ and continue executing $T_{c}$ cycles until terminated by READY.

## READY Operation

The current bus master and 82288 bus controller terminate each bus operation simultaneously to achieve maximum bus operation bandwidth. Both are informed in advance by READY active (opencollector output from 82C284) which identifies the last $T_{C}$ cycle of the current bus operation. The bus master and bus controller must see the same sense
of the $\overline{R E A D Y}$ signal, thereby requiring $\overline{\text { READY }}$ be synchronous to the system clock.

## Synchronous Ready

The 82C284 clock generator provides $\overline{\text { READY syn- }}$ chronization from both synchronous and asynchronous sources (see Figure 25). The synchronous ready input (SRDY) of the clock generator is sampled with the falling edge of CLK at the end of phase 1 of each $T_{c}$. The state of SRDY is then broadcast to the bus master and bus controller via the READY output line.

## Asynchronous Ready

Many systems have devices or subsystems that are asynchronous to the system clock. As a result, their ready outputs cannot be guaranteed to meet the 82C284 SRDY setup and hold time requirements. But the 82C284 asynchronous ready input ( $\overline{\mathrm{ARDY}}$ ) is designed to accept such signals. The ARDY input is sampled at the beginning of each $T_{C}$ cycle by 82C284 synchronization logic. This provides one system CLK cycle time to resolve its value before broadcasting it to the bus master and bus controller.


Figure 25. Synchronous and Asynchronous Ready
$\overline{\text { ARDY }}$ or $\overline{\text { ARDYEN }}$ must be HIGH at the end of Ts. ARDY cannot be used to terminate bus cycle with no wait states.

Each ready input of the 82C284 has an enable pin (SRDYEN and ARDYEN) to select whether the current bus operation will be terminated by the synchronous or asynchronous ready. Either of the ready inputs may terminate a bus operation. These enable inputs are active low and have the same timing as their respective ready inputs. Address decode logic usually selects whether the current bus operation should be terminated by $\overline{\text { ARDY }}$ or $\overline{\text { SRDY. }}$

## Data Bus Control

Figures 26, 27, and 28 show how the DT/ $\overline{\mathrm{R}}, \mathrm{DEN}$, data bus, and address signals operate for different combinations of read, write, and idle bus operations. DT/侣 goes active (LOW) for a read operation. DT/信 remains HIGH before, during, and between write operations.

The data bus is driven with write data during the second phase of $T_{s}$. The delay in write data timing allows the read data drivers, from a previous read cycle, sufficient time to enter 3-state OFF before the 80286 CPU begins driving the local data bus for write operations. Write data will always remain valid for one system clock past the last $T_{c}$ to provide sufficient hold time for Multibus or other similar memory or I/O systems. During write-read or write-idle sequences the data bus enters 3 -state OFF during the second phase of the processor cycle after the last $T_{c}$. In a write-write sequence the data bus does not enter 3-state OFF between $T_{c}$ and $T_{s}$.

## Bus Usage

The 80286 local bus may be used for several functions: instruction data transfers, data transfers by other bus masters, instruction fetching, processor extension data transfers, interrupt acknowledge, and halt/shutdown. This section describes local bus activities which have special signals or requirements.


Figure 26. Back to Back Read-Write Cycles


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Figure 27. Back to Back Write-Read Cycles


Figure 28. Back to Back Write-Write Cycles

## HOLD and HLDA

HOLD AND HLDA allow another bus master to gain control of the local bus by placing the 80286 bus into the $T_{h}$ state. The sequence of events required to pass control between the 80286 and another local bus master are shown in Figure 29.

In this example, the 80286 is initially in the $T_{h}$ state as signaled by HLDA being active. Upon leaving $T_{h}$, as signaled by HLDA going inactive, a write operation is started. During the write operation another local bus master requests the local bus from the 80286 as shown by the HOLD signal. After completing the write operation, the 80286 performs one $T_{i}$ bus cycle, to guarantee write data hold time, then enters $T_{h}$ as signaled by HLDA going active.

The CMDLY signal and $\overline{\text { ARDY }}$ ready are used to start and stop the write bus command, respectively. Note that SRDY must be inactive or disabled by $\overline{\text { SRDYEN }}$ to guarantee $\overline{\text { ARDY will terminate the cy- }}$ cle.

HOLD must not be active during the time from the leading edge of RESET until 34 CLKs following the trailing edge of RESET.

## Lock

The CPU asserts an active lock signal during Inter-rupt-Acknowledge cycles, the XCHG instruction, and during some descriptor accesses. Lock is also asserted when the LOCK prefix is used. The LOCK prefix may be used with the following ASM-286 assembly instructions; MOVS, INS, and OUTS. For bus
cycles other than Interrupt-Acknowledge cycles, Lock will be active for the first and subsequent cycles of a series of cycles to be locked. Lock will not be shown active during the last cycle to be locked. For the next-to-last cycle, Lock will become inactive at the end of the first $T_{c}$ regardiess of the number of wait-states inserted. For Interrupt-Acknowledge cycles, Lock will be active for each cycle, and will become inactive at the end of the first $T_{c}$ for each cycle regardless of the number of wait-states inserted.

## Instruction Fetching

The 80286 Bus Unit (BU) will fetch instructions ahead of the current instruction being executed. This activity is called prefetching. It occurs when the local bus would otherwise be idle and obeys the following rules:

A prefetch bus operation starts when at least two bytes of the 6 -byte prefetch queue are empty.

The prefetcher normally performs word prefetches independent of the byte alignment of the code segment base in physical memory.

The prefetcher will perform only a byte code fetch operation for control transfers to an instruction beginning on a numerically odd physical address.

Prefetching stops whenever a control transfer or HLT instruction is decoded by the IU and placed into the instruction queue.
In real address mode, the prefetcher may fetch up to 6 bytes beyond the last control transfer or HLT instruction in a code segment.

In protected mode, the prefetcher will never cause a segment overrun exception. The prefetcher stops at the last physical memory word of the code segment. Exception 13 will occur if the program attempts to execute beyond the last full instruction in the code segment.

If the last byte of a code segment appears on an even physical memory address, the prefetcher will read the next physical byte of memory (perform a word code fetch). The value of this byte is ignored and any attempt to execute it causes exception 13.


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NOTES:

1. Status lines are not driven by 80286 , yet remain high due to pullup resistors in 82288 and 82289 during HOLD state. 2. Address, $M / \overline{I O}$ and $C O D / \mathbb{N T A}$ may start floating during any $T_{C}$ depending on when internal 80286 bus arbiter decides to release bus to external HOLD. The float starts in $\phi 2$ of $T_{C}$.
2. $\overline{B H E}$ and $\overline{L O C K}$ may start floating after the end of any $T_{C}$ depending on when internal 80286 bus arbiter decides to release bus to external HOLD. The float starts in $\phi 1$ of $T_{C}$.
3. The minimum HOLD to HLDA time is shown. Maximum is one $T_{H}$ longer.
4. The earliest HOLD time is shown. It will always allow a subsequent memory cycle if pending is shown.
5. The minimum HOLD to HLDA time is shown. Maximum is a function of the instruction, type of bus cycle and other machine state (i.e., Interrupts, Waits, Lock, etc.).
6. Asynchronous ready allows termination of the cycie. Synchronous ready does not signal ready in this example. Synchronous ready state is ignored after ready is signaled via the asynchronous input.

Figure 29. MULTIBUS* Write Terminated by Asynchronous Ready with Bus Hold

## Processor Extension Transfers

The processor extension interface uses I/O port addresses $00 \mathrm{F8}(\mathrm{H}), 00 \mathrm{FA}(\mathrm{H})$, and $00 \mathrm{FC}(\mathrm{H})$ which are part of the I/O port address range reserved by intel. An ESC instruction with Machine Status Word bits $\mathrm{EM}=0$ and $\mathrm{TS}=0$ will perform I/O bus operations to one or more of these I/O port addresses independent of the value of IOPL and CPL.

ESC instructions with memory references enable the CPU to accept PEREQ inputs for processor extension operand transfers. The CPU will determine the operand starting address and read/write status of the instruction. For each operand transfer, two or three bus operations are performed, one word transfer with I/O port address 00FA(H) and one or two bus operations with memory. Three bus operations are required for each word operand aligned on an odd byte address.

## NOTE:

Odd-aligned numerics operands should be avoided when using an 80286 system running six or more memory-write wait states. The 80286 can generate an incorrect numerics address if all the following conditions are met:

- Two floating point (FP) instructions are fetched and in the 80286 queue.
- The first FP instruction is any floating point store except FSTSW AX.
- The second FP instruction accesses memory.
- The operand of the first instruction is aligned on an odd memory address.
- Six or more wait states are inserted during either of the last two memory write (odd aligned operands are transferred as two bytes) transfers of the first instruction.

The second FP operand's address will be incremented by one if these conditions are met. These conditions are most likely to occur in a multi-master system. For a hardware solution, contact your local Intel representative.

Commands to the numerics coprocessor should not be delayed by nine or more $T$-states. Excessive (nine or more) command-delays can cause the 80286 and 80287 to lose synchronization.

## Interrupt Acknowledge Sequence

Figure 30 illustrates an interrupt acknowledge sequence performed by the 80286 in response to an

INTR input. An interrupt acknowledge sequence consists of two INTA bus operations. The first allows a master 8259A Programmable Interrupt Controller (PIC) to determine which if any of its slaves should return the interrupt vector. An eight bit vector is read on D0-D7 of the 80286 during the second INTA bus operation to select an interrupt handler routine from the interrupt table.

The Master Cascade Enable (MCE) signal of the 82288 is used to enable the cascade address drivers, during INTA bus operations (See Figure 30), onto the local address bus for distribution to slave interrupt controllers via the system address bus. The 80286 emits the LOCK signal (active LOW) during $T_{s}$ of the first INTA bus operation. A local bus "hold" request will not be honored until the end of the second INTA bus operation.

Three idle processor clocks are provided by the 80286 between INTA bus operations to allow for the minimum INTA to INTA time and CAS (cascade address) out delay of the 8259A. The second INTA bus operation must always have at least one extra $T_{c}$ state added via logic controlling READY. This is needed to meet the 8259A minimum INTA pulse width.

## Local Bus Usage Priorities

The 80286 local bus is shared among several internal units and external HOLD requests. In case of simultarieous requests, their relative priorities are:
(Highest) Any transfers which assert LOCK either explicitly (via the LOCK instruction prefix) or implicitly (i.e. some segment descriptor accesses, interrupt acknowledge sequence, or an XCHG with memory).
The second of the two byte bus operations required for an odd aligned word operand.
The second or third cycle of a processor extension data transfer.
Local bus request via HOLD input.
Processor extension data operand transfer via PEREQ input.
Data transfer performed by EU as part of an instruction.
(Lowest) An instruction prefetch request from BU. The EU will inhibit prefetching two processor clocks in advance of any data transfers to minimize waiting by EU for a prefetch to finish.


## NOTES:

1. Data is ignored, upper data bus, $D_{8}-D_{15}$, should not change state during this time.
2. First INTA cycle should have at least one wait state inserted to meet 8259A minimum INTA pulse width.
3. Second INTA cycle should have at least one wait state inserted to meet 8259A minimum INTA pulse width.
4. LOCK is active for the first INTA cycle to prevent the 82289 from releasing the bus between INTA cycles in a multimaster system. LOCK is also active for the second INTA cycle.
5. $A_{23}-A_{0}$ exits 3 -state OFF during $\phi 2$ of the second $T_{C}$ in the INTA cycle.
6. Upper data bus should not change state during this time.

Figure 30. Interrupt Acknowledge Sequence

## Halt or Shutdown Cycles

The 80286 externally indicates halt or shutdown conditions as a bus operation. These conditions occur due to a HLT instruction or multiple protection exceptions while attempting to execute one instruction. A halt or shutdown bus operation is signalled when $\overline{\mathrm{S} 1}, \overline{\mathrm{SO}}$ and COD/INTA are LOW and M/IO is HIGH. A $A_{1}$ HIGH indicates halt, and $A_{1}$ LOW indicates shutdown. The 82288 bus controller does not
issue ALE, nor is $\overline{\text { READY required to terminate a halt }}$ or shutdown bus operation.

During halt or shutdown, the 80286 may service PEREQ or HOLD requests. A processor extension segment overrun exception during shutdown will inhibit further service of PEREQ. Either NMI or RESET will force the 80286 out of either halt or shutdown. An INTR, if interrupts are enabled, or a processor extension segment overrun exception will also force the 80286 out of halt.


Figure 31. Basic 80286 System Configuration

## SYSTEM CONFIGURATIONS

The versatile bus structure of the 80286 microsystem, with a full complement of support chips, allows flexible configuration of a wide range of systems. The basic configuration, shown in Figure 31, is similar to an 8086 maximum mode system. It includes the CPU plus an 8259A interrupt controller, 82C284 clock generator, and the 82288 Bus Controller.

As indicated by the dashed lines in Figure 31, the ability to add processor extensions is an integral feature of 80286 microsystems. The processor extension interface allows external hardware to perform special functions and transfer data concurrent with CPU execution of other instructions. Full system integrity is maintained because the 80286 supervises all data transfers and instruction execution for the processor extension.

The 80287 has all the instructions and data types of an 8087. The 80287 NPX can periorm numeric calculations and data transfers concurrently with CPU program execution. Numerics code and data have the same integrity as all other information protected by the 80286 protection mechanism.

The 80286 can overlap chip select decoding and address propagation during the data transfer for the previous bus operation. This information is latched by ALE during the middle of a $\mathrm{T}_{\mathrm{s}}$ cycle. The latched chip select and address information remains stable during the bus operation while the next cycle's address is being decoded and propagated into the system. Decode logic can be implemented with a high speed bipolar PROM.

The optional decode logic shown in Figure 31 takes advantage of the overlap between address and data of the 80286 bus cycle to generate advanced memory and IO-select signals. This minimizes system


Figure 32. MULTIBUS* System Bus Interface
performance degradation caused by address propagation and decode delays. In addition to selecting memory and I/O, the advanced selects may be used with configurations supporting local and system buses to enable the appropriate bus interface for each bus cycle. The COD/INTA and M/Ī signals are applied to the decode logic to distinguish between interrupt, I/O, code and data bus cycles.

By adding the 82289 bus arbiter chip, the 80286 provides a MULTIBUS system bus interface as shown in Figure 32. The ALE output of the 82288 for the

MULTIBUS bus is connected to its CMDLY input to delay the start of commands one system CLK as required to meet MULTIBUS address and write data setup times. This arrangement will add at least one extra $T_{c}$ state to each bus operation which uses the MULTIBUS.

A second 82288 bus controller and additional latches and transceivers could be added to the local bus of Figure 32. This configuration allows the 80286 to support an on-board bus for local memory and peripherals, and the MULTIBUS for system bus interfacing.


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Figure 33. 80286 System Configuration with Dual-Ported Memory

Figure 33 shows the addition of dual ported dynamic memory between the MULTIBUS system bus and the 80286 local bus. The dual port interface is provided by the 8207 Dual Port DRAM Controller. The 8207 runs synchronously with the CPU to maximize throughput for local memory references. It also arbitrates between requests from the local and system buses and performs functions such as refresh,
initialization of RAM, and read/modify/write cycles. The 8207 combined with the 8206 Error Checking and Correction memory controller provide for single bit error correction. The dual-ported memory can be combined with a standard MULTIBUS system bus interface to maximize performance and protection in multiprocessor system configurations.

Table 16. 80286 Systems Recommended Pull Up Resistor Values

| 80286 Pin and Name | Pullup Value | Purpose |
| :---: | :---: | :---: |
| 4-51 | $20 \mathrm{~K} \Omega \pm 10 \%$ | Pull $\overline{\mathrm{So}}, \overline{\mathrm{S} 1}$, and $\overline{\text { PEACK }}$ inactive during 80286 hold periods(1) |
| 5-- ${ }^{5}$ |  |  |
| 6--PEACK |  |  |
| 63-READY | 910 $\Omega \pm 5 \%$ | Pull READY inactive within required minimum time ( $C_{L}=150 \mathrm{pF}$, $\mathrm{I}_{\mathrm{R}} \leq 7 \mathrm{~mA}$ ) |

## NOTE:

1. Pull-up resistors are not required on $\overline{\mathrm{SO}}$ and $\overline{\mathrm{S} 1}$ when the corresponding pins of the 82 C 284 are connected to $\overline{\mathrm{SO}}$ and $\overline{\mathrm{S} 1}$.

## I2ICETM-286 System Design $^{2}$ Considerations

One of the advantages of using the 80286 is that full in-circuit emulation debugging support is provided through the I2ICE system 80286 probe. To utilize this powerful tool it is necessary that the system designer be aware of a few minor parametric and
functional differences between the 80286 and ${ }^{12}$ ICE system 80286 probe. The I2ICE data sheet (I2ICE Integrated Instrumentation and In-Circuit Emulation System, order \#210469) contains a detailed description of these design considerations. It is recommended that this document be reviewed by the 80286 system designer to determine whether or not these differences affect his design.

## ABSOLUTE MAXIMUM RATINGS*

Ambient Temperature Under Bias $\ldots .0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$
Storage Temperature $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ Voltage on Any Pin with

Respect to Ground............... -1.0 V to +7 V
Power Dissipation. . . . . . . . . . . . . . . . . . . . . . . . . . .3.3W
*Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## D.C. CHARACTERISTICS $\left(V_{C C}=5 \mathrm{~V} \pm 5 \%, \mathrm{~T}_{\text {CASE }}=0^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C}\right)^{*}$

| Symbol | Parameter | Min | Max | Unit | Test Condition |
| :--- | :--- | :---: | :---: | :---: | :---: |
| ICC | Supply Current ( $0^{\circ} \mathrm{C}$ Turn On) |  | 600 | mA | (Note 1) |
| $\mathrm{C}_{\mathrm{CLK}}$ | CLK Input Capacitance |  | 20 | pF | (Note 2) |
| $\mathrm{C}_{\mathrm{IN}}$ | Other Input Capacitance |  | 10 | pF | (Note 2) |
| $\mathrm{C}_{\mathrm{O}}$ | Input/Output Capacitance |  | 20 | pF | (Note 2) |

## NOTES:

1. Tested at worst case load and maximum frequency.
2. These are not tested. They are guaranteed by design characterization.

## D.C. CHARACTERISTICS

$\left(V_{C C}=5 \mathrm{~V} \pm 5 \%, T_{\text {CASE }}=0^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C}\right)^{*}$ Tested at the minimum operating frequency of the part.

| Symbol | Parameter | Min | Max | Unit | Test Condition |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{IL}}$ | Input LOW Voltage | -0.5 | 0.8 | V |  |
| $\mathrm{V}_{\mathrm{IH}}$ | Input HIGH Voltage | 2.0 | $\mathrm{V}_{\mathrm{CC}}+0.5$ | V |  |
| $\mathrm{V}_{\text {ILC }}$ | CLK Input LOW Voltage | -0.5 | 0.6 | V |  |
| $\mathrm{V}_{\text {IHC }}$ | CLK Input HIGH Voltage | 3.8 | $\mathrm{V}_{\text {CC }}+0.5$ | V |  |
| V O | Output LOW Voltge |  | 0.45 | V | $\mathrm{l} \mathrm{OL}=2.0 \mathrm{~mA}$ |
| VOH | Output HIGH Voltage | 2.4 |  | V | $\mathrm{IOH}=-400.0 \mu \mathrm{~A}$ |
| L | Input Leakage Current |  | $\pm 10$ | $\mu \mathrm{A}$ | $0 \mathrm{~V} \leq \mathrm{V}_{\text {IN }} \leq \mathrm{V}_{\text {CC }}$ |
| ILCR | Input CLK, RESET Leakage Current |  | $\pm 10$ | $\mu \mathrm{A}$ | $0.45 \leq \mathrm{V}_{\text {IN }} \leq \mathrm{V}_{\text {CC }}$ |
| ILCR | Input CLK, RESET Leakage Current |  | $\pm 1$ | mA | $0 \leq \mathrm{V}_{\mathbb{N}}<0.45$ |
| ILL | input Sustaining Current on BUSY and ERROR Pins | 30 | 500 | $\mu \mathrm{A}$ | $V_{\text {IN }}=0 \mathrm{~V}$ |
| Lo | Output Leakage Current |  | $\pm 10$ | $\mu \mathrm{A}$ | $0.45 \leq \mathrm{V}_{\text {OUT }} \leq \mathrm{V}_{\text {CC }}$ |
| Lo | Output Leakage Current |  | $\pm 1$ | mA | $0 \leq V_{\text {OUT }}<0.45$ |

${ }^{*} T_{A}$ is guaranteed from $0^{\circ} \mathrm{C}$ to $+55^{\circ} \mathrm{C}$ as long as TCASE is not exceeded.
A.C. CHARACTERISTICS $\left(V_{C C}=5 \mathrm{~V} \pm 5 \%, \text { TCASE }=0^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C}\right)^{*}$

AC timings are referenced to 0.8 V and 2.0 V points of signals as illustrated in datasheet waveforms, unless otherwise noted.

| Symbol | Parameter | 8 MHz |  | 10 MHz |  | $\begin{gathered} 12.5 \mathrm{MHz} \\ \text { (Preliminary) } \\ \hline \end{gathered}$ |  | Unit | Test Condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{array}{\|c\|} \hline-8 \\ \text { Min } \\ \hline \end{array}$ | $\begin{array}{\|c\|} \hline-8 \\ \text { Max } \end{array}$ | $\begin{aligned} & -10 \\ & \text { Min } \\ & \hline \end{aligned}$ | $\begin{aligned} & -10 \\ & \text { Max } \end{aligned}$ | $\begin{aligned} & -12 \\ & \text { Min } \\ & \hline \end{aligned}$ | $\begin{aligned} & -12 \\ & \operatorname{Max} \end{aligned}$ |  |  |
| 1 | System Clock (CLK) Period | 62 | 250 | 50 | 250 | 40 | 250 | ns |  |
| 2 | System Clock (CLK) LOW Time | 15 | 225 | 12 | 232 | 11 | 237. | ns | at 1.0 V |
| 3 | System Clock (CLK) HIGH Time | 25 | 235 | 16 | 239 | 13 | 239 | ns | at 3.6 V |
| 17 | System Clock (CLK) Rise Time |  | 10 | . | 8 | - | 8 | ns | 1.0 V to 3.6 V , (Note 7) |
| 18 | System Clock (CLK) Fall Time |  | 10 |  | 8 | - | 8 | ns | 3.6 V to 1.0 V , (Note 7) |
| 4 | Asynch. Inputs Setup Time | 20 |  | 20 |  | 15 |  | ns | (Note 1) |
| 5 | Asynch. Inputs Hold Time | 20 |  | 20 |  | 15 |  | ns | (Note 1) |
| 6 | RESET Setup Time | 28 |  | 23 |  | 18 |  | ns |  |
| 7 | RESET Hold Time | 5 |  | 5 |  | 5 |  | ns |  |
| 8 | Read Data Setup Time | 10 |  | 8 |  | 5 |  | ns |  |
| 9 | Read Data Hold Time | 8 |  | 8 |  | 6 |  | ns |  |
| 10 | READY Setup Time | 38 |  | 26 |  | 22 |  | ns |  |
| 11 | READY Hold Time | 25 |  | 25 |  | 20 |  | ns |  |
| 12 | Status/产EACK Valid Delay | 1 | 40 | - | - | - | - | ns | (Notes 2, 3) |
| 12 a 1 | Status Active Delay | - | - | 1 | 22 | 3 | 18 | ns | (Notes 2, 3) |
| 12a2 | PEACK Active Delay | - | - | 1 | 22 | 3 | 20 | ns | (Notes 2, 3) |
| 12b | Status/PEACK Inactive Delay | - | - | 1 | 30 | 3 | 22 | ns | (Notes 2, 3) |
| 13 | Address Valid Delay | 1 | 60 | 1 | 35 | 1 | 32 | ns | (Notes 2, 3) |
| 14 | Write Data Valid Delay | 0 | 50 | 0 | 30 | 0 | 30 | ns | (Notes 2, 3) |
| 15 | Address/Status/Data Float Delay | 0 | 50 | 0 | 47 | 0 | 32 | ns | (Notes 2, 4, 7) |
| 16 | HLDA Valid Delay | 0 | 50 | 0 | 47 | 0 | 27 | ns | (Notes 2, 3) |
| 19 | Address Valid To Status Valid Setup Time | 38 | - | 27 |  | 22 |  | ns | (Notes 3, 5, 6) |

${ }^{\circ} \mathrm{T}_{\mathrm{A}}$ is guaranteed from $0^{\circ} \mathrm{C}$ to $+55^{\circ} \mathrm{C}$ as long as $\mathrm{T}_{\text {CASE }}$ is not exceeded.

## NOTES:

1. Asynchronous inputs are INTR, NMI, HOLD, PEREQ, ERROR, and $\overline{B U S Y}$. This specification is given only for testing purposes, to assure recognition at a specific CLK edge.
2. Delay from 1.0 V on the CLK, to 0.8 V or 2.0 V or float on the output as appropriate for valid or floating condition.
3. Output load: $C_{L}=100 \mathrm{pF}$.
4. Float condition occurs when output current is less than ILO in magnitude.
5. Delay measured from address either reaching 0.8 V or 2.0 V (valid) to status going active reaching 2.0 V or status going inactive reaching 0.8 V .
6. For load capacitance of 10 pF or more on STATUS/PEACK lines, subtract typically 7 ns for $8 \mathrm{MHz}, 10 \mathrm{MHz}$ and 12.5 MHz spec.
7. These are not tested. They are guaranteed by design characterization.

## A.C. CHARACTERISTICS (Continued)



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NOTE 8:
AC Test Loading on Outputs


210253-39
NOTE 10:
AC Setup, Hold and Delay Time Measurement-General

## A.C. CHARACTERISTICS (Continued)

82C284 Timing Requirements

| Symbol | Parameter | 82C284-8 |  | 82C284-10 |  | 82C284-12 |  | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max |  |  |
| 11 | SRDY/SRDYEN Setup Time | 17 |  | 15 |  | 15 |  | ns |  |
| 12 | SRDY/SRDYEN Hold Time | 0 |  | 2 |  | 2 |  | ns |  |
| 13 | $\overline{\text { ARDY/ARDYEN Setup Time }}$ | 0 |  | 0 |  | 0 |  | ns | (Note 1) |
| 14 | $\overline{\text { ARDY/ARDYEN }}$ Hold Time | 30 | $\square$ | 30 |  | 25 |  | ns | (Note 1) |
| 19 | PCLK Delay | 0 | 45 | 0 | 35 | 0 | 23 | ns | $\begin{aligned} & \mathrm{C}_{\mathrm{L}}=75 \mathrm{pF} \\ & \mathrm{l}_{\mathrm{OL}}=5 \mathrm{~mA} \\ & \mathrm{IOH}_{\mathrm{OH}}=-1 \mathrm{~mA} \end{aligned}$ |

NOTE 1:
These times are given for testing purposes to assure a predetermined action.

82288 Timing Requirements

| Symbol | Parameter |  | 822 | 88-8 | 8228 | 8-10 | 8228 | 88-12 | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Max | Min | Max | Min | Max |  |  |
| 12 | CMDLY Setup Time |  | 20 |  | 15 |  | 15 |  | ns |  |
| 13 | CMDLY Hold Time |  | 1 |  | 1 |  | 1 |  | ns |  |
| 30 | Command Delay from CLK | Command Inactive | 5 | 20 | 5 | 20 | 5 | 20 | ns | $\begin{aligned} & \mathrm{C}_{\mathrm{L}}=300 \mathrm{pF} \text { max } \\ & \mathrm{I}_{\mathrm{OL}}=32 \mathrm{~mA} \text { max } \end{aligned}$ |
| 29 |  | Command Active | 3 | 25 | 3 | 21 | 3 | 21 |  | $\mathrm{l}_{\mathrm{OH}}=-5 \mathrm{~mA}$ max |
| 16 | ALE Active Delay |  | 3 | 20 | 3 | 16 | 3 | 16 | ns |  |
| 17 | ALE Inactive Delay |  |  | 25 |  | 19 |  | 19 | ns |  |
| 19 | DT/可 Read Active Delay |  |  | 25 |  | 23 |  | 23 | ns |  |
| 22 | DT/信Read Inactive Delay |  | 5 | 35 | 5 | 20 | 5 | 18 | ns | $01=16 \mathrm{~mA} \mathrm{~m}$ |
| 20 | DEN Read Active Delay |  | 5 | 35 | 5 | 21 | 5 | 21 | ns | $I_{O H}=-1 \mathrm{~mA} \text { max }$ |
| 21 | DEN Read Inactive Delay |  | 3 | 35 | 3 | 21 | 3 | 19 | ns |  |
| 23 | DEN Write Active Delay |  |  | 30 |  | 23 |  | 23 | ns |  |
| 24 | DEN Write Inactive Delay |  | 3 | 30 | 3 | 19 | 3 | 19 | ns |  |

## WAVEFORMS

MAJOR CYCLE TIMING


NOTE:

1. The modified timing is due to the $\overline{\text { CMDLY }}$ signal being active.

WAVEFORMS (Continued)

## 80286 ASYNCHRONOUS INPUT SIGNAL TIMING



## NOTES:

1. PCLK indicates which processor cycle phase will occur on the next CLK. PCLK may not indicate the correct phase until the first bus cycle is performed.
2. These inputs are asynchronous. The setup and hold times shown assure recognition for testing purposes.

## 80286 RESET INPUT TIMING AND SUBSEQUENT PROCESSOR CYCLE PHASE



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NOTE:
When RESET meets the setup time shown, the next CLK will start or repeat $\phi 2$ of a processor cycle.


## NOTES:

1. These signals may not be driven by the 80286 during the time shown. The worst case in terms of latest float time is shown.
2. The data bus will be driven as shown if the last cycle before $T_{1}$ in the diagram was a write $T_{C}$.
3. The 80286 floats its status pins during $T_{H}$. External $20 \mathrm{~K} \Omega$ resistors keep these signals high (see Table 16).
4. For HOLD request set up to HLDA, refer to Figure 29.
5. BHE and LOCK are driven at this time but will not become valid until $T_{S}$.
6. The data bus will remain in 3 -state OFF if a read cycle is performed.

## WAVEFORMS (Continued)

## 80286 PEREQ/PEACK TIMING FOR ONE TRANSFER ONLY



## NOTES:

1. PEACK always goes active during the first bus operation of a processor extension data operand transfer sequence. The first bus operation will be either a memory read at operand address or I/O read at port address OOFA(H).
2. To prevent a second processor extension data operand transfer, the worst case maximum time (Shown above) is: $3 \times$ (1)-12a $\left.\mathrm{a}_{\text {max. }}-{ }^{-1}\right)_{\text {min. }}$. The actual, configuration dependent, maximum time is: $3 \times(1)-12 \mathrm{a}_{2 \text { max. }}-{ }^{(1)}$ min. $+\mathrm{A} \times 2 \times$ (1). $A$ is the number of extra $T_{C}$ states added to either the first or second bus operation of the processor extension data operand transfer sequence.

## INITIAL 80286 PIN STATE DURING RESET



## NOTES:

1. Setup time for RESET $\uparrow$ may be violated with the consideration that $\phi 1$ of the processor clock may begin one system CLK period later.
2. Setup and hold times for RESET $\downarrow$ must be met for proper operation, but RESET $\downarrow$ may occur during $\phi 1$ or $\phi 2$.
3. The data bus is only guaranteed to be in 3 -state OFF at the time shown.


Figure 35. 80286 Instruction Format Examples

## 80286 INSTRUCTION SET SUMMARY

## Instruction Timing Notes

The instruction clock counts listed below establish the maximum execution rate of the 80286 . With no delays in bus cycles, the actual clock count of an 80286 program will average $5 \%$ more than the calculated clock count, due to instruction sequences which execute faster than they can be fetched from memory.

To calculate elapsed times for instruction sequences, multiply the sum of all instruction clock counts, as listed in the table below, by the processor clock period. An 8 MHz processor clock has a clock period of 125 nanoseconds and requires an 80286 system clock (CLK input) of 16 MHz .

## Instruction Clock Count Assumptions

1. The instruction has been prefetched, decoded, and is ready for execution. Control transfer instruction clock counts include all time required to fetch, decode, and prepare the next instruction for execution.
2. Bus cycles do not require wait states.
3. There are no processor extension data transfer or local bus HOLD requests.
4. No exceptions occur during instruction execution.

## Instruction Set Summary Notes

Addressing displacements selected by the MOD field are not shown. If necessary they appear after the instruction fields shown.

Above/below refers to unsigned value Greater refers to positive signed value Less refers to less positive (more negative) signed values
if $\mathbf{d}=1$ then to register; if $\mathbf{d}=0$ then from register
if $\mathbf{w}=1$ then word instruction; if $\mathbf{w}=0$ then byte instruction
if $s=0$ then 16 -bit immediate data form the operand
if $\mathbf{s}=1$ then an immediate data byte is sign-extended to form the 16 -bit operand
$x$ don't care
$z$ used for string primitives for comparison with ZF FLAG

If two clock counts are given, the smaller refers to a register operand and the larger refers to a memory operand

* = add one clock if offset calculation requires summing 3 elements
$\mathrm{n}=$ number of times repeated
$\mathrm{m}=$ number of bytes of code in next instruction Level (L)-Lexical nesting level of the procedure

The following comments describe possible exceptions, side effects, and allowed usage for instructions in both operating modes of the 80286 .

## REAL ADDRESS MODE ONLY

1. This is a protected mode instruction. Attempted execution in real address mode will result in an undefined opcode exception (6).
2. A segment overrun exception (13) will occur if a word operand reference at offset $\operatorname{FFFF}(\mathrm{H})$ is attempted.
3. This instruction may be executed in real address mode to initialize the CPU for protected mode.
4. The IOPL and NT fieids will remain 0.
5. Processor extension segment overrun interrupt (9) will occur if the operand exceeds the segment limit.

## EITHER MODE

6. An exception may occur, depending on the value of the operand.
7. $\overline{\text { LOCK }}$ is automatically asserted regardless of the presence or absence of the LOCK instruction prefix.
8. LOCK does not remain active between all operand transfers.

## PROTECTED VIRTUAL ADDRESS MODE ONLY

9. A general protection exception (13) will occur if the memory operand cannot be used due to either a segment limit or access rights violation. If a stack segment limit is violated, a stack segment overrun exception (12) occurs.
10. For segment load operations, the CPL; RPL, and DPL must agree with privilege rules to avoid an exception. The segment must be present to avoid a not-present exception (11). If the SS register is the destination, and a segment not-present violation occurs, a stack exception (12) occurs.
11. All segment descriptor accesses in the GDT or LDT made by this instruction will automatically assert LOCK to maintain descriptor integrity in multiprocessor systems.
12. JMP, CALL, INT, RET, IRET instructions referring to another code segment will cause a general protection exception (13) if any privilege rule is violated.
13. A general protection exception (13) occurs if $\mathrm{CPL} \neq 0$.
14. A general protection exception (13) occurs if CPL > IOPL.
15. The IF field of the flag word is not updated if CPL $>$ IOPL. The IOPL field is updated only if $\mathrm{CPL}=0$.
16. Any violation of privilege rules as applied to the selector operand do not cause a protection exception; rather, the instruction does not return a result and the zero flag is cleared.
17. If the starting address of the memory operand violates a segment limit, or an invalid access is attempted, a general protection exception (13) will occur before the ESC instruction is executed. A stack segment overrun exception (12) will occur if the stack limit is violated by the operand's starting address. If a segment limit is violated during an attempted data transfer then a processor extension segment overrun exception (9) occurs.
18. The destination of an INT, JMP, CALL, RET or IRET instruction must be in the defined limit of a code segment or a general protection exception (13) will occur.

## 80286 INSTRUCTION SET SUMMARY

| FUNCTION | FORMAT |  |  |  | CLOCK COUNT |  | COMMENTS |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Real Address Mode |  | Protected Virtual Address Mode | Real <br> Address Mode | Protected <br> Virtual <br> Address <br> Mode |
| DATA TRANSFER MOV = Move: |  |  |  |  |  | 2,3* | 2,3* | 2 | 9 |
| Register to Register/Memory | 1000100w | modreg r/m |  |  |  |  |  |  |
| Register/memory to register | 1000101 w | modreg r/m | $\cdots$ |  | 2,5* | 2,5* | 2 | 9 |  |
| mmediate to register/memory | 1100011 w | $\bmod 000 \mathrm{r} / \mathrm{m}$ | data | data if $w=1$ | 2,3* | 2,3* | 2 | 9 |  |
| frmediate to register | 1011w reg | data | data if $w=1$ |  | 2 | 2 |  |  |  |
| Memory to accumulator | 1010000 w | addr-low | addr-high |  | 5 | 5 | 2 | 9 |  |
| Accumulator to memory | 1010001 w | addr-low | addr-high |  | 3 | 3 | 2 | 9 |  |
| Register/memory to segment register | 10001110 | $\bmod 0 \mathrm{reg} \mathrm{r} / \mathrm{m}$ |  |  | 2,5* | 17,19* | 2 | 9,10,11 |  |
| Segment register to register/memory | 10001100 | $\bmod 0$ reg r/m |  |  | 2,3* | 2,3* | 2 | 9 |  |
| PUSH $=$ Push: |  |  |  |  |  |  |  |  |  |
| Memory | 11111111 | $\bmod 110 \mathrm{r} / \mathrm{m}$ |  |  | 5* | 5* | 2 | 9 |  |
| Register | 01010 reg |  |  |  | 3 | 3 | 2 | 9 |  |
| Segment register | 000 reg 110 |  |  |  | 3 | 3 | 2 | 9 |  |
| Immediate | 01101080 | data | $\text { data its }=0$ |  | $\text { - } 3$ | 3 | $2$ | V. 日 |  |
| Pusha - Push Ah | $01100000$ |  |  |  | $17$ |  | $\mid$ | $\|x\|$ |  |
| POP $=$ POP: |  |  |  |  |  |  |  |  |  |
| Memory | 10001111 | $\bmod 000 \mathrm{r} / \mathrm{m}$ |  |  | 5* | $5^{*}$ | 2 | 9 |  |
| Register | 01011 reg |  |  |  | 5 | 5 | 2 | 9 |  |
| Segment register | 000 reg 111 | $(\mathrm{reg} * 01$ ) |  |  | 5 | 20 | 2 | 9,10,11 |  |
| POPA = POp All | 011000011 |  |  |  | $19$ | $19$ |  |  |  |
| $\mathrm{XCHG}=$ Exhcange: |  |  |  |  |  |  | : |  |  |
| Register/memory with register | 1000011 w | mod reg r/m |  | , | 3,5* | 3,5* | 2,7 | 7,9 |  |
| Register with accumulator | 10010 reg |  |  |  | 3 | 3 |  |  |  |
| $\mathrm{N}=$ Input from: |  |  |  |  |  |  |  |  |  |
| Fixed port | 1110010 w | port |  |  | 5 | 5 |  | 14 |  |
| Variable port | 1110110 w |  |  |  | 5 | 5 |  | 14 |  |
| OUT $=$ Output to: |  |  |  |  |  |  |  | . |  |
| Fixed port | 1110011 w | port |  |  | 3 | 3 | . | 14 |  |
| Variable port | 1110111 w |  |  |  | 3 | 3 |  | 14 |  |
| KLAT $=$ Transiate byte to AL | 11010111 |  |  |  | 5 | 5 | . | 9 |  |
| $E A=$ Load EA to register | 10001101 | modreg $\mathrm{r} / \mathrm{m}$ |  |  | 3* | $3 *$ |  |  |  |
| DS = Load pointer to DS | 11000101 | mod reg r/m | $(\bmod \geqslant 11)$ |  | 7* | $21^{*}$ | 2 | 9,10,11 |  |
| ES = Load pointer to ES | 11000100 | modreg $\quad \mathrm{r} / \mathrm{m}$ | $(\bmod \neq 1)$ |  | $7{ }^{*}$ | $21^{*}$ | 2 | 9,10,11 |  |

Shaded areas indicate instructions not available in 8086, 88 microsystems.

80286 INSTRUCTION SET SUMMARY (Continued)

| FUNCTION | FORMAT |  |  |  | CLOCK COUNT |  | COMMENTS |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | Real Addrese Mode | Protected Virtual Addres: Mode | Real Address Mode | Protected Virtual Address Mode |
| PATA TRANSFER (Continued) |  |  |  |  |  |  |  |  |
| LAHF Load AH with flags | 10011111 |  |  |  | 2 | 2 |  |  |
| SAHF $=$ Store AH into flags | 10011110 |  |  |  | 2 | 2 |  |  |
| PUSHF= Push flags | 10011100 |  |  |  | 3 | 3 | 2 | 9 |
| POPF $=$ Pop flags | 10011101 |  |  |  | 5 | 5 | 2,4 | 9,15 |
| ARITHMETIC ADD = Add: |  | . |  |  |  |  |  |  |
| Reg/memory with register to either | 000000 dw | modreg r/m |  |  | 2,7* | 2,7* | 2 | 9 |
| mmediate to register/memory | 100000sw | $\bmod 000 \mathrm{r} / \mathrm{m}$ | data | data if s w $=01$ | 3,7* | 3,7* | 2 | 9 |
| mmediate to accumulator | 0000010 w | data | data if $w=1$ |  | 3 | 3 |  |  |
| $A D C=$ Add with carry: |  |  |  |  |  |  |  |  |
| Reg/memory with register to either | 000100 dw | modreg r/m |  |  | 2,7* | 2,7* | 2 | 9 |
| Immediate to register/memory | 100000 sw | $\bmod 010 \mathrm{r} / \mathrm{m}$ | data | data if $s w=01$ | 3,7* | 3,7* | 2 | 9 |
| Immediate to accumulator | 0001010 w | data | data if $w=1$ |  | 3 | 3 |  |  |
| NC = increment: |  |  |  |  |  |  |  |  |
| Register/memory | 1111111 w | $\bmod 000 \mathrm{r} / \mathrm{m}$ |  |  | 2,7* | 2,7* | 2 | 9 |
| Register | 01000 reg |  |  |  | 2 | 2 |  |  |
| SUB = Subtract: |  |  |  | . |  |  |  |  |
| Reg/memory and register to either | 001010 dw | modreg r/m |  |  | 2,7* | 2,7* | 2 | 9 |
| Immediate from register/memory | 100000 sw | $\bmod 101 \mathrm{r} / \mathrm{m}$ | data | data if $s w=01$ | $3,7{ }^{\circ}$ | 3,7* | 2 | 9 |
| Immediate from accumulator | 0010110 w | data | data if $w=1$ | . | 3 | 3 |  |  |
| SBB = Subtract with borrow: |  |  |  |  |  |  |  |  |
| Reg/memory and register to either | 000110 dw | modreg r/m |  |  | 2,7* | 2,7* | 2 | 9 |
| mmediate from register/memory | 100000 sw | $\bmod 011 \mathrm{r} / \mathrm{m}$ | data | data if $\mathrm{s} \mathbf{w}=01$ | $3.7 *$ | 3,7* | 2 | 9 |
| mmediate from accumulator | 0001110 w | data | data if $w=1$ |  | 3 | 3 |  |  |
| DEC $=$ Decrement |  |  | - | $\cdots$ |  |  |  |  |
| Pegister/memory | 1111111 w | $\bmod 001 \mathrm{r} / \mathrm{m}$ |  |  | $2.7 *$ | 2,7* | 2 | 9 |
| Pegister | 01001 reg |  |  |  | 2 | 2 |  |  |
| CMP = Compare |  |  |  |  |  |  |  |  |
| Register/memory with register | 0011101 w | modreg r/m |  |  | 2,6* | 2,6* | 2 | 9 |
| Pegister with register/memory | 0011100 w | modreg r/m |  |  | 2,7* | 2,7* | 2 | 9 |
| mmediate with register/memory | 100000 sw | $\bmod 111 \mathrm{r} / \mathrm{m}$ | data | data if $s w=01$ | 3,6* | 3,6* | 2 | 9 |
| mmediate with accumulator | 0011110 w | data | data if $w=1$ |  | 3 | 3 | . |  |
| VEG = Change sign | 1111011 w | $\bmod 011 \mathrm{r} / \mathrm{m}$ | - | * . . . | 2 | $7{ }^{*}$ | 2 | 9 |
| AAA $=$ ASCII adjust for add | 00110111 |  |  | - | 3 | 3 |  | , |
| DAA = Decimal adjust for add | 00100111 |  |  | - | 3 | 3 |  |  |

80286 INSTRUCTION SET SUMMARY (Continued)


Shaded areas indicate instructions not available in 8086,88 microsystems.

80286 INSTRUCTION SET SUMMARY (Continued)

| FUNCTION | FORMAT |  |  |  | CLOCK COUNT |  | COMMENTS |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | Real Address Mode | Protected <br> Virtual <br> Address <br> Mode | Real Address Mode Mod | Protected Virtual Address Mode |
| ARITHMETIC (Continued) AND = And: |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  |
| Reg/memory and register to either | 001000 dw | modreg $\mathrm{f} / \mathrm{m}$ |  |  | 2,7* | 2.7* | 2 | 9 |
| mmediate to register/memory | 1000000w | $\bmod 100 \mathrm{t} / \mathrm{m}$ | data | data if $w=1$ | $3,7 *$ | 3,7* | 2 | 9 |
| mmediate to accumulator | 0010010 w | data | data if $\mathbf{w}=1$ |  | 3 | 3 |  |  |
| TEST = And function to flags, no result: |  |  |  |  |  |  |  |  |
| Register/memory and register | 1000010 w | modreg r/m |  |  | 2,6* | $2,6^{*}$ | 2 | 9 |
| Immediate data and register/memory | 1111011 w | $\bmod 000 \mathrm{r} / \mathrm{m}$ | data | data if $w=1$ | 3,6* | $3.6{ }^{*}$ | 2 | 9 |
| Immediate data and accumulator | 1010100 w | data | data if $w=1$ |  | 3 | 3 |  |  |
| $O R=O r:$ |  |  |  |  |  |  |  |  |
| Reg/memory and register to either | 000010dw | modreg $\mathrm{r} / \mathrm{m}$ |  |  | 2,7* | 2,7* | 2 | 9 |
| mmediate to register/memory | 1000000 w | $\bmod 001 \mathrm{r} / \mathrm{m}$ | data | data if $w=1$ | 3,7* | 3,7* | 2 | 9 |
| mmediate to accumulator | 0000110 w | data | data if $w=1$ |  | 3 | 3 |  |  |
| XOR = Exclusive or: |  |  |  |  |  |  |  |  |
| Reg/memory and register to either | 001100 dw | modreg 1/m |  |  | 2,7* | 2,7* | 2 | 9 |
| Immediate to register/memory | 1000000w | $\bmod 110 \mathrm{r} / \mathrm{m}$ | data | data if $w=1$ | 3,7* | 3,7* | 2 | 9 |
| Immediate to accumulator | 0011010 w | data | data if $w=1$ |  | 3 | 3 |  |  |
| NOT = Invert register/memory | 1111011 w | $\bmod 010 \mathrm{r} / \mathrm{m}$ |  |  | 2,7* | $2,7^{*}$ | 2 | 9 |
| STRING MANIPULATION: |  |  |  |  |  |  |  |  |
| MOVS $=$ Move byte/word | 1010010 w |  |  |  | 5 | 5 | 2 | 9 |
| CMPS = Compare byte/word | 1010011 w |  |  |  | 8 | 8 | 2 | 9 |
| SCAS $=$ Scan byte/word | 1010111 w |  |  |  | 7 | 7 | 2 | 9 |
| LODS = Load byte/wd to AL/AX | 1010110 w |  |  |  | 5 | 5 | 2 | 9 |
| STOS = Stor byte/wd from AL/A | 1010101 w |  |  |  | 3 | 3 | 2 | 9 |
| NS = input byle/wd from OX port Outs = Output bytelwd to DX pot | $\frac{0110110 w}{01101116}$ |  |  |  | $\begin{aligned} & 5 \\ & 5 \\ & \sqrt{5} \end{aligned}$ | $\left\lvert\, \begin{aligned} & 5 \\ & 4 \\ & 5 \end{aligned}\right.$ | $\left\|\begin{array}{ll} 2 \\ k & 2 \\ 2 \end{array}\right\|$ | $0,14 \% 4$ $8.14 \%$ 9,14 |
| Repeated by count in CX |  |  |  |  |  |  |  |  |
| $\mathrm{MOV}_{5}=$ Move string | 11110011 | 1010010 w |  |  | $5+4 n$ | $5+4 n$ | 2 | 9 |
| CMPS $=$ Compare string | 11110012 | 101001 mw |  |  | $5+9 n$ | $5+9 n$ | 2,8 | 8,9 |
| SCAS $=$ Scan string | 11110012 | 1010111 w |  |  | $5+8 n$ | $5+8 n$ | 2,8 | 8,9 |
| ODS $=$ Load string | 11110011 | 1010110 w |  |  | $5+4 n$ | $5+4 n$ | 2,8 | 8,9 |
| STOS $=$ Store string | 11110011 | 1010101 w |  |  | $4+3 n$ | $4+3 n$ | 2,8 | 8,9 |
| His ingut string <br> oufs=output tring | $\frac{1+11100116}{211100115}$ | $\frac{01+0110 \mathrm{w}}{0110111 \mathrm{w}}$ |  |  | $\begin{aligned} & 5+4 n \\ & 5+4 n \end{aligned}$ | $\begin{aligned} & 5+4 n \\ & 5+4 n \end{aligned}$ | 2 <br> 2 |  |

Shaded areas indicate instructions not available in 8086, 88 microsystems.

## 80286 INSTRUCTION SET SUMMARY (Continued)



80286 INSTRUCTION SET SUMMARY (Continued)


Shaded areas indicate instructions not available in 8086, 88 microsystems.

80286 INSTRUCTION SET SUMMARY (Continued)


Shaded areas indicate instructions not available in 8086, 88 microsystems.

## 80286 INSTRUCTION SET SUMMARY (Continued)

| FUNCTION | FORMAT |  |  | CLOCK COUNT |  | COMmENTS |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | Real Address Mode | Protected <br> Virtual <br> Address Mode | Real Address Mode | Protected <br> Virtual <br> Address <br> Mode |
| PROTECTION CONTROL (COMINUEA) <br> LTA-Locel lisk register trom register/menion <br> STR = Store tatk regleter to reglater mention <br> LuSN = Losd maching status word trom registet/mentory <br> 5wsw - Store machirie status wort <br> LAh - Load acceas rights forn register/memory <br> LSL - Load segment timit trom registor/memory <br> ARPL - Aduct requestad privilege leval: from regiater/memory <br> VERR - Verty read accoss: register/menory <br> VERR = Verify write access: | 00001111 <br> 00001111 <br> 00001111 <br> 000011111 <br> 00001111 <br> 00001111 <br> 00001111 <br> 00001111 | 00000000 <br> 00000000 <br> 00000001 <br> 00000001 <br> 00000010 <br> 00000011 <br> 01100011 <br> 00000000 <br> 00000000 | $\bmod 0111 \mathrm{ym}$ <br> modo 01 sm <br> nod 110 gtm <br> mod $100 \mathrm{~J} / \mathrm{m}$ <br> mod reg. $\mathrm{r} / \mathrm{m}$ <br> modreg. I/m <br> modren $1 / \mathrm{m}$ <br> mod 100 $/ \mathrm{m}$ <br> mod 1011 m | $\square$ |  |  |  |

Shaded areas indicate instructions not available in 8086, 88 microsystems.

## Footnotes

The Effective Address (EA) of the memory operand is computed according to the mod and $\mathrm{r} / \mathrm{m}$ fields:
if $\mathrm{mod}=11$ then $\mathrm{r} / \mathrm{m}$ is treated as a REG field
if mod $=00$ then DISP $=0^{*}$, disp-low and disp-high are absent
if mod $=01$ then DISP $=$ disp-low sign-extended to 16 bits, disp-high is absent
if mod $=10$ then DISP $=$ disp-high: disp-low
if $\mathrm{r} / \mathrm{m}=000$ then $E A=(B X)+(S I)+$ DISP
if $\mathrm{r} / \mathrm{m}=001$ then $E A=(B X)+(D I)+D I S P$
if $\mathrm{r} / \mathrm{m}=010$ then $E A=(B P)+(S I)+D I S P$
if $r / m=011$ then $E A=(B P)+(D i)+D I S P$
if $\mathrm{r} / \mathrm{m}=100$ then $\mathrm{EA}=(\mathrm{SI})+$ DISP
if $\mathrm{r} / \mathrm{m}=101$ then $E A=(\mathrm{DI})+$ DISP
if $r / m=110$ then $E A=(B P)+D I S P *$
if $\mathrm{r} / \mathrm{m}=111$ then EA $=(B X)+$ DISP
DISP follows 2nd byte of instruction (before data if required)
*except if $\bmod =00$ and $\mathrm{r} / \mathrm{m}=110$ then $\mathrm{EQ}=$ disp-high: disp-low.

## SEGMENT OVERRIDE PREFIX

001 reg 110
reg is assigned according to the following:

## Segment

reg Register
00 ES
01 CS
10 SS
11 DC

REG is assigned according to the following table:

| 16-Bit $(w=1)$ | $8-\mathrm{Bit}(w=0)$ |
| :---: | :---: |
| 000 AX | 000 AL |
| 001 CX | 001 CL |
| 010 DX | 010 DL |
| 011 BX | 011 BL |
| 100 SP | 100 AH |
| 101 BP | 101 CH |
| 101 SI | 110 DH |
| 111 DI | 111 BH |

The physical addresses of all operands addressed by the BP register are computed using the SS segment register. The physical addresses of the destination operands of the string primitive operations (those addressed by the DI register) are computed using the ES segment, which may not be overridden.

## DATA SHEET REVISION REVIEW

The following list represents key differences between this and the -012 data sheet. Please review this summary carefully.

1. Specifications for the 6 MHz version of the part have been deleted. Intel no longer manufactures an 80286-6.
2. The system diagrams (Figures 31 and 32) have been modified. The circuit which drives the RES input of the 82C284 has been modified in order to allow the 82C284 to correctly generate a system reset signal. See the 82 C 284 data sheet (Order No. 210453) for further information.

80287
80-BIT HMOS NUMERIC PROCESSOR EXTENSION (80287-3, 80287-6, 80287-8, 80287-10)

## - High Performance 80-Bit Internal Architecture <br> - Implements Proposed IEEE Floating Point Standard 754 <br> - Expands 80286 Data types to Include <br> \% 32-, 64-, 80-Bit Floating Point, 32-, 64Bit Integers and 18-Digit BCD Operands * <br> - Object Code Compatible with 8087 <br> - Built-in Exception Handling. <br> - Operates in Both Real and Protected Mode 80286 Systems <br> ■ 8x80-Bit, Individually Addressable, Numeric Register Stack

- Protected Mode Operation Completely Conforms to the $\mathbf{8 0 2 8 6}$ Memory Management and Protection Mechanisms
- Directly Extends 80286 Instruction Set to Trigonometric, Logarithmic, Exponential and Arithmetic Instructions for All Data types
- Operates with 80386 CPU without Software Modification
■ Available in EXPRESS-Standard Temperature Range
- Available in 40 pin-CERDIP package (see Packaging Spec: Order \# \#231369)

The Intel 80287 is a high performance numerics processor extension that extends the 80286 architecture with floating point, extended integer and BCD data types. The 80286/80287 computing system fully conforms to the proposed IEEE Floating Point Standard. Using a numerics oriented architecture, the 80287 adds over fifty mnemonics to the $80286 / 80287$ instruction set, making the $80286 / 80287$ a complete solution for high performance numeric processing. The 80287 is implemented in N -channel, depletion load, silicon gate technology (HMOS) and packaged in a 40-pin cerdip package. The $80286 / 80287$ is object code compatible with the 8086/8087 and 8088/8087.


Figure 1. 80287 Block Diagram


NOTE:
N/C Pins should not be connected
Figure 2.
80287 Pin Configuration

Table 1. 80287 Pin Description

| Symbols | Type | Name and Functon |
| :---: | :---: | :---: |
| CLK | 1 | CLOCK INPUT: this clock provides the basic timing for internal 80287 operations. Special MOS level inputs are required. The 82284 or 8284A CLK outputs are compatible to this input. |
| CKM | $1$ | CLOCK MODE SIGNAL: indicates whether CLK input is to be divided by 3 or used directly. A HIGH input will cause CLK to be used directly. This input must be connected to $\mathrm{V}_{\mathrm{CC}}$ or $\mathrm{V}_{\mathrm{SS}}$ as appropriate. This input must be either HIGH or LOW 20 CLK cycles before RESET goes LOW. |
| RESET : |  | SYSTEM RESET: causes the 80287 to immediately terminate its present activity and enter a dormant state. RESET is required to be HIGH for more than 480287 CLK cycles. For proper initialization the HIGH-LOW transition must occur no sooner than $50 \mu \mathrm{~s}$ after $\mathrm{V}_{\mathrm{CC}}$ and CLK meet their D.C. and A.C. specifications. |
| D15-D0 | $1 / 0$ | DATA: 1 -bit bidirectional data bus. Inputs to these pins may be applied asynchronous to the 80287 clock. |
| BUSY | 0 | BUSY STATUS: asserted by the 80287 to indicate that it is currently executing a command. |
| ERROR | 0 | ERROR STATUS: reflects the ES bit of the status word. This signal indicates that an unmasked error condition exists. |
| PEREQ | $0$ | PROCESSOR EXTENSION DATA CHANNEL OPERAND TRANSFER REQUEST: a HIGH on this output indicates that the 80287 is ready to transfer data. PEREQ will be disabled upon assertion of PEACK or upon actual data transfer, whichever occurs first, if no more transfers are required. |
| PEACK | 1 | PROCESSOR EXTENSION DATA CHANNEL OPERAND TRANSFER ACKNOWLEDGE: acknowledges that the request signal (PEREQ) has been recognized. Will cause the request (PEREQ) to be withdrawn in case there are no more transfers required. PEACK may be asynchronous to the 80287 clock. |
| $\overline{N P R D}$ | 1 | NUMERIC PROCESSOR READ: Enables transfer of data from the 80287. This input may be asynchronous to the 80287 clock. |
| $\overline{N P W R}$ | 1 | NUMERIC PROCESSOR READ: Enables transfer of data from the 80287. This input may be asynchronous to the 80287 clock. |
| $\overline{\mathrm{NPS} 1}$, NPS2 | 1 | NUMERIC PROCESSOR SELECTS: indicate the CPU is performing an ESCAPE instruction. Concurrent assertion of these signals (i.e., NPS1 is LOW and NPS2 is HIGH) enables the 80287 to perform floating point instrucctions. No data transfers involving the 80287 will occur unless the device is selected via these lines. These inputs may be asynchronous to the 80287 clock. |
| CMD1, CMDO | 1 | COMMAND LINES: These, along with select inputs, allow the CPU to direct the operation of the 80287. <br> These inputs may be asynchronous to the 80287 clock. |

Table 1.80187 Pin Description (Continued)

| Symbols | Type $_{1}$ | Name and Function |
| :---: | :---: | :--- |
| $V_{S S}$ | 1 | System ground, both pins must be connected to ground. |
| $V_{C C}$ | 1 | +5 V supply |

## FUNCTIONAL DESCRIPTION

The 80287 Numeric Processor Extension (NPX) provides arithmetic instructions for a variety of numeric data types in 80286/80287 systems. It also executes numerous built-in transcendental functions. (e.g., tangent and log functions). The 80287 executes instructions in parallel with an 80286. It effec-
tively extends the register and instruction set of an 80286 system for existing 80286 data types and adds several new data types as well. Figure 3 presents the program visible register model of the 80286/80287. Essentially, the 80287 can be treated as an additional resource or an extension to the 80286 that can be used as a single unified system, - the 80286/80287.


Figure 3. 80286/80287 Architecture

The 80287 has two operating modes similar to the two modes of the 80286. When reset, 80287 is in the real address mode. It can be placed in the protected virtual address mode by executing the SETPM ESC instruction. The 80287 cannot be switched back to the real address mode except by reset. In the real address mode, the 80286/80287 is completely software compatible with 8086/8087 and 8088/8087.

Once in protected mode, all references to memory for numerics data or status information, obey the 80286 memory management and protection rules giving a fully protected extension of the 80286 CPU. In the protected mode, 80286/80287 numerics software is also completely compatible with 8086/8087 and 8088/8087.

## SYSTEM CONFIGURATION WITH 80286

As a processor extension to an 80286, the 80287 can be connected to the CPU as shown in Figure 4A. The data channel control signals (PEREQ, PEACK), the BUSY signal and the NPRD, NPWR signals, allow the NPX to receive instructions and data from the CPU. When in the protected mode, all information received by the NPX is validated by the 80286 memory management and protection unit. Once started, the 80287 can process in parallel with and independent of the host CPU. When the NPX detects an error or exception, it will indicate this to the CPU by asserting the ERROR signal.

The NPX uses the processor extension request and acknowledge pins of the 80286 CPU to implement data transfers with memory under the protection model of the CPU. The full virtual and physical address space of the 80286 is available. Data for the 80287 in memory is addressed and represented in the same manner as for an 8087.

The 80287 can operate either directly from the CPU clock or with a dedicated clock. For operation with the CPU clock (CKM $=0$ ), the 80287 works at onethird the frequency of the system clock (i.e., for an 8 MHz 80286 , the 16 MHz system clock is divided down to 5.3 MHz ). The 80287 provides a capability to internally divide the CPU clock by three to produce the required internal clock ( $33 \%$ duty cycle). To use a higher performance 80287 ( 8 MHz ), an 8284A clock driver and appropriate crystal may be used to directly drive the 80287 with a $1 / 3$ duty cycle clock on the CLK input (CKM = 1). The following table describes the relationship between the clock speed and the 287 speed version needed as a function of the CKM state.

| 287 Speed <br> Version | CLK Speed |  |
| :---: | :---: | :---: |
|  | CKM $=0$ | CKM $=1$ |
| 5 MHz | 12 MHz | 5 MHz |
| 6 MHz | 16 MHz | 6 MHz |
| 8 MHz | 20 MHz | 8 MHz |
| 10 MHz | 25 MHz | 10 MHz |

## SYSTEM CONFIGURATION WITH 80386

The 80287 can also be connected as a processor extension to the 80386 CPU as shown in Figure 4b. All software written for 8086/8087 and 80286/ 80287 is object code compatible with $80386 / 80287$ and can benefit from the increased speed of the 80386 CPU.

Note that the PEACK input pin is pulled high. This is because the 80287 is not required to keep track of the number of words transferred during an operand transfer when it is connected to the 80386 CPU. Unlike the 80286 CPU, the 80386 CPU knows the exact length of the operand being transferred to/from the 80287. After an ESC instruction has been sent to the 80287, the 80386 processor extension data channel will initiate the data transfer as soon as it receives the PEREQ signal from the 80287. The transfer is automatically terminated by the 80386 CPU as soon as all the words of the operand have been transferred.

Because of the very high speed local local bus of the 80386 CPU, the 80287 cannot reside directly on the CPU local bus. A local bus controller logic is used to generate the necessary read and write cycle timings as well as the chip select timings for the 80287. The 80386 CPU uses I/O addresses 800000F8 through 800000FF to communicate with the 80287. This is beyond the normal I/O address space of the CPU and makes it easier to generate the chip select signals using A31 and M/IO. It may also be noted that the 80386 CPU automatically generates 16 -bit bus cycles whenever it communicates with the 80287.

## HARDWARE INTERFACE

Communication of instructions and data operands between the 80286 and 80287 is handled by the CMD0, CMD1, $\overline{\text { NPS1, NPS2, }}$ NPRD, and NPWR signais. I/O port addresses $00 \mathrm{~F} 8 \mathrm{H}, 00 \mathrm{FAH}$, and 00 FCH are used by the 80286 for this communication. When any of these addresses are used, the NPS1 input must be LOW and NPS2 input HIGH. The $\overline{\text { ORC }}$ and $\overline{\text { IOWC }}$ outputs of the 82288 identify I/O space transfers (see Figure 4A). CMDO should be connected to latched 80286 A1 and CMD1 should be connected to latched 80286 A2.

I/O ports 00 F 8 H to 00FFH are reserved for the 80286/80287 interface. To guarantee correct operation of the 80287, programs must not perform any 1/O operations to these ports.

The PEREQ, $\overline{P E A C K}, \overline{B U S Y}$, and ERROR signals of the 80287 are connected to the same-named 80286 input. The data pins of the 80287 should be directly connected to the 80286 data bus. Note that all bus drivers connected to the 80286 local bus must be inhibited when the 80286 reads from the 80287. The use of $M / \overline{10}$ in the decoder prevents INTA bus cycles from disabling the data transceivers.

## PROGRAMMING INTERFACE

Table 2 lists the seven data types the 80287 supports and presents the format for each type. These
values are stored in memory with the least significant digits at the lowest memory address. Programs retrieve these values by generating the lowest address. All values should start at even addresses for maximum system performance.

Internally the 80287 holds all numbers in the temporary real format. Load instructions automatically convert operands represented in memory as 16 -, 32-, or 64 -bit integers, 32 - or 64 -bit floating point number or

18-digit packed BCD numbers into temporary real format. Store instructions perform the reverse type conversion.

80287 computations use the processor's register stack. These eight 80 -bit registers provide the equivalent capacity of 4016 -bit registers. The 80287 register set can be accessed as a stack, with instructions operating on the top one or two stack elements, or as a fixed register set, with instructions operating on explicitly designated registers.


Figure 4A. 80286/80287 System Configuration


210920-5
Figure 4B. 80386/80287 System Configuration

Table 2. 80287 Data Type Representation in Memory


## NOTES:

1. $S=\operatorname{Sign}$ bit $(0=$ positive, $1=$ negative $)$
2. $d_{n}=$ Decimal digit (two per byte)
3. $X=$ Bits have no significance; 8087 ignores when loading, zeros when storing.
4. $\Delta=$ Position of implicit binary point
5. 1 = Integer bit of significant; stored in temporary real, implicit in short and tong real.
6. Exponent Bias (normalized values):

Short Real: 127 (7FH)
Long Real: 1023 (3FFH)
Temporary Real: 16383 (3FFFH)
7. Packed BCD: $(-1)^{s}\left(D_{17} \ldots D_{0}\right)$
8. Reat: $(-1)^{\mathrm{S}}\left(\mathbf{2 E}^{\mathrm{E}} \mathrm{BIAS}\right)\left(\mathrm{F}_{0} \mathrm{~F}_{1} \ldots\right)$

Table 6 lists the 80287's instructions by class. No special programming tools are necessary to use the 80287 since all new instructions and data types are directly supported by the 80286 assembler and
appropriate high level languages. All 8086/8088 development toois which support the 8087 can also be used to develop software for the $80286 / 80287$ in real address mode.

## SOFTWARE INTERFACE

The $80286 / 80287$ is programmer as a single processor, All communication between the 80286 and the 80287 is transparent to software. The CPU automatically controls the 80287 whenever a numeric instruction is executed. All memory addressing modes, physical memory, and virtual memory of the CPU are available for use by the NPX.

Since the NPX operates in parallel with the CPU, any errors detected by the NPX may be reported after the CPU has executed the ESCAPE instruction which caused it. To allow identification of the failing numeric instruction, the NPX contains two pointer registers which identify the address of the failing numeric instruction and the numeric memory operand if appropriate for the instruction encountering this error.

## INTERRUPT DESCRIPTION

Several interrupts of the 80286 are used to report exceptional conditions while executing numeric programs in either real or protected mode. The interrupts and their functions are shown in Table 3.

## PROCESSOR ARCHITECTURE

As shown in Figure 1, the NPX is internally divided into two processing elements, the bus interface unit (BIU) and the numeric execution unit (NEU). The NEU executes all numeric instructions, while the BIU receives and decodes instructions, requests operand transfers to and from memory and executes processor control instructions. The two units are able to operate independently of one another allowing the BIU to maintain asynchronous communication with the CPU while the NEU is busy processing a numeric instruction.

## BUS INTERFACE UNIT

The BIU decodes the ESC instruction executed by the CPU. If the ESC code defines a math instruction, the BIU transmits the formatted instruction to the NEU. If the ESC code defines an administrative instruction, the BIU executes it independently of the NEU. The parallel operation of the NPX with the CPU is normally transparent to the user. The BIU generates the BUSY and ERROR signals for 80826/ 80287 processor synchronization and error notification, respectively.

The 80287 executes a single numeric instruction at a time. When executing most ESC instructions, the

Table 3. 80286 Interrupt Vectors Reserved for NPX

| Interrupt Number | Interrupt Function |
| :---: | :--- |
| $7:$ | An ESC instruction was encountered when EM or TS of the 80286 MSW was <br> set. EM $=1$ indicates that software emulation of the instruction is required. <br> When TS is set, either an ESC or WAIT instruction will cause interrupt 7. This <br> indicates that the current NPX context may not belong to the current task. |
| 9 | The second or subsequent words of a numeric operand in memory. exceeded a <br> segment's limit. This interrupt occurs after executing an ESC instruction. The <br> saved return address will not point at the numeric instruction causing this <br> interrupt. After processing the addressing error, the 80286 program, can be <br> restarted at the return address with IRET. The address of the failing numeric. <br> instruction and numeric operand and saved in the 80287. An interrupt handler <br> for this interrupt must execute FNiNIT before any other ESC or WAIT. <br> instruction. |
| 13 | The starting address of a numeric operand is not in the segment's limit. The <br> return address will point at the ESC instruction, including prefixes, causing this <br> error. The 80287 has not executed this instruction. The instruction and data <br> address is 80287 refer to a previous, correctly executed, instruction. |
| 16 | The previous numeric instruction caused an unmasked numeric error. The <br> address of the faulty numeric instruction or numeric data operand is stored in <br> the 80287. Only ESC or WAIT instructions can cause this interrupt. The 80286 <br> return address will point at a WAIT or ESC instruction, including prefixes, which <br> may be restarted after clearing the error condition in the NPX. |

80286 tests the $\overline{B U S Y}$ pin and waits until the 80287 indicates that it is not busy before initiating the command. Once initiated, the 80286 continues program execution while the 80287 executes the ESC instruction. In 8086/8087 systems, this synchronization is achieved by placing a WAIT instruction before an ESC instruction. For most ESC instructions, the 80287 does not require a WAIT instruction before the ESC opcode. However, the 80287 will operate correctly with these WAIT instruction. In all cases, a WAIT or ESC instruction should be inserted after any 80287 store to memory (except FSTSW and FSTCW) or load from memory (except FLDENV or FRSTOR) before the 80286 reads or changes the value to be sure the numeric value has already been wrtten or read by the NPX.

Data transfers between memory and the 80287 , when needed, are controlled by the PEREQ PEACK, $\overline{\text { NPRD, }}$ NPWR, $\overline{\text { NPS1, NPS2 signals. The } 80286}$ does the actual data transfer with memory through its processor extension data channel. Numeric data transfers with memory performed by the 80286 use the same timing as any other bus cycle. Control signal for the 80287 are generated by the 80826 as
shown in Figure 4a, and meet the timing requirements shown in the AC requirements section.

## NUMERIC EXECUTION UNIT

The NEU executes all instructions that involve the register stack; these include arithmetic, logical, transcendental, constant and data transfer instructions. The data path in the NEU is 84 bits wide ( 68 significand bits, 15 exponent bits and a sign bit) which allows internal operand transfers to be performed at very high speeds.

When the NEU begins executing an instruction, it activated the BIU BUSY signal. This signal is used in conjunction with the CPU WAIT instruction or automatically with most of the ESC instructions to synchronize both processors.

## REGISTER SET

The 80287 register set is shown in Figure 5. Each of the eight data registers in the 8087's register stack

OVEva


Figure 5. 80287 Register Set
is 80 bits wide and is divided into "fields" corresponding to the NPX's temporary real data type.

At a given point in time the TOP field in the status word identifies the current top-of-stack register. A "push" operation decrements TOP by 1 and loads a value into the new top register. A "pop" operation stores the value from the current top register and then increments TOP by 1 . Like 80286 stacks in memory, the 80287 register stack grows "down" toward lower-addressed registers.

Instructions may address the data registers either implicitly or explicitly. Many instructions operate on the register at the TOP of the stack. These instructions implicitly address the register pointed by the TOP. Other instructions allow the programmer to explicitly specify the register which is to be used. This explicit register addressing is also "top-relative."

## STATUS WORD

The 16-bit status word (in the status register) shown in Figure 6 reflects the overall state of the 80287. It may be read and inspected by CPU code. The busy bit (bit 15) indicates whether the NEU is executing an instruction $(B=1)$ or is idle $(B=0)$.

The instructions FSTSW, FSTSW AX, FSTENV, and FSAVE which store the status word are executed exclusively by the BIU and do not set the busy bit themselves or require the Busy bit be cleared in order to be executed.

The four numeric condition code bits $\left(\mathrm{C}_{0}-\mathrm{C}_{3}\right)$ are similar to the flags in a CPU: instructions that perform arithmetic operations update these bits to reflect the outcome of NPX operations. The effect of these instructions on the condition code is summarized in Tables 4a and 4b.

Bits 14-12 of the status word point to the 80287 register that is the current top-of-stack (TOP) as described above. Figure 6 shows the six error flags in bits 5-0 of the status word. Bits 5-0 are set to indicate that the NEU has detected an exception while executing an instruction. The section on exception handling explains how they are set and used.

Bit 7 is the error summary status bit. This bit is set if any unmasked exception bit is set and cleared otherwise. If this bit is set, the ERROR signal is asserted.


Figure 6. 80287 Status Word

## TAG WORD

The tag word marks the content of each register as shown in Figure 7. The principal function of the tag word is to optimize the NPX's performance. The eight two-bit tags in the tag word can be used, however, to interpret the contents of 80287 registers.

## INSTRUCTION AND DATA POINTERS

The instruction and data pointers (See Figures 8a and 8 b ) are provided for user-written error handlers. Whenever the 80287 executes a new instruction, the BIU saves the instruction address, the operand address (if present) and the instruction opcode. 80287 instructions can store this data into memory.

The instruction and data pointers appear in one of two formats depending on the operating mode of the 80287. In real mode, these values are the 20 -bit physical address and 11-bit opcode formatted like the 8087. In protection mode, these values are the

32-bit virtual address used by the program which executed an ESC instruction. The same FLDENV/ FSTENV/FSAVE/FRSTOR instructions as those of the 8087 are used to transfer these values between the 80287 registers and memory.

The saved instruction address in the 80287 will point at any prefixes which preceded the instruction. This is different than in the 8087 which only pointed at the ESCAPE instruction opcode.

## CONTROL WORD

The NPX provides several processing options which are selected by loading a word from memory into the control word. Figure 9 shows the format and encoding of fields in the control word.

The low order byte of this control word configures the 80287 error and exception masking. Bits 5-0 of the control word contain individual masks for each of the six exceptions that the 80287 recognizes. The high order byte of the control word configures the

Table 4a. Condition Code Interpretation


## NOTES:

1. ST $=$ Top of Stack
2. $X=$ value is not affected by instruction
3. $U=$ value is undefined following instruction
4. $\mathrm{Q}_{\mathrm{n}}=$ Quotient bit n

Table 4b. Condition Code Interpretation after FPREM (See Note 1) Instruction as a Function of Dividend Value

| Dividend Range | $\mathbf{Q}_{2}$ | $\mathbf{Q}_{1}$ | $\mathbf{Q}_{0}$ |
| :---: | :--- | :--- | :--- |
| Dividend $<2$ * Modulus | $\mathrm{C}_{3}$ | $\mathrm{C}_{1}$ | $\mathbf{Q}_{0}$ |
| Dividend $<4^{\circ}$ Modulus | $\mathrm{C}_{3}$ | $\mathrm{Q}_{1}$ | $\mathrm{Q}_{0}$ |
| Dividend $24^{*}$ Modulus | $Q_{2}$ | $Q_{1}$ | $Q_{0}$ |

## NOTE:

1. Previous value of indicated bit, not affected by FPREM instruction execution.

80287 operating mode including precision, rounding, and infinity control. The precision control bits (bits 9-8) can be used to set the 80287 internal operating precision at less than the default of temporary real ( 80 -bit) precision. This can be useful in providing compatibility with the early generation arithmetic processors of smaller precision than the 80287. The rounding control bits (bits $11-10$ ) provide for directed rounding and true chop as well as the unbiased round to nearest even mode specified in the IEEE standard. Control over closure of the number space at infinity is also provided (either affine closure: $\pm$ $\infty$, or projective closure: $\infty$, is treated as unsigned, may be specified).


Figure 7. 80287 Tag Word

| - | 0 MEMORY OFFSET |  |  | $\because$ |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |
|  | CONTROL WORD | +0 | $\because$ |  |
|  | STATUS WORD | +2 |  |  |
|  | TAG WORD | +4 |  |  |
|  | IP OFFSET | +6 |  |  |
|  | CS SELECTOR | +8 |  |  |
|  | DATA OPERAND OFFSET | +10 |  |  |
|  | DATA OPERAND SELECTOR | $+12$ |  |  |

Figure 8a. Protected Mode 80287 Instruction and Data Pointer Image in Memory

## EXCEPTION HANDLING

The 80287 detects six different exception conditions that can occur during instruction execution. Any or all exceptions will cause the assertion of external ERROR signal and ES bit of the Status Word if the appropriate exception masks are not set.

The exceptions that the 80287 detects and the 'default' procedures that will be carried out if the exception is masked, are as follows:

Invalid Operation: Stack overflow, stack underflow, indeterminate form ( $0 / 0, \infty,-\infty$, etc) or the use of a Non-Number (NAN) as an operand. An exponent value of all ones and non-zero significand is reserved to identify NANs. If this exception is masked, the 80287 default response is to generate a specific

NAN called INDEFINITE, or to propogate already existing NANs as the calculation result.

Overflow: The result is too large in magnitude to fit the specified format. The 80287 will generate an encoding for infinity if this exception is masked.

Zero Divisor: The divisor is zero while the dividend is a non-infinite, non-zero number. Again, the 80287 will generate an encoding for infinity if this exception is masked.

Underflow: The result in non-zero but too small in magnitude to fit in the specified format. If this exception is masked the 80287 will denormalize (shift right) the fraction until the exponent is in range. The process is called gradual underflow.


Figure 9. 80287 Control Word

Denormalized Operand: At least one of the operands is denormalized; it has the smallest exponent but a non-zero significand. Normal processing continues if this exception is masked off.

Inexact Result: The true result is not exactly representable in the specified format, the result is rounded according to the rounding mode, and this flag is set. If this exception is masked, processing will simply continue.

If the error is not masked, the corresponding error bit and the error status bit (ES) in the control word will be set, and the ERROR output signal will be asserted. If the CPU attempts to execute another ESC or WAIT instruction, exception 7 will occur.

The error condition must be resolved via an interrupt senvice routine. The 80287 saves the address of the floating point instruction causing the error as well as the address of the lowest memory location of any memory operand required by that instruction.

## 8086/8087 COMPATIBILITY:

The 80286/80287 supports portability of $8086 / 8087$ programs when it is in the real address mode. However, because of differences in the numeric error handling techniques, error handling routines may need to be changed. The differences between an 80286/80287 and 8086/8087 are:

1. The NPX error signal does not pass through an interrupt controller ( 8087 INT signal does).

Therefore, any interrupt controller oriented instructions for the 8086/8087 may have to be deleted.
2. Interrupt vector 16 must point at the numeric error handler routine.
3. The saved floating point instruction address in the 80287 includes any leading prefixes before the ESCAPE opcode. The corresponding saved address of the 8087 does not include leading prefixes.
4. In protected mode, the format of the saved instruction and operand pointers is different than for

- the 8087. The instruction opcode is not saved-it must be read from memory if needed.

5. Interrupt 7 will occur when executing ESC instructions with either TS or EM or $\mathrm{MSW}=1$. If TS of MSW $=1$ then WAIT will also cause interrupt 7 . An interrupt handler should be added to handle this situation.
6. Interrupt 9 will occur if the second or subsequent words of a floating point operand fall outside a segment's size. Interrupt 13 will occur if the starting address of a numeric operand falls outside a segment's size. An interrupt handier should be added to report these programming errors.

In the protected mode, 8086/8087 application code can be directly ported via recompilation if the 80286 memory protection rules are not violated.

## ABSOLUTE MAXIMUM RATINGS*

Ambient Temperature Under Bias ...... $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$
Storage Temperature $\qquad$ $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Case Temperature $\qquad$
Voltage on any Pin with
Respect to Ground . . . . . . . . . . . . . . - 1.0 to +7 V
Power Dissipation... . ....................3.0 Watt
*Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
D.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, T_{C}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, V_{C C}=5 \mathrm{~V} \pm 5 \%$

ALL SPEEDS SELECTIONS

| Symbol | Parameter | Min | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {IL }}$ : O | Input LOW Voltage | -0.5 | 0.8 - | - V | --.... |
| $\mathrm{V}_{\mathrm{IH}}$ | Input HIGH Voltage | 2.0 | $\mathrm{V}_{\mathrm{CC}}+0.5$ | V |  |
| $V_{\mathrm{IHC}}$ | Clock Input HIGH Voltage $\begin{aligned} & C K M=1: \\ & C K M=0: \end{aligned}$ | $\begin{aligned} & 2.0 \\ & 3.8 \end{aligned}$ | $\begin{aligned} & v_{C C}+1 \\ & v_{C C}+1 \end{aligned}$ | $\begin{aligned} & v \\ & v \end{aligned}$ |  |
| $V_{\text {ILC }}$ | Clock Input LOW Voltage $\begin{aligned} & \text { CKM }=1 . \\ & C K M=0 \end{aligned}$ | $\begin{array}{r} -0.5 \\ -0.5 \end{array}$ | $\begin{aligned} & 0.8 \\ & 0.6 \end{aligned}$ | $\begin{aligned} & v \\ & v \end{aligned}$ | $\cdots$ |
| $\mathrm{V}_{\mathrm{OL}}$ | Output LOW Voltage | - | 0.45 | $V$ | $\mathrm{lOL}=3.0 \mathrm{~mA}$ |
| VOH | Output HIGH Voitage | 2.4 |  | V | $\mathrm{l}^{\mathrm{OH}}=-400 \mu \mathrm{~A}$ |
| ILI | Input Leakage Current | " | $\pm 10$ | $\mu \mathrm{A}$ | $\mathrm{OV} \leq \mathrm{V}_{\mathbb{I}} \leq \mathrm{V}_{\mathrm{CC}}$ |
| Lo | Output Leakage Current - - | $\cdots$ | $\pm 10$ | $\mu \mathrm{A}$ | $0.45 \mathrm{~V} \leq \mathrm{V}_{\text {OUT }} \leq \mathrm{V}_{\text {CC }}$ |
|  | Power Supply Current | - | $\begin{aligned} & 600 \\ & 475 \\ & 375 \\ & \hline \end{aligned}$ | mA <br> mA <br> mA | $\begin{aligned} & T_{A}=0^{\circ} \mathrm{C} \\ & T_{A}=25^{\circ} \mathrm{C} \\ & T_{A}=70^{\circ} \mathrm{C} \end{aligned}$ |
| $\mathrm{ClN}_{\mathrm{I}}$ | Input Capacitance | - | 10 | pF | $\mathrm{F}_{\mathrm{C}}=\mathrm{MHz}$ |
| $\mathrm{Co}^{-}$ | Input/Output Capacitance (D0-D15) | - | 20 | pF | $\mathrm{V}_{\mathrm{C}}=1 \mathrm{MHz}$ |
| $\mathrm{C}_{\text {CLK }}$ | CLK Capacitance | $\bullet$ | $12^{-}$ | pF | $\mathrm{F}_{\mathrm{C}}=1 \mathrm{MHz}$ |

A.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{T}_{\text {CASE }}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V} \pm 5 \%$. T . gat

## TIMING REQUIREMENTS

A.C. timings are referenced to 0.8 V and 2.0 V points on signals unless otherwise noted.

| Symbol | Parameter | $\begin{gathered} 80287-3 \\ 5 \mathrm{MHz} \end{gathered}$ |  | $\begin{aligned} & 80287-6 \\ & 6 \mathrm{MHz} \end{aligned}$ |  | $\begin{gathered} 80287-8 \\ 8 \mathrm{MHz} \end{gathered}$ |  | $\begin{gathered} 80287-10 \\ 10 \mathrm{MHz} \\ \text { Preliminary } \\ \hline \end{gathered}$ |  | Units | Test Condifions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max | Min | Max |  |  |
| TCLCL | $\begin{gathered} \text { CLK Period } \\ \text { CKM }=1: \\ \text { CKM }=0: \end{gathered}$ | $\begin{array}{r} 200 \\ 62.5 \\ \hline \end{array}$ | $\begin{array}{\|l} 500 \\ 250 \\ \hline \end{array}$ | $\begin{array}{r} 166 \\ 62.5 \\ \hline \end{array}$ | $\begin{aligned} & 500 \\ & 166 \end{aligned}$ | $\begin{gathered} 125 \\ 50 \end{gathered}$ | $\begin{aligned} & 500 \\ & 166 \end{aligned}$ | $\begin{gathered} 100 \\ 40 \end{gathered}$ | $\begin{array}{r} 500 \\ .166 \\ \hline \end{array}$ | ns ns | $\cdots$ |
| TCLCH | $\begin{gathered} \text { CLK LOW Time } \\ \text { CKM }=1: \\ \text { CKM }=0: \\ \hline \end{gathered}$ | $\begin{array}{r} 118 \\ 15 \\ \hline \end{array}$ | 230 | $\begin{gathered} 100 \\ 15 \\ \hline \end{gathered}$ | $\begin{aligned} & 343 \\ & 146 \end{aligned}$ | $\begin{aligned} & 68 \\ & 15 \end{aligned}$ | $\begin{array}{r} 343 \\ 146 \\ \hline \end{array}$ | $\begin{aligned} & 62 \\ & 11 \end{aligned}$ | $\begin{aligned} & 343 \\ & 146 \end{aligned}$ | $\begin{array}{r} \mathrm{ns} \\ \quad \mathrm{~ns} \\ \hline \end{array}$ | At 0.8 V At 0.6 V |
| TCHCL | $\begin{gathered} \text { CLK HIGH Time } \\ \text { CKM }=1: \\ \text { CKM }=0: \\ \hline \end{gathered}$ | $\begin{aligned} & 69 \\ & 20 \\ & \hline \end{aligned}$ | $235$ | $\begin{array}{r} 50 \\ 20 \\ \hline \end{array}$ | $\begin{array}{\|l\|} 230 \\ 151 \\ \hline \end{array}$ | $\begin{aligned} & 43 \\ & 20 \\ & \hline \end{aligned}$ | $\begin{aligned} & 230 \\ & 151 \\ & \hline \end{aligned}$ | $\begin{aligned} & 28 \\ & 18 \end{aligned}$ | $\begin{aligned} & 230 \\ & 151 \end{aligned}$ | $\begin{aligned} & \text { ns } \\ & \text { ns } \end{aligned}$ | At 2.0 V At 3.6 V |
| $\mathrm{T}_{\mathrm{CH} 1 \mathrm{CH} 2}$ | CLK Rise Time |  | 10 |  | 10 |  | 10 |  | 10 | ns | $\begin{aligned} & 1.0 \mathrm{~V} \text { to } 3.6 \mathrm{~V} \\ & \text { if } \mathrm{CKM}=0 \end{aligned}$ |
| TCL2CL1 | CLK Fall Time |  | 10 |  | 10 |  | 10 | $\cdots$ | 10 | ns. | $\begin{aligned} & 3.6 \mathrm{~V} \text { to } 1.0 \mathrm{~V} \\ & \text { if } \mathrm{CKM}=0 \end{aligned}$ |
| Torwh | Data Setup to NPWR Inactive | 75 |  | 75 |  | 75 | -- | 75 | $\cdots$ | ns | $\ldots$ |
| TWHDX | Data Hold from NPWR Inactive | 30 |  | 30 |  | 18 | $\cdots$ | 18 | $\cdots$ | ns | - $\quad \therefore$ |
| TWLWH TRLRH | $\overline{\text { NPWR }} \overline{\text { NPRD }}$ Active Time | 95 |  | 95 |  | 90 |  | 90 | $\cdots$ | ns | At 0.8 V |
| TAVWL $T_{\text {AVRL }}$ | Command Valid to NPWR or NPRDActive | 0 |  | 0 |  | 0 |  | 0 | . | ns | $\because$ |
| TMHRL | Minimum Delay from PEREQ Active to $\overline{\text { NPRD Active }}$ | 130 |  | 130 |  | $130$ |  | 100 | ... | ns | - |
| TKLKH | PEAK Active Time | 85 |  | 85 |  | 85 |  | 60 |  | ns | At 0.8 V |
| TKHKL | $\overline{\text { PEAK }}$ Inactive Time | 250 |  | 250 |  | 250 |  | 200 |  | ns | At 2.0 V |
| TKHCH | $\overline{\text { PEAK }}$ Inactive to NPWR, NPRD Inactive | 50 |  | 50 | $\cdots$ | 40 |  | 40 |  | ns | $\cdots$ |
| $T_{\text {CHKL }}$ | $\overline{N P W R}, \overline{\text { NPRD }}$ Inactive to PEAK Active | -30 |  | -30 |  | -30 |  | -30 |  | ns |  |
| TwHAX TrHAX | Command Hold from NPWR, NPRD Inactive | 30 |  | 30 |  | 30 |  | 22 |  | ns |  |
| TKLCL | $\overline{\text { PEAK Active }}$ Setup to NPWR NPRD Active | 50 |  | 50 |  | 40 |  | 40 |  | ns |  |

## A.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{T}_{\text {CASE }}=0^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V} \pm 5 \%$ (Continued)

TIMING REQUIREMENTS (Continued)
A.C. timings are referenced to 0.8 V and 2.0 V points on signals unless otherwise noted.

| Symbol | Parameter | $\begin{gathered} 80287-3 \\ 5 \mathrm{MHz} \end{gathered}$ |  | $\begin{gathered} 80287-6 \\ 6 \mathrm{MHz} \end{gathered}$ |  | 80287-8 <br> 8 MHz |  | $\begin{array}{r} 80287-10 \\ -10 \mathrm{MHz} \\ \text { Prelliminary } \end{array}$ |  | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max | Min | Max |  |  |
| TIVCL: | $\overline{N P W R}, \overline{N P R D}$ to CLK Setup Time | 70 | $\therefore$ | 70 | $\cdots$ | 70 | … | 53. |  | ns | (Note 1) |
| TCLIH | $\overline{\text { NPWR, }} \overline{\text { NPRD }}$ from CLK Hold Time | 45. |  | 45 |  | 45 | $\because$ | 37. | . $\quad .$. | ns | (Note 1) |
| TRSCL | RESET to CLK Setup Time | 20 |  | 20 |  | 20 |  | 20 |  | ns | (Note 1) |
| TClRS | RESET from CLK Hold Time | 20 |  | 20 |  | 20 | $\because$ | 20 |  | ns | (Note 1) |

## TIMING RESPONSES

| Symbol | Parameter | $\begin{gathered} 80287-3 \\ 5 \mathrm{MHz} \end{gathered}$ |  | $\begin{gathered} 80287-6 \\ 6 \mathrm{MHz} \end{gathered}$ |  | $\begin{gathered} 80287-8 \\ 8 \mathrm{MHz} \end{gathered}$ |  | $\begin{aligned} & 80287-10 \\ & 10 \mathrm{MHz} \\ & \text { Preliminary } \\ & \hline \end{aligned}$ |  | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max | Min | Max |  |  |
| TrHaz | $\overline{\text { NPRD }}$ Inactive to Data Float |  | 37.5 |  | 37.5 |  | 35 |  | 21 | ns | (Note 2) |
| TrLov | $\overline{\text { NPRD Active to }}$ Data Valid |  | 60 |  | 60 |  | 60 |  | 60 | ns | (Note 3) |
| TILBH | ERROR Active to BUSY Inactive | 100 |  | $100$ |  | 100 |  | 100 |  | ns | (Note 4) |
| TWLBV | $\overline{\text { NPWR }}$ Active to BUSY Active | ... | 100 | . | 100 |  | 100 | - | 100 | ns | (Note 5) |
| TKLML | PEAK Active to PEREQ Inactive |  | 127 |  | 127 |  | 127 |  | 100 | ns | (Note 6) |
| TCMDI | Command Inactive Time <br> Write-to-Write Read-to-Read Write-to-Read Read-to-Write | $\begin{gathered} 95 \\ 250 \\ 105 \\ 95 \\ \hline \end{gathered}$ | - | $\begin{aligned} & 95 \\ & 95 \\ & 95 \\ & 95 \\ & \hline \end{aligned}$ |  | $\begin{aligned} & 95 \\ & 95 \\ & 95 \\ & 95 \\ & \hline \end{aligned}$ | $\cdots$ | $\begin{aligned} & 75 \\ & 75 \\ & 75 \\ & 75 \\ & \hline \end{aligned}$ |  | ns <br> ns <br> ns <br> ns | At 2.0 V At 2.0 V At 2.0 V At 2.0 V |
| $\mathrm{T}_{\text {RHQH }}$ | Data Hold from NPRD Inactive | 5 |  | 3 |  | 3 |  | 3 |  | ns | (Note 7) |

## NOTES:

1. This is an asynchronous input. This specification is given for testing purposes only, to assure recognition at a specific CLK edge.
2. Float condition occurs when output current is less than LLO on DO-D15.
3. $\mathrm{DO} 0 \mathrm{D} 15 \mathrm{loSINF} ¢: \mathrm{XL}=100 \mathrm{pF}$.
4. BUSY loading: $\mathrm{CL}=100 \mathrm{pF}$.
5. $\overline{\text { BUSY loading: }} \mathrm{CL}=100 \mathrm{pF}$.
6. On last data transfer on numeric instruction.
7. DO -D15 loading: $\mathrm{CL}=100 \mathrm{pF}$.

## WAVEFORMS

DATA TRANSFER TIMING (Initiated by 80286)


DATA CHANNEL TIMING (Initiated by 80287)


WAVEFORMS (Continued) an argerangity

ERROR OUTPUT TIMING



## CLK, RESET TIMING (CKM = 1)



NOTE:
Reset, $\overline{\text { NPWR, }} \overline{\text { NPRD }}$ are inputs asynchronous to CLK. Timing requirements on this page are given for testing purposes only, to assure recognition at a specific CLK edge.

## WAVEFORMS (Continued)

CLK, $\overline{\text { NPRD, }}$ NPWR TIMING (CKM $=1$ )


## CLK, RESET TIMING (CKM $=0$ )



NOTE:
Reset must meet timing shown to guarantee known phase of internal $\div 3$ circuit.

CLK, $\overline{\text { NPRD }}$, $\overline{\text { NPWR }}$ TIMING (CKM $=0$ )


Table 6. 80287 Extensions to the 80286 Instruction Set



AC Drive and Measurement Points-CLK Input


AC Setup, Hold and Delay Time Measurement-General


210920-11
AC Test Loading on Outputs

Table 6. 80287 Extensions to the 80286 Instruction Set (Continued)


## NOTE:

1. If $P=1$ then add 5 clocks.

Table 6. 80287 Extensions to the $\mathbf{8 0 2 8 6}$ Instruction Set (Continued)


Table 6.80287 Extensions to the 80286 Instruction Set (Continued)

| Clock Coum Range |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |

## NOTES:

1. if $\bmod =00$ then DISP $=0^{*}$, disp-low and disp-high are absent
if mod $=01$ then DISP $=$ disp-low sign-extended to 16-bits, disp-high is absent
if $\bmod =10$ then DISP $=$ disp-high; disp-low
if $\mathrm{mod}=11$ then $\mathrm{r} / \mathrm{m}$ is treated as an $\mathrm{ST}(\mathrm{i})$ field
2. if $\mathrm{r} / \mathrm{m}=000$ then $E A=(B X)+(S I)+D I S P$
if $\mathrm{r} / \mathrm{m}=001$ then $E A=(B X)+(D I)+D I S P$
if $\mathrm{r} / \mathrm{m}=010$ then $E A=(B P)+(S I)+D I S P$
if $\mathrm{r} / \mathrm{m}=011$ then $E A=(B P)+(D I)+$ DISP
if $\mathrm{r} / \mathrm{m}=100$ then $E A=(\mathrm{SI})+$ DISP
if $\mathrm{r} / \mathrm{m}=101$ then EA $=(\mathrm{DI})+$ DISP
if $\mathrm{r} / \mathrm{m}=110$ then EA $=(B P)+$ DISP
if $\mathrm{r} / \mathrm{m}=111$ then EA $=(\mathrm{BX})+$ DISP
*except if mod $=000$ and $\mathrm{r} / \mathrm{m}=110$ then $E A=$ disp-high; disp-low.
3. MF = Memory Format

00-32-bit Real
01-32-bit Integer
10-64-bit Real
11-16-bit Integer
4. $\mathrm{ST}(0)=$ Current stack top

ST(i) $=$ jth register below stack top
5. $d=$ Destination

0-Destination is ST(0)
1-Destination is $\mathrm{ST}(\mathrm{i})$
6. $P=P O P$

O-No pop
1-Pop ST(0)
7. $R=$ Reverse: When $d=1$ reverse the sense of $R$

0 -Destination (op) Source
1-Source (op) Destination
8. For FSQRT: $\quad-0 \leq \mathrm{ST}(0) \leq+\infty$

For FSCALE: $\quad-215 \leq S T(1)<+215$ and $S T(1)$ integer
For F2XM1: $\quad 0 \leq S T(0) \leq 2^{-1}$
For FYL2X: $\quad 0<\operatorname{ST}(0)<\infty$

$$
-\infty<\operatorname{ST}(1)<+\infty
$$

For FYL2XP1: $\quad 0 \leq|S T(0)|<(2-\sqrt{ } 2) / 2$

$$
-\infty<S T(1)<\infty
$$

For FPTAN: $\quad 0 \leq S T(0) \leq \pi / 4$
For FPATAN: $\quad 0 \leq \operatorname{ST}(0)<\operatorname{ST}(1)<+\infty$
9. ESCAPE bit pattern is 11011.

## DATA SHEET REVISION REVIEW

The following list represents the key differences between this and the -006 80287 Data Sheet. Please review the summary carefully.

1. The CLK speed table in the section entitled "SYSTEM CONFIGURATION WITH 80286" was modified to show the required CLK frequencies in the divide-by-3 mode (CKM = 0) for the 287 speeds tabulated.
2. Obsolete components were replaced with readily available components in Figure 4A.
3. In the AC TIMING REQUIREMENTS table, the timing symbols, $T_{\text {AVRL }}$ and $T_{\text {AVWL }}$ were reversed in order to match the parameter description.
2.78087 and 80287 Compatibility

This section summarizes the differences between the 80387 and the 80287. Any migration from the 8087 directly to the 80387 must also take into account the differences between the 8087. and the 80287 as listed in Appendix A.

Many changes have been designed into the 80387 to directly support the IEEE standard in hardware. These changes result in increased performance by eliminating the need for software that supports the standard.
2.7.1 GENERAL DIFFERENCES

The 80387 supports only affine closure for infinity arithmetic, not projective closure. Bit 12 of the Control Word (CW) no longer defines infinity control. It is a reserved bit; but it is initialized to zero after RESET or FINIT and is changeable upon loading the CW. Programs must ignore this bit.

Operands for FSCALE and FPATAN are no longer restricted in range (except for $\pm \infty$ ); F2XM1 and FPTAN accept a wider range of operands.

The results of transcendental operations may be slightly different from those computed by 80287.

In the case of FPTAN, the 80387 supplies a true tangent result in ST(1), and (always) a floating point 1 in ST.

Rounding control is in effect for FLD constant.
Software cannot change entries of the tag word to values (other than empty) that do not reflect the actual register contents.

After reset, FINIT, and incomplete FPREM, the 80387 resets to zero the condition code bits $\mathrm{C}_{3}-\mathrm{C}_{0}$ of the status word.

In conformance with the IEEE standard, the 80387 does not support the special data formats: pseudozero, pseudo-NaN, pseudoinfinity, and unnormal.

Table 2.7. Exceptions


Note: the $80287 \times 1$ and most 287-compatible chips use the 387 instruction set.

### 2.7.2 EXCEPTIONS

A number of differences exist due to changes in the IEEE standard and to functional improvements to the architecture of the 80387:

1. When the overflow or underflow exception is masked, the 80387 differs from the 80287 in rounding when overflow or underflow occurs. The 80387 produces results that are consistent with the rounding mode.
2. When the underfiow exception is masked, the 80387 sets its undertlow flag only if there is also a loss of accuracy during denormalization.
3. Fewer invalid-operation exceptions due to denormal operands, because the instructions FSQRT, FDIV, FPREM, and conversions to BCD or to integer normalize denormal operands before proceeding.
4. The FSQRT, FBSTP, and FPREM instructions may cause underflow, because they support denormal operands.
5. The denormal exception can occur during the transcendental instructions and the FXTRACT instruction.
6. The denormal exception no longer takes precedence over all other exceptions.
7. When the denormal exception is masked, the 80387 automatically normalizes denormal operands. The 8087/80287 performs unnormal arithmetic, which might produce an unnormal result.
8. When the operand is zero, the FXTRACT instruction reports a zero-divide exception and leaves $-\infty$ in ST(1).
9. The status word has a new bit (SF) that signals when invalid-operation exceptions are due to stack underflow or overflow.
10. FLD extended precision no longer reports denormal exceptions, because the instruction is not numeric.
11. FLD single/double precision when the operand is denormal converts the number to extended precision and signals the denormalized operand exception. When loading a signaling $\mathrm{NaN}, \mathrm{FLD}$ single/double precision signals an invalid-operand exception.
12. The 80387 only generates quiet NaNs (as on the 80287); however, the 80387 distinguishes between quiet NaNs and signaling NaNs . Signaling NaNs trigger exceptions when they are used as operands; quiet NaNs do not (except for FCOM, FIST, and FBSTP which also raise IE for quiet NaNs ).
13. When stack overflow occurs during FPTAN and overflow is masked, both $\mathrm{ST}(0)$ and $\mathrm{ST}(1)$ contain quiet NaNs. The 80287/8087 leaves the original operand in ST (1) intact.
14. When the scaling factor is $\pm \infty$, the FSCALE (ST(0), ST(1)) instruction behaves as follows ( $\mathrm{ST}(0)$ and $\mathrm{ST}(1)$ contain the scaled and scaling operands respectively):

- $\operatorname{FSCALE}(0, \infty)$ generates the invalid operation exception.
- FSCALE(finite, $-\infty$ ) generates zero with the same sign as the scaled operand.
- FSCALE (finite, $+\infty$ ) generates $\infty$ with the same sign as the scaled operand.
The 8087/80287 returns zero in the first case and raises the invalid-operation exception in the other cases.

15. The 80387 returns signed infinity/zero as the unmasked response to massive overflow/underflow. The 8087 and 80287 support a limited range for the scaling factor; within this range either massive overflow/underflow do not occur or undefined results are produced.

### 3.0 HARDWARE INTERFACE

In the following description of hardware interface, the \# symbol at the end of a signal name indicates that the active or asserted state occurs when the signal is at a low voltage. When no \# is present after the signal name, the signal is asserted when at the high voltage level.

### 3.1 Signal Description

In the following signal descriptions, the 80387 pins are grouped by function as follows:

1. Execution control-CPUCLK2, NUMCLK2, CKM, RESETIN
2. NPX handshake-PEREQ, BUSY\#, ERROR\#
3. Bus interface pins-D31-D0, W/R\#, ADS\#, READY \#, READYO\#
4. Chip/Port Select-STEN, NPS1\#, NPS2, CMDO \#
5. Power supplies- $V_{C C}, V_{S S}$

Table 3.1 lists every pin by its identifier, gives a brief description of its function, and lists some of its characteristics. All output signals are tristate; they leave floating state only when STEN is active. The output buffers of the bidirectional data pins D31-D0 are also tristate; they leave floating state only in read cycles when the 80387 is selected (i.e. when STEN, NPS1 \#, and NPS2 are all active).

Figure 3.1 and Table 3.2 together show the location of every pin in the pin grid array.

|  | Instruction |  |  |  |  |  |  |  |  | Optional Fields |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | First Byte |  |  |  | Second Byte |  |  |  |  |  |  |
| 1 | 11011 | OPA |  | 1 | MOD |  | 1 | OPB | R/M | SIB | DISP |
| 2 | 11011 | MF |  | OPA | MOD |  | OPB |  | R/M | SIB | DISP |
| 3 | 11011 | d | P | OPA | 1 | 1 | OPB |  | ST(i) |  |  |
| 4 | 11011 | 0 | 0 | 1 | 1 | 1 | 1 | OP |  |  |  |
| 5 | 11011 | 0 | 1 | 1 | 1 | 1 | 1 | OP |  |  |  |
|  | 15-11 | 10 | 9 | 8 | 7 | 6 | 5 | 43 | 10 |  |  |

### 6.080387 EXTENSIONS TO THE 386TM CPU INSTRUCTION SET

Instructions for the 80387 assume one of the five forms shown in the following table. In all cases, instructions are at least two bytes long and begin with the bit pattern 11011B, which identifies the ESCAPE class of instruction. Instructions that refer to memory operands specify addresses using the 386 CPU addressing modes.
$\mathrm{OP}=$ Instruction opcode, possible split into two fields OPA and OPB

$$
\begin{aligned}
& \text { MF }=\text { Memory Format } \\
& 00-32 \text {-bit real } \\
& 01-32 \text {-bit integer } \\
& 10-64 \text {-bit real } \\
& 11-16 \text {-bit integer }
\end{aligned}
$$

$P=P o p$
0-Do not pop stack
1-Pop stack after operation
$E S C=11011$
$d=$ Destination
0 -Destination is ST(0)
1-Destination is ST(i)
R XOR $d=0$-Destination (op) Source
R XOR $d=1$-Source (op) Destination
$\mathrm{ST}(\mathrm{i})=$ Register stack element $i$
$000=$ Stack top
$001=$ Second stack element
$111=$ Eighth stack element

MOD (Mode field) and R/M (Register/Memory specifier) have the same interpretation as the corresponding fields of the 386 Microprocessor instructions (refer to 386 TM Microprocessor Programmer's Reference Manual).

SIB (Scale Index Base) byte and DISP (displacement) are optionally present in instructions that have MOD and R/M fields. Their presence depends on the values of MOD and R/M, as for 386 Microprocessor instructions.

The instruction summaries that follow assume that the instruction has been prefetched, decoded, and is ready for execution; that bus cycles do not require wait states; that there are no local bus HOLD request delaying processor access to the bus; and that no exceptions are detected during instruction execution. If the instruction has MOD and R/M fields that call for both base and index registers, add one clock.

80387 Extensions to the 386TM CPU Instruction Set


Shaded areas indicate instructions not available in 8087/80287.

## NOTE:

a. When loading single- or double-precision zero from memory, add 5 clocks.

80387 Extensions to the 386TM CPU Instruction Set (Continued)


Shaded areas indicate instructions not available in $8087 / 80287$.

## NOTES:

b. Add 3 clocks to the range when $d=1$.
c. Add 1 clock to each range when $R=1$.
d. Add 3 clocks to the range when $d=0$.
e. typical $=52($ When $d=0,46-54$, typical $=49)$.
f. Add 1 clock to the range when $R=1$.
g. 135-141 when $\mathrm{R}=1$.
h. Add 3 clocks to the range when $d=1$.
i. $-0 \leq S T(0) \leq+\infty$.

80387 Extensions to the 386TM CPU Instruction Set (Continued)

| Instruction | Encoding |  |  | Clock Count Range |
| :---: | :---: | :---: | :---: | :---: |
|  | $\begin{gathered} \text { Byte } \\ 0 \end{gathered}$ | $\begin{gathered} \text { Byte } \\ 1 \end{gathered}$ | Optional <br> Bytes 2-6 |  |
| THANSCENDENTAL |  |  |  |  |
| FCOs ${ }^{\text {k }}$ - Cosine of ST(0) | ESC001 | 11111111 | 48\% | 123-772 |
| FPTANk $=$ Partial tangent of ST(0) | ESC 001 | 11110010 |  | 191-497i |
| FPATAN $=$ Partial arctangent | ESC 001 | 11110011 |  | 314-487 |
| FSINK = Sine of ST(0) | ESC 001 | 11111110 |  | 122-7711 |
| FSINCOS ${ }^{\text {k }}=$ Sine and cosine of ST(0) | ESC001 | 11111011 | $\cdots$ | 194-8091 |
| F2XM1 ${ }^{\text {l }}=2^{S T}(0)-1$ | ESC001 | 11110000 |  | 211-476 |
| FYL2Xm $=$ ST(1) $\cdot \log _{2}(S T(0))$ | ESC 001 | 11110001 |  | 120-538 |
|  | ESC 001 | 11111001 |  | 257-547 |
| PROCESSOR CONTROL |  |  |  |  |
| FINIT $=$ Initialize NPX | ESC 011 | 11100011 |  | 33 |
| FSTSW AX = Store status word | ESC 111 | 11100000 |  | 13 |
| FLDCW = Load control word | ESC 001 | MOD 101 R/M | SIB/DISP | 19 |
| FSTCW = Store control word | ESC 101 | MOD $111 \mathrm{R} / \mathrm{M}$ | SIB/DISP | 15 |
| FSTSW = Store status word | ESC 101 | MOD 111R/M | SIB/DISP | 15 |
| FCLEX = Clear exceptions | ESC011 | 11100010 |  | 11 |
| FSTENV $=$ Store environment | ESC 001 | MOD 110 R/M | SIB/DISP | 103-104 |
| FLDENV = Load environment | ESC001 | MOD $100 \mathrm{R} / \mathrm{M}$ | SIB/DISP | 71 |
| FSAVE = Save state | ESC 101 | MOD $110 \mathrm{R} / \mathrm{M}$ | SIB/DISP | 375-376 |
| FRSTOR $=$ Restore state | ESC 101 | MOD $100 \mathrm{R} / \mathrm{M}$ | SIB/DISP | 308 |
| FINCSTP $=$ Increment stack pointer | ESC 001 | 11110111 |  | 21 |
| FDECSTP $=$ Decrement stack pointer | ESC 001 | 11110110 |  | 22 |
| FFREE $=$ Free ST(i) | ESC 101 | 11000 ST( $)$ |  | 18 |
| FNOP $=$ No operations | ESC 001 | 11010000 |  | 12 |

Shaded areas indicate instructions not available in $8087 / 80287$.

## NOTES:

$j$. These timings hold for operands in the range $|x|<\pi / 4$. For operands not in this range, up to 76 additional clocks may be needed to reduce the operand.
k. $0 \leq|\mathrm{ST}(0)|<2^{63}$.
l. $-1.0 \leq \mathrm{ST}(0) \leq 1.0$.
$\mathrm{m} .0 \leq \mathrm{ST}(0)<\infty,-\infty<\mathrm{ST}(1)<+\infty$.
n. $0 \leq|\mathrm{ST}(0)|<(2-\operatorname{SQRT}(2)) / 2,-\infty<\mathrm{ST}(1)<+\infty$.

# 8237A <br> HIGH PERFORMANCE PROGRAMMABLE DMA CONTROLLER (8237A, 8237A-4, 8237A-5) 

\author{

- Enable/Disable Control of Individual DMA Requests <br> - Four Independent DMA Channels <br> - Independent Autoinitialization of All Channels <br> - Memory-to-Memory Transfers <br> - Memory Block Initiallzation <br> - Address Increment or Decrement <br> - High Performance: Transfors up to 1.6M Bytes/Second with 5 MHz 8237A-5
}
- Directly Expandable to Any Number of Channels
- End of Process Input for Terminating Tranafers
- Software DMA Requeste
- Independent Polarty Control for DREQ and DACK Signals
- Avallable in EXPRESS -Standard Temperature Range
- Avallable in 40-Lead Cerdip and Plastle Packages
(See Packning Spec, Order 2313e0)

The 8237A Muttimode Direct Memory Access (DMA) Controller is a peripheral interface circuit for microprocessor systems. It is designed to improve system performance by allowing external devices to directly transfer information from the system memory. Memory-to-memory transfor capability is also provided. The 8237A offers a wide variety of programmable control features to enhance data throughput and system optimization and to allow dynamic reconfiguration under program control.

The 8237A is designed to be used in conjunction with an external 8 -bit address latch. It contains four independent channels and may be expanded to any number of channels by cascading additional controller chips. The three basic transfer modes allow programmability of the types of DMA service by the user. Each channel can be individually programmed to Autoinitialize to its original condition following an End of Process (EOP). Each channel has a full 64K address and word count capability.

The 8273A-4 and 8237A-5 are 4 MHz and 5 MHz versions of the standard 3 MHz 8237A respectively.


Figure 1. Block Dlegram

Table 1. Pin Deacription

| Symbol | Type | Name and Function |
| :---: | :---: | :---: |
| $V_{\text {cc }}$ |  | POWER: + 5V supply. |
| $V_{\text {ss }}$ |  | OROUND: Ground. |
| CLK | 1 | CLOCK INPUT: Clock Input controls the internal operations of the 8237A and its rate of data transfers. The input may be driven at up to 3 MHz for the standard 8237A and up to 5 MHz for the 8237A-5. |
| CS | 1 | CHIP SELECT: Chip Select is an active low input used to select the 8237A as an I/O device during the Idle cycle. This allows CPU communication on the data bus. |
| RESET | 1 | RESET: Reset is an active high input which clears the Command, Status, Request and Temporary registers. It also clears the first/ last flip/flop and sets the Mask register. Following a Reset the device is in the Idle cycle. |
| READY | 1 | READY: Ready is an input used to extend the memory read and write pulses from the 8237A to accommodate slow memories or //O peripheral devices. Ready must not make transitions during its specified setup/hold time. |
| HLDA | 1 | HOLD ACKNOWLEDGE: The active high Hold Acknowledge from the CPU indicates that it has relinquished control of the system busses. |
| DREC0-DREQ3 | 1 | DIMA REQUEST: The DMA Request lines are individual asynchronous channel request inputs used by peripheral circuits to obtain DMA service. In fixed Priority, DRECO has the highest priority and DREC3 has the lowest priority. A request is generated by activating the DREQ line of a channel. DACK will acknowledge the recognition of DREQ signal. Polarity of DREQ is programmable. Reset initializes these lines to active high. DREQ must be maintained until the corresponding DACK goes active. |
| D80-D87 | 1/0 | DATA BUS: The Data Bus lines are bidirectional three-state signals connected to the system data bus. The outputs are enabled in the Program condition during the I/O Read to output the contents of an Address register, a Status register, the Temporary register or a Word Count register to the CPU. The outputs are disabled and the inputs are read during an I/O Write cycle when the CPU is programming the 8237A control registers. During DMA cycles the most significant 8 bits of the address are output onto the data bus to be strobed into an external latch by ADSTB. In memory-to-memory operations, data from the memory comes into the 8237A on the data bus during the read-frommemory transfer. In the write-to-memory transfer, the data bus outputs place the data into the new memory location. |
| IOR | 1/0 | I/O READ: I/O Read is a bidirectional active low three-state line. In the ldie cycle, it is an input control signal used by the CPU to read the control registers. In the Active cycle, it is an output control signal used by the 8237A to access data from a peripheral during a DMA Write transfer. |
| IOW | 1/0 | I/O WRITE: I/O Write is a bidirectional active low three-state line. In the idle cycle, it is an input control signal used by the CPU to load information into the 8237A. In the Active cycle, it is an output control signal used by the 8237A to load data to the peripheral during a DMA Read transfer. |

## 8237A

Table 1. Pin Deseription (Continued)

| Symbol | Type | Name and Function |
| :---: | :---: | :---: |
| EOP | 1/0 | END OF PROCESs: End of Process is an active low bidirectional signal. Information concerning the complation of DMA services is available at the bidirectional EOP pin. The 8237A allows an external signal to terminate an active DMA service. This is accomplished by pulling the EOP input low with an external EOP signal. The 8237A also generates a pulse when the terminal count (TC) for any channel is reached. This generates an EOP signal which is output through the EOP line. The reception of EOP, either internal or external, will cause the 8237A to terminate the service, reset the request, and, it Autoinitialize is enabled, to write the base registers to the current registers of that channel. The mask bit and TC bit in the status word will be set for the currently active channel by EOP unless the channel is programmed for Autoinitialize. In that case, the mask bit remains unchanged. During memory-to-memory transfers, EOP will be output when the TC for channel 1 occurs. EOP should be tied high with a pull-up resistor if it is not used to prevent erroneous end of process inputs. |
| A0-A3 | 1/0 | ADDRESS: The four least significant address lines are bidirectional three-state signals. In the idte cycle they are inputs and are used by the CPU to address the register to be loaded or read. In the Active cycle they are outputs and provide the lower 4 bits of the output address. |
| A4-A7 | 0 | ADDRESS: The four most significant address lines are three-state outputs and provide 4 bits of address. These lines are enabled only during the DMA service. |
| HRQ | 0 | HOLD REQUEST: This is the Hold Request to the CPU and is used to request control of the system bus. If the corresponding mask bit is clear, the presence of any valid DREQ causes 8237A to issue the HRQ. |
| DACKO-DACK3 | 0 | DMA ACKNOWLEDCE: DMA ACknowledge is used to notify the individual peripherals when one has been granted a DMA cycle. The sense of these lines is programmable. Reset initializes them to active low. |
| AEN | 0 | ADDRESS ENABLE: Address Enable enables the 8 -bit latch containing the upper 8 address bits onto the system address bus. AEN can also be used to disable other system bus drivers during DMA transfers. AEN is active HIGH. |
| ADSTB | 0 | ADDRESS STROBE: The active high, Address Strobe is used to strobe the upper address byte into an external latch. |
| MEMR | 0 | MEMORY READ: The Memory Read signal is an active low threestate output used to access data from the selected memory location during a DMA Read or a memory-to-memory transfer. |
| MEEMW | 0 | MEMORY WRITE: The Memory Write is an active low three-state output used to write data to the selected memory location during a DMA Write or a memory-to-memory transfer. |
| PIN5 | 1 | PIN5: This pin should always be at a logic HIGH level. An internal pull-up resistor will establish a logic high when the pin is left floating. It is recommended however, that PIN5 be connected to V C. |

## FUNCTIONAL DESCRIPTION

The 8237A block diagram includes the major logic blocks and all of the internal registers. The data interconnection paths are also shown. Not shown are the various control signals between the blocks. The 8237A contains 344 bits of internal memory in the form of registers. Figure 3 lists these registers by name and shows the size of each. A detailed description of the registers and their functions can be found under Register Description.

| Name | Ske | Number |
| :---: | :---: | :---: |
| Beae Addrees Reglaters | 18 bres | 4 |
| Bewe Word Count Rogietors | 16 bts | 4 |
| Currem Addreee Regleters | 16 blts | 4 |
| Current Word Coum Reginters | 16 bite | 4 |
| Temporary Addrese Reginter | 16 bla | 1 |
| Temporay Word Count Regiater | 18 bis | 1 |
| Statue Regreter | 8 bits | 1 |
| Command Regiater | 8 bits | 1 |
| Temporary Regiter | 8 bits | 1 |
| Mode Requaters | 6 bis | 4 |
| Mask Reglater | 4 bris | 1 |
| Requeet Regiter | 4 bits | 1 |

Figure 3. 8237A Internal Reglaters
The 8237A contains three basic blocks of control logic. The Timing Control block generates internal timing and external controi signals for the 8237A. The Program Command Control block decodes the various commands given to the 8237A by the microprocessor prior to servicing a DMA Request. It also decodes the Mode Control word used to select the type of DMA during the servicing. The Priority Enzoder block resolves priority contention between MA channels requesting service simultaneously.

The Timing Control block derives internal timing from the clock input. in 8237A systems, this input will usually be the $\$ 2$ TTL clock from an 8224 or CLK from an 8085AH or 8284A. 33\% duty cycle clock generators, however, may not meet the clock high time requirement of the 8237A of the same frequency. For example, 82C84A-5 CLK output violates the clock high time requirement of 8237A-5. In this case 82C84A CLK can simply be inverted to meet 8237A-5 clock high and low time requirements. For 8085AH-2 systems above 3.9 MHz , the 8085 CLK(OUT) does not satisfy 8237A-5 clock LOW and HIGH time requirements. In this case, an external clock should be used to drive the 8237A-5.

## DMA OPERATION

The 8237A is designed to operate in two major cycles. These are called Idle and Active cycles. Each device cycle is made up of a number of states. The 8237A can assume seven separate states, each composed of one fuli clock period. State $1(\mathrm{SI})$ is the inactive state. It is entered when the 8237A has no
valid DMA requests pending. While in SI, the DMA controller is inactive but may be in the Program Condition, being programmed by the processor. State SO (SO) is the first state of a DMA service. The 8237A has requested a hold but the processor has not yet returned an acknowledge. The 8237A may still be programmed until it receives HLDA from the CPU. An acknowledge from the CPU will signal that DMA transfers may begin. S1, S2, S3 and S4 are the working states of the DMA service. If more time is needed to complete a transfer than is available with normal timing, wait states (SW) can be inserted between S2 or S3 and S4 by the use of the Ready line on the 8237A. Note that the data is transferred directly from the I/O device to memory (or vice versa) with ORR and MENW (or MEMR and IOW) being active at the same time. The data is not read into or driven out of the 8237A in 1/O-to-memory or memo-ry-to-I/O DMA transfers.

Memory-to-memory transfers sequire a read-from and a write-to-memory to complete each transfer. The states, which resemble the normal working states, use two digit numbers for identification. Eight states are required for a single transter. The first four states (S11, S12, S13, S14) are used for the read-from-memory half and the last four states (S21, S22, $\mathbf{S 2 3}, \mathbf{S 2 4}$ ) for the write-to-memory half of the transfer.

## IDLE CYCLE

When no channel is requesting service, the 8237A will enter the Idle cycle and perform "Si" states. In this cycle the 8237A will sample the DREQ lines every clock cycle to determine if any channel is requesting a DMA service. The device will also sample CS, looking for an attempt by the microprocessor to write or read the internal registers of the 8237A. When CS is low and HLDA is low, the 8237A enters the Program Condition. The CPU can now establish, change or inspect the internal definition of the part by reading from or writing to the internal registers. Address lines AO-A3 are inputs to the device and select which registers will be read or written. The IOR and IOW lines are used to select and time reads or writes. Due to the number and size of the internal registers, an internal flip-flop is used to generate an additional bit of address. This bit is used to determine the upper or lower byte of the 16 -bit Address and Word Count registers. The flip-flop is reset by Master Clear or Reset. A separate software command can also reset this flip-flop.

Special software commands can be executed by the 8237A in the Program Condition. These commands are decoded as sets of addresses with the CS and IOW. The commands do not make use of the data bus. Instructions include Clear First/Last Flip-Flop and Master Clear.

## aCtive cycle

When the 8237A is in the ldie cycle and a nonmasked channel requests a DMA service, the device will output an HRQ to the microprocessor and enter the Active cycle. It is in this cycle that the DMA service will take place, in one of four modes:

Single Tranafer Mode-In Single Transfer mode the device is programmed to make one transfer only. The word count will be decremented and the address decremented or incremented following each transfer. When the word count "rolls over" from zero to FFFFH, a Terminal Count (TC) will cause an Autoinitialize if the channel has been programmed to do so.

DREQ must be held active until DACK becomes active in order to be recognized. If DREQ is held active throughout the single transfer, HRQ will go inactive and release the bus to the system. It will again go active and, upon receipt of a new HLDA, another single transfer will be performed. In 8080A, 8085AH, 8088, or 8086 system, this will ensure one full machine cycle execution between DMA transfers. Details of timing between the 8237A and other bus control protocols will depend upon the characteristics of the microprocessor involved.

Block Transfer Mode-In Block Transfer mode the device is activated by DREQ to continue making transfers during the service until a TC, caused by word count going to FFFFH, or an external End of

Process (EOP) is encountered. DREQ need only be held active until DACK becomes active. Again, an Autoinitialization will occur at the end of the service if the channel has been programmed for it.

Demand Transfer Mode-In Demand Transfer mode the device is programmed to continue making transfers until a TC or external EOP is encountered or until DREQ goes inactive. Thus transfers may continue until the I/O device has exhausted its data capacity. After the I/O device has had a chance to catch up, the DMA service is re-established by means of a DREQ. During the time between services when the microprocessor is allowed to operate, the intermediate values of address and word count are stored in the 8237A Current Address and Current Word Count registers. Only an EOP can cause an Autoinitialize at the end of the service. EOP is generated either by TC or by an external signal. DREQ has to be low before S4 to prevent another Transfer.

Cascade Mode-This mode is used to cascade more than one 8237A together for simple system expansion. The HRQ and HLDA signals from the additional 8237A are connected to the DREQ and DACK signals of a channel of the initial 8237A. This allows the DMA requests of the additional device to propagate through the priority network circuitry of the preceding device. The priority chain is preserved and the new device must wait for its turn to acknowledge requests. Since the cascade channel of the initial 8237A is used only for prioritizing the additional device, it does not output any address or control


Figure 4. Cascaded 8237As
signals of its own. These could conflict with the outputs of the active channel in the added device. The 8237A will respond to DREQ and DACK but all other outputs except HRQ will be disabled. The ready input is ignored.

Figure 4 shows two additional devices cascaded into an initial device using two of the previous channels. This forms a two level DMA system. More 8237As could be added at the second level by using the remaining channels of the first level. Additional devices can also be added by cascading into the channels of the second level device, forming a third level.

## TRANSFER TYPES

Each of the three active transfer modes can perform three different types of transfers. These are Read, Write and Verify. Write transfers move data from an I/O device to the memory by activating MENW and 1OA. Read transfers move data from memory to an 1/O device by activating MEMR and IOW. Verify transfers are pseudo transfers. The 8237A operates as in Read or Write transfers generating addresses, and responding to EOP, etc. However, the memory and I/O control lines all remain inactive. The ready input is ignored in verity mode.

Mennory-to-Memory-To perform block moves of data from one mernory address space to another with a minimum of program effort and time, the 8237A includes a memory-to-memory transfer feature. Programming a bit in the Command register selects channels 0 and 1 to operate as memory-tomemory transfer channels. The transfor is initiated by setting the software DREQ for channel 0 . The 8237A requests a DMA service in the normal manner. After HLDA is true, the device, using four state transfers in Block Transfer mode, reads data from the memory. The channel 0 Current Address register is the source for the address used and is decremented or incremented in the normal manner. The data byte read from the memory is stored in the 8237A internal Temporary register. Channel 1 then performs a four-state transfer of the data from the Temporary register to memory using the address in its Current Address register and incrementing or decrementing it in the normal manner. The channel 1 current Word Count is decremented. When the word count of channel 1 goes to FFFFH, a TC is generated causing an EOP output terminating the service.

Channel 0 may be programmed to retain the same address for all transfers. This allows a single word to be written to a block of memory.

The 8237A will respond to external EOP signals during memory-to-memory transfers. Data comparators in block search schemes may use this input to terminate the service when a match is found. The timing of memory-to-memory transfers is found in Figure 12. Memory-to-memory operations can be detected as an active AEN with no DACK outputs.

Autoinitiallze-By programming a bit in the Mode register, a channel may be set up as an Autoinitialize channel. During Autoinitialize initialization, the original values of the Current Address and Current Word Count registers are automatically restored from the Base Address and Base Word count registers of that channel following EOP. The base registers are loaded simultaneously with the current registers by the microprocessor and remain unchanged throughout the DMA service. The mask bit is not altered when the channel is in Autoinitialize. Following Autoinitialize the channel is ready to perform another DMA service, without CPU intervention, as soon as a valid DREQ is detected. In order to Autoinitialize both channels in a memory-to-memory transfer, both word counts should be programmed identically. If interrupted externally, EOP pulses should be applied in both bus cycles.

Priority-The 8237A has two types of priority encoding available as software selectable options. The first is Fixed Priority which fixes the channels in priority order based upon the descending value of their number. The channel with the lowest priority is 3 followed by 2, 1 and the highest prionity channel, 0 . After the recognition of any one channel for service, the other channels are prevented from interfering with that service until it is completed.

After completion of a service, HRQ will go inactive and the 8237A will wait for HLDA to go low before activating HRQ to service another channel.

The second scheme is Rotating Priority. The last channel to get service becomes the lowest priority channel with the others rotating accordingly.


With Rotating Priority in a single chip DMA system, any device requesting service is guaranteed to be recognized after no more than three higher priority services have occurred. This prevents any one channel from monopolizing the system.

Compressed Timing-In order to achieve even greater throughput where system characteristics permit, the 8237A can compress the transfer time to two clock cycles. From Figure 11 it can be seen that state S3 is used to extend the access time of the read pulse. By removing state S3, the read pulse width is made equal to the write pulse width and a transfer consists only of state S2 to change the address and state S4 to perform the read/write. S1 states will still occur when A8-A15 need updating (see Address Generation). Timing for compressed transfers is found in Figure 14.

Address Generation-In order to reduce pin count, the 8237A multiplexes the eight higher order address bits on the data lines. State S1 is used to output the higher order address bits to an external latch from which they may be placed on the address bus. The falling edge of Address Strobe (ADSTB) is used to load these bits from the data lines to the latch. Address Enable (AEN) is used to enable the bits onto the address bus through a three-state enable. The lower order address bits are output by the 8237A directly. Lines A0-A7 should be connected to the address bus. Figure 11 shows the time relationships between CLK, AEN, ADSTB, DB0-D87 and AO-A7.

During Block and Demand Transfer mode services, which include multiple transfers, the addresses generated will be sequential. For many transfers the data held in the external address latch will remain the same. This data need only change when a carry or borrow from A7 to A8 takes place in the normal sequence of addresses. To save time and speed transfers, the 8237A executes S1 states only when updating of A8-A15 in the latch is necessary. This means for long services, S1 states and Address Strobes may occur only once every 256 transfers, a savings of 255 ctock cycles for each 256 transfers.

## REGISTER DESCRIPTION

Current Addreas Register-Each channel has a 16-bit Current Address register. This register holds the value of the address used during DMA transfers. The address is automatically incremented or decremented after each transfer and the intermediate values of the address are stored in the Current Address register during the transfer. This register is written or read by the microprocessor in successive 8-bit bytes. It may also be reinitialized by an Autoinitialize back to its original value. Autoinitialize takes place only after an EOP.

Current Word Register-Each channel has a 16bit Current Word Count register. This register determines the number of transfers to be performed. The actual number of transfers will be one more than the number programmed in the Current Word Count register (i.e., programming a count of 100 will result in 101 transfers). The word count is decremented after each transfer. The intermediate value of the word count is stored in the register during the transfer. When the value in the register goes from zero to FFFFH, a TC will be generated. This register is loaded or read in successive 8 -bit bytes by the microprocessor in the Program Condition. Following the end of a DMA service it may also be reinitialized by an Autoinitialization back to its original value. Autoinitialize can occur only when an EOP occurs. If it is not Autoinitialized, this register will have a count of FFFFH after TC.

Base Address and Base Word Count ReglatersEach channel has a pair of Base Address and Base Word Count registers. These 16 -bit registers store the original value of their associated current registers. During Autoinitialize these values are used to restore the current registers to their original values. The base registers are written simultaneously with their corresponding current register in 8 -bit bytes in the Program Condition by the microprocessor. These registers cannot be read by the microprocessor.

Command Reglater-This 8 -bit register controls the operation of the 8237A. It is programmed by the microprocessor in the Program Condition and is cleared by Reset or a Master Clear instruction. The following table lists the function of the command bits. See Figure 6 for address coding.

Mode Reglater-Each channel has a 6-bit Mode register associated with it. When the register is being written to by the microprocessor in the Program Condition, bits 0 and 1 determine which channel Mode register is to be written.

Request Reglater-The 8237A can respond to requests for DMA service which are initiated by software as well as by a DREQ. Each channel has a request bit associated with it in the 4-bit Request register. These are non-maskable and subject to prioritization by the Priority Encoder network. Each register bit is set or reset separately under software control or is cleared upon generation of a TC or external EOP. The entire register is cleared by a Reset. To set or reset a bit, the software loads the proper form of the data word. See Figure 5 for register address coding. In order to make a software request, the channel must be in Block Mode.

Command Reglater


Mode Regleter


Roquest Regitater


Mask Reglster-Each channel has associated with it a mask bit which can be set to disable the incoming DREQ. Each mask bit is set when its associated channel produces an EOP if the channel is not programmed for Autoinitialize. Each bit of the 4-bit Mask register may also be set or cleared separately under software control. The entire register is also set by a Reset. This disables all DMA requests until a clear Mask register instruction allows them to occur. The instruction to separately set or clear the mask bits is simitar in form to that used with the Request register. See Figure 5 for instruction addressing.


All four bits of the Mask register may also be written with a single command.


| Reglater | Operation | Signale |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  | CS | IOF | IOW | A3 A2 A1 | AO |  |
| Command | Write | 0 | 1 | 0 | 1 | 0 | 0 | 0 |
| Mode | Write | 0 | 1 | 0 | 1 | 0 | 1 | 1 |
| Request | Write | 0 | 1 | 0 | 1 | 0 | 0 | 1 |
| Mask | Set/Reset | 0 | 1 | 0 | 1 | 0 | 1 | 0 |
| Mask | Write | 0 | 1 | 0 | 1 | 1 | 1 | 1 |
| Temporary | Read | 0 | 0 | 1 | 1 | 1 | 0 | 1 |
| Status | Read | 0 | 0 | 1 | 1 | 0 | 0 | 0 |

Flgure 5. Definition of Register Codes
Status Regleter-The Status register is available to be read out of the 8237A by the microprocessor. It contains information about the status of the devices at this point. This information includes which channels have reached a terminal count and which chan-

nels have pending DMA requests. Bits 0-3 are set every time a TC is reached by that channel or an external EOP is applied. These bits are cleared upon Reset and on each Status Read. Bits 4-7 are set whenever their corresponding channel is requesting service.

Temporary Regleter-The Temporary register is used to hold data during memory-to-memory transfers. Following the completion of the transfers, the last word moved can be read by the microprocessor in the Program Condition. The Temporary register always contains the last byte transferred in the previous memory-to-memory operation, unless cleared by a Reset.

Software Commands-These are additional special software commands which can be executed in the Program Condition. They do not depend on any specific bit pattern on the data bus. The three software commands are:

Clear First/Last Flip-Flop: This command must be executed prior to writing or reading new address or word count information to the 8237A. This initializes the flip-flop to a known state so that subsequent accesses to register contents by the microprocessor will address upper and lower bytes in the correct sequence.

Master Clear. This software instruction has the same effect as the hardware Reset. The Command, Status, Request, Temporary, and Internal First/Last Flip-Flop registers are cleared and the Mask register is set. The 8237A will enter the Idle cycle.

Clear Mask Register: This command clears the mask bits of all four channels, enabling them to accept DMA requests.

Figure 6 lists the address codes for the software commands.

| Stanale |  |  |  |  |  | Operation |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A3 | A2 | A1 | AO | FOP | H0w |  |
| 1 | 0 | 0 | 0 | 0 | 1 | Read Status Register |
| 1 | 0 | 0 | 0 | 1 | 0 | Write Command Register |
| 1 | 0 | 0 | 1 | 0 | 1 | Illegal |
| 1 | 0 | 0 | 1 | 1 | 0 | Write Request Register |
| 1 | 0 | 1 | 0 | 0 | 1 | Illegal |
| 1 | 0 | 1 | 0 | 1 | 0 | Write Single Mask Register Bit |
| 1 | 0 | 1 | 1 | 0 | 1 | Illegal |
| 1 | 0 | 1 | 1 | 1 | 0 | Write Mode Register |
| 1 | 1 | 0 | 0 | 0 | 1 | Iliegal |
| 1 | 1 | 0 | 0 | 1 | 0 | Clear Byte Pointer Flip/Flop |
| 1 | 1 | 0 | 1 | 0 | 1 | Read Temporary Register |
| 1 | 1 | 0 | 1 | 1 | 0 | Master Clear |
| 1 | 1 | 1 | 0 | 0 | 1 | lliegal |
| 1 | 1 | 1 | 0 | 1 | 0 | Clear Mask Register |
| 1 | 1 | 1 | 1 | 0 | 1 | Illegal |
| 1 | 1 | 1 | 1 | 1 | 0 | Write All Mask Register Bits |

Figure 6. Software Command Codes

| Channel | Reglater | Operation | Signats |  |  |  |  |  |  | Internal Filp-Fiop | Data Bus D80-D87 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | C | IO' | IOW | A3 | 12 | A1 | AO |  |  |
| 0 | Base and Current Address | Write | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | A0-A7 |
|  |  |  | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | A8-A15 |
|  | Current Address | Read | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | A0-A7 |
|  |  |  | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | A8-A15 |
|  | Base and Current Word Count | Write | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 0 | W0-W7 |
|  |  |  | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | W8-W15 |
|  | Current Word Count | Read | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | W0-W7 |
|  |  |  | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 1 | W8-W15 |
| 1 | Base and Current Address | Write | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 0 | A0-A7 |
|  |  |  | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 1 | A8-A15 |
|  | Current Address | Read | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | A0-A7 |
|  |  |  | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 1 | A8-A15 |
|  | Base and Current Word Count | Write | 0 | 1 | 0 | 0 | 0 | 1 | 1 | 0 | W0-W7 |
|  |  |  | 0 | 1 | 0 | 0 | 0 | 1 | 1 | 1 | W8-W15 |
|  | Current Word Count | Read | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 0 | W0-W7 |
|  |  |  | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 1 | W8-W15 |
| 2 | Base and Current Address | Write | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 0 | A0-A7 |
|  |  |  | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 1 | A8-A15 |
|  | Current Address | Read | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 0 | A0-A7 |
|  |  |  | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 1 | A8-A15 |
|  | Base and Current Word Count | Write | 0 | 1 | 0 | 0 | 1 | 0 | 1 | 0 | W0-W7 |
|  |  |  | 0 | 1 | 0 | 0 | 1 | 0 | 1 | 1 | W8-W15 |
|  | Current Word Count | Read | 0 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | W0-W7 |
|  |  |  | 0 | 0 | 1 | 0 | 1 | 0 | 1 | 1 | W8-W15 |
| 3 | Base and Current Address | Write | 0 | 1 | 0 | 0 | 1 | 1 | 0 | 0 | A0-A7 |
|  |  |  | 0 | 1 | 0 | 0 |  |  | 0 | 1 | A8-A15 |
|  | Current Address | Read | 0 | 0 | 1 | 0 | 1 |  | 0 | 0 | A0-A7 |
|  |  |  | 0 | 0 | 1 | 0 |  |  | 0 | 1 | A8-A15 |
|  | Base and Current Word Count | Write | 0 | 1 | 0 | 0 | 1 | 1 | 1 | 0 | W0-W7 |
|  |  |  | 0 | 1 | 0 | 0 | 1 |  | 1 | 1 | W8-W15 |
|  | Current Word Count | Read | 0 | 0 | 1 | 0 | 1 | 1 | 1 | 0 | W0-W7 |
|  |  |  | 0 | 0 | 1 | 0 | 1 | 1 | 1 | 1 | W8-W15 |

Figure 7. Word Count and Address Regleter Commend Codes

## PROCRAMIMING

The 8237A will accept programming from the host processor any time that HLDA is inactive; this is true even if HRQ is active. The responsibility of the host is to assure that programming and HLDA are mutually exclusive. Note that a problem can occur if a DMA request occurs, on an unmasked channel while the 8237A is being programmed. For instance, the CPU may be starting to reprogram the two byte Address register of channel 1 when channel 1 receives a DMA request. If the 8237A is enabled (bit 2 in the command register is 0 ) and channel 1 is unmasked, a DMA service will occur after only one byte of the Address register has been reprogrammed. This can be avoided by disabling the controlier (setting bit 2 in the command register) or masking the channel before programming any other registers. Once the programming is complete, the controller can be enabled/unmasked.

After power-up it is suggested that all internal locations, especially the Mode registers, be loaded with some valid value. This should be done even if some
channels are unused. An invalid mode may force all control signals to go active at the same time.

## APPLICATION INFORMATION (Note 1)

Figure 8 shows a convenient method for configuring a DMA system with the 8237A controller and an 8080A/8085AH microprocessor system. The multimode DMA controller issues a HRQ to the processor whenever there is at least one valid DMA request from a peripheral device. When the processor replies with a HLDA signal, the 8237A takes control of the address bus, the data bus and the control bus. The address for the first transfer operation comes out in two bytes-the least significant 8 bits on the eight address outputs and the most significant 8 bits on the data bus. The contents of the data bus are then latched into an 8 -bit latch to complete the full 16 bits of the address bus. The 8282 is a high speed, 8 -bit, three-state latch in a 20 -pin package. After the initial transfer takes place, the latch is updated only after a carry or borrow is generated in the least significant address byte. Four DMA channels are provided when one 8237A is used.


Figure 8.8237A System Interface

## NOTE:

1. See Application Note AP-67 for $\mathbf{8 0 8 6}$ design information. 8237A

## ABSOLUTE MAXIMUM RATINGS*

Ambient Temperature under Bias $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$
Case Temperature $\qquad$ $0^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$
Storage Temperature $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Voltage on Any Pin with
Respect to Ground. . . . . . . . . . . . . . -0.5 V to +7 V
Power Dissipation 1.5 Watt
"Notice: Stresses above those listed under "AbsoIute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operationa/ sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## D.C. CHARACTERISTICS

$T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$, $\mathrm{T}_{\text {CASE }}=0^{\circ} \mathrm{C}$ to $75^{\circ} \mathrm{C}, \mathrm{V}_{\text {CC }}=+5.0 \mathrm{~V} \pm 5 \%, \mathrm{GND}=0 \mathrm{~V}$

| Symbol | Parameter | Min | Typ (Note 1) | Max | Unit | Teat Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VOH | Output High Voltage | 2.4 |  |  | V | $\mathrm{IOH}^{\prime}=-200 \mu \mathrm{~A}$ |
|  |  | 3.3 |  |  | V | $\mathrm{I}_{\mathrm{OH}}=-100 \mu \mathrm{~A}$ (HRQ Only) |
| V OL | Output LOW Voltage |  |  | 0.40 | $V$ | $\mathrm{IOL}=3.2 \mathrm{~mA}$ |
| $\mathrm{V}_{\mathrm{H}}$ | Input HIGH Voltage | 2.0 |  | $V_{C C}+0.5$ | V |  |
| $V_{\text {IL }}$ | Input LOW Voltage | -0.5 |  | 0.8 | V |  |
| lu | Input Load Current |  |  | $\pm 10$ | $\mu \mathrm{A}$ | $\mathrm{OV} \leq \mathrm{V}_{\mathbb{I}} \leq \mathrm{V}_{\text {CC }}$ |
| Lo | Output Leakage Current |  |  | $\pm 10$ | $\mu A$ | $0.45 \mathrm{~V} \leq \mathrm{V}_{\text {OUT }} \leq \mathrm{V}_{\text {CC }}$ |
| Icc | Vcc Supply Current |  | 110 | 130 | mA | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |
|  |  |  | 130 | 150 | mA | $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ |
| $\mathrm{C}_{0}$ | Output Capacitance |  | 4 | 8 | pF |  |
| $C_{1}$ | Input Capacitance |  | 8 | 15 | pF | fc $=1.0 \mathrm{MHz}$, Inputs $=0 \mathrm{~V}$ |
| $\mathrm{CrO}_{10}$ | I/O Capacitance |  | 10 | 18 | pF |  |

## NOTE:

1. Typical values are for $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, nominal supply voltage and nominal processing parameters.
A.C. CHARACTERISTICS-DMA (MASTER) MODE
$T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$. $T_{\text {CASE }}=0^{\circ} \mathrm{C}$ to $75^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=+5 \mathrm{~V} \pm 5 \%, \mathrm{GND}=0 \mathrm{~V}$

| Symbel | Parameter | 1237A |  | 3237A-4 |  | 0337A-6 |  | Unt |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | min | Max | min | Max | min | max |  |
| TAEL | AEN HIGH from CLK LOW (81) Delay Time |  | 300 |  | 225 |  | 200 | ns |
| TAET | AEN LOW from CLK HIGH (8) Delay Time |  | 200 |  | 150 |  | 130 | ns |
| TAFAB | ADR Active to Float Delay from CLK HIGH |  | 180 |  | 120 |  | 0 | nis |
| TAFC | READ or WFITE Float from CLK HICH |  | 150 |  | 120 |  | 120 | ns |
| TAFDB | DB Active to Float Dellay from CLK HIGH |  | 250 |  | 100 |  | 170 | ns |
| taha | ADR from RIEAD HIGH Hold Time | TCY-100 |  | TCY-100 |  | TCY-100 |  | $n \mathrm{~s}$ |
| TAHS | DB from ADSTB LOW Hold Time | 40 |  | 40 |  | 30 |  | ns |
| TAHW | ADR from WRITE HIGH Hold Time | TCY-50 |  | TCY-50 |  | TCY. 50 |  | n |
| tak | DACK Valld from CLK LOW Delay Time (Note 1) |  | 250 |  | 220 |  | 170 | ns |
|  | EOP HIGH from CLK HIGH Delay Tlme (Note 2) |  | 260 |  | 100 |  | 170 | ne |
|  | EOP LOW from CLK HIGH Delay Time |  | 250 |  | 100 |  | 170 | ns |
| TASM | ADR Stable from CLK HIGH |  | 250 |  | 100 |  | 170 | nis |
| TASS | DB to ADSTB LOW Setup Time | 100 |  | 100 |  | 100 |  | ns |
| TCH | Clock High Time (Tranaltions $\leq 10 \mathrm{ma}$ ) | 120 |  | 100 |  | 80 |  | ns |
| TCL | Clock LOW Time (Tranalitions $\leq 10$ ns) | 150 |  | 110 |  | 68 |  | ne |
| TCY | CLK Cycle Tine | 320 |  | 250 |  | 200 |  | nm |
| TDCL | CLK HIGH to READ or WFITE LOW Dolay (Note 3) |  | 270 |  | 200 |  | 100 | nm |
| TDCTR | FEXD HIGH from CLK HIGH (84) Delay Time (Note 3) |  | 270 |  | 210 |  | 100 | ns |
| TDCTW | WFITE HICH from CLK HIGH (S4) Delay Time (Note 3) |  | 200 |  | 150 |  | 130 | ne |
| $\begin{aligned} & \text { TDO1 } \\ & \text { TDO2 } \end{aligned}$ | HRQ Valld from CLK HICH Dolay Thme (Note 4) |  | 160 |  | 120 |  | 120 | ns |
|  |  |  | 280 |  | 180 |  | 120 | ns |
| TEPS | EOP' LOW from CLK LOW Setup Time | 60 |  | 46 |  | 40 |  | ne |
| TEPW | EOP Pulse Width | 300 |  | 226 |  | 220 |  | $n$ |
| TFAAB | ADR Float to Active Delay from CLK HIGH |  | 250 |  | 100 |  | 170 | ne |
| TFAC | FIEAD or WRITE Activo from CLK HIGH |  | 200 |  | 160 |  | 160 | m |
| TFADB | DB Float to Active Deday from CLK HIGH |  | 300 |  | 225 |  | 200 | m |
| THS | HLDA Valld to CLK HIGH Setup Time | 100 |  | 75 |  | 76 |  | ns |
| TIDH | Input Data from MEENIF HIGH Hold Time | 0 |  | 0 |  | 0 |  | ns |
| TIDS | Input Deta to MENIF HICH Setup Trme | 250 |  | 100 |  | 170 |  | ne |
| TODH | Output Data from MiENWW HIGH Hold Time | 20 |  | 20 |  | 10 |  | ne |
| TODV | Output Data Valld to MEENWW HIGH | 200 |  | 125 |  | 125 |  | $\mathrm{nc}^{8}$ |
| Tas | DREQ to CLK LOW (SI, S4) Setup Time (Note 1) | 0 |  | 0 |  | 0 |  | nt |
| TRH | CLK to READY LOW Hold Time | 20 |  | 20 |  | 20 |  | n |
| TRS | READY to CLK LOW Setup Time | 100 |  | 60 |  | 60 |  | $n$ |
| TSTL | ADSTB HIGH from CLK HIGH Delay Time |  | 200 |  | 150 |  | 130 | ne |
| TSTT | ADSTB LOW from CLK HIGH Delay Time |  | 140 |  | 110 |  | 00 | ns |

A.C. CHARACTERISTICS-PERIPHERAL (SLAVE) MODE
$T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{T}_{\text {CASE }}=0^{\circ} \mathrm{C}$ to $75^{\circ} \mathrm{C}, V_{C C}=+5 \mathrm{~V} \pm 5 \%, G N D=0 \mathrm{~V}$

| 8ymbel | Paramoter | 2237A |  | $8237 \mathrm{~A}-4$ |  | 3237A-6 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | mm | max | Min | max |  |
| TAR | ADR Valid or CS LOW to READ LOW | 50 |  | 50 |  | 50 |  | ns |
| TAW | ADR Valid to WFITE HIGH Setup Time | 200 |  | 150 |  | 130 |  | n8 |
| TCW | CS LOW to WFITE HIGH Setup Time | 200 |  | 150 |  | 130 |  | ns |
| TOW | Data Valid to WFITE HIGH Setup Time | 200 |  | 150 |  | 130 |  | ns |
| TRA | ADR or CS Hold from FEEAD HIGH | 0 |  | 0 |  | 0 |  | n8 |
| TRDE | Data Access from FIEAD LOW (Note 5) |  | 200 |  | 200 |  | 140 | ns |
| TRDF | DB Float Delay from FIEAD HIGH | 20 | 100 | 20 | 100 | 0 | 70 | ns |
| TRSTD | Power Supply HIGH to RESET LOW Setup Time | 500 |  | 500 |  | 500 |  | ns |
| TRSTS | RESET to First HOWR | 2TCY |  | 2 TCY |  | 2TCY |  | ns |
| TRSTW | RESET Pulse Wioth | 300 |  | 300 |  | 300 |  | ns |
| TRW | FEADO Width | 300 |  | 250 |  | 200 |  | ns |
| TWA | ADR from WFiTE HIGH Hold Time | 20 |  | 20 |  | 20 |  | ns |
| TWC | CS HIGH from WRiTE HIGH Hold Time | 20 |  | 20 |  | 20 |  | ns |
| TWD | Deta from WFilTE HIGH Hold Time | 30 |  | 30 |  | 30 |  | ns |
| TWWS | Witte Width | 200 |  | 200 |  | 160 |  | ns |
| TWR | End of Write to End of Read in DMA Tranafor | 0 |  | 0 |  | 0 |  | ns |

## MOTE

1. DREO end DACK cignale may be ective high or ective bw. Thining diegrams eseurne the active hidh mode.
2. EOp io en open colfector output. This perameter aeeumes the prosence of a 2.2 K puliup to $\mathrm{V} C$.
 The me FOiA or MEEMI pute with for normal reed will be $2 T C Y-50$ ne and for comprosced reed will be TCY - 50 ne.
3. TDO in apecined for two dinerent Output HIGH Iovile TDO1 is meenurd at 2.0 V . TDO2 is meavired at 3.3 V . The velve



## A.C. TESTMM IMPUT/OUTPUT WAVEPORM



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## WAVEFORMS

8LAVE MODE WRITE TIMIMG


231408-13
MOTE:

1. Succeestve read and/or witte operations by the external proceesor to program or examine the controlier mued be timed to allow at least $\mathbf{6 0 0} \mathrm{ne}$ for the 8237 A , at loast 500 ns for the $8237 \mathrm{~A}-4$, and at luast 400 ne for the 8237A-6 ass recovery time between active read or witte pulses. The same recovery time is needed between an sctive read or witio pulse followed by a DMA transfer.

Figure 9. Siave Mode Witte
SLAVE MODE READ TIMING


## NOTE:

1. Succeselve read anci/or withe operations by the external processor to program or examine the controller must be timed to allow at least 600 ne for the 8237A, at least 500 ns for the 8237A-4, and at least 400 ns for the 8237A-6 ae recovery time between active read or write pulses. The same recovery time is needed between an active reed or write pulse followed by a DMA transfer.

Figure 10. Slave Mode Read

WAVEFORMS (Continued)
DMA TRANBFER TIMMNG


Figure 11. Dima Tranoter

## WAVEFORMS (Contimad)

## MEMORY-TO-MEMORY TRANSFER TMMNG



Figure 12 Memory-do-memory Tranetor
READY timang


Figure 13. Reedy

## WAVEFORMS (Continued)

## Comprizesed TRANSFER Timmonc



Figure 14. Compreesed Tranefor
REset thume


Figure 15. neent

## DESIGN CONSIDERATIONS

1. Cascading from channel zero. When using multiple 8237s, always start cascading with channel zero. Channel zero of the 8237 will operate incorrectly if one or more of channels 1,2 , or 3 are used in the cascade mode while channel zero is used in a mode other than cascade.
2. Do not troet the DREQ signal as an asynchronows input whille the channel is in the "domand" or "cascade" modes. If DREQ becomes inactive at any time during state S4, an illegal state may occur causing the 8237 to operate improperty.
3. HPQ must remaln active untll HLDA becomes active. If HRQ goes inactive before HLDA is received the 8237 can enter an illegal state causing it to operate improperly.
4. Meke sure the MEMAF / Ine has 50 pF loading capactance on th. When doing memory to memory transfers, the 8237 requires at least 50 pF loading capacitance on the MEMR * signal for proper operation. In most cases board capacitance is sufficient.
5. Treat the READY input as a synchronous Input. If a transition occurs during the setup/hold window, erratic operation may result.

## DATA SHEET REVISION REVIEW

The following list represents key differences between this and the -002 data sheet. Please review this summary carefully.

1. Major cleanup on the "NOTE" sections of this data sheet.
a. Pin 5 no longer references a note. It is now included in the pin description area under the name "PIN5".
b. The note placed in the "typical" section of the D.C. Characteristics table is now referenced to a note section included with that table.
c. Notes in the A.C. Characteristics table have been renumbered and are included in a notes section for the A.C. Characteristics.
d. The note that was previously referenced in the A.C. TESTING INPUT/OUTPUT WAVEFORM diagram has been replaced with the actual note.
e. The note that was previously referenced in the SLAVE MODE WRITE TIMING diagram has been included in a "NOTE" section with the diagram.
f. The note that was previously referenced in the SLAVE MODE READ TIMING diagram has been included in a "NOTE" section with the diagram.
g. The note that was previously referenced in the DMA TRANSFER TIMING diagram has been included in a "NOTE" section with the diagram.
2. A "Design Considerations" section was added to alert designers to certain design aspects of the 8237.
3. The timing parameters TAR for the 8237A-4 and 8237A-5 have been changed from 50 ns to 0 ns .

## 8254 <br> PROGRAMMABLE INTERVAL TIMER

- Compatible with All Intel and Most Other Microprocessors
- Handles Inputs from DC to 10 MHz
- 5 MHz 8254-5
- 8 MHz 8254
- 10 MHz 8254-2
- Status Read-Back Command
- Six Programmable Counter Modes
- Three Independent 16-Bit Counters
- Binary or BCD Counting
- Single +5 V Supply
- Available in EXPRESS
- Standard Temperature Range

The Intel® 8254 is a counter/timer device designed to solve the common timing control problems in microcomputer system design. It provides three independent 16-bit counters, each capable of handiing cicck inputs up to 10 MHz . All modes are software programmable. The 8254 is a superset of the 8253 .

The 8254 uses HMOS technology and comes in a 24 -pin plastic or CERDIP package.



231164-2
Figure 2. Pin Configuration

Figure 1. 8254 Block Diagram

Table 1. Pin Description


## FUNCTIONAL DESCRIPTION

## General

The 8254 is a programmable interval timer/counter designed for use with Intel microcomputer systems. It is a general purpose, multi-timing element that can be treated as an array of $1 / O$ ports in the system software.

The 8254 solves one of the most common problems in any microcomputer system, the generation of accurate time delays under software control. Instead of setting up timing loops in software, the programmer configures the 8254 to match his requirements and programs one of the counters for the desired delay. After the desired delay, the 8254 will interrupt the CPU. Software overhead is minimal and variable length delays can easily be accommodated.

Some of the other counter/timer functions common to microcomputers which can be implemented with the 8254 are:

- Real time clock
- Event-counter
- Digital one-shot
- Programmable rate generator
- Square wave generator
- Binary rate multiplier
- Complex waveform generator
- Complex motor controller


## Block Diagram

## DATA BUS BUFFER

This 3-state, bi-directional, 8-bit buffer is used to interface the 8254 to the system bus (see Figure 3).


231164-3
Figure 3. Block Diagram Showing Data Bus Buffer and Read/Write Logic Functions

## READ/WRITE LOGIC

The Read/Write Logic accepts inputs from the system bus and generates control signals for the other functional blocks of the 8254. $A_{1}$ and $A_{0}$ select one of the three counters or the Control Word Register to be read from/written into. A "low" on the $\overline{R D}$ input tells the 8254 that the CPU is reading one of the counters. A "low" on the WR input tells the 8254 that the CPU is writing either a Control Word or an initial count. Both $\overline{\mathrm{RD}}$ and $\overline{\mathrm{WR}}$ are qualified by $\overline{\mathrm{CS}}$; $\overline{\mathrm{R}} \overline{\mathrm{D}}$ and $\overline{\mathrm{WR}}$ are ignored unless the 8254 has been selected by holding $\overline{\mathrm{CS}}$ low.

## CONTROL WORD REGISTER

The Control Word Register (see Figure 4) is selected by the Read/Write Logic when $A_{1}, A_{0}=11$. If the CPU then does a write operation to the 8254 , the data is stored in the Control Word Register and is interpreted as a Control Word used to define the operation of the Counters.

The Control Word Register can only be written to; status information is available with the Read-Back Command.

## COUNTER 0, COUNTER 1, COUNTER 2

These three functional blocks are identical in operation, so only a single Counter will be described. The internal block diagram of a single counter is shown in Figure 5.

The Counters are fully independent. Each Counter may operate in a different Mode.

The Control Word Register is shown in the figure; it is not part of the Counter itself, but its contents determine how the Counter operates.

The status register, shown in Figure 5, when latched, contains the current contents of the Control Word Register and status of the output and null count flag. (See detailed explanation of the ReadBack command.)

The actual counter is labelled CE (for "Counting Element"). It is a 16 -bit presettable synchronous down counter.
$\mathrm{OL}_{M}$ and $O L_{L}$ are two 8 -bit latches. OL stands for "Output Latch"; the subscripts M and L stand for "Most significant byte" and "Least significant byte"


Figure 4. Block Diagram Showing Control Word Register and Counter Functions


Figure 5. Internal Block Diagram of a Counter
respectively. Both are normally referred to as one unit and called just OL. These latches normally "follow" the CE, but if a suitable Counter Latch Command is sent to the 8254, the latches "latch" the present count until read by the CPU and then return to "following" the CE. One latch at a time is enabled by the counter's Control Logic to drive the internal bus. This is how the 16 -bit Counter communicates over the 8 -bit internal bus. Note that the CE itself cannot be read; whenever you read the count, it is the OL that is being read.

Similarly, there are two 8 -bit registers called $\mathrm{CR}_{M}$ and $\mathrm{CR}_{\mathrm{L}}$ (for "Count Register"). Both are normally referred to as one unit and called just CR. When a new count is written to the Counter, the count is stored in the CR and later transferred to the CE. The Control Logic allows one register at a time to be loaded from the internal bus. Both bytes are transferred to the $C E$ simultaneously. $C R_{M}$ and $C R_{L}$ are cleared when the Counter is programmed. In this way, if the Counter has been programmed for one byte counts (either most significant byte only or least significant byte only) the other byte will be zero. Note that the CE cannot be written into; whenever a count is written, it is written into the CR.

The Control Logic is also shown in the diagram. CLK n, GATE $n$, and OUT $n$ are all connected to the outside world through the Control Logic.

## 8254 SYSTEM INTERFACE

The 8254 is a component of the Intel Microcomputer Systems and interfaces in the same manner as all
other peripherals of the family. It is treated by the system's software as an array of peripheral 1/O ports; three are counters and the fourth is a control register for MODE programming.

Basically, the select inputs $A_{0}, A_{1}$ connect to the $A_{0}$, $A_{1}$ address bus signals of the CPU. The CS can be derived directly from the address bus using a linear select method. Or it can be connected to the output of a decoder, such as an Intel 8205 for larger systems.

## OPERATIONAL DESCRIPTION

## General

After power-up, the state of the 8254 is undefined. The Mode, count value, and output of all Counters are undefined.

How each Counter operates is determined when it is programmed. Each Counter must be programmed before it can be used. Unused counters need not be programmed.

## Programming the $\mathbf{8 2 5 4}$

Counters are programmed by writing a Control Word and then an initial count.

The Control Words are written into the Control Word Register, which is selected when $A_{1}, A_{0}=11$. The Control Word itself specifies which Counter is being programmed.


Figure 6. 8254 System Interface

## Control Word Format

$A_{1}, A_{0}=11 \quad \overline{C S}=0 \quad \overline{R D}=1 \quad \overline{W R}=0$


| $\begin{array}{r} \mathrm{SC}-\mathrm{S} \\ \mathrm{SC} \end{array}$ | SCO |  |
| :---: | :---: | :---: |
| 0 | 0 | Select Counter 0 |
| 0 | 1 | Select Counter 1 |
| 1 | 0 | Select Counter 2 |
| 1 | 1 | Read-Back Command (see Read Operations) |


| M-Mode <br> M2 |
| :--- |
| 0 0 0 M1 Mode 0 |
| 0 |

## RW—Read/Write

RW1 RW0

| 0 | 0 | Counter Latch Command (see Read <br> Operations) |
| :---: | :---: | :--- |
| 0 | 1 | Read/Write least significant byte only |
| 1 | 0 | Read/Write most significant byte only |
| 1 | 1 | Read/Write least significant byte first, <br> then most significant byte |

BCD

| 0 | Binary Counter 16-bits |
| :--- | :--- |
| 1 | Binary Coded Decimal (BCD) Counter <br> (4 Decades) |

NOTE:
Don't care bits $(\mathrm{X})$ should be 0 to insure compatibility with future Intel products.
Figure 7. Control Word Format

By contrast, initial counts are written into the Counters, not the Control Word Register. The $A_{1}, A_{0}$ inputs are used to select the Counter to be written into. The format of the initial count is determined by the Control Word used.

## Write Operations

The programming procedure for the 8254 is very flexible. Only two conventions need to be remembered:

1) For each Counter, the Control Word must be written before the initial count is written.
2) The initial count must follow the count format specified in the Control Word (least significant byte only, most significant byte only, or least significant byte and then most significant byte).

Since the Control Word Register and the three Counters have separate addresses (selected by the $A_{1}, A_{0}$ inputs), and each Control Word specifies the Counter it applies to (SC0,SC1 bits), no special instruction sequence is required. Any programming sequence that follows the conventions in Figure 7 is acceptable.

A new initial count may be written to a Counter at any time without affecting the Counter's programmed Mode in any way. Counting will be affected as described in the Mode definitions. The new count must follow the programmed count format.

If a Counter is programmed to read/write two-byte counts, the following precaution applies: A program must not transfer control between writing the first and second byte to another routine which also writes into that same Counter. Otherwise, the Counter will be loaded with an incorrect count.

|  | $\mathrm{A}_{1}$ | $\mathrm{A}_{0}$ |  | $\mathrm{A}_{1}$ | $A_{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Control Word-Counter 0 | 1 | 1 | Control Word-Counter 2 | 1 | 1 |
| LSB of count-Counter 0 | 0 | 0 | Control Word-Counter 1 | 1 | 1 |
| MSB of count-Counter 0 | 0 | 0 | Control Word-Counter 0 | 1 | 1 |
| Control Word-Counter 1 | 1 | 1 | LSB of count-Counter 2 | 1 | 0 |
| LSB of count-Counter 1 | 0 | 1 | MSB of count-Counter 2 | 1 | 0 |
| MSB of count-Counter 1 | 0 | 1 | LSB of count-Counter 1 | 0 | 1 |
| Control Word-Counter 2 | 1 | 1 | MSB of count-Counter 1 | 0 | 1 |
| LSB of count-Counter 2 | 1 | 0 | LSB of count-Counter 0 | 0 | 0 |
| MSB of count-Counter 2 | 1 | 0 | MSB of count-Counter 0 | 0 | 0 |
|  | $\mathrm{A}_{1}$ | $\mathrm{A}_{0}$ |  | $\mathrm{A}_{1}$ | $\mathrm{A}_{0}$ |
| Control Word-Counter 0 | 1 | 1 | Control Word-Counter 1 | 1 | 1 |
| Control Word-Counter 1 | 1 | 1 | Control Word-Counter 0 | 1 | 1 |
| Control Word-Counter 2 | 1 | 1 | LSB of count-Counter 1 | 0 | 1 |
| LSB of count-Counter 2 | 1 | 0 | Control Word-Counter 2 | 1 | 1 |
| LSB of count-Counter 1 | 0 | 1 | LSB of count-Counter 0 | 0 | 0 |
| LSB of count-Counter 0 | 0 | 0 | MSB of count-Counter 1 | 0 | 1 |
| MSB of count-Counter 0 | 0 | 0 | LSB of count-Counter 2 | 1 | 0 |
| MSB of count-Counter 1 | 0 | 1 | MSB of count-Counter 0 | 0 | 0 |
| MSB of count-Counter 2 | 1 | 0 | MSB of count-Counter 2 | 1 | 0 |

NOTE:
In all four examples, all Counters are programmed to read/write two-byte counts. These are only four of many possible programming sequences.

Figure 8. A Few Possible Programming Sequences

## Read Operations

It is often desirable to read the value of a Counter without disturbing the count in progress. This is easily done in the 8254.

There are three possible methods for reading the counters: a simple read operation, the Counter Latch Command, and the Read-Back Command. Each is explained below. The first method is to perform a simple read operation. To read the Counter, which is selected with the A1, AO inputs, the CLK input of the selected Counter must be inhibited by using either the GATE input or external logic. Otherwise, the count may be in the process of changing when it is read, giving an undefined result.

## COUNTER LATCH COMMAND

The second method uses the "Counter Latch Command". Like a Control Word, this command is written to the Control Word Register, which is selected when $A_{1}, A_{0}=11$. Also like a Control Word, the SC0, SC1 bits select one of the three Counters, but two other bits, D5 and D4, distinguish this command from a Control Word.
$A_{1}, A_{0}=11 ; C S=0 ; R D=1 ; W R=0$

| $\mathrm{D}_{7}$ | $\mathrm{D}_{6}$ | $\mathrm{D}_{5}$ | $\mathrm{D}_{4}$ | $\mathrm{D}_{3}$ | $\mathrm{D}_{2}$ | $\mathrm{D}_{1}$ | $\mathrm{D}_{\mathbf{0}}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SC 1 | SCO | O | O | X | X | X | X |

SC1,SC0-specify counter to be latched

| SC1 | SC0 | Counter |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 2 |
| 1 | 1 | Read-Back Command |

D5,D4-00 designates Counter Latch Command
X-don't care
NOTE:
Don't care bits ( X ) should be 0 to insure compatibility with future intel products.

Figure 9. Counter Latching Command Format

The selected Counter's output latch (OL) latches the count at the time the Counter Latch Command is received. This count is held in the latch untilit is read by the CPU (or until the Counter is reprogrammed). The count is then unlatched automatically and the OL returns to "following" the counting element (CE). This allows reading the contents of the Counters "on the fly" without affecting counting in progress. Multiple Counter Latch Commands may be used to latch more than one Counter. Each latched Counter's OL holds its count until it is read. Counter Latch Commands do not affect the programmed Mode of the Counter in any way.

If a Counter is latched and then, some time later, latched again before the count is read, the second Counter Latch Command is ignored. The count read will be the count at the time the first Counter Latch Command was issued.

With either method, the count must be read according to the programmed format; specifically, if the Counter is programmed for two byte counts, two bytes must be read. The two bytes do not have to be read one right after the other; read or write or programming operations of other Counters may be inserted between them.

Another feature of the 8254 is that reads and writes of the same Counter may be interleaved; for example, if the Counter is programmed for two byte counts, the following sequence is valid.

1) Read least significant byte.
2) Write new least significant byte.
3) Read most significant byte.
4) Write new most significant byte.

If a Counter is programmed to read/write two-byte counts, the following precaution applies: A program must not transfer control between reading the first and second byte to another routine which also reads from that same Counter. Otherwise, an incorrect count will be read.

## READ-BACK COMMAND

The third method uses the Read-Back Command. This command allows the user to check the count value, programmed Mode, and current states of the OUT pin and Null Count flag of the selected counter(s).

The command is written into the Control Word Register and has the format shown in Figure 10. The command applies to the counters selected by setting their corresponding bits D3, D2, D1 $=1$.

$\mathrm{D}_{5}: 0=$ Latch count of selected counter(s)
$\mathrm{D}_{4}: 0=$ Latch status of selected counters(s)
$D_{3}: 1=$ Select Counter 2
$D_{2}: 1=$ Select Counter 1
$D_{1}: 1=$ Select Counter 0
$D_{0}$ : Reserved for future expansion; Must be 0
Figure 10. Read-Back Command Format

- The read-back command may be used to latch multiple counter output latches (OL) by setting the COUNT bit D5 $=0$ and selecting the desired counter(s). This single command is functionally equivalent to several counter latch commands, one for each counter latched. Each counter's latched count is held until it is read (or the counter is reprogrammed). The counter is automatically unlatched when read, but other counters remain latched until they are read. If multiple count read-back commands are issued to the same counter without reading the count, all but the first are ignored; i.e., the count which will be read is the count at the time the first read-back command was issued.

The read-back command may also be used to latch status information of selected counter(s) by setting STATUS bit D4 $=0$. Status must be latched to be read; status of a counter is accessed by a read from that counter.

The counter status format is shown in Figure 11. Bits D5 through DO contain the counter's programmed Mode exactly as written in the last Mode Control Word. OUTPUT bit D7 contains the current state of the OUT pin. This allows the user to monitor the counter's output via software, possibly eliminating some hardware from a system.

| $\mathrm{D}_{7}$ | $\mathrm{D}_{6}$ | $\mathrm{D}_{5}$ | $\mathrm{D}_{4}$ | $\mathrm{D}_{3}$ | $\mathrm{D}_{2}$ | $\mathrm{D}_{1}$ | $\mathrm{D}_{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output | $\begin{aligned} & \text { Null } \\ & \text { Count } \end{aligned}$ | RW1 | RW0 | M2 | M1 | M0 | BCD |
| $\begin{array}{ll} D_{7} \quad 1 & =\text { OUT Pin is } 1 \\ & 0=\text { OUT Pin is } 0 \end{array}$ |  |  |  |  |  |  |  |
| $\begin{array}{ll} D_{6} & 1=\text { Null Count } \\ & 0=\text { Count available for reading } \end{array}$ |  |  |  |  |  |  |  |
| $D_{5}-D_{0}$ Counter programmed mode (see Figure 7) |  |  |  |  |  |  |  |

Figure 11. Status Byte

NULL COUNT bit D6 indicates when the last count written to the counter register (CR) has been loaded into the counting element (CE). The exact time this happens depends on the Mode of the counter and is described in the Mode Definitions, but until the count is loaded into the counting element (CE), it can't be read from the counter. If the count is latched or read before this time, the count value will not reflect the new count just written. The operation of Null Count is shown in Figure 12.

## This Action

A. Write to the control word register; (1)
B. Write to the count register (CR);(2)
C. New Count is loaded into CE (CR $\longrightarrow C E)$;

## NOTE:

1. Only the counter specified by the control word will have its Null Count set to 1. Null count bits of other counters are unaffected.
2. If the counter is programmed for two-byte counts (least significant byte then most significant byte) Null Count goes to 1 when the second byte is written.

Figure 12. Null Count Operation
If multiple status latch operations of the counter(s) are performed without reading the status, all but the first are ignored; i.e., the status that will be read is the status of the counter at the time the first status read-back command was issued.

Both count and status of the selected counter(s) may be latched simultaneously by setting both
$\overline{\text { COUNT }}$ and STATUS bits D5,D4 $=0$. This is functionally the same as issuing two separate read-back commands at once, and the above discussions apply here also. Specifically, if multiple count and/or status read-back commands are issued to the same counter(s) without any intervening reads, all but the first are ignored. This is illustrated in Figure 13.

If both count and status of a counter are latched, the first read operation of that counter will return latched status, regardless of which was latched first. The next one or two reads (depending on whether the counter is programmed for one or two type counts) return latched count. Subsequent reads return unlatched count.

| $\overline{\mathbf{C S}}$ | $\overline{\mathbf{R D}}$ | $\overline{\mathbf{W R}}$ | $\mathbf{A}_{\mathbf{1}}$ | $\mathbf{A}_{\mathbf{0}}$ |  |
| :---: | :---: | :---: | :---: | :---: | :--- |
| 0 | 1 | 0 | 0 | 0 | Write into Counter 0 |
| 0 | 1 | 0 | 0 | 1 | Write into Counter 1 |
| 0 | 1 | 0 | 1 | 0 | Write into Counter 2 |
| 0 | 1 | 0 | 1 | 1 | Write Control Word |
| 0 | 0 | 1 | 0 | 0 | Read from Counter 0 |
| 0 | 0 | 1 | 0 | 1 | Read from Counter 1 |
| 0 | 0 | 1 | 1 | 0 | Read from Counter 2 |
| 0 | 0 | 1 | 1 | 1 | No-Operation (3-State) |
| 1 | X | X | X | X | No-Operation (3-State) |
| 0 | 1 | 1 | X | X | No-Operation (3-State) |

Figure 14. Read/Write Operations Summary


Figure 13. Read-Back Command Example

## Mode Definitions

The following are defined for use in describing the operation of the 8254.
CLK Pulse: a rising edge, then a falling edge, in that order, of a Counter's CLK input.
Trigger: $\quad$ a rising edge of a Counter's GATE input.
Counter loading: the transfer of a count from the CR to the CE (refer to the "Functional Description')

## MODE O: INTERRUPT ON TERMINAL COUNT

Mode 0 is typically used for event counting. After the Control Word is written, OUT is initially low, and will remain low until the Counter reaches zero. OUT then goes high and remains high until a new count or a new Mode 0 Control Word is written into the Counter.

GATE $=1$ enables counting; GATE $=0$ disables counting. GATE has no effect on OUT.

After the Control Word and initial count are written to a Counter, the initial count will be loaded on the next CLK pulse. This CLK pulse does not decrement the count, so for an initial count of N, OUT does not go high until $N+1$ CLK pulses after the initial count is written.

If a new count is written to the Counter, it will be loaded on the next CLK puise and counting will continue from the new count. If a two-byte count is written, the following happens:

1) Writing the first byte disables counting. OUT is set low immediately (no clock pulse required)
2) Writing the second byte allows the new count to be loaded on the next CLK pulse.

This allows the counting sequence to be synchronized by software. Again, OUT does not go high until $N+1$ CLK pulses after the new count of $N$ is written.

If an initial count is written while GATE $=0$, it will still be loaded on the next CLK pulse. When GATE goes high, OUT will go high N CLK pulses later; no CLK pulse is needed to load the Counter as this has already been done.

## MODE 1: HARDWARE RETRIGGERABLE ONE-SHOT

OUT will be initially high. OUT will go low on the CLK pulse following a trigger to begin the one-shot pulse, and will remain low until the Counter reaches zero.

OUT will then go high and remain high until the CLK pulse after the next trigger.

After writing the Control Word and initial count, the Counter is armed. A trigger results in loading the Counter and setting OUT low on the next CLK pulse, thus starting the one-shot pulse. An initial count of $N$ will result in a one-shot pulse N CLK cycles in duration. The one-shot is retriggerable, hence OUT will remain low for N CLK pulses after any trigger. The one-shot pulse can be repeated without rewriting the same count into the counter. GATE has no effect on - OUT.

If a new count is written to the Counter during a oneshot pulse, the current one-shot is not affected unless the counter is retriggered. In that case, the Counter is loaded with the new count and the oneshot pulse continues until the new count expires.

## MODE 2: RATE GENERATOR

This Mode functions like a divide-by- N counter. It is typically used to generate a Real Time Clock interrupt. OUT will initialiy be high. When the initial count has decremented to 1 , OUT goes low for one CLK pulse. OUT then goes high again, the Counter reloads the initial count and the process is repeated. Mode 2 is periodic; the same sequence is repeated indefinitely. For an initial count of $N$, the sequence repeats every N CLK cycles.

GATE $=1$ enables counting; GATE $=0$ disables counting. If GATE goes low during an output pulse, OUT is set high immediately. A trigger reloads the Counter with the initial count on the next CLK pulse; OUT goes low N CLK pulses after the trigger. Thus the GATE input can be used to synchronize the Counter.

After writing a Control Word and initial count, the Counter will be loaded on the next CLK pulse. OUT goes low N CLK Pulses after the initial count is written. This allows the Counter to be synchronized by software also.

Writing a new count while counting does not affect the current counting sequence. If a trigger is received after writing a new count but before the end of the current period, the Counter will be loaded with the new count on the next CLK pulse and counting will continue from the new count. Otherwise, the new count will be loaded at the end of the current counting cycle. In mode 2, a COUNT of 1 is illegal.

## MODE 3: SQUARE WAVE MODE

Mode 3 is typically used for Baud rate generation. Mode 3 is similar to Mode 2 except for the duty cycle of OUT. OUT will initially be high. When half the


NOTE:
The following conventions apply to all mode timing diagrams:

1. Counters are programmed for binary (not BCD) counting and for reading/writing least significant byte (LSB) only.
2. The counter is always selected ( $\overline{\mathrm{CS}}$ always low).
3. CW stands for "Control Word"; CW = 10 means a control word of 10 HEX is written to the counter.
4. LSB stands for "Least Significant Byte" of count.
5. Numbers below diagrams are count values. The lower number is the least significant byte. The upper number is the most significant byte. Since the counter is programmed to read/write LSB only, the most significant byte cannot be read. N stands for an undefined count.
Vertical lines show transitions between count values.
Figure 15. Mode 0


Figure 16. Mode 1
initial count has expired, OUT goes low for the remainder of the count. Mode 3 is periodic; the sequence above is repeated indefinitely. An initial count of $N$ results in a square wave with a period of N CLK cycles.

GATE $=1$ enables counting; GATE $=0$ disables counting. If GATE goes low while OUT is low, OUT is set high immediately; no CLK pulse is required. A trigger reloads the Counter with the initial count on the next CLK pulse. Thus the GATE input can be used to synchronize the Counter.

After writing a Control Word and initial count, the Counter will be loaded on the next CLK pulse. This allows the Counter to be synchronized by software also.

Writing a new count while counting does not affect the current counting sequence. If a trigger is received after writing a new count but before the end of the current half-cycle of the square wave, the Counter will be loaded with the new count on the next CLK pulse and counting will continue from the


Figure 17. Mode 2
new count. Otherwise, the new count will be loaded at the end of the current half-cycle.

Mode 3 is implemented as follows:
Even counts: OUT is initially high. The initial count is loaded on one CLK pulse and then is decremented by two on succeeding CLK pulses. When the count expires OUT changes value and the Counter is reloaded with the initial count. The above process is repeated indefinitely.

Odd counts: OUT is initially high. The initial count minus one (an even number) is loaded on one CLK pulse and then is decremented by two on succeeding CLK pulses. One CLK pulse after the count expires, OUT goes low and the Counter is reloaded with the initial count minus one. Succeeding CLK puises decrement the count by two. When the count expires, OUT goes high again and the Counter is reloaded with the initial count minus one. The above process is repeated indefinitely. So for odd counts, OUT will be high for $(\mathrm{N}+1) / 2$ counts and low for ( $\mathrm{N}-1$ )/2 counts.












NOTE:
A GATE transition should not occur one clock prior to terminal count.
Figure 18. Mode 3

## MODE 4: SOFTWARE TRIGGERED STROBE

OUT will be initially high. When the initial count expires, OUT will go low for one CLK pulse and then go high again. The counting sequence is "triggered" by writing the initial count.

GATE $=1$ enables counting; GATE $=0$ disables counting. GATE has no effect on OUT.

After writing a Control Word and initial count, the Counter will be loaded on the next CLK pulse. This CLK pulse does not decrement the count, so for an
initial count of $N$, OUT does not strobe low until $N+$ 1 CLK pulses after the initial count is written.

If a new count is written during counting, it will be loaded on the next CLK pulse and counting will continue from the new count. If a two-byte count is written, the following happens:

1) Writing the first byte has no effect on counting.
2) Writing the second byte allows the new count to be loaded on the next CLK pulse.

This allows the sequence to be "retriggered" by software. OUT strobes low $N+1$ CLK pulses after the new count of $N$ is written.


Figure 19. Mode 4

## MODE 5: HARDWARE TRIGGERED STROBE (RETRIGGERABLE)

OUT will initially be high. Counting is triggered by a rising edge of GATE. When the initial count has expired, OUT will go low for one CLK pulse and then go high again.

After writing the Control Word and initial count, the counter will not be loaded until the CLK pulse after a trigger. This CLK pulse does not decrement the count, so for an initial count of N, OUT does not strobe low until $\mathrm{N}+1$ CLK pulses after a trigger.

A trigger results in the Counter being loaded with the initial count on the next CLK pulse. The counting sequence is retriggerable. OUT will not strobe low for $N+1$ CLK pulses after any trigger. GATE has no effect on OUT.

If a new count is written during counting, the current counting sequence will not be affected. If a trigger occurs after the new count is written but before the current count expires, the Counter will be loaded with the new count on the next CLK pulse and counting will continue from there.


Figure 20. Mode 5

| Signal Status <br> Modes | Low Or Going Low | Rising | High |
| :---: | :---: | :---: | :---: |
| 0 | Disables Counting | - - | Enables Counting |
| 1 | - - | 1) Initiates Counting <br> 2) Resets Output after Next Clock | -- |
| 2 | 1) Disables Counting <br> 2) Sets Output Immediately High | Initiates <br> Counting | Enables <br> Counting |
| 3 | 1) Disables Counting <br> 2) Sets Output Immediately High | Initiates Counting | Enables <br> Counting |
| 4 | Disables Counting | -- | Enables Counting |
| 5 | - - | Initiates <br> Counting | - - |

Figure 21. Gate Pin Operations Summary

| Mode | Min <br> Count | Max <br> Count |
| :---: | :---: | :---: |
| 0 | 1 | 0 |
| 1 | 1 | 0 |
| 2 | 2 | 0 |
| 3 | 2 | 0 |
| 4 | 1 | 0 |
| 5 | 1 | 0 |

NOTE:
0 is equivalent to $2^{16}$ for binary counting and $10^{4}$ for $B C D$ counting.

Figure 22. Minimum and Maximum Initial Counts

## Operation Common to All Modes

## PROGRAMMING

When a Control Word is written to a Counter, all Control Logic is immediately reset and OUT goes to a known initial state; no CLK pulses are required for this.

## GATE

The GATE input is always sampled on the rising edge of CLK. In Modes $0,2,3$, and 4 the GATE input is level sensitive, and the logic level is sampled on the rising edge of CLK. In Modes 1, 2, 3, and 5 the GATE input is rising-edge sensitive. In these Modes, a rising edge of GATE (trigger) sets an edge-sensitive flip-flop in the Counter. This flip-flop is then sampled on the next rising edge of CLK; the flip-flop is reset immediately after it is sampled. In this way, a trigger will be detected no matter when it occurs-a high logic level does not have to be maintained until the next rising edge of CLK. Note that in Modes 2 and 3 , the GATE input is both edge- and level-sensitive. In Modes 2 and 3, if a CLK source other than the system clock is used, GATE should be pulsed immediately following $\overline{W R}$ of a new count value.

## COUNTER

New counts are loaded and Counters are decremented on the falling edge of CLK.

The largest possible initial count is 0 ; this is equivalent to $2^{16}$ for binary counting and 104 for $B C D$ counting.

The Counter does not stop when it reaches zero. In Modes $0,1,4$, and 5 the Counter "wraps around" to the highest count, either FFFF hex for binary counting or 9999 for BCD counting, and continues counting. Modes 2 and 3 are periodic; the Counter reloads itself with the initial count and continues counting from there.

## ABSOLUTE MAXIMUM RATINGS*

Ambient Temperature Under Bias $\qquad$ $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$
Storage Temperature $\qquad$ $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Voltage on Any Pin with
Respect to Ground............... -0.5 V to +7 V
Power Dissipation 1W
*Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## D.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V} \pm 10 \%$

| Symbol | Parameter | Min | Max | Units | Test Conditions |
| :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{IL}}$ | Input Low Voltage | -0.5 | 0.8 | V |  |
| $\mathrm{~V}_{\text {IH }}$ | Input High Voltage | 2.0 | $\mathrm{~V}_{\mathrm{CC}}+0.5 \mathrm{~V}$ | V |  |
| $\mathrm{~V}_{\mathrm{OL}}$ | Output Low Voltage |  | 0.45 | V | $\mathrm{I}_{\mathrm{OL}}=2.0 \mathrm{~mA}$ |
| $\mathrm{~V}_{\mathrm{OH}}$ | Output High Voltage | 2.4 |  | V | $\mathrm{I}_{\mathrm{OH}}=-400 \mu \mathrm{~A}$ |
| $\mathrm{I}_{\mathrm{IL}}$ | Input Load Current |  | $\pm 10$ | $\mu \mathrm{~A}$ | $\mathrm{~V}_{\mathrm{IN}}=\mathrm{V}_{\mathrm{CC}}$ to 0 V |
| $\mathrm{I}_{\mathrm{OFL}}$ | Output Float Leakage |  | $\pm 10$ | $\mu \mathrm{~A}$ | $\mathrm{~V}_{\mathrm{OUT}}=\mathrm{V}_{\mathrm{CC}}$ to 0.45 V |
| $\mathrm{I}_{\mathrm{CC}}$ | $\mathrm{V}_{\mathrm{CC}}$ Supply Current |  | 170 | mA |  |
| $\mathrm{C}_{\text {IN }}$ | Input Capacitance |  | 10 | pF | $\mathrm{f}_{\mathrm{C}}=1 \mathrm{MHz}$ |
| $\mathrm{C}_{\text {I/O }}$ | I/O Capacitance |  | 20 | pF | Unmeasured pins <br> returned to $\mathrm{V}_{\mathrm{SS}}{ }^{(4)}$ |

## A.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V} \pm 10 \%, \mathrm{GND}=0 \mathrm{~V}$

## Bus Parameters(1)

READ CYCLE

| Symbol | Parameter | 8254-5 |  | 8254 |  | 8254-2 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max |  |
| $t_{\text {AR }}$ | Address Stable Before $\overline{\mathrm{RD}} \downarrow$ | 45 |  | 45 |  | 30 |  | ns |
| ${ }_{\text {t }}$ R | $\overline{\mathrm{CS}}$ Stable Before $\overline{\mathrm{RD}} \downarrow$ | 0 |  | 0 |  | 0 |  | ns |
| $\mathrm{t}_{\mathrm{RA}}$ | Address Hold Time After $\overline{\mathrm{R}} \mathrm{D} \uparrow$ | 0 |  | 0 |  | 0 |  | ns |
| $t_{\text {RR }}$ | $\overline{\mathrm{RD}}$ Pulse Width | 150 |  | 150 |  | 95 |  | ns |
| $t_{\text {RD }}$ | Data Delay from $\overline{\mathrm{RD}} \downarrow$ |  | 120 |  | 120 |  | 85 | ns |
| $t_{A D}$ | Data Delay from Address |  | 220 |  | 220 |  | 185 | ns |
| $t_{\text {dF }}$ | $\overline{\mathrm{RD}} \uparrow$ to Data Floating | 5 | 90 | 5 | 90 | 5 | 65 | ns |
| $t_{\text {t }}$ | Command Recovery Time | 200 |  | 200 |  | 165 |  | ns |

## NOTE:

1. AC timings measured at $\mathrm{V}_{\mathrm{OH}}=2.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{OL}}=0.8 \mathrm{~V}$.

## A.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V} \pm 10 \%, \mathrm{GND}=0 \mathrm{~V}$ (Continued)

WRITE CYCLE

| Symbol | Parameter | 8254-5 |  | 8254 |  | 8254-2 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max |  |
| ${ }^{\text {taw }}$ | Address Stable Before $\overline{\mathrm{WR}} \downarrow$ | 0 |  | 0 |  | 0 |  | ns |
| ${ }_{\text {t }}$ W | $\overline{\mathrm{CS}}$ Stable Before $\overline{\mathrm{WR}} \downarrow$ | 0 |  | 0 |  | 0 |  | ns |
| tWA | Address Hold Time After $\overline{\mathrm{WR}} \downarrow$ | 0 |  | 0 |  | 0 |  | ns |
| tww | $\overline{\text { WR Pulse Width }}$ | 150 |  | 150 |  | 95 |  | ns |
| $t_{\text {bw }}$ | Data Setup Time Before $\overline{W R} \uparrow$ | 120 |  | 120 |  | 95 |  | ns |
| twD | Data Hoid Time After $\bar{W}$ R $\uparrow$ | 0 |  | 0 |  | 0 |  | ns |
| $t_{\text {RV }}$ | Command Recovery Time | 200 |  | 200 |  | 165 |  | ns |

CLOCK AND GATE

| Symbol | Parameter | 8254-5 |  | 8254 |  | 8254-2 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max |  |
| ${ }_{\text {t }}^{\text {CLK }}$ | Clock Period | 200 | DC | 125 | DC | 100 | DC | ns |
| tpwh | High Pulse Width | 60(3) |  | 60(3) |  | 30(3) |  | ns |
| tpwL | Low Pulse Width | 60(3) |  | 60(3) |  | 50(3) |  | ns |
| $t_{R}$ | Clock Rise Time |  | 25 |  | 25 |  | 25 | ns |
| $t_{\text {F }}$ | Clock Fall Time |  | 25 |  | 25 |  | 25 | ns |
| $\mathrm{t}_{\mathrm{GW}}$ | Gate Width High | 50 |  | 50 |  | 50 |  | ns |
| $\mathrm{t}_{\mathrm{GL}}$ | Gate Width Low | 50 |  | 50 |  | 50 |  | ns |
| $t_{\text {GS }}$ | Gate Setup Time to CLK $\uparrow$ | 50 |  | 50 |  | 40 |  | ns |
| $\mathrm{t}_{\mathrm{GH}}$ | Gate Setup Time After CLK $\uparrow$ | 50(2) |  | 50(2) |  | 50(2) |  | ns |
| tod | Output Delay from CLK $\downarrow$ |  | 150 |  | 150 |  | 100 | ns |
| todg | Output Delay from Gate $\downarrow$ |  | 120 |  | 120 |  | 100 | ns |
| $t_{\text {w }}$ | CLK Delay for Loading $\downarrow$ | 0 | 55 | 0 | 55 | 0 | 55 | ns |
| twg | Gate Delay for Sampling | -5 | 50 | -5 | 50 | -5 | 40 | ns |
| two | OUT Delay from Mode Write |  | 260 |  | 260 |  | 240 | ns |
| $\mathrm{t}_{\mathrm{CL}}$ | CLK Set Up for Count Latch | -40 | 45 | -40 | 45 | -40 | 40 | ns |

NOTES:
2. In Modes 1 and 5 triggers are sampled on each rising clock edge. A second trigger within 120 ns ( 70 ns for the $8254-2$ ) of the rising clock edge may not be detected.
3. Low-going glitches that violate tPWH, tPWL may cause errors requiring counter reprogramming.
4. Sampled, not $100 \%$ tested. $T_{A}=25^{\circ} \mathrm{C}$.
5. If CLK present at TWC min then Count equals $N+2$ CLK pulses, TWC max equals Count $N+1$ CLK pulse. TWC min to TWC max, count will be either $N+1$ or $N+2$ CLK pulses.
6. In Modes 1 and 5, if GATE is present when writing a new Count value, at TWG min Counter will not be triggered, at TWG max Counter will be triggered.
7. If CLK present when writing a Counter Latch or ReadBack Command, at TCL min CLK will be reflected in count value latched, at TCL max CLK will not be reflected in the count value latched.

## WAVEFORMS

WRITE


READ


WAVEFORMS (Continued)
RECOVERY


CLOCK AND GATE


## A.C. TESTING INPUT, OUTPUT WAVEFORM


A.C. Testing: Inputs are driven at 2.4 V for a Logic "1" and 0.45 V for a Logic " 0 ." Timing measurements are made at 2.0 V for a Logic " 1 " and 0.8 V for a Logic " 0 ".

## A.C. TESTING LOAD CIRCUIT



231164-18
$C_{L}=150 \mathrm{pF}$
$C_{L}$ Inctudes Jig Capacitance

# 8255A/8255A-5 PROGRAMMABLE PERIPHERAL INTERFACE 

- MCS-85TM Compatible 8255A-5 24 Programmable 1/O Pins
- Completely TTL Compatible
- Fully Compatible with Intel Microprocessor Families
- Improved Timing Characteristics
- Direct Bit Set/Reset Capability Easing Control Application Interface
- Reduces System Package Count
- Improved DC Driving Capability
- Available in EXPRESS
- Standard Temperature Range
- Extended Temperature Range
- 40 Pin DIP Package or 44 Lead PLCC
(See Intel Packaging: Order Number: 231369) The Intel 8255A is a general purpose programmable I/O device designed for use with Intel microprocessors. It has 24 I/O pins which may be individually programmed in 2 groups of 12 and used in 3 major modes of operation. In the first mode (MODE 0), each group of $12 \mathrm{l} / \mathrm{O}$ pins may be programmed in sets of 4 to be input or output. In MODE 1, the second mode, each group may be programmed to have 8 lines of input or output. Of the remaining 4 pins, 3 are used for handshaking and interrupt control signals. The third mode of operation (MODE 2) is a bidirectional bus mode which uses 8 lines for a bidirectional bus, and 5 lines, borrowing one from the other group, for handshaking.


Figure 1. 8255A Block Diagram

## 8255A FUNCTIONAL DESCRIPTION

## General

The 8255 A is a programmable peripheral interface (PPI) device designed for use in Intel microcomputer systems. Its function is that of a general purpose I/O component to interface peripheral equipment to the microcomputer system bus. The functional configuration of the 8255 A is programmed by the system software so that normally no external logic is necessary to interface peripheral devices or structures.

## Data Bus Buffer

This 3-state bidirectional 8-bit buffer is used to interface the 8255A to the system data bus. Data is transmitted or received by the buffer upon execution of input or output instructions by the CPU. Control words and status information are also transferred through the data bus buffer.

## Read/Write and Control Logic

The function of this block is to manage all of the internal and external transfers of both Data and Control or Status words. It accepts inputs from the

CPU Address and Control busses and in turn, issues commands to both of the Control Groups.

## (CS)

Chip Select. A "low" on this input pin enables the communication between the 8255A and the CPU.

## ( $\overline{\mathrm{RD}})$

Read. A "low" on this input pin enables the 8255A to send the data or status information to the CPU on the data bus. In essence, it allows the CPU to "read from' the 8255A.

## (WR)

Write. A "low" on this input pin enables the CPU to write data or control words into the 8255A.

## ( $A_{0}$ and $A_{1}$ )

Port Select 0 and Port Select 1. These input signals, in conjunction with the RD and WR inputs, control the selection of one of the three ports or the control word registers. They are normally connected to the least significant bits of the address bus ( $\mathrm{A}_{0}$ and $A_{1}$ ).


Figure 3. 8255A Block Diagram Showing Data Bus Buffer and Read/Write Control Logic Functions

8255A BASIC OPERATION

| $A_{1}$ | $A_{\mathbf{0}}$ | $\overline{\mathrm{RD}}$ | $\overline{\mathbf{W}} \overline{\mathbf{C S}}$ | $\overline{\text { Input Operation (READ) }}$ |  |
| :---: | :---: | :---: | :---: | :---: | :--- |
| 0 | 0 | 0 | 1 | 0 | Port A $\rightarrow$ Data Bus |
| 0 | 1 | 0 | 1 | 0 | Port B $\rightarrow$ Data Bus |
| 1 | 0 | 0 | 1 | 0 | Port $\mathrm{C} \rightarrow$ Data Bus |
|  |  |  |  |  | Output Operation <br> (WRITE) |
| 0 | 0 | 1 | 0 | 0 | Data Bus $\rightarrow$ Port A |
| 0 | 1 | 1 | 0 | 0 | Data Bus $\rightarrow$ Port B |
| 1 | 0 | 1 | 0 | 0 | Data Bus $\rightarrow$ Port C |
| 1 | 1 | 1 | 0 | 0 | Data Bus $\rightarrow$ Control |
|  |  |  |  |  | Disable Function |
| X | X | X | X | 1 | Data Bus $\rightarrow$ 3-State |
| 1 | 1 | 0 | 1 | 0 | Illegal Condition |
| X | X | 1 | 1 | 0 | Data Bus $\rightarrow$ 3-State |

## (RESET)

Reset. A "high" on this input clears the control register and all ports ( $\mathrm{A}, \mathrm{B}, \mathrm{C}$ ) are set to the input mode.

## Group A and Group B Controls

The functional configuration of each port is programmed by the systems software. In essence, the CPU "outputs" a control word to the 8255A. The control word contains information such as "mode", "bit set", "bit reset", etc., that initializes the functional configuration of the 8255A.

Each of the Control blocks (Group A and Group B) accepts "commands" from the Read/Write Control Logic, receives "control words" from the internal data bus and issues the proper commands to its associated ports.

Control Group A-Port A and Port C upper (C7-C4)
Control Group B-Port B and Port C lower (C3-C0)
The Control Word Register can Only be written into. No Read operation of the Control Word Register is allowed.

## Ports A, B, and C

The 8255A contains three 8 -bit ports ( $\mathrm{A}, \mathrm{B}$, and C ). All can be configured in a wide variety of functional characteristics by the system software but each has its own special features or "personality" to further enhance the power and flexibility of the 8255A.

Port A. One 8-bit data output latch/buffer and one 8 -bit data input latch.

Port B. One 8-bit data input/output latch/buffer and one 8 -bit data input buffer.

Port C. One 8-bit data output latch/buffer and one 8 -bit data input buffer (no latch for input). This port can be divided into two 4-bit ports under the mode control. Each 4 -bit port contains a 4 -bit latch and it can be used for the control signal outputs and status signal inputs in conjunction with ports A and B .


231308-4
Figure 4. 8225A Block Diagram Showing Group A and Group B Control Functions


Pin Names

| $D_{7}-D_{0}$ | Data Bus (Bi-Directional) |
| :--- | :--- |
| RESET | Reset Input |
| $\overline{\mathrm{CS}}$ | Chip Select |
| $\overline{\mathrm{RD}}$ | Read Input |
| $\overline{\mathrm{WR}}$ | Write Input |
| $\mathrm{A0}, \mathrm{~A} 1$ | Port Address |
| PA7-PA0 | Port A (BIT) |
| PB7-PB0 | Port B (BIT) |
| PC7-PC0 | Port C (BIT) |
| $V_{C C}$ | +5 Volts |
| GND | 0 Volts |

## 8255A OPERATIONAL DESCRIPTION

## Mode Selection

There are three basic modes of operation that can be selected by the system software:

## Mode 0-Basic Input/Output

## Mode 1-Strobed Input/Output

Mode 2-Bi-Directional Bus
When the reset input goes "high" all ports will be set to the input mode (i.e., all 24 lines will be in the high impedance state). After the reset is removed the 8255A can remain in the input mode with no additional initialization required. During the execution of the system program any of the other modes may be selected using a single output instruction. This allows a single 8255A to service a variety of peripheral devices with a simple software maintenance routine.

The modes for Port A and Port B can be separately defined, while Port $C$ is divided into two portions as required by the Port A and Port B definitions. All of the output registers, including the status flip-flops, will be reset whenever the mode is changed. Modes may be combined so that their functional definition can be "tailored" to almost any 1/O structure. For instance; Group B can be programmed in Mode 0 to monitor simple switch closings or display computational results, Group A could be programmed in Mode 1 to monitor a keyboard or tape reader on an interrupt-driven basis.


Figure 5. Basic Mode Definitions and Bus Interface


Figure 6. Mode Definition Format
The mode definitions and possible mode combinations may seem confusing at first but after a cursory review of the complete device operation a simple, logical I/O approach will surface. The design of the 8255A has taken into account things such as efficient PC board layout, control signal definition vs PC layout and complete functional flexibility to support almost any peripheral device with no external logic. Such design represents the maximum use of the - available pins.

## Single Bit Set/Reset Feature

Any of the eight bits of Port C can be Set or Reset using a single OUTput instruction. This feature reduces software requirements in Control-based applications.


Figure 7. Bit Set/Reset Format
When Port C is being used as status/control for Port A or B, these bits can be set or reset by using the Bit Set/Reset operation just as if they were data output ports.

## Interrupt Control Functions

When the 8255A is programmed to operate in mode 1 or mode 2, control signals are provided that can be used as interrupt request inputs to the CPU. The interrupt request signals, generated from port C , can be inhibited or enabled by setting or resetting the associated INTE flip-flop, using the bit set/reset function of port C .

This function allows the Programmer to disallow or allow a specific I/O device to interrupt the CPU without affecting any other device in the interrupt structure.

INTE flip-flop definition:
(BIT-SET)-INTE is set-Interrupt enable
(BIT-RESET)-INTE is RESET-Interrupt disable

## NOTE:

All Mask flip-flops are automatically reset during mode selection and device Reset.

## Operating Modes

MODE 0 (Basic Input/Output). This functional configuration provides simple input and output operations for each of the three ports. No "handshaking" is required, data is simply written to or read from a specified port.

Mode 0 Basic Functional Definitions:

- Two 8-bit ports and two 4-bit ports.
- Any port can be input or output.
- Outputs are latched.
- Inputs are not latched.
- 16 different Input/Output configurations are possible in this Mode.

MODE 0 (BASIC INPUT)


MODE 0 (BASIC OUTPUT)


MODE 0 PORT DEFINITION

| A |  | B |  | Group A |  |  | Group B |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{D}_{4}$ | $\mathbf{D}_{3}$ | $\mathbf{D}_{1}$ | $\mathbf{D}_{\mathbf{0}}$ | Port A | Port C <br> (Upper) | $\#$ | Port B | Port C <br> (Lower) |
| 0 | 0 | 0 | 0 | OUTPUT | OUTPUT | 0 | OUTPUT | OUTPUT |
| 0 | 0 | 0 | 1 | OUTPUT | OUTPUT | 1 | OUTPUT | INPUT |
| 0 | 0 | 1 | 0 | OUTPUT | OUTPUT | 2 | INPUT | OUTPUT |
| 0 | 0 | 1 | 1 | OUTPUT | OUTPUT | 3 | INPUT | INPUT |
| 0 | 1 | 0 | 0 | OUTPUT | INPUT | 4 | OUTPUT | OUTPUT |
| 0 | 1 | 0 | 1 | OUTPUT | INPUT | 5 | OUTPUT | INPUT |
| 0 | 1 | 1 | 0 | OUTPUT | INPUT | 6 | INPUT | OUTPUT |
| 0 | 1 | 1 | 1 | OUTPUT | INPUT | 7 | INPUT | INPUT |
| 1 | 0 | 0 | 0 | INPUT | OUTPUT | 8 | OUTPUT | OUTPUT |
| 1 | 0 | 0 | 1 | INPUT | OUTPUT | 9 | OUTPUT | INPUT |
| 1 | 0 | 1 | 0 | INPUT | OUTPUT | 10 | INPUT | OUTPUT |
| 1 | 0 | 1 | 1 | INPUT | OUTPUT | 11 | INPUT | INPUT |
| 1 | 1 | 0 | 0 | INPUT | INPUT | 12 | OUTPUT | OUTPUT |
| 1 | 1 | 0 | 1 | INPUT | INPUT | 13 | INPUTPUT | INPUT |
| 1 | 1 | 1 | 0 | INPUT | INPUT | 14 | INPUT | OUTPUT |
| 1 | 1 | 1 | 1 | INPUT | INPUT | 15 | INPUT | INPUT |

MODE CONFIGURATIONS

| CONTROL WORD \#0$\begin{array}{llllllll} D_{3} & D_{6} & D_{5} & D_{4} & D_{3} & D_{2} & D_{1} & D_{0} \end{array}$ |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 0 | 0 |  |  | 0 |  |
|  |  |  |  |  |  |  |

CONTROL WORD \#1


CONTROL WORD \#4


CONTROL WORD \#2


CONTROL WORD \#3


CONTROL WORD \#5

| $\mathrm{D}_{7}$ | $\mathrm{D}_{6}$ | $\mathrm{D}_{5}$ | $\mathrm{D}_{4}$ | $\mathrm{D}_{3}$ | $\mathrm{D}_{2}$ | $\mathrm{D}_{1}$ | $\mathrm{D}_{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | 0 | 0 | 0 | 1 | 0 | 0 | 1 |



CONTROL WORD $\# 6$

| $D_{7}$ | $D_{6}$ | $D_{5}$ | $D_{4}$ | $D_{3}$ | $D_{2}$ | $D_{1}$ | $D_{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | 0 | 0 | 0 | 1 | 0 | 1 | 0 |



CONTROL WORD \#7

| $D_{7}$ | $D_{6}$ | $D_{5}$ | $D_{4}$ | $D_{3}$ | $D_{2}$ | $D_{1}$ | $D_{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | 0 | 0 | 0 | 1 | 0 | 1 | 1 |



CONTROL WORD \#9

| $\mathrm{D}_{7}$ | $\mathrm{D}_{6}$ | $\mathrm{D}_{5}$ | $\mathrm{D}_{4}$ | $\mathrm{D}_{3}$ | $\mathrm{D}_{2}$ | $\mathrm{D}_{1}$ | $\mathrm{D}_{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | 0 | 0 | 1 | 0 | 0 | 0 | 1 |

$\xrightarrow{8255 A}$

CONTROL WORD $=10$

| $D_{7}$ | $D_{6}$ | $D_{5}$ | $D_{4}$ | $D_{3}$ | $D_{2}$ | $D_{1}$ | $D_{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | 0 | 0 | 1 | 0 | 0 | 1 | 0 |



CONTROL WORD \#11

| $D_{7}$ | $D_{6}$ | $D_{5}$ | $D_{4}$ | $D_{3}$ | $D_{2}$ | $D_{1}$ | $D_{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | 0 | 0 | 1 | 0 | 0 | 1 | 1 |




## Operating Modes

MODE 1 (Strobed Input/Output). This functional configuration provides a means for transferring 1/O data to or from a specified port in conjunction with strobes or "handshaking" signals. In mode 1, port A and port $B$ use the lines on port $C$ to generate or accept these "handshaking" signals.

## Mode 1 Basic Functional Definitions:

- Two Groups (Group A and Group B)
- Each group contains one 8-bit data port and one 4-bit control/data port.
- The 8 -bit data port can be either input or output. Both inputs and outputs are latched.
- The 4 -bit port is used for control and status of the 8 -bit data port.


CONTROL WORD \#15


## Input Control Signal Definition

$\overline{\text { STB }}$ (Strobe Input). A "low" on this input loads data into the input latch.

## IBF (Input Buffer Full F/F)

A "high" on this output indicates that the data has been loaded into the input latch; in essence, an acknowledgement. IBF is set by STB input being low and is reset by the rising edge of the RD input.

## INTR (Interrupt Request)

A "high" on this output can be used to interrupt the CPU when an input device is requesting service. INTR is set by the STB is a "one", IBF is a "one" and INTE is a "one". It is reset by the falling edge of $\overline{\mathrm{RD}}$. This procedure allows an input device to request service from the CPU by simply strobing its data into the port.

## INTE A

Controlled by bit set/reset of $\mathrm{PC}_{4}$.


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## INTE B

Controlled by bit set/reset of $\mathrm{PC}_{2}$.


Figure 8. MODE 1 Input


Figure 9. MODE 1 (Strobed Input)

## Output Control Signal Definition

$\overline{\mathrm{OBF}}$ (Output Buffer Full F/F). The $\overline{\mathrm{OBF}}$ output will go "low" to indicate that the CPU has written data out to the specified port. The OBF F/F will be set by the rising edge of the $\overline{W R}$ input and reset by $\overline{A C K}$ input being low.
$\overline{A C K}$ (Acknowledge Input). A "low" on this input informs the 8255A that the data from port A or port $B$ has been accepted. In essence, a response from the peripheral device indicating that it has received the data output by the CPU.

INTR (Interrupt Request). A "high" on this output can be used to interrupt the CPU when an output
device has accepted data transmitted by the CPU. INTR is set when $\overline{A C K}$ is a "one", $\overline{O B F}$ is a "one", and INTE is a "one". It is reset by the falling edge of WR.

## INTE A

Controlled by bit set/reset of $\mathrm{PC}_{6}$.

INTE B
Controlled by bit set/reset of $\mathrm{PC}_{2}$.


Figure 10. MODE 1 Output


Figure 11. MODE 1 (Strobed Output)


Figure 12. Combinations of MODE 1

## Combinations of MODE 1

Port A and Port B can be individually defined as input or output in MODE 1 to support a wide variety of strobed I/O applications.

## Operating Modes

MODE 2 (Strobed Bidirectional Bus I/O). This functional configuration provides a means for communicating with a peripheral device or structure on a single 8 -bit bus for both transmitting and receiving data (bidirectional bus I/O). "Handshaking" signals are provided to maintain proper bus flow discipline in a similar manner to MODE 1. Interrupt generation and enable/disable functions are also available.

MODE 2 Basic Functional Definitions:

- Used in Group A only.
- One 8 -bit, bi-directional bus Port (Port A) and a.5bit control Port (Port C).
- Both inputs and outputs are latched.
- The 5 -bit control port (Port C) is used for control and status for the 8 -bit, bi-directional bus port (Port A).


## Bidirectional Bus I/O Control Signal Definition

INTR (Interrupt Request). A high on this output can be used to interrupt the CPU for both input or output operations.

## Output Operations

$\overline{\mathrm{OBF}}$ (Output Buffer Full). The $\overline{\mathrm{OBF}}$ output will go "low" to indicate that the CPU has written data out to port A .
$\overline{\text { ACK (Acknowledge). A "low" on this input enables }}$ the tri-state output buffer of port A to send out the data. Otherwise, the output buffer will be in the high impedance state.

INTE 1 (The INTE Flip-Flop Associated with OBF). Controlled by bit set/reset of $\mathrm{PC}_{6}$.

## Input Operations

$\overline{\text { STB }}$ (Strobe Input). A "low" on this input loads data into the input latch.

IBF (Input Buffer Full F/F). A "high" on this output indicates that data has been loaded into the input latch.


Figure 13. MODE Control Word

INTE 2 (The INTE Flip-Flop Associated with IBF). Controlled by bit set/reset of $\mathrm{PC}_{4}$.


Figure 14. MODE 2


Figure 15. MODE 2 (Bidirectional)


Figure 16. MODE $1 / 4$ Combinations

Mode Definition Summary

|  | MODE O |  |
| :---: | :---: | :---: |
|  | $\mathbb{I N}$ | OUT |
| $\mathrm{PA}_{0}$ | $\mathbb{I N}$ | OUT |
| $\mathrm{PA}_{1}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PA}_{2}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PA}_{3}$ | $\mathbb{I N}$ | OUT |
| $\mathrm{PA}_{4}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PA}_{5}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PA}_{6}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PA}_{7}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PB}_{0}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PB}_{1}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PB}_{2}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PB}_{3}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PB}_{4}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PB}_{5}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PB}_{6}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PB}_{7}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PC}_{0}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PC}_{1}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PC}_{2}$ | $\mathbb{I N}$ | OUT |
| $\mathrm{PC}_{3}$ | $\mathbb{I N}$ | OUT |
| $\mathrm{PC}_{4}$ | $\mathbb{I N}$ | OUT |
| $\mathrm{PC}_{5}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PC}_{6}$ | $\mathbb{N}$ | OUT |
| $\mathrm{PC}_{7}$ | $\mathbb{I N}$ | OUT |


| MODE 1 |  |
| :---: | :---: |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| $\mathbb{N}$ | OUT |
| $\underline{N}$ | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| IN | OUT |
| $\mathrm{INTR}_{\mathrm{B}}$ | $\mathrm{INTR}_{\text {B }}$ |
| $1 \mathrm{IBF}_{\mathrm{B}}$ | $\overline{\mathrm{OBF}}_{\mathrm{B}}$ |
| $\mathrm{STB}_{\mathrm{S}}$ | $\overline{\mathrm{ACK}}_{\mathrm{B}}$ |
| ${ }^{\mathrm{NNTR}_{A}}$ | $\mathrm{INTR}_{\text {A }}$ |
| $\overline{\text { STB }}_{\text {A }}$ | $1 / 0$ |
| $18 F_{\text {A }}$ | 1/O |
| 1/O | $\overline{\mathrm{ACK}}_{\mathrm{A}}$ |
| 1/0 | $\overline{\mathrm{OBF}}_{\mathrm{A}}$ |


| MODE 2 |
| :---: |
| GROUP A ONLY |
|  |

## Special Mode Combination Considerations

There are several combinations of modes when not all of the bits in Port C are used for control or status. The remaining bits can be used as follows:

## If Programmed as Inputs-

All input lines can be accessed during a normal Port $C$ read.

## If Programmed as Outputs-

Bits in C upper $\left(\mathrm{PC}_{7}-\mathrm{PC}_{4}\right)$ must be individually accessed using the bit set/reset function.

Bits in $C$ lower ( $\mathrm{PC}_{3}-\mathrm{PC}_{0}$ ) can be accessed using the bit set/reset function or accessed as a threesome by writing into Port C.

## Source Current Capability on Port B and Port C

Any set of eight output buffers, selected randomly from Ports $B$ and $C$ can source 1 mA at 1.5 volts.

This feature allows the 8255 to directly drive Darlington type drivers and high-voltage displays that require such source current.

## Reading Port C Status

In Mode 0, Port C transfers data to or from the peripheral device. When the 8255 is programmed to function in Modes 1 or 2, Port C generates or accepts "hand-shaking" signals with the peripheral device. Reading the contents of Port $C$ allows the programmer to test or verify the "status" of each peripheral device and change the program flow accordingly.

There is no special instruction to read the status information from Port C. A normal read operation of Port $C$ is executed to perform this function.


Figure 17. MODE 1 Status Word Format


Figure 18. MODE 2 Status Word Format

## APPLICATIONS OF THE 8255A

The 8255A is a very powerful tool for interfacing peripheral equipment to the microcomputer system. It represents the optimum use of available pins and is flexible enough to interface almost any $1 / O$ device without the need for additional external logic.

Each peripheral device in a microcomputer system usually has a "service routine" associated with it. The routine manages the software interface between the device and the CPU. The functional definition of the 8255A is programmed by the I/O service routine and becomes an extension of the system software. By examining the I/O devices interface characteristics for both data transfer and timing, and matching this information to the examples and tables in the detailed operational description, a control word can easily be developed to initialize the 8255A to exactly "fit" the application. Figures 19 through 25 represent a few examples of typical applications of the 8255 A .


Figure 19. Printer Interface


Figure 20. Keyboard and Display Interface


Figure 21. Keyboard and Terminal Address Interface


Figure 22. Digital to Analog, Analog to Digital


Figure 23. Basic Floppy Disk Interface


Figure 24. Basic CRT Controller Interface


Figure 25. Machine Tool Controller Interface

## ABSOLUTE MAXIMUM RATINGS*

Ambient Temperature Under Bias ...... $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ Storage Temperature $\qquad$ $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Voltage on Any Pin
with Respect to Ground $\qquad$
Power Dissipation 1 Watt
*Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
D.C. CHARACTERISTICS $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=+5 \mathrm{~V} \pm 10 \%$, $\mathrm{GND}=0 \mathrm{~V}^{*}$

| Symbol | Parameter | Min | Max | Unit | Test Conditions |
| :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{IL}}$ | Input Low Voltage | -0.5 | 0.8 | V |  |
| $\mathrm{~V}_{\mathrm{IH}}$ | Input High Voltage | 2.0 | $\mathrm{~V}_{\mathrm{CC}}$ | V |  |
| $\mathrm{V}_{\mathrm{OL}}$ (DB) | Output Low Voltage (Data Bus) |  | $0.45^{*}$ | V | $\mathrm{I}_{\mathrm{OL}}=2.5 \mathrm{~mA}$ |
| $\mathrm{~V}_{\mathrm{OL}}$ (PER) | Output Low Voltage (Peripheral Port) |  | $0.45^{*}$ | V | $\mathrm{I}_{\mathrm{OL}}=1.7 \mathrm{~mA}$ |
| $\mathrm{~V}_{\mathrm{OH}}$ (DB) | Output High Voltage (Data Bus) | 2.4 |  | V | $\mathrm{I}_{\mathrm{OH}}=-400 \mu \mathrm{~A}$ |
| $\mathrm{~V}_{\mathrm{OH}}(\mathrm{PER})$ | Output High Voltage (Peripheral Port) | 2.4 |  | V | $\mathrm{I}_{\mathrm{OH}}=-200 \mu \mathrm{~A}$ |
| $\mathrm{I}_{\mathrm{DAR}}{ }^{(1)}$ | Darlington Drive Current | -1.0 | -4.0 | mA | $\mathrm{R}_{\mathrm{EXT}}=750 \Omega ; \mathrm{V}_{\mathrm{EXT}}=1.5 \mathrm{~V}$ |
| $\mathrm{I}_{\mathrm{CC}}$ | Power Supply Current |  | 120 | mA |  |
| $\mathrm{I}_{\mathrm{IL}}$ | Input Load Current |  | $\pm 10$ | $\mu \mathrm{~A}$ | $\mathrm{~V}_{\mathrm{IN}}=\mathrm{V}_{\mathrm{CC}}$ to 0 V |
| $\mathrm{I}_{\mathrm{OFL}}$ | Output Float Leakage |  | $\pm 10$ | $\mu \mathrm{~A}$ | $\mathrm{~V}_{\mathrm{OUT}}=\mathrm{V}_{\mathrm{CC}}$ to 0.45 V |

## NOTE:

1. Available on any 8 pins from Port B and C.

CAPACITANCE $T_{A}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=\mathrm{GND}=0 \mathrm{~V}$

| Symbol | Parameter | Min | Typ | Max | Unit | Test Conditions |
| :--- | :--- | :---: | :---: | :---: | :---: | :--- |
| $\mathrm{C}_{\text {IN }}$ | Input Capacitance |  |  | 10 | pF | $\mathrm{f}_{\mathrm{C}}=1 \mathrm{MHz}^{(4)}$ |
| $\mathrm{C}_{\mathrm{I} / \mathrm{O}}$ | I/O Capacitance |  |  | 20 | pF | Unmeasured pins returned to GND(4) |

## A.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=+5 \mathrm{~V} \pm 10 \%, \mathrm{GND}=0 \mathrm{~V}^{*}$

## Bus Parameters

READ

| Symbol | Parameter | 8255A |  | 8255A-5 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |
| $t_{\text {AR }}$ | Address Stable before READ | 0 |  | 0 |  | ns |
| tra | Address Stable after READ | 0 |  | 0 |  | ns |
| $t_{\text {RR }}$ | READ Pulse Width | 300 |  | 300 |  | ns |
| $t_{\text {RD }}$ | Data Valid from READ (1) |  | 250 |  | 200 | ns |
| $\mathrm{t}_{\mathrm{DF}}$ | Data Float after READ | 10 | 150 | 10 | 100 | ns |
| $t_{\text {PV }}$ | Time between READs and/or WRITEs | 850 |  | 850 |  | ns |

WRITE

| Symbol | Parameter | 8255A |  | 8255A-5 |  | Unit |
| :--- | :--- | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |
| $t_{\text {AW }}$ | Address Stable before WRITE | 0 |  | 0 |  | ns |
| $t_{\text {WA }}$ | Address Stable after WRITE | 20 |  | 20 |  | ns |
| $t_{\text {WW }}$ | WRITE Pulse Width | 400 |  | 300 |  | ns |
| $t_{\text {DW }}$ | Data Valid to WRITE (T.E.) | 100 |  | 100 |  | ns |
| $t_{\text {WD }}$ | Data Valid after WRITE | 30 |  | 30 |  | ns |

OTHER TIMINGS

| Symbol | Parameter | 8255A |  | 8255A-5 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |
| twB | WR $=1$ to Output(1) |  | 350 |  | 350 | ns |
| $\mathrm{t}_{1 \mathrm{R}}$ | Peripheral Data before RD | 0 |  | 0 |  | ns |
| $t_{\text {HR }}$ | Peripheral Data after RD | 0 |  | 0 |  | ns |
| $t_{\text {AK }}$ | ACK Pulse Width | 300 |  | 300 |  | ns |
| $\mathrm{t}_{\text {ST }}$ | STB Pulse Width | 500 | $\cdot$ | 500 |  | ns |
| $t_{\text {PS }}$ | Per. Data before T.E. of STB | 0 |  | 0 | $\ldots$ | ns |
| ${ }_{\text {tPH }}$ | Per. Data after T.E. of STB | 180 |  | 180 |  | ns |
| $t_{A D}$ | ACK $=0$ to Output(1) |  | 300 |  | 300 | ns |
| $t_{K D}$ | ACK $=1$ to Output Float | 20 | 250 | 20 | 250 | ns |

## A.C. CHARACTERISTICS (Continued)

OTHER TIMINGS (Continued)

| Symbol | Parameter | 8255A |  | 8255A-5 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |
| twob | $W R=1$ to OBF $=0$ (1) |  | 650 |  | 650 | ns |
| $\mathrm{t}_{\mathrm{AOB}}$ | $A C K=0$ to $O B F=1(1)$ |  | 350 |  | 350 | ns |
| ${ }_{\text {t }}{ }_{\text {IIB }}$ | $S T B=0$ to $1 B F=1(1)$ |  | 300 |  | 300 | ns |
| $\mathrm{t}_{\text {RIB }}$ | $\mathrm{RD}=1$ to $\mathrm{IBF}=0(1)$ |  | 300 |  | 300 | ns |
| trit $^{\text {tit }}$ | RD $=0$ to $\operatorname{NTR}=0(1)$ |  | 400 |  | 400 | ns |
| ${ }_{\text {t }}^{\text {SIT }}$ | STB = 1 to $\operatorname{NTR}=1(1)$ |  | 300 |  | 300 | ns |
| $t_{\text {AIT }}$ | $A C K=1$ to $\operatorname{NTR}=1(1)$ |  | 350 |  | 350 | ns |
| $t_{\text {WIT }}$ | $W R=0$ to $\operatorname{INTR}=0(1,3)$ |  | 850 |  | 850 | ns |

## NOTES:

1. Test Conditions: $C_{L}=150 \mathrm{pF}$.
2. Period of Reset pulse must be at least $50 \mu \mathrm{~s}$ during or after power on. Subsequent Reset pulse can be 500 ns min.
3. INTR $\uparrow$ may occur as early as $\overline{W R} \downarrow$.
4. Sampled, not $100 \%$ tested.
*For Extended Temperature EXPRESS, use M8255A electrical parameters.

## A.C. TESTING INPUT, OUTPUT WAVEFORM

A.C. Testing: Inputs are driven at 2.4 V for a Logic " 1 " and 0.45 V
for a Logic " 0 ". Timing measurements are made at 2.0 V for a
Logic " 1 " and 0.8 V for a Logic " 0 ".
A.C. TESTING LOAD CIRCUIT


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* $V_{E X Y}$ is set at various voltages during testing to guarantee the specification. $C_{L}$ includes jig capacitance.


## WAVEFORMS

MODE 0 (BASIC INPUT)


MODE 0 (BASIC OUTPUT)


WAVEFORMS (Continued)
MODE 1 (STROBED INPUT)


MODE 1 (STROBED OUTPUT)


## WAVEFORMS (Continued)

## MODE 2 (BIDIRECTIONAL)



## WRITE TIMING



READ TIMING


- 8086, 8088 Compatible

■ MCS-80®, MCS-85® Compatible

- Eight-Level Priority Controller
- Expandable to 64 Levels
- Programmable Interrupt Modes
- Individual Request Mask Capability
- Single +5 V Supply (No Clocks)

E Available in 28-Pin DIP and 28-Lead PLCC Package
(See Packaging Spec., Order *231369)

- Available in EXPRESS
- Standard Temperature Range
-Extended Temperature Range

The Intel 8259A Programmable Interrupt Controller handles up to eight vectored priority interrupts for the CPU. It is cascadable for up to 64 vectored priority interrupts without additional circuitry. It is packaged in a 28 -pin DIP, uses NMOS technology and requires a single +5 V supply. Circuitry is static, requiring no clock input.

The 8259A is designed to minimize the software and real time overhead in handling multi-level priority interrupts. It has several modes, permitting optimization for a variety of system requirements.

The 8259A is fully upward compatible with the Intel 8259 . Software originally written for the 8259 will operate the 8259A in all 8259 equivalent modes (MCS-80/85, Non-Buffered, Edge Triggered).


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Figure 1. Block Diagram


PLCC


Figure 2. Pin Configurations

8259A

Table 1. Pin Description

| Symbol | Pln No. | Type | Name and Function |
| :---: | :---: | :---: | :---: |
| $V_{C C}$ | 28 | 1 | SUPPLY: + 5V Supply. |
| GND | 14 | 1 | GROUND |
| $\overline{C S}$ | 1 | 1 | CHIP SELECT: A low on this pin enables $\overline{\mathrm{RD}}$ and $\overline{\mathrm{WR}}$ communication between the CPU and the 8259A. INTA functions are independent of CS. |
| $\overline{W R}$ | 2 | 1 | WRITE: A low on this pin when CS is low enables the 8259A to accept command words from the CPU. |
| $\overline{\mathrm{RD}}$ | 3 | 1 | READ: A low on this pin when CS is low enables the 8259A to release status onto the data bus for the CPU. |
| $\mathrm{D}_{7}-\mathrm{D}_{0}$ | 4-11 | 1/0 | BIDIRECTIONAL DATA BUS: Control, status and interrupt-vector information is transferred via this bus. |
| $\mathrm{CAS}_{0}-\mathrm{CAS}_{2}$ | 12, 13, 15 | $1 / 0$ | CASCADE LINES: The CAS lines form a private 8259A bus to control a multiple 8259A structure. These pins are outputs for a master 8259A and inputs for a slave 8259A. |
| $\overline{\text { SP/EN }}$ | 16 | $1 / 0$ | SLAVE PROGRAM/ENABLE BUFFER: This is a dual function pin. When in the Buffered Mode it can be used as an output to control buffer transceivers (EN). When not in the buffered mode it is used as an input to designate a master $(S P=1)$ or slave ( $\mathrm{SP}=0$ ). |
| INT | 17 | 0 | INTERRUPT: This pin goes high whenever a valid interrupt request is asserted. It is used to interrupt the CPU, thus it is connected to the CPU's interrupt pin. |
| $\mathrm{IR}_{0}-\mathrm{R}_{7}$ | 18-25 | 1 | INTERRUPT REQUESTS: Asynchronous inputs. An interrupt request is executed by raising an IR input (low to high), and holding it high until it is acknowledged (Edge Triggered Mode), or just by a high level on an IR input (Level Triggered Mode). |
| INTA | 26 | 1 | INTERRUPT ACKNOWLEDGE: This pin is used to enable 8259A interrupt-vector data onto the data bus by a sequence of interrupt acknowiedge pulses issued by the CPU. |
| $A_{0}$ | 27 | 1 | AO ADDRESS LINE: This pin acts in conjunction with the $\overline{\mathrm{CS}}, \overline{\mathrm{WR}}$, and $\overline{R D}$ pins. It is used by the 8259A to decipher various Command Words the CPU writes and status the CPU wishes to read. It is typically connected to the CPU AO address line (A1 for 8086, 8088). |

## FUNCTIONAL DESCRIPTION

## Interrupts in Microcomputer Systems

Microcomputer system design requires that 1.0 devices such as keyboards, displays, sensors and other components receive servicing in a an efficient manner so that large amounts of the total system tasks can be assumed by the microcomputer with little or no effect on throughput.

The most common method of servicing such devices is the Polled approach. This is where the processor must test each device in sequence and in effect "ask" each one if it needs servicing. It is easy to see that a large portion of the main program is looping through this continuous polling cycle and that such a method would have a serious detrimental effect on system throughput, thus limiting the tasks that could be assumed by the microcomputer and reducing the cost effectiveness of using such devices.

A more desirable method would be one that would allow the microprocessor to be executing its main program and only stop to service peripheral devices when it is told to do so by the device itself. In effect, the method would provide an external asynchronous input that would inform the processor that it should complete whatever instruction that is currently being executed and fetch a new routine that will service the requesting device. Once this servicing is complete, however, the processor would resume exactly where it left off.

This method is called Interrupt. It is easy to see that system throughput would drastically increase, and thus more tasks could be assumed by the microcomputer to further enhance its cost effectiveness.

The Programmable Interrupt Controller (PIC) functions as an overall manager in an Interrupt-Driven system environment. It accepts requests from the peripheral equipment, determines which of the incoming requests is of the highest importance (priority), ascertains whether the incoming request has a higher priority value than the level currently being serviced, and issues an interrupt to the CPU based on this determination.

Each peripheral device or structure usually has a special program or "routine" that is associated with its specific functional or operational requirements; this is referred to as a "service routine". The PIC, after issuing an Interrupt to the CPU, must somehow input information into the CPU that can "point" the Program Counter to the service routine associated with the requesting device. This "pointer" is an address in a vectoring table and will often be referred to, in this document, as vectoring data.


Figure 3a. Polled Method


Figure 3b. Interrupt Method

The 8259A is a device specifically designed for use in real time, interrupt driven microcomputer systems. It manages eight levels or requests and has built-in features for expandability to other 8259A's (up to 64 levels). It is programmed by the system's software as an 1/O peripheral. A selection of priority modes is available to the programmer so that the manner in which the requests are processed by the 8259A can be configured to match his system requirements. The priority modes can be changed or reconfigured dynamically at any time during the main program. This means that the complete interrupt structure can be defined as required, based on the total system environment.

## INTERRUPT REQUEST REGISTER (IRR) AND IN-SERVICE REGISTER (ISR)

The interrupts at the $\mathbb{R}$ input lines are handled by two registers in cascade, the Interrupt Request Register (IRR) and the In-Service (ISR). The IRR is used to store all the interrupt levels which are requesting service; and the ISR is used to store all the interrupt levels which are being serviced.

## PRIORITY RESOLVER

This logic block determines the priorites of the bits set in the IRR. The highest priority is selected and strobed into the corresponding bit of the ISR during INTA pulse.

## INTERRUPT MASK REGISTER (IMR)

The IMR stores the bits which mask the interrupt lines to be masked. The IMR operates on the IRR. Masking of a higher priority input will not affect the interrupt request lines of lower quality.

## INT (INTERRUPT)

This output goes directly to the CPU interrupt input. The $\mathrm{V}_{\mathrm{OH}}$ level on this line is designed to be fully compatible with the 8080A, 8085A and 8086 input levels.

## INTA (INTERRUPT ACKNOWLEDGE)

$\overline{\text { INTA }}$ pulses will cause the 8259A to release vectoring information onto the data bus. The format of this data depends on the system mode ( $\mu \mathrm{PM}$ ) of the 8259A.

## DATA BUS BUFFER

This 3 -state, bidirectional 8 -bit buffer is used to interface the 8259A to the system Data Bus. Control words and status information are transferred through the Data Bus Buffer.

## READ/WRITE CONTROL LOGIC

The function of this block is to accept OUTput commands from the CPU. It contains the Initialization Command Word (ICW) registers and Operation Command Word (OCW) registers which store the various control formats for device operation. This function block also allows the status of the 8259A to be transferred onto the Data Bus.

## $\overline{C S}$ (CHIP SELECT)

A LOW on this input enables the 8259A. No reading or writing of the chip will occur unless the device is selected.

## $\overline{\text { WR }}$ (WRITE)

A LOW on this input enables the CPU to write control words (ICWs and OCWs) to the 8259A.

## $\overline{R D}$ (READ)

A LOW on this input enables the 8259A to send the status of the Interrupt Request Register (IRR), In Service Register (ISR), the Interrupt Mask Register (IMR), or the Interrupt level onto the Data Bus.

## $A_{0}$

This input signal is used in conjunction with $\overline{W R}$ and $\overline{R D}$ signals to write commands into the various command registers, as well as reading the various status registers of the chip. This line can be tied directly to one of the address lines.
incel


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Figure 4a. 8259A Block Diagram


Figure 4b. 8259A Block Diagram

## THE CASCADE BUFFER/COMPARATOR

This function block stores and compares the IDs of all 8259A's used in the system. The associated three I/O pins (CASO-2) are outputs when the 8259A is used as a master and are inputs when the 8259A is used as a slave. As a master, the 8259A sends the ID of the interrupting slave device onto the CASO-2 lines. The slave thus selected will send its preprogrammed subroutine address onto the Data Bus during the next one or two consecutive INTA pulses. (See section "Cascading the 8259A".)

## INTERRUPT SEQUENCE

The powerful features of the 8259A in a microcomputer system are its programmability and the interrupt routine addressing capability. The latter allows direct or indirect jumping to the specific interrupt routine requested without any polling of the interrupting devices. The normal sequence of events during an interrupt depends on the type of CPU being used.

The events occur as follows in an MCS-80/85 system:

1. One or more of the INTERRUPT REQUEST lines (IR7-0) are raised high, setting the corresponding IRR bit(s).
2. The 8259 A evaluates these requests, and sends an INT to the CPU, if appropriate.
3. The CPU acknowledges the INT and responds with an INTA pulse.
4. Upon receiving an INTA from the CPU group, the highest priority ISR bit is set, and the corresponding IRR bit is reset. The 8259A will also release a CALL instruction code (11001101) onto the 8 -bit Data Bus through its D7-0 pins.
5. This CALL instruction will initiate two more INTA pulses to be sent to the 8259A from the CPU group.
6. These two $\overline{I N T A}$ pulses allow the 8259A to release its preprogrammed subroutine address onto the Data Bus. The lower 8 -bit address is re-
leased at the first $\overline{\mathbb{N T A}}$ pulse and the higher 8 -bit address is released at the second INTA pulse.
7. This completes the 3-byte CALL instruction released by the 8259A. In the AEOI mode the ISR bit is reset at the end of the third INTA pulse. Otherwise, the ISR bit remains set until an appropriate EOI command is issued at the end of the interrupt sequence.

The events occuring in an 8086 system are the same until step 4.
4. Upon receiving an $\overline{N T A}$ from the CPU group, the highest priority ISR bit is set and the corresponding IRR bit is reset. The 8259A does not drive the Data Bus during this cycle.
5. The 8086 will initiate a second INTA pulse. During this pulse, the 8259A releases an 8-bit pointer onto the Data Bus where it is read by the CPU.
6. This completes the interrupt cycle. In the AEOI mode the ISR bit is reset at the end of the second INTA pulse. Otherwise, the ISR bit remains set until an appropriate EOI command is issued at the end of the interrupt subroutine.

If no interrupt request is present at step 4 of either sequence (i.e., the request was too short in duration) the 8259A will issue an interrupt level 7. Both the vectoring bytes and the CAS lines will look like an interrupt level 7 was requested.

When the 8259A PIC receives an interrupt, INT becomes active and an interrupt acknowledge cycle is started. If a higher priority interrupt occurs between the two INTA pulses, the INT line goes inactive immediately after the second INTA pulse. After an unspecified amount of time the INT line is activated again to signify the higher priority interrupt waiting for service. This inactive time is not specified and can vary between parts. The designer should be aware of this consideration when designing a system which uses the 8259A. It is recommended that proper asynchronous design techniques be followed.


Figure 4c. 8259A Block Diagram


Figure 5. 8259A Interface to Standard System Bus

## INTERRUPT SEQUENCE OUTPUTS

## MCS-80 ${ }^{\circledR}$, MCS-85 ${ }^{\circledR}$

This sequence is timed by three INTA pulses. During the first $\overline{\operatorname{NTA}}$ pulse the CALL opcode is enabled onto the data bus.

Content of First Interrupt Vector Byte D7 D6 D5 D4 D3 D2 D1 D0
CALL CODE


During the second $\overline{\mathrm{NT} A}$ pulse the lower address of the appropriate service routine is enabled onto the data bus. When Interval $=4$ bits $A_{5}-A_{7}$ are programmed, while $A_{0}-A_{4}$ are automatically inserted by the 8259A. When interval $=8$ only $A_{6}$ and $A_{7}$ are programmed, while $A_{0}-A_{5}$ are automatically inserted.

Content of Second Interrupt Vector Byte

| IR | Interval $=\mathbf{4}$ |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| 7 | A7 | A6 | A5 | 1 | 1 | 1 | 0 | 0 |
| 6 | A7 | A6 | A5 | 1 | 1 | 0 | 0 | 0 |
| 5 | A7 | A6 | A5 | 1 | 0 | 1 | 0 | 0 |
| 4 | A7 | A6 | A5 | 1 | 0 | 0 | 0 | 0 |
| 3 | A7 | A6 | A5 | 0 | 1 | 1 | 0 | 0 |
| 2 | A7 | A6 | A5 | 0 | 1 | 0 | 0 | 0 |
| 1 | A7 | A6 | A5 | 0 | 0 | 1 | 0 | 0 |
| 0 | A7 | A6 | A5 | 0 | 0 | 0 | 0 | 0 |


| IR | Interval $=\mathbf{8}$ |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| 7 | A7 | A6 | 1 | 1 | 1 | 0 | 0 | 0 |
| 6 | A7 | A6 | 1 | 1 | 0 | 0 | 0 | 0 |
| 5 | A7 | A6 | 1 | 0 | 1 | 0 | 0 | 0 |
| 4 | A7 | A6 | 1 | 0 | 0 | 0 | 0 | 0 |
| 3 | A7 | A6 | 0 | 1 | 1 | 0 | 0 | 0 |
| 2 | A7 | A6 | 0 | 1 | 0 | 0 | 0 | 0 |
| 1 | A7 | A6 | 0 | 0 | 1 | 0 | 0 | 0 |
| 0 | A7 | A6 | 0 | 0 | 0 | 0 | 0 | 0 |

During the third INTA pulse the higher address of the appropriate service routine, which was programmed as byte 2 of the initialization sequence $\left(A_{8}-A_{15}\right)$, is enabled onto the bus.

Content of Third Interrupt Vector Byte

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A15 | A14 | A13 | A12 | A11 | A10 | A9 | A8 |

## 8086, 8088

8086 mode is similar to MCS-80 mode except that only two Interrupt Acknowledge cycles are issued by the processor and no CALL opcode is sent to the processor. The first interrupt acknowledge cycle is similar to that of MCS-80, 85 systems in that the 8259A uses it to internally freeze the state of the interrupts for priority resolution and as a master it issues the interrupt code on the cascade lines at the end of the INTA pulse. On this first cycle it does not issue any data to the processor and leaves its data bus buffers disabled. On the second interrupt acknowledge cycle in 8086 mode the master (or slave if so programmed) will send a byte of data to the processor with the acknowledged interrupt code
composed as follows (note the state of the ADI mode control is ignored and $\mathrm{A}_{5}-\mathrm{A}_{11}$ are unused in 8086 mode):

Content of Interrupt Vector Byte for 8086 System Mode

|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IR7 | T7 | T6 | T5 | T4 | T3 | 1 | 1 | 1 |
| IR6 | T7 | T6 | T5 | T4 | T3 | 1 | 1 | 0 |
| IR5 | T7 | T6 | T5 | T4 | T3 | 1 | 0 | 1 |
| IR4 | T7 | T6 | T5 | T4 | T3 | 1 | 0 | 0 |
| IR3 | T7 | T6 | T5 | T4 | T3 | 0 | 1 | 1 |
| IR2 | T7 | T6 | T5 | T4 | T3 | 0 | 1 | 0 |
| IR1 | T7 | T6 | T5 | T4 | T3 | 0 | 0 | 1 |
| IR0 | T7 | T6 | T5 | T4 | T3 | 0 | 0 | 0 |

## PROGRAMMING THE 8259A

The 8259A accepts two types of command words generated by the CPU:

1. Initialization Command Words (ICWs): Before normal operation can begin, each 8259A in the system must be brought to a starting point-by a sequence of 2 to 4 bytes timed by WR pulses.
2. Operation Command Words (OCWs): These are the command words which command the 8259A to operate in various interrupt modes. These modes are:
a. Fully nested mode
b. Rotating priority mode
c. Special mask mode
d. Polled mode

The OCWs can be written into the 8259A anytime after initialization.

## INITIALIZATION COMMAND WORDS (ICWS)

## General

Whenever a command is issued with $\mathrm{AO}=0$ and D4 $=1$, this is interpreted as Initialization Command Word 1 (ICW1). ICW1 starts the intiitalization sequence during which the following automatically occur.
a. The edge sense circuit is reset, which means that following initialization, an interrupt request (IR) input must make a low-to-high transistion to generate an interrupt.
b. The Interrupt Mask Register is cleared.
c. $\operatorname{IR} 7$ input is assigned priority 7.
d. The slave mode address is set to 7 .
e. Special Mask Mode is cleared and Status Read is set to IRR.
f. If IC4 $=0$, then all functions selected in ICW4 are set to zero. (Non-Buffered mode*, no AutoEOI, MCS-80, 85 system).
*NOTE:
Master/Slave in ICW4 is only used in the buffered mode.

## Initialization Command Words 1 and 2 (ICW1, ICW2)

$\mathrm{A}_{5}-\mathrm{A}_{15}$ : Page starting address of service routines. In an MCS 80/85 system, the 8 request levels will generate CALLs to 8 locations equally spaced in memory. These can be programmed to be spaced at intervals of 4 or 8 memory locations, thus the 8 routines will occupy a page of 32 or 64 bytes, respectively.

The address format is 2 bytes long $\left(A_{0}-A_{15}\right)$. When the routine interval is $4, A_{0}-A_{4}$ are automatically inserted by the 8259A, while $A_{5}-A_{15}$ are programmed externally. When the routine interval is $8, A_{0}-A_{5}$ are automatically inserted by the 8259A, while $A_{6}-A_{15}$ are programmed externally.

The 8 -byte interval will maintain compatibility with current software, while the 4 -byte interval is best for a compact jump table.

In an 8086 system $A_{15}-A_{11}$ are inserted in the five most significant bits of the vectoring byte and the 8259A sets the three least significant bits according to the interrupt level. $A_{10}-A_{5}$ are ignored and ADI (Address interval) has no effect.
LTIM: If LTIM $=1$, then the 8259A will operate in the level interrupt mode. Edge detect logic on the interrupt inputs will be disabled.
ADI: CALL address interval. $A D I=1$ then inter$\mathrm{val}=4 ; A D I=0$ then interval $=8$.
SNGL: Single. Means that this is the only 8259A in the system. If SNGL $=1$ no ICW3 will be issued.
1C4: If this bit is set-ICW4 has to be read. If ICW4 is not needed, set IC4 $=0$.

## Initialization Command Word 3 (ICW3)

This word is read only when there is more than one 8259A in the system and cascading is used, in which
case $\mathrm{SNGL}=0$. It will load the 8 -bit slave register. The functions of this register are:
a. In the master mode (either when $S P=1$, or in buffered mode when $M / S=1$ in ICW4) a " 1 " is set for each slave in the system. The master then will release byte 1 of the call sequence (for MCS80/85 system) and will enable the corresponding slave to release bytes 2 and 3 (for 8086 only byte 2) through the cascade lines.
b. In the slave mode (either when $\overline{\mathrm{SP}}=0$, or if BUF $=1$ and $M / S=0$ in ICW4) bits $2-0$ identify the slave. The slave compares its cascade input with these bits and, if they are equal, bytes 2 and 3 of the call sequence (or just byte 2 for 8086) are released by it on the Data Bus.


Figure 6. Initialization Sequence

Initialization Command Word 4 (ICW4)
SFNM: If $S F N M=1$ the special fully nested mode is programmed.
BUF: If BUF $=1$ the buffered mode is programmed. In buffered mode $\overline{S P} / \overline{E N}$ becomes an enable output and the master/ slave determination is by M/S.
$\mathrm{M} / \mathrm{S}$ : If buffered mode is selected: $\mathrm{M} / \mathrm{S}=1$ means the 8259A is programmed to be a
master, $M / S=0$ means the 8259A is programmed to be a slave. If $B U F=0, M / S$ has no function.
$\mathrm{AEOI}:$ If $\mathrm{AEO}=1$ the automatic end of interrupt mode is programmed.
$\mu$ PM: Microprocessor mode: $\mu \mathrm{PM}=0$ sets the 8259A for MCS-80, 85 system operation, $\mu \mathrm{PM}=1$ sets the 8259 A for 8086 system operation.


231468-10

$A_{15} A_{8}$ OF INTERRUPT VECTOR ADDRESS (MCS80/85 MODE) $\mathrm{T}_{7}-\mathrm{T}_{3}$ OF INTERRUPT VECTOR ADDRESS ( 8086 /8088 MODE)

Figure 7. Initialization Command Word Format

ICW3 (MASTERDEVICE)



231468-13


231468-14

NOTE:
Slave ID is equal to the corresponding master IR input.
Figure 7. Initialization Command Word Format (Continued)

## OPERATION COMMAND WORDS (OCWS)

After the Initialization Command Words (ICWs) are programmed into the 8259A, the chip is ready to accept interrupt requests at its input lines. However, during the 8259A operation, a selection of algorithms can command the 8259A to operate in various modes through the Operation Command Words (OCWs).

Operation Control Words (OCWs) OCW1

| A0 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | M7 | M6 | M5 | M4 | M3 | M2 | M1 | M0 |

OCW2
0

| $R$ | SL | EOI | 0 | 0 | L2 | L1 | LO |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

OCW3
0

OCWI


OCW 2


Figure 8. Operation Command Word Format

## Operation Control Word 1 (OCW1)

OCW1 sets and clears the mask bits in the interrupt Mask Register (IMR). $M_{7}-M_{0}$ represent the eight mask bits. $M=1$ indicates the channel is masked (inhibited), $M=0$ indicates the channel is enabled.

## Operation Control Word 2 (OCW2)

R, SL, EOI-These three bits control the Rotate and End of Interrupt modes and combinations of the two. A chart of these combinations can be found on the Operation Command Word Format.
$L_{2}, L_{1}, L_{0}$ - These bits determine the interrupt level acted upon when the SL bit is active.


Figure 8. Operation Command Word Format (Continued)

## Operation Control Word 3 (OCW3)

ESMM—Enable Special Mask Mode. When this bit is set to 1 it enables the SMM bit to set or reset the Special Mask Mode. When ESMM $=0$ the SMM bit becomes a "don't care".

SMM—Special Mask Mode. If ESMM $=1$ and SMM $=1$ the 8259A will enter Special Mask Mode. If $E S M M=1$ and $S M M=0$ the 8259A will revert to normal mask mode. When ESMM $=0$, SMM has no effect.

## Fully Nested Mode

This mode is entered after initialization unless another mode is programmed. The interrupt requests are ordered in priority from 0 through 7 ( 0 highest). When an interrupt is acknowledged the highest priority request is determined and its vector placed on the bus. Additionally, a bit of the interrupt Service register (ISO-7) is set. This bit remains set until the microprocessor issues an End of Interrupt (EOI) command immediately before returning from the service routine, or if AEOI (Automatic End of Interrupt) bit is set, until the trailing edge of the last INTA. While the IS bit is set, all further interrupts of the same or lower priority are inhibited, while higher levels will generate an interrupt (which will be acknowledged only if the microprocessor internal Interupt enable flip-flop has been re-enabled through software).

After the initialization sequence, IRO has the highest prioirity and IR7 the lowest. Priorities can be changed, as will be explained, in the rotating priority mode.

## End of Interrupt (EOI)

The In Service (IS) bit can be reset either automatically following the trailing edge of the last in sequence INTA pulse (when AEOI bit in ICW1 is set) or by a command word that must be issued to the 8259A before returning from a service routine (EOI command). An EOI command must be issued twice if in the Cascade mode, once for the master and once for the corresponding slave.

There are two forms of EOI command: Specific and Non-Specific. When the 8259A is operated in modes which perserve the fully nested structure, it can determine which IS bit to reset on EOI. When a NonSpecific EOI command is issued the 8259A will automatically reset the highest is bit of those that are set, since in the fully nested mode the highest IS level was necessarily the last level acknowledged and serviced. A non-specific EOI can be issued with OCW2 ( $\mathrm{EOI}=1, \mathrm{SL}=0, \mathrm{R}=0$ ).

When a mode is used which may disturb the fully nested structure, the 8259A may no longer be able to determine the last level acknowledged. In this case a Specific End of Interrupt must be issued which includes as part of the command the IS level to be reset. A specific EOI can be issued with OCW2 ( $\mathrm{EOI}=1, \mathrm{SL}=1, \mathrm{R}=0$, and LO-L2 is the binary level of the IS bit to be reset).

It should be noted that an IS bit that is masked by an IMR bit will not be cleared by a non-specific EOI if the 8259A is in the Special Mask Mode.

## Automatic End of Interrupt (AEOI) Mode

If AEOI $=1$ in ICW4, then the 8259A will operate in AEOI mode continuously until reprogrammed by ICW4. in this mode the 8259A will automatically perform a non-specific EOI operation at the trailing edge of the last interrupt acknowledge pulse (third pulse in MCS-80/85, second in 8086). Note that from a system standpoint, this mode should be used only when a nested multilevel interrupt structure is not required within a single 8259A.

The AEOI mode can only be used in a master 8259A and not a slave. 8259As with a copyright date of 1985 or later will operate in the AEOI mode as a master or a slave.

## Automatic Rotation (Equal Priority Devices)

In some applications there are a number of interrupting devices of equal priority. In this mode a device, after being serviced, receives the lowest priority, so a device requesting an interrupt will have to wait, in the worst case until each of 7 other devices are serviced at most once. For example, if the priority and "in service" status is:

Before Rotate (IR4 the highest prioirity requiring service)


After Rotate (IR4 was serviced, all other priorities rotated correspondingly)


There are two ways to accomplish Automatic Rotation using OCW2, the Rotation on Non-Specific EOI Command $(R=1, S L=0, E O I=1)$ and the Rotate in Automatic EOI Mode which is set by ( $\mathrm{R}=1$. $\mathrm{SL}=0, \mathrm{EOI}=0)$ and cleared by $(\mathrm{R}=0, \mathrm{SL}=0$, $E O I=0$ ).

## Specific Rotation (Specific Priority)

The programmer can change priorities by programming the bottom priority and thus fixing all other priorities; i.e., if IR5 is programmed as the bottom priority device, then IR6 will have the highest one.

The Set Priority command is issued in OCW2 where: $R=1, S L=1, L 0-L 2$ is the binary priority level code of the bottom priority device.

Observe that in this mode internal status is updated by software control during OCW2. However, it is independent of the End of Interrupt (EOI) command (also executed by OCW2). Priority changes can be executed during an EOI command by using the Rotate on Specific EOI command in OCW2 $R=1, S L$ $=1, \mathrm{EOI}=1$ and $\mathrm{LO}-\mathrm{L} 2=\mathbb{R}$ level to receive bottom priority).

## Interrupt Masks

Each !nterrupt Request input can bem masked individually by the Interrupt Mask Register (IMR) programmed through OCW1. Each bit in the IMR masks one interrupt channel if it is set (1). Bit 0 masks IRO, Bit 1 masks IR1 and so forth. Masking an IR channel does not affect the other channels operation.

## Special Mask Mode

Some applications may require an interrupt service routine to dynamically alter the system priority struc-
ture during its execution under software control. For example, the routine may wish to inhibit lower priority requests for a portion of its execution but enable some of them for another portion.

The difficulty here is that if an Interrupt Request is acknowledged and an End of Interrupt command did not reset its IS bit (i.e., while executing a service routine), the 8259A would have inhibited all lower priority requests with no easy way for the routine to enable them.

That is where the Special Mask Mode comes in. In the special Mask Mode, when a mask bit is set in OCW1, it inhibits further interrupts at that level and enables interrupts from all other levels (lower as well as higher) that are not masked.

Thus, any interrupts may be selectively enabled by loading the mask register.

The special Mask Mode is set by OWC3 where: SSMM $=1$, SMM $=1$, and cleared where SSMM $=$ $1, S M M=0$.

## Poll Command

In Poll mode the INT output functions as it normally does. The microprocessor should ignore this output. This can be accomplished either by not connecting the INT output or by masking interrupts within the microprocessor, thereby disabling its interrupt input. Service to devices is achieved by software using a Poll command.

The Poll command is issued by setting $P=1$ ' in OCW3. The 8259A treats the next $\overline{\mathrm{RD}}$ pulse to the 8259A (i.e., $\overline{R D}=0, \overline{C S}=0$ ) as an interrupt acknowledge, sets the appropriate IS bit if there is a request, and reads the priority level. Interrupt is frozen from $\overline{W R}$ to $\overline{R D}$.

The word enabled onto the data bus during $\overline{\mathrm{RD}}$ is:

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | - | - | - | - | W2 | W1 | W0 |

W0-W2: Binary code of the highest priority level requesting service.
I : Equal to " 1 " if there is an interrupt.
This mode is useful if there is a routine command common to several levels so that the $\overline{\text { NTA }}$ sequence is not needed (saves ROM space). Another application is to use the poll mode to expand the number of priority levels to more than 64.

## Reading the 8259A Status

The input status of several internal registers can be read to update the user information on the system.


Figure 9. Priority Cell-Simplified Logic Diagram

The following registers can be read via OCW3 (IRR and ISR or OCW1 [IMR]).

Interrupt Request Register (IRR): 8-bit register which contains the levels requesting an interrupt to be acknowledged. The highest request level is reset from the IRR when an interrupt is acknowledged. (Not affected by IMR.)

In-Service Register (ISR): 8-bit register which contains the priority levels that are being serviced. The ISR is updated when an End of Interrupt Command is issued.

Interrupt Mask Register: 8-bit register which contains the interrupt request lines which are masked.

The IRR can be read when, prior to the RD pulse, a Read Register Command is issued with OCW3 (RR $=1$, RIS $=0$.)

The ISR can be read, when, prior to the RD pulse, a Read Register Command is issued with OCW3 (RR $=1$, RIS $=1$ ).

There is no need to write an OCW3 before every status read operation, as long as the status read corresponds with the previous one; i.e., the 8259A "remembers" whether the IRR or ISR has been previously selected by the OCW3. This is not true when poll is used.

After initialization the 8259A is set to IRR.
For reading the IMR, no OCW3 is needed. The output data bus will contain the IMR whenever $\overline{\mathrm{RD}}$ is active and $A 0=1$ (OCW1).

Polling overrides status read when $P=1, R R=1$ in OCW3.

## Edge and Level Triggered Modes

This mode is programmed using bit 3 in ICW1.
If LTIM = ' 0 ', an interrupt request will be recognized by a low to high transition on an $\mathbb{R}$ input. The $\mathbb{R}$ input can remain high without generating another interrupt.


Figure 10. IR Triggering Timing Requirements

If $L T I M=$ ' 1 ', an interrupt request will be recognized by a 'high' level on IR Input, and there is no need for an edge detection. The interrupt request must be removed before the EOI command is issued or the CPU interrupts is enabled to prevent a second interrupt from occurring.

The priority ceil diagram shows a conceptual circuit of the level sensitive and edge sensitive input circuitry of the 8259A. Be sure to note that the request latch is a transparent $D$ type latch.

In both the edge and level triggered modes the IR inputs must remain high until after the falling edge of the first INTA. If the IR input goes low before this time a DEFAULT IR7 will occur when the CPU acknowledges the interrupt. This can be a useful safeguard for detecting interrupts caused by spurious noise glitches on the IR inputs. To implement this feature the IR7 routine is used for "clean up" simply executing a return instruction, thus ignoring the interrupt. If IR7 is needed for other purposes a default IR7 can still be detected by reading the ISR. A normal IR7 interrupt will set the corresponding ISR bit, a default IR7 won't. If a default IR7 routine occurs during a normal IR7 routine, however, the ISR will remain set. In this case it is necessary to keep track of whether or not the IR7 routine was previously entered. If another IR7 occurs it is a default.

## The Special Fully Nest Mode

This mode will be used in the case of a big system where cascading is used, and the priority has to be conserved within each slave. In this case the fully nested mode will be programmed to the master (us-
ing ICW4). This mode is similar to the normal nested mode with the following exceptions:
a. When an interrupt request from a certain slave is in service this slave is not locked out from the master's priority logic and further interrupt requests from higher priority IR's within the slave will be recognized by the master and will initiate interrupts to the processor. (In the normal nested mode a slave is masked out when its request is in service and no higher requests from the same slave can be serviced.)
b. When exiting the Interrupt Service routine the software has to check whether the interrupt serviced was the only one from that slave. This is done by sending a non-specific End of Interrupt (EOI) command to the slave and then reading its In-Service register and checking for zero. If it is empty, a non-specific EOI can be sent to the master too. If not, no EOI should be sent.

## Buffered Mode

When the 8259A is used in a large system where bus driving buffers are required on the data bus and the cascading mode is used, there exists the problem of enabling buffers.

The buffered mode will structure the 8259A to send an enable signal on $\overline{S P} / \overline{E N}$ to enable the buffers. In this mode, whenever the 8259A's data bus outputs are enabled, the $\overline{\mathrm{SP}} / \overline{\mathrm{EN}}$ output becomes active.

This modification forces the use of software programming to determine whether the 8259A is a master or a slave. Bit 3 in ICW4 programs the buffered mode, and bit 2 in ICW4 determines whether it is a master or a slave.

## CASCADE MODE

The 8259A can be easily interconnected in a system of one master with up to eight slaves to handle up to 64 priority levels.

The master controls the slaves through the 3 line cascade bus. The cascade bus acts like chip selects to the slaves during the INTA sequence.

In a cascade configuration, the slave interrupt outputs are connected to the master interrupt request inputs. When a slave request line is activated and afterwards acknowledged, the master will enable the corresponding slave to release the device routine address during bytes 2 and 3 of $\overline{\mathrm{NTA}}$. (Byte 2 only for $8086 / 8088$ ).

The cascade bus lines are normally low and will contain the slave address code from the trailing edge of the first INTA pulse to the trailing edge of the third pulse. Each 8259A in the system must follow a separate initialization sequence and can be programmed to work in a different mode. An EOI command must be issued twice: once for the master and once for the corresponding slave. An address decoder is required to activate the Chip Select (CS) input of each 8259A.

The cascade lines of the Master 8259A are activated only for slave inputs, non-slave inputs leave the cascade line inactive (low).


Figure 11. Cascading the 8259A

## ABSOLUTE MAXIMUM RATINGS*

Ambient Temperature Under Bias ...... $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$
Storage Temperature ........... $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Voltage on Any Pin
with Respect to Ground. ......... -0.5 V to +7 V
Power Dissipation 1W
*Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
D.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{V}_{C C}=5 \mathrm{~V} \pm 10 \%$

| Symbol | Parameter | Min | Max | Units | Test Conditions |
| :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{IL}}$ | Input Low Voltage | -0.5 | 0.8 | V |  |
| $\mathrm{~V}_{\mathrm{IH}}$ | Input High Voltage | $2.0^{*}$ | $\mathrm{~V}_{\mathrm{CC}}+0.5 \mathrm{~V}$ | V |  |
| $\mathrm{~V}_{\mathrm{OL}}$ | Output Low Voltage |  | 0.45 | V | $\mathrm{I}_{\mathrm{OL}}=2.2 \mathrm{~mA}$ |
| $\mathrm{~V}_{\mathrm{OH}}$ | Output High Voltage | 2.4 |  | V | $\mathrm{I}_{\mathrm{OH}}=-400 \mu \mathrm{~A}$ |
| $\mathrm{~V}_{\mathrm{OH}(\mathrm{INT})}$ | Interrupt Output High <br> Voltage | 3.5 |  | V | $\mathrm{I}_{\mathrm{OH}}=-100 \mu \mathrm{~A}$ |
|  | 2.4 |  | V | $\mathrm{I}_{\mathrm{OH}}=-400 \mu \mathrm{~A}$ |  |
| $\mathrm{I}_{\mathrm{LI}}$ | Input Load Current | -10 | +10 | $\mu \mathrm{~A}$ | $0 \mathrm{~V} \leq \mathrm{V}_{I N} \leq \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{I}_{\text {LOL }}$ | Output Leakage Current | -10 | +10 | $\mu \mathrm{~A}$ | $0.45 \mathrm{~V} \leq \mathrm{V}_{\mathrm{OUT}} \leq \mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{I}_{\mathrm{CC}}$ | $\mathrm{V}_{\mathrm{CC}}$ Supply Current |  | 85 | mA |  |
| $\mathrm{I}_{\mathrm{LIR}}$ | IR Input Load Current |  | -300 | $\mu \mathrm{~A}$ | $\mathrm{~V}_{\mathrm{IN}}=0$ |

*NOTE:
For Extended Temperature EXPRESS $\mathrm{V}_{\mathbb{H}}=2.3 \mathrm{~V}$.

CAPACITANCE $T_{A}=25^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{CC}}=\mathrm{GND}=0 \mathrm{~V}$

| Symbol | Parameter | Min | Typ | Max | Unit | Test Conditions |
| :--- | :--- | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{C}_{\mathbb{N}}$ | Input Capacitance |  |  | 10 | pF | $\mathrm{fc}=1 \mathrm{MHz}$ |
| $\mathrm{C}_{\mathrm{I} / \mathrm{O}}$ | I/O Capacitance |  |  | 20 | pF | Unmeasured Pins Returned to $V_{S S}$ |

## A.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V} \pm 10 \%$

TIMING REQUIREMENTS

| Symbol | Parameter | 8259A |  | 8259A-2 |  | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |  |
| TAHRL | AO/ $\overline{\mathrm{CS}}$ Setup to $\overline{\mathrm{RD}} / \overline{\mathrm{NTA}} \downarrow$ | 0 |  | 0 |  | ns |  |
| TRHAX | AO/ $\overline{\mathrm{CS}}$ Hold after $\overline{\mathrm{RD}} / \overline{\mathrm{NTA}} \uparrow$ | 0 |  | 0 |  | ns |  |
| TRLRH | $\overline{\mathrm{RD}}$ Pulse Width | 235 |  | 160 |  | ns |  |
| TAHWL | AO/ $\overline{\mathrm{CS}}$ Setup to $\overline{W R} \downarrow$ | 0 |  | 0 |  | ns |  |
| TWHAX | AO/ $\overline{C S}$ Hold after $\overline{W R} \uparrow$ | 0 |  | 0 |  | ns |  |
| TWLWH | $\overline{\text { WR Pulse Width }}$ | 290 |  | 190 |  | ns |  |
| TDVWH | Data Setup to WR $\uparrow$ | 240 |  | 160 |  | ns |  |
| TWHDX | Data Hold after WR $\uparrow$ | 0 |  | 0 |  | ns |  |
| TJLJH | Interrupt Request Width (Low) | 100 |  | 100 |  | ns | See Note 1 |
| TCVIAL | Cascade Setup to Second or Third $\overline{\text { INTA }} \downarrow$ (Slave Only) | 55 |  | 40 |  | ns |  |
| TRHRL | End of $\overline{R D}$ to Next $\overline{R D}$ End of INTA to Next INTA within an INTA Sequence Only | 160 |  | 100 |  | ns |  |
| TWHWL | End of WR to Next $\overline{W R}$ | 190 |  | 100 |  | ns |  |
| *TCHCL | End of Command to Next Command (Not Same Command Type) | 500 500 |  | 150 |  | ns |  |
|  | End of INTA Sequence to Next INTA Sequence. | 500 |  | 300 |  |  |  |

-Worst case timing for TCHCL in an actual microprocessor system is typically much greater than 500 ns (i.e. $8085 \mathrm{~A}=$ $1.6 \mu \mathrm{~s}, 8085 \mathrm{~A}-2=1 \mu \mathrm{~s}, 8086=1 \mu \mathrm{~s}, 8086-2=625 \mathrm{~ns})$

NOTE:
This is the low time required to clear the input latch in the edge triggered mode.

## TIMING RESPONSES

| Symbol | Parameter | 8259A |  | 8259A-2 |  | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |  |
| TRLDV | Data Valid from $\overline{\mathrm{RD}} / \overline{\mathrm{NTA}} \downarrow$ |  | 200 |  | 120 | ns | $\begin{aligned} & \text { C of Data Bus }= \\ & 100 \mathrm{pF} \\ & \text { C of Data Bus } \\ & \text { Max Test } \mathrm{C}=100 \mathrm{pF} \\ & \text { Min Test } \mathrm{C}=15 \mathrm{pF} \\ & \mathrm{C}_{\text {INT }}=100 \mathrm{pF} \\ & \mathrm{C}_{\text {CASCADE }}=100 \mathrm{pF} \end{aligned}$ |
| TRHDZ | Data Float after $\overline{\mathrm{RD}} / \overline{\mathrm{NTA}} \uparrow$ | 10 | 100 | 10 | 85 | ns |  |
| TJHIH | Interrupt Output Delay |  | 350 |  | 300 | ns |  |
| TIALCV | Cascade Valid from First INTA $\downarrow$ (Master Only) |  | 565 |  | 360 | ns |  |
| TRLEL | Enable Active from $\overline{R D} \downarrow$ or INTA $\downarrow$ |  | 125 |  | 100 | ns |  |
| TRHEH | Enable Inactive from $\overline{\mathrm{RD}} \uparrow$ or $\overline{\mathrm{INTA}} \uparrow$ |  | 150 |  | 150 | ns |  |
| TAHDV | Data Valid from Stable Address |  | 200 |  | 200 | ns |  |
| TCVDV | Cascade Valid to Valid Data |  | 300 |  | 200 | ns |  |

## A.C. TESTING INPUT/OUTPUT WAVEFORM


A.C. Testing: inputs are driven at 2.4 V for a logic " 1 " and 0.45 V for a logic " 0 '. Timing measurements are made at 2.0 V for a logic " 1 " and 0.8 V for a logic " 0 ".

## A.C. TESTING LOAD CIRCUIT



231468-26
$C_{L}=100 \mathrm{pF}$
$\mathrm{C}_{\mathrm{L}}$ Includes Jig Capacitance

## WAVEFORMS

## WRITE



## WAVEFORMS (Continued)

READ/INTA


OTHER TIMING


WAVEFORMS (Continued)
INTA SEQUENCE


## NOTES:

Interrupt output must remain HIGH at least until leading edge of first INTA.

1. Cycle 1 in 8086,8088 systems, the Data Bus is not active.

## Data Sheet Revision Review

The following changes have been made since revision 2 of the 8259A data sheet.

1. The first paragraph of the Poll Command section was rewritten to clarify the status of the INT pin.
2. A paragraph was added to the Interrupt Sequence section to indicate the status of the INT pin during multiple interrupts.
3. A reference to PLCC packaging was added.
4. All references to the 8259A-8 have been deleted.

## inter

SINGLE/DOUBLE DENSITY FLOPPY DISK CONTROLLER

- IBM Compatible in Both Single and Double Density Recording Formats
- Programmable Data Record Lengths: 128, 256, 512, or 1024 Bytes/Sector
- Multi-Sector and Multi-Track Transfer Capability
- Drives Up to 4 Floppy or Mini-Floppy Disks
- Controls 8 " , $51 / 4^{\prime \prime}$ and $31 / 2^{\prime \prime}$ Floppy Disk Drives
- Data Transfers in DMA or Non-DMA Mode
- Paralle! Seek Operations on Up to Four Drives
- Compatible with all Intel and Most Other Microprocessors
- Single-Phase 8 MHz Clock
- Single +5V Power Supply ( $\pm 10 \%$ )
- Plastic $\mathbf{4 0}$ Pin DIP or $\mathbf{4 0}$ Pin CERDIP Packages

The 8272A is an LSI Floppy Disk Controller (FDC) Chip, which contains the circuitry and control functions for interfacing a processor to 4 Floppy Disk Drives. It is capable of supporting either IBM 3740 single density format (FM), or IBM System 34 Double Density format (MFM) including double sided recording. The 8272A provides control signals which simplify the design of an external phase locked loop and write precompensation circuitry. The FDC simplifies and handies most of the burdens associated with implementing a Floppy Disk Drive Interface. The 8272A is a pin-compatible upgrade to the 8272.


210606-1

Figure 1. 8272A Internal Block Dlagram


Figure 2. Pin Configuration

Table 1. Pin Description

| Symbol | Pin No. | Type | Connection To | Name and Function |
| :---: | :---: | :---: | :---: | :---: |
| RESET | 1 | 1 | $\mu \mathrm{P}$ | RESET: Places FDC in idle state. Resets output lines to FDD to " 0 " (low). Does not clear the last specify command. |
| $\overline{\text { RD }}$ | 2 | (1) | $\mu \mathrm{P}$ | READ: Control signal for transfer of data from FDC to Data Bus, when "0" (low). |
| $\overline{\text { WR }}$ | 3 | (1) | $\mu \mathrm{P}$ | WRITE: Control signal for transfer of data to FDC via Data Bus, when "0" (low). |
| CS | 4 | 1 | $\mu \mathrm{P}$ | CHIP SELECT: IC selected when "0" (low) allowing FD and WR to be enabled. |
| $A_{0}$ | 5 | (1) | $\mu \mathrm{P}$ | DATA/STATUS REGISTER SELECT: Selects Data Reg ( $\mathrm{A}_{0}=1$ ) or Status Reg $\left(A_{0}=0\right)$ contents to be sent to Data Bus. |
| $\mathrm{DB}_{0}-\mathrm{DB}_{7}$ | 6-13 | 1/0(1) | $\mu \mathrm{P}$ | DATA BUS: Bidirectional 8-Bit Data Bus. |
| DRQ | 14 | 0 | DMA | DATA DMA REQUEST: DMA Request is being made by FDC when DRQ "1".(3) |
| DACK | 15 | 1 | DMA | DMA ACKNOWLEDGE: DMA cycle is active when " 0 " (low) and Controller is performing DMA transfer. |
| TC | 16 | 1 | DMA | TERMINAL COUNT: Indicates the termination of a DMA transfer when " 1 " (high)(2). |
| IDX | 17 | 1 | FDD | INDEX: Indicates the beginning of a disk track. |
| INT | 18 | 0 | $\mu \mathrm{P}$ | INTERRUPT: Interrupt Request Generated by FDC. |
| CLK | 19 | 1 |  | CLOCK: Single Phase 8 MHz ( 4 MHz for mini floppies) Squarewave Clock. |
| GND | 20 |  |  | GROUND: D.C. Power Return. |
| $\mathrm{V}_{\text {CC }}$ | 40 |  |  | D.C. POWER: + 5V |
| RW/SEEK | 39 | 0 | FDD | READ WRITE/SEEK: When "1" (high) Seek mode selected and when " 0 " (low) Read/Write mode selected. |
| LCT/DIR | 38 | 0 | FDD | LOW CURRENT/DIRECTION: Lowers Write current on inner tracks in Read/Write mode, determines direction head will step in Seek mode. |
| FR/STP | 37 | 0 | FDD | FAULT RESET/STEP: Resets fault FF in FDD in Read/Write mode, provides step pulses to move head to another cylinder in Seek mode. |
| HDL | 36 | 0 | FDD | HEAD LOAD: Command which causes Read/Write head in FDD to contact diskette. |
| RDY | 35 | 1 | FDD | READY: Indicates FDD is ready to send or receive data. Must be tied high (gated by the index pulse) for mini floppies which do not normally have a Ready line. |
| WP/TS | 34 | 1 | FDD | WRITE PROTECT/TWO-SIDE: Senses Write Protect status in Read/ Write mode, and Two Side Media in Seek mode. |
| FLT/TRKO | 33 | 1 | FDD | FAULT/TRACK 0: Senses FDD fault condition in Read/Write mode and Track 0 condition in Seek mode. |
| $\mathrm{PS}_{1}, \mathrm{PS}_{0}$ | 31, 32 | 0 | FDD | PRECOMPENSATION (PRE-SHIFT): Write precompensation status during MFM mode. Determines early, late, and normal times. |
| WR DATA | 30 | 0 | FDD | WRITE DATA: Serial clock and data bits to FDD. |
| $\underline{D S}{ }_{1}, \mathrm{DS}_{0}$ | 28,29 | 0 | FDD | DRIVE SELECT: Selects FDD unit. |

Table 1. Pin Description (Continued)

| Symbol | Pln <br> No. | Type | Connec- <br> tion To | Name and Function |
| :--- | :---: | :---: | :---: | :--- |
| HDSEL | 27 | 0 | FDD | HEAD SELECT: Head 1 selected when "1" (high) Head 0 selected <br> when " 0 " (low). |
| MFM | 26 | 0 | PLL | MFM MODE: MFM mode when "1," FM mode when "0". |
| WE | 25 | 0 | FDD | WRITE ENABLE: Enables write data into FDD. |
| VCO | 24 | 0 | PLL | VCO SYNC: Inhibits VCO in PLL when " "" (low), enables VCO when <br> "1." |
| RD DATA | 23 | 1 | FDD | READ DATA: Read data from FDD, containing clock and data bits. |
| DW | 22 | 1 | PLL | DATA WINDOW: Generated by PLL, and used to sample data from <br> FDD. |
| WR CLK | 21 | 1 |  | WRITE CLOCK: Write data rate to FDD FM = 500 kHz, MFM $=1$ <br> MHz, with a pulse width of 250 ns for both FM and MFM. <br> Must be enabled for all operations, both Read and Write. |

## NOTES:

1. Disabled when $\overline{C S}=1$.
2. TC must be activated to terminate the Execution Phase of any command.
3. DRO is also an inpur for certain test modes. It should have a $5 \mathrm{k} \Omega$ pull-up resistor to prevent activation.


Figure 3. 8272A System Block Diagram

## DESCRIPTION

Hand-shaking signals are provided in the 8272A which make DMA operation easy to incorporate with the aid of an external DMA Controller chip, such as the 8237A. The FDC will operate in either DMA or Non-DMA mode. In the Non-DMA mode, the FDC generates intermpts to the processor for every transfer of a data byte between the CPU and the 8272A. In the DMA mode, the processor need only
load a command into the FDC and all data transfers occur under control of the 8272A and DMA controller.

There are 15 separate commands which the 8272A will execute. Each of these commands require multiple 8-bit bytes to fully specify the operation which the processor wishes the FDC to perform. The following commands are available.

Read Data
Read ID
Read Deleted Data
Read a Track
Scan Equal
Scan High or Equal
Scan Low or Equal Specity

Write Data
Format a Track
Write Deleted Data
Seek
Recalibrate (Restore to Track 0)
Sense Interrupt Status
Sense Drive Status

For more information see the Intel Application Notes AP-116 and AP-121.

## FEATURES

Address mark detection circuitry is internal to the FDC which simplifies the phase locked loop and read electronics. The track stepping rate, head load time, and head unload time may be programmed by the user. The 8272A offers many additional features such as multiple sector transfers in both read and write modes with a single command, and full IBM compatibility in both single (FM) and double density (MFM) modes.

## 8272A ENHANCEMENTS

On the 8272A, after detecting the Index Pulse, the VCO Sync output stays low for a shorter period of time. See Figure 4a.

On the 8272 there can be a problem reading data when Gap 4A is 00 and there is no IAM. This occurs on some older floppy formats. The 8272A cures this problem by adjusting the VCO Sync timing so that it is not low during the data field. See Figure 4b.


Figure 4. 8272A Enhancements over the 8272

## 8272A REGISTERS-CPU INTERFACE

The 8272A contains two registers which may be accessed by the main system processor; a Status Register and a Data Register. The 8 -bit Main Status Register contains the status information of the FDC, and may be accessed at any time. The 8-bit Data Register (actually consists of several registers in a stack with only one register presented to the data bus at a time), stores data, commands, parameters, and FDD status information. Data bytes are read out of, or written into, the Data Register in order to program or obtain the results after execution of a command. The Status Register may only be read and is used to facilitate the transfer of data between the processor and 8272A.

The relationship between the Status/Down registers and the signals $\overline{R D}, \overline{W R}$, and $A_{0}$ is shown in Table 2.

Table 2. AO, $\overline{R D}, \overline{W R}$ Decoding for the Selection of Status/Data Register Functions.

| $A_{0}$ | $\overline{R D}$ | $\overline{W R}$ | Function |
| :---: | :---: | :---: | :--- |
| 0 | 0 | 1 | Read Main Status Register |
| 0 | 1 | 0 | Illegal(1) |
| 0 | 0 | 0 | Illegal(1) |
| 1 | 0 | 0 | Illegal(1) |
| 1 | 0 | 1 | Read from Data Register |
| 1 | 1 | 0 | Write into Data Register |

## NOTE:

1. Design must guarantee that the 8272A is not subjected to illegal inputs.

The Main Status Register bits are defined in Table 3.

Table 3. Main Status Regiater Bit Description

| Bit Number | Name | Symbol | Description |
| :---: | :--- | :---: | :--- |
| $D_{0}$ | FDD 0 Busy | $D_{0} B$ | FDD number 0 is in the Seek mode. |
| $D_{1}$ | FDD 1 Busy | $D_{1} B$ | FDD number 1 is in the Seek mode. |
| $D_{2}$ | FDD 2 Busy | $D_{2} B$ | FDD number 2 is in the Seek mode. |
| $D_{3}$ | FDD 3 Busy | $\mathrm{D}_{3} B$ | FDD number 3 is in the Seek mode. |
| $D_{4}$ | FDC Busy | CB | A read or write command is in process. |
| $D_{5}$ | Non-DMA Mode | NDM | The FDC is in the non-DMA mode. This bit is set only <br> during the execution phase in non-DMA mode. <br> Transition to " 0 " state indicates execution phase has <br> ended. |
| $D_{6}$ | Data Input/Output | DIO | Indicates direction of data transfer between FDC and <br> Data Register. If DIO = " $1 "$ then transfer is from Data <br> Register to the Processor. If DIO $=$ " 0 ", then transfer <br> is from the Processor to Data Register. |
| $D_{7}$ | Request for Master | RQM | Indicates Data Register is ready to send or receive <br> data to or from the Processor. Both bits DIO and RQM <br> should be used to perform" the handshaking functions <br> of "ready" and "direction" to the processor. |

The DIO and RQM bits in the Status Register indicate when Data is ready and in which direction data will be transferred on the Data Bus.

## NOTE:

There is a $12 \mu \mathrm{~s}$ or $\mathbf{2 4} \mu \mathrm{s}$ RQM flag delay when using an 8 or 4 MHz clock respectively.


210008-6

## MOTES:

A-Data register ready to be written into by processor
B-Data register not ready to be written into by processor
C-Data register ready for next data byte to be read by the processor
D-Data register not ready for next data byte to be read by processor
Figure 5. Status Register Timing

The 8272A is capable of executing 15 different commands. Each command is initiated by a multi-byte transfer from the processor, and the result after execution of the command may also be a multi-byte transfer back to the processor. Because of this mul-ti-byte interchange of information between the 8272A and the processor, it is convenient to consider each command as consisting of three phases:
Command Phase: The FDC receives all information required to perform a particular operation from the processor.
Execution Phase: The FDC performs the operation it was instructed to do.
Result Phase: After completion of the operation, status and other housekeeping information are made available to the processor.

During Command or Result Phases the Main Status Register (described in Table 3) must be read by the processor before each byte of information is written into or read from the Data Register. Bits D6 and D7 in the Main Status Register must be in a 0 and 1 state, respectively, before each byte of the command word may be written into the 8272A. Many of the commands require multiple bytes, and as a result the Main Status Register must be read prior to each byte transfer to the 8272A. On the other hand, during the Result Phase, D6 and D7 in the Main Status Register must both be 1's (D6 = 1 and D7 = 1) before reading each byte from the Data Register.

## NOTE:

This reading of the Main Status Register before each byte transfer to the 8272A is required in only the Command and Result Phases, and NOT during the Execution Phase.

During the Execution Phase, the Main Status Register need not be read. If the 8272A is in the non-DMA Mode, then the receipt of each data byte (if 8272A is reading data from FDD) is indicated by an interrupt signal on pin 18 (INT $=1$ ). The generation of a Read signal ( $\mathrm{RD}=0$ ) will reset the interrupt as well as output the Data onto the Data Bus. For example, if the processor cannot handle interrupts fast enough (every $13 \mu \mathrm{~s}$ for MFM mode) then it may poll the Main Status Register and then bit D7 (ROM) functions just like the Interrupt signal. If a Write Command is in process, then the WR signal performs the reset to the interrupt signal.

The 8272A always operates in a multi-sector transfer mode. It continues to transfer data until the TC input is active. In Non-DMA Mode, the system must supply the TC input.

If the 8272A is in the DMA Mode, no Interrupts are generated during the Execution Phase. The 8272A generates DRQ's (DMA Requests) when each byte of data is available. The DMA Controller responds to this request with both a $\overline{D A C K}=0$ (DMA Acknowledge) and a $\overline{\text { RD }}=0$ (Read signal). When the DMA Acknowledge signal goes low ( $\overline{D A C K}=0$ ) then the DMA Request is reset ( $\mathrm{DRQ}=0$ ). If a Write Command has been programmed then a $\overline{W A}$ signal will appear instead of $\bar{R} \bar{D}$. After the Execution Phase has been completed (Terminal Count has occurred) then an Interrupt will occur (INT $=1$ ). This signifies the beginning of the Result Phase. When the first byte of data is read during the Result Phase, the Interrupt is automatically reset ( $\operatorname{INT}=0$ ).

It is important to note that during the Result Phase all bytes shown in the Command Table must be read. The Read Data Command, for example has seven bytes of data in the Result Phase. All seven bytes must be read in order to successfully complete the Read Data Command. The 8272A will not accept a new command until all seven bytes have been read. Other commands may require fewer bytes to be read during the Result Phase.

The 8272A contains five Status Registers. The Main Status Register mentioned above may be read by the processor at any time. The other four Status Registers (STO, ST1, ST2, and ST3) are only available during the Result Phase, and may be read only after successfully completing a command. The particular command which has been executed determines how many of the Status Registers will be read.

The bytes of data which are sent to the 8272A to form the Command Phase, and are read out of the 8272A in the Result Phase, must occur in the order shown in the Table 4. That is, the Command Code must be sent first and the other bytes sent in the prescribed sequence. No foreshortening of the Command or Result Phases are allowed. After the last byte of data in the Command Phase is sent to the 8272A, the Execution Phase automatically starts. In a similar fashion, when the last byte of

Table 4. 8272A Command Set


8272A

Table 4. 8272A Command Set (Continued)


8272A

Table 4. 8272A Command Set (Continued)


8272A

Table 4. 8272A Command Set (Continued)


Table 4. 8272A Command Set (Continued)


Table 4. 8272A Command Set (Continued)


Table 4. 8272A Command Set (Continued)

\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[t]{2}{*}{Phase} \& \multirow{2}{*}{R/W} \& \multicolumn{8}{|c|}{Data Bus} \& \multirow{2}{*}{Remarks} \\
\hline \& \& D7 \& \(\mathrm{D}_{6}\) \& \(\mathrm{D}_{5}\) \& \(\mathrm{D}_{4}\) \& \(\mathrm{D}_{3}\) \& \(\mathrm{D}_{2}\) \& \(\mathrm{D}_{1}\) \& \(\mathrm{D}_{0}\) \& \\
\hline \multicolumn{11}{|l|}{SPECIFY} \\
\hline Command \& w
w
w \& 0
-
- \& 0
SRT
HLT \& 0 \& \(\longrightarrow\) \& \& 0 \& \(\xrightarrow{\text { HUT }}\) \& \begin{tabular}{c}
1 \\
\hline ND
\end{tabular} \& Command Codes \\
\hline \multicolumn{11}{|l|}{SENSE DRIVE STATUS} \\
\hline \begin{tabular}{l}
Command \\
Result
\end{tabular} \& \[
\begin{aligned}
\& \mathbf{W} \\
\& \mathbf{W} \\
\& R
\end{aligned}
\] \& \[
\begin{aligned}
\& 0 \\
\& 0
\end{aligned}
\] \& \[
\begin{aligned}
\& 0 \\
\& 0
\end{aligned}
\] \& \[
\begin{aligned}
\& 0 \\
\& 0
\end{aligned}
\] \& \[
\begin{aligned}
\& 0 \\
\& 0 \\
\& \hline
\end{aligned}
\] \& \[
\begin{aligned}
\& 0 \\
\& 0
\end{aligned}
\] \& \[
\begin{gathered}
1 \\
\text { HDS } \\
\hline
\end{gathered}
\] \& \[
\begin{gathered}
0 \\
\text { DS1 } \\
\hline
\end{gathered}
\] \& \[
\begin{gathered}
0 \\
\text { DSO }
\end{gathered}
\] \& \begin{tabular}{l}
Command Codes \\
Status Information about FDD
\end{tabular} \\
\hline \multicolumn{11}{|l|}{SEEK} \\
\hline \begin{tabular}{l}
Command \\
Execution
\end{tabular} \& \[
\begin{aligned}
\& \mathbf{w} \\
\& \mathbf{w} \\
\& \mathbf{w}
\end{aligned}
\] \& \[
\begin{aligned}
\& 0 \\
\& 0
\end{aligned}
\] \& \[
\begin{aligned}
\& 0 \\
\& 0
\end{aligned}
\] \& \[
\begin{aligned}
\& 0 \\
\& 0
\end{aligned}
\] \& \[
\begin{aligned}
\& 0 \\
\& 0 \\
\&
\end{aligned}
\] \& \[
\begin{aligned}
\& 1 \\
\& 0
\end{aligned}
\] \& \[
\begin{gathered}
1 \\
\text { HDS }
\end{gathered}
\] \& \[
\begin{gathered}
1 \\
\mathrm{DS} 1
\end{gathered}
\] \& \[
\begin{gathered}
1 \\
\text { DSO }
\end{gathered}
\] \& \begin{tabular}{l}
Command Codes \\
Head is Positioned Over Proper Cylinder on Diskette
\end{tabular} \\
\hline \multicolumn{11}{|l|}{INVALID} \\
\hline \begin{tabular}{l}
Command \\
Result
\end{tabular} \& W
R \& \& \& \& Invalio

S \& des \& \&  \& - \& | Invalid Command |
| :--- |
| Codes (NoOp-FDC |
| Goes Into Standby |
| State) $\text { ST } 0=80$ |
| (16) | <br>

\hline
\end{tabular}

8272A

Table 5. Command Mnemonics

| Symbol | Name | Description |
| :---: | :---: | :---: |
| $A_{0}$ | Address Line 0 | $A_{0}$ controls selection of Main Status Register ( $A_{0}=0$ ) or Data Register ( $A_{0}=1$ ). |
| C | Cylinder Number | C stands for the current selected Cylinder track number 0 through 76 of the medium. |
| D | Data | D stands for the data pattern which is going to be written into a Sector. |
| $D_{7}-D_{0}$ | Data Bus | 8 -bit Data bus where $D_{7}$ is the most significant bit, and $D_{0}$ is the least significant bit. |
| DS0, DS1 | Drive Select | DS stands for a selected drive number 0 or 1 . |
| DTL | Data Length | When N is defined as 00 , DTL stands for the data length which users are going to read out or write into the Sector. |
| EOT | End of Track | EOT stands for the final Sector number of a Cylinder. |
| GPL | Gap Length | GPL stands for the length of Gap 3 (spacing between Sectors excluding VCO Sync Field). |
| H | Head Address | H stands for head number 0 or 1, as specified in ID field. |
| HDS | Head Select | HDS stands for a selected head number 0 or $1(H=H D S$ in all command words). |
| HLT | Head Load Time | HLT stands for the head load time in the FDD (2 to 254 ms in 2 ms increments). |
| HUT | Head Unload Time | HUT stands for the head unload time after a read or write operation has occurred ( 16 to 240 ms in 16 ms increments). |
| MFM | FM or MFM Mode | If MF is low, FM mode is selected and if it is high, MFM mode is selected. |
| MT | Multi-Track | If MT is high, a multi-track operation is to be peformed (a cylinder under both HD0 and HD1 will be read or written). |
| N | Number | N stands for the number of data bytes written in a Sector. |
| NCN | New Cylinder Number | NCN stands for a new Cylinder number, which is going to be reached as a result of the Seek operation. Desired position of Head. |
| ND | Non-DMA Mode | ND stands for operation in the Non-DMA Mode. |
| PCN | Prosent Cylinder Number | PCN stands for the Cylinder number at the completion of SENSE INTERRUPT STATUS Command. Position of Head at present time. |
| R | Record | R stands for the Sector number, which will be read or written. |
| R/W | Read/Write | R/W stands for either Read (R) or Write (W) signal. |
| SC | Sector | SC indicates the number of Sectors per Cylinder. |
| SK | Skip | SK stands for Skip Deleted Data Address Mark. |
| SRT | Step Rate Time | SRT stands for the Stepping Rate for the FDD ( 1 to 16 ms in 1 ms increments). The same Stepping Rate applies to all drives ( $F=1 \mathrm{~ms}$, $E=2 \mathrm{~ms}$, etc.). |

Table 5. Command Mnemonics (Continued)

| Symbol | Name | Description |
| :---: | :---: | :---: |
| ST 0 <br> ST 1 <br> ST 2 <br> ST 3 | Status 0 <br> Status 1 <br> Status 2 <br> Status 3 | ST 0-3 stand for one of four registers which store the status information after a command has been executed. This information is available during the result phase after command execution. These registers should not be confused with the main status register (selected by $A_{0}=0$ ). ST $0-3$ may be read only after a command has been executed and contain information relevant to that particular command. |
| STP |  | During a Scan operation, if STP $=1$, the data in contiguous sectors is compared byte by byte with data sent from the processor (or DMA), and if STP $=2$, then alternate sectors are read and compared. |

data is read out in the Result Phase, the command is automatically ended and the 8272A is ready for a new command. A command may be aborted by simply sending a Terminal Count signal to pin 16 ( $\mathrm{TC}=1$ ). This is a convenient means of ensuring that the processor may always get the 8272A's attention even if the disk system hangs up in an abnormal manner.

## POLLING FEATURE OF THE 8272A

After power-up RESET, the Drive Select Lines DSO and DS1 will automatically go into a polling mode. In between commands (and between step pulses in the SEEK command) the 8272A polls all four FDDs looking for a change in the Ready line from any of the drives. If the Ready line changes state (usually due to a door opening or closing) then the 8272A will generate an interrupt. When Status Register 0 (STO) is read (after Sense Interrupt Status is issued), Not Ready (NR) will be indicated. The polling of the Ready line by the 8272A occurs continuously between instructions, thus notifying the processor which drives are on or off line. Approximate scan timing is shown in Table 6.

Table 6. Scan Timing

| DS1 | DS0 | Approximate Scan Timing |
| :---: | :---: | :---: |
| 0 | 0 | $220 \mu \mathrm{~s}$ |
| 0 | 1 | $220 \mu \mathrm{~s}$ |
| 1 | 0 | $220 \mu \mathrm{~s}$ |
| 1 | 1 | $440 \mu \mathrm{~s}$ |

## COMMAND DESCRIPTIONS

During the Command Phase, the Main Status Register must be polled by the CPU before each byte is
written into the Data Register. The DIO (DB6) and RQM (BD7) bits in the Main Status Register must be in the " 0 " and " 1 " states respectively, before each byte of the command may be written into the 8272A. The beginning of the execution phase for any of these commands will cause DIO and ROM to switch to " 1 " and " 0 " states respectively.

## READ DATA

A set of nine (9) byte words are required to place the FDC into the Read Data Mode. After the Read Data command has been issued the FDC loads the head (if it is in the unloaded state), waits the specified head setting time (defined in the Specify Command), and begins reading ID Address Marks and ID fields. When the current sector number ("R") stored in the ID Register (IDR) compares with the sector number read off the diskette, then the FDC outputs data (from the data field) byte-by-byte to the main system via the data bus.

After completion of the read operation from the current sector, the Sector Number is incremented by one, and the data from the next sector is read and output on the data bus. This continuous function is called a "Multi-Sector Read Operation." The Read Data Command must be terminated by the receipt of a Terminal Count signal. Upon receipt of this signal, the FDC stops outputting data to the processor, but will continue to read data from the current sector, check CRC (Cyclic Redundancy Count) bytes, and then at the end of the sector terminate the Read Data Command.

The amount of data which can be handled with a single command to the FDC depends upon MT (mul-ti-track), MFM (MFM/FM), and N (Number of Bytes/ Sector). Table 7 on the next page shows the Transfor Capacity.

Table 7. Transfer Capacity


The "multi-track" function (MT) allows the FDC to read data from both sides of the diskette. For a particular cylinder, data will be transferred starting at Sector 1, Side 0 and completing at Sector L, Side 1 (Sector $\mathrm{L}=$ last sector on the side).

## NOTE:

This function pertains to only one cylinder (the same track) on each side of the diskette.

When $\mathbf{N}=\mathbf{0}$, then DTL defines the data length which the FDC must treat as a sector. If DTL is smaller than the actual data length in a Sector, the data beyond DTL in the Sector is not sent to the Data Bus. The FDC reads (internally) the complete Sector performing the CRC check, and depending upon the manner of command termination, may perform a Multi-Sector Read Operation. When N is nonzero, then OTL has no meaning and should be set to OFFH.

At the completion of the Read Data Command, the head is not unloaded until after Head Unload Time Interval (specified in the Specify Command) has elapsed. If the processor issues another command before the head unioads then the head setting time may be saved between subsequent reads. This time out is particularly valuable when a diskette is copied from one drive to another.

If the FDC detects the Index Hole twice without finding the right sector, (indicated in "R"), then the FDC sets the ND (No Data) flag in Status Register 1 to a 1 (high), and terminates the Read Data Command. (Status Register 0 also has bits 7 and 6 set to 0 and 1 respectively.)

After reading the ID and Data Fields in each sector, the FDC checks the CRC bytes. If a read error is detected (incorrect CRC in ID field), the FDC sets the DE (Data Error) flag in Status Register 1 to a 1 (high), and if a CRC error occurs in the Data Field the FDC also sets the DD (Data Error in Data Field) flag in Status Register 2 to a 1 (high), and terminates the Read Data Command. (Status Register 0 also has bits 7 and 6 set to 0 and 1 respectively.)

If the FDC reads a Deleted Data Address Mark off the diskette, and the SK bit (bit D5 in the first Command Word) is not set $(S K=0)$, then the FDC sets the CM (Control Mark) flag in Status Register 2 to a 1 (high), and terminates the Read Data Command, after reading all the data in the Sector. If SK $=1$, the FDC skips the sector with the Deleted Data Address Mark and reads the next sector.

During disk data transfers between the FDC and the processor, via the data bus, the FDC must be serviced by the processor every $27 \mu$ s in the FM Mode, and every $13 \mu \mathrm{~s}$ in the MFM Mode, or the FDC sets the OR (Over Run) flag in Status Register 1 to a 1 (high), and terminates the Read Data Command.

If the processor terminates a read (or write) operation in the FDC, then the ID Information in the Result Phase is dependent upon the state of the MT bit and EOT byte. Table 5 shows the values for C, H, R, and N , when the processor terminates the Command.

Table 8. ID Information When Processor Terminates Command

| MT | EOT | Final Sector Transferred to Proceseor | ID Information at Result Phase |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | C | H | R | $N$ |
| 0 | 1A <br> OF $08$ | Sector 1 to 25 at Side 0 <br> Sector 1 to 14 at Side 0 <br> Sector 1 to 7 at Side 0 | NC | NC | $R+1$ | NC |
|  | 1A <br> OF <br> 08 | Sector 26 at Side 0 Sector 15 at Side 0 Sector 8 at Side 0 | $C+1$ | NC | $R=01$ | NC |
|  | 1A <br> OF $08$ | $\begin{aligned} & \text { Sector } 1 \text { to } 25 \text { at Side } 1 \\ & \text { Sector } 1 \text { to } 14 \text { at Side } 1 \\ & \text { Sector } 1 \text { to } 7 \text { at Side } 1 \end{aligned}$ | NC | NC | $R+1$ | NC |
|  | 1A OF <br> 08 | Sector 26 at Side 1 <br> Sector 15 at Side 1 <br> Sector 8 at Side 1 | $C+1$ | NC | $\mathrm{R}=01$ | NC |
| 1 | 1A OF | $\begin{aligned} & \text { Sector } 1 \text { to } 25 \text { at Side } 0 \\ & \text { Sector } 1 \text { to } 14 \text { at Side } 0 \\ & \text { Sector } 1 \text { to } 7 \text { at Side } 0 \end{aligned}$ | NC | NC | $R+1$ | NC |
|  | 1A <br> OF <br> 08 | Sector 26 at Side 0 <br> Sector 15 at Side 0 <br> Sector 8 at Side 0 | NC | LSB | $R=01$ | NC |
|  | 1A <br> OF <br> 08 | $\begin{aligned} & \text { Sector } 1 \text { to } 25 \text { at Side } 1 \\ & \text { Sector } 1 \text { to } 14 \text { at Side } 1 \\ & \text { Sector } 1 \text { to } 7 \text { at Side } 1 \end{aligned}$ | NC | NC | $R+1$ | NC |
|  | $1 A$ OF $08$ | Sector 26 at Side 1 <br> Sector 15 at Side 1 <br> Sector 8 at Side 1 | C +1 | LSB | $R=01$ | NC |

NOTES:

1. NC (No Change): The same value as the one at the beginning of command execution.
2. LSB (Least Significant Bit): The least significant bit of H is complemented.

## WRITE DATA

A set of nine (9) bytes are required to set the FDC into the Write Data mode. After the Write Data command has been issued the FDC loads the head (if it is in the unloaded state), waits the specified head settling time (defined in the Specity Command), and begins reading ID Fields. When the current sector number ("R"), stored in the ID Register (IDR) compares with the sector number read off the diskette, then the FDC takes data from the processor byte-by-byte via the data bus, and outputs it to the FDD.

After writing data into the current sector, the Sector Number stored in "R" is incremented by one, and the next data field is written into. The FDC continues this "Multi-Sector Write Operation" until the issu-
ance of a Terminal Count signal. If a Terminal Count signal is sent to the FDC it continues writing into the current sector to complete the data field. If the Terminal Count signal is received while a data field is being written then the remainder of the data field is filled with 00 (zeros).

The FDC reads the ID field of each sector and checks the CRC bytes. If the FDC detects a read error (incorrect CRC) in one of the ID Fields, it sets the DE (Data Error) flag of Status Register 1 to a 1 (high), and terminates the Write Data Command. (Status Register 0 also has bits 7 and 6 set to 0 and 1 respectively.)

The Write Command operates in much the same manner as the Read Command. The following items

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are the same; refer to the Read Data Command for details:

- Transfer Capacity
- EN (End of Cylinder) Flag
- ND (No Data) Flag
- Head Unload Time Interval
- ID Information when the processor terminates command (see Table 2)
- Definition of DTL when $\mathbf{N}=0$ and when $\mathbf{N} \neq 0$

In the Write Data mode, data transfers between the processor and FDC must occur every $31 \mu$ s in the FM mode, and every $15 \mu \mathrm{~s}$ in the MFM mode. If the time interval between data transfers is longer than this then the FDC sets the OR (Over Run) flag in Status Register 1 to a 1 (high), and terminates the Write Data Command.

For mini-floppies, multiple track writes are usually not permitted. This is because of the turn-off time of the erase head coils-the head switches tracks before the erase head turns off. Therefore the system should typically wait 1.3 ms before attempting to step or change sides.

## WRITE DELETED DATA

This command is the same as the Write Data Command except a Deleted Data Address Mark is written at the beginning of the Data Field instead of the normal Data Address Mark.

## READ DELETED DATA

This command is the same as the Read Data Command except that when the FDC detects a Data Address Mark at the beginning of a Data Field (and SK $=\mathbf{0}$ (low)), it will read all the data in the sector and set the CM flag in Status Register 2 to a 1 (high), and then terminate the command. If $\mathrm{SK}=1$, then the FDC skips the sector with the Data Address Mark and reads the next sector.

## READ A TRACK

This command is similar to READ DATA Command except that the entire data field is read continuously from each of the sectors of a track. Immediately after encountering the INDEX HOLE, the FDC starts reading all data fields on the track as continuous blocks of data. If the FDC finds an error in the ID or DATA CRC check bytes, it continues to read data from the track. The FDC compares the ID information read from each sector with the value stored in the IDR, and sets the ND flag of Status Register 1 to
a 1 (high) if there is no comparison. Multi-track or skip operations are not allowed with this command.

This command terminates when EOT number of sectors have been read. If the FDC does not find an ID Address Mark on the diskette after it encounters the INDEX HOLE for the second time, then it sets the MA (missing address mark) flag in Status Register 1 to a 1 (high), and terminates the command. (Status Register 0 has bits 7 and 6 set to 0 and 1 respectively.)

## READ ID

The READ ID Command is used to give the present position of the recording head. The FDC stores the values from the first ID Field it is able to read. If no proper ID Address Mark is found on the diskette, before the INDEX HOLE is encountered for the second time then the MA (Missing Address Mark) fiag in Status Register 1 is set to a 1 (high), and if no data is found then the ND (No Data) flag is also set in Status Register 1 to a 1 (high) and the command is terminated.

## FORMAT A TRACK

The Format Command allows an entire track to be formatted. After the INDEX HOLE is detected, Data is written on the Diskette: Gaps, Address Marks, ID Fields and Data Fields, all per the IBM System 34 (Double Density) or System 3740 (Single Density) Format are recorded. The particular format which will be written is controlled by the values programmed into N (number of bytes/sector), SC (sectors/cylinder), GPL (Gap Length), and D (Data Pattern) which are supplied by the processor during the Command Phase. The Data Field is filled with the Byte of data stored in D. The ID Field for each sector is supplied by the processor; that is, four data requests per sector are made by the FDC for C (Cylinder Number), H(Head Number), R(Sector Number) and $N$ (Number of Bytes/Sector). This allows the diskette to be formatted with nonsequential sector numbers, if desired.

After formatting each sector, the processor must send new values for C, H, R, and N to the 8272A for each sector on the track. The contents of the R Register is incremented by one after each sector is formatted, thus, the R register contains a value of R + I when it is read during the Result Phase. This incrementing and formatting continues for the whole track until the FDC encounters the INDEX HOLE for the second time, whereupon it terminates the command.

If a FAULT signal is received from the FDD at the end of a write operation, then the FDC sets the EC flag of Status Register 0 to a 1 (high), and terminates the command after setting bits 7 and 6 of Status Register 0 to 0 and 1 respectively. Also the loss of a READY signal at the beginning of a com-
mand execution phase causes command termination.

Table 9 shows the relationship between N, SC, and GPL for various sector sizes:

Table 9. Sector Size Relationships

| Format | Eytes/ Sector | $8{ }^{\text {8 }}$ Floppy |  |  |  | Bytes/ Sector | 51/4" Fioppy |  |  |  | Bytes/ <br> Sector | 31/2" Mind Floppy |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | N | SC | OPL(1) | GPL(2) |  | N | SC | GPL (1) | GPL(2) |  | N | Sc | GPL(1) | GPL(2) |
| FM Mode | 128 | 00 | 1A | 07 | 1 B | 128 | 00 | 12 | 07 | 09 | 128 | 0 | OF | 07 | 18 |
|  | 256 | 01 | OF | OE | 2A | 128 | 00 | 10 | 10 | 19 | - | - | - | - | - |
|  | 512 | 02 | 08 | 18 | 3A | 256 | 01 | 08 | 18 | 30 | 256 | 1 | 09 | OF | 2 A |
|  | 1024 | 03 | 04 | 47 | 8A | 512 | 02 | 04 | 46 | 87 | 512 | 2 | 05 | 18 | 3 A |
|  | 2048 | 04 | 02 | C8 | FF | 1024 | 03 | 02 | C8 | FF | - | - | - | - | - |
|  | 4096 | 05 | 01 | C8 | FF | 2048 | 04 | 01 | C8 | FF | - |  |  | - |  |
| MPM Mode | 256 | 01 | 1A | OE | 36 | 256 | 01 | 12 | OA | OC | 258 | 1 | OF | CE | 36 |
|  | 512 | 02 | OF | 18 | 54 | 256 | 01 | 10 | 20 | 32 | - | - | - | - | - |
|  | 1024 | 03 | 08 | 35 | 74 | 512 | 02 | 08 | 2A | 50 | 512 | 2 | 09 | 18 | 54 |
|  | 2048 | 04 | 04 | 99 | FF | 1024 | 03 | 04 | 80 | F0 | 1024 | 3 | 05 | 35 | 74 |
|  | 4096 | 05 | 02 | C8 | FF | 2048 | 04 | 02 | C8 | FF | - | - | - | - | - |
|  | 8192 | 08 | 01 | C8 | FF | 4096 | 05 | 01 | C8 | FF |  |  |  |  |  |

NOTES:

1. Suggested values of GPL in Read or Write Commands to avoid splice point between data field and ID field of contiguous sections.
2. Suggested values of GPL in format command.

## SCAN COMMANDS

The SCAN Commands allow data which is being read from the diskette to be compared against data which is being supplied from the main system (Processor in NON-DMA mode, and DMA Controller in DMA mode). The FDC compares the data on a byte-by-byte basis, and looks for a sector of data which meets the conditions of $D_{\text {FDD }}=D_{\text {Processor, }} D_{\text {FDD }} \leq$ $D_{\text {Processor, }}$ or $D_{F D D} \geq D_{\text {Processor }}$. Ones complement arithmetic is used for comparison (FF = largest number, $00=$ smallest number). After a whole
sector of data is compared, if the conditions are not met, the sector number is incremented (R + STP $\rightarrow$ R), and the scan operation is continued. The scan operation continues until one of the following conditions occur, the conditions for scan are met (equal, low, or high), the last sector on the track is reached (EOT), or the terminal count signal is received.

If the conditions for scan are met then the FDC sets the SH (Scan Hit) flag of Status Register 2 to a 1 (high), and terminates the Scan Command. If the

Table 10. Scan Status Codes

| Command | Status Regiater 2 |  | Comments |
| :---: | :---: | :---: | :---: |
|  | Bit 2 $=\mathbf{S N}$ | Bit 3 = SH |  |
| Scan Equal | 0 | 1 | $D_{\text {FDD }}=D_{\text {Processor }}$ |
|  | 1 | 0 | $D_{\text {FDD }} \neq D_{\text {Processor }}$ |
|  | 0 | 1 | $D_{\text {FDD }}=D_{\text {Processor }}$ |
|  | 0 | 0 | $D_{\text {FDD }}<D_{\text {Processor }}$ |
|  | 1 | 0 | $D_{\text {FDD }}>D_{\text {Processor }}$ |
| Scan High or Equal | 0 | 1 | $D_{\text {FDD }}=D_{\text {Processor }}$ |
|  | 0 | 0 | $D_{\text {FDD }}>D_{\text {Processor }}$ |
|  | 1 | 0 | $D_{\text {FDD }} \& D_{\text {Processor }}$ |

conditions for scan are not met between the starting sector (as specified by R) and the last sector on the cylinder (EOT), then the FDC sets the SN (Scan Not Satisfied) flag of Status Register 2 to a 1 (high), and terminates the Scan Command. The receipt of a TERMINAL COUNT signal from the Processor or DMA Controller during the scan operation will cause the FDC to complete the comparison of the particular byte which is in process, and then to terminate the command. Table 10 shows the status of bits SH and SN under various conditions of SCAN.

If the FDC encounters a Deleted Data Address Mark on one of the sectors (and SK $=0$ ), then it regards the sector as the last sector on the cylinder, sets CM (Control Mark) flag of Status Register 2 to a 1 (high) and terminates the command. If $S K=1$, the FDC skips the sector with the Deleted Address Mark, and reads the next sector. In the second case (SK = 1), the FDC sets the CM (Control Mark) flag of Status Register 2 to a 1 (high) in order to show that a Deleted Sector had been encountered.

When either the STP (contiguous sectors STP $=01$, or alternate sectors STP = 02 sectors are read) or the MT (Multi-Track) are programmed, it is necessary to remember that the last sector on the track must be read. For example, if $S T P=02, M T=0$, the sectors are numbered sequentially 1 through 26 , and we start the Scan Command at sector 21; the following will happen. Sectors 21, 23, and 25 will be read, then the next sector (26) will be skipped and the Index Hole will be encountered before the EOT value of 26 can be read. This will result in an abnormal termination of the command. If the EOT had been set at 25 or the scanning started at sector 20 , then the Scan Command would be completed in a normal manner.

During the Scan Command data is supplied by either the processor or DMA Controller for comparison against the data read from the diskette. In order to avoid having the OR (Over Run) flag set in Status Register 1, it is necessary to have the data available in less than $27 \mu s$ (FM Mode) or $13 \mu s$ (MFM Mode). If an Overrun occurs the FDC terminates the command.

## SEEK

The read/write within the FDD is moved from cylinder to cylinder under control of the Seek Command. The FDC compares the PCN (Present Cylinder Number) which is the current head position with the NCN
(New Cylinder Number), and performs the following operation if there is a difference:

PCN < NCN: Direction signal to FDD set to a 1 (high), and Step Pulses are issued. (Step In.)

PCN > NCN: Direction signal to FDD set to a 0 (low), and Step Pulses are issued. (Step Out.)

The rate at which Step Pulses are issued is controlled by SRT (Stepping Rate Time) in the SPECIFY Command. After each Step Pulse is issued NCN is compared against PCN, and when NCN = PCN, then the SE (Seek End) flag is set in Status Register 0 to a 1 (high), and the command is terminated.

During the Command Phase of the Seek operation the FDC is in the FDC BUSY state, but during the Execution Phase it is in the NON BUSY state. While the FDC is in the NON BUSY state, another Seek Command may be issued, and in this manner parallel seek operations may be done on up to 4 Drives at once.

If an FDD is in a NOT READY state at the beginning of the command execution phase or during the seek operation, then the NR (NOT READY) flag is set in Status Register 0 to a 1 (high), and the command is terminated.

Note that the 8272A Read and Write Commands do not have implied Seeks. Any R/W command should be preceded by: 1) Seek Command; 2) Sense Interrupt Status; and 3) Read ID.

## RECALIBRATE

This command causes the read/write head within the FDD to retract to the Track 0 position. The FDC clears the contents of the PCN counter, and checks the status of the Track 0 signal from the FDD. As long as the Track 0 signal is low, the Direction signal remains 1 (high) and Step Pulses are issued. When the Track 0 signal goes high, the SE (SEEK END) flag in Status Register 0 is set to a 1 (high) and the command is terminated. If the Track 0 signal is still low after 77 Step Pulses have been issued, the FDC sets the SE (SEEK END) and EC (EQUIPMENT CHECK) flags of Status Register 0 to both 1 s (highs), and terminates the command.

The ability to overlap RECALIBRATE Commands to multiple FDDs, and the loss of the READY signal, as described in the SEEK Command, also applies to the RECALIBRATE Command.

## SENSE INTERRUPT STATUS

An interrupt signal is generated by the FDC for one of the following reasons:

1) Upon entering the Result Phase of:
a) Read Data Command
b) Read a Track Command
c) Read ID Command
d) Read Deleted Data Command
e) Write Data Command
f) Format a Cylinder Command
g) Write Deleted Data Command
h) Scan Commands
2) Ready Line of FDD changes state
3) End of Seek or Recalibrate Command
4) During Execution Phase in the NON-DMA Mode

Interrupts caused by reasons 1 and 4 above occur during normal command operations and are easily discernible by the processor. However, interrupts caused by reasons 2 and 3 above may be uniquely identified with the aid of the Sense Interrupt Status Command. This command when issued resets the interrupt signal and via bits 5, 6, and 7 of Status Register 0 identifies the cause of the interrupt.

Neither the Seek or Recalibrate Command have a Result Phase. Therefore, it is mandatory to use the Sense Interrupt Status Command after these commands to effectively terminate them and to provide verification of the head position (PCN).

Table 11. Seek, Interrupt Codes

| Seek End <br> Bit 5 | Interrupt Code |  | Cause |
| :---: | :---: | :---: | :--- |
|  | Bit 6 | Bit 7 |  |
| 1 | 0 | 1 | Ready Line Changed <br> State, Either Polarity |
| 1 | 1 | 0 | Normal Termination <br> of Seek or Recalibrate <br> Command |
|  | 0 | Abnormal Termination <br> of Seek or Recalibrate <br> Command |  |

## SPECIFY

The Specify Command sets the intitial values for each of the three internal timers. The HUT (Head Unload Time) defines the time from the end of the Execution Phase of one of the Read/Write Com-
mands to the head unload state. This timer is programmable from 16 to 240 ms in increments of $16 \mathrm{~ms}(01=16 \mathrm{~ms}, 02=32 \mathrm{~ms} \ldots \mathrm{OF}=$ 240 ms ). The SRT (Step Rate Time) defines the time interval between adjacent step pulses. This timer is programmable from 1 to 16 ms in increments of 1 ms ( $F=1 \mathrm{~ms}, E=2 \mathrm{~ms}, \mathrm{D}=3 \mathrm{~ms}$, etc.). The HLT (Head Load Time) defines the time between when the Head Load signal goes high and when the Read/Write operation starts. This timer is programmable from 2 to 254 ms in increments of $2 \mathrm{~ms}(01=$ $2 \mathrm{~ms}, 02=4 \mathrm{~ms}, 03=6 \mathrm{~ms} \ldots . . \mathrm{FE}=254 \mathrm{~ms}$ ).

The step rate should be programmed 1 ms longer than the minimum time required by the drive.

The time intervals mentioned above are a direct function of the clock (CLK on pin 19). Times indicated above are for an 8 MHz clock, if the clock was reduced to 4 MHz (mini-floppy application) then all time intervals are increased by a factor of 2.

The choice of DMA or NON-DMA operation is made by the ND (NON-DMA) bit. When this bit is high ( $\mathrm{ND}=1$ ) the NON-DMA mode is selected, and when ND $=0$ the DMA mode is selected.

## SENSE DRIVE STATUS

This command may be used by the processor whenever it wishes to obtain the status of the FDDs. Status Register 3 contains the Drive Status information.

## INVALID

If an invalid command is sent to the FDC (a command not defined above), then the FDC will terminate the command. No interrupt is generated by the 8272A during this condition. Bit 6 and bit 7 (DIO and RQM) in the Main Status Register are both high ("1") indicating to the processor that the 8272A is in the Result Phase and the contents of Status Register 0 (STO) must be read. When the processor reads Status Register 0 it will find an $\mathbf{8 0 H}$ indicating an invalid command was received.

A Sense Interrupt Status Command must be sent after a Seek or Recalibrate interrupt, otherwise the FDC will consider the next command to be an Invalid Command.

In some applications the user may wish to use this command as a No-Op command, to place the FDC in a standby or no operation state.

Table 12. Status Registers

| Bit |  |  | Description |
| :---: | :---: | :---: | :---: |
| No. | Name | Symbol |  |
| STATUS REGISTER 0 |  |  |  |
| $\mathrm{D}_{7}$ | Interrupt Code | IC | $D_{7}=0 \text { and } D_{6}=0$ <br> Normal Termination of Command, (NT). Command was completed and properly executed. |
| $\mathrm{D}_{6}$ |  |  | $D_{7}=0 \text { and } D_{6}=1$ <br> Abnormal Termination of Command, (AT). Execution of Command was started, but was not successfully completed. |
|  |  |  | $D_{7}=1 \text { and } D_{6}=0$ <br> Invalid Command issue, (IC). Command which was issued was never started. |
|  |  |  | $D_{7}=1 \text { and } D_{6}=1$ <br> Abnormal Termination because during command execution the ready signal from FDD changed state. |
| $\mathrm{D}_{5}$ | Seek End | SE | When the FDC completes the SEEK Command, this flag is set to 1 (high). |
| $\mathrm{D}_{4}$ | Equipment Check | EC | If a fault Signal is received from the FDD, or if the Track 0 Signal fails to occur after 77 Step Pulses (Recalibrate Command) then this flag is set. |
| $\mathrm{D}_{3}$ | Not Ready | NR | When the FDD is in the not-ready state and a read or write command is issued, this flag is set. If a read or write command is issued to Side 1 of a single sided drive, then this flag is set. |
| $\mathrm{D}_{2}$ | Head Address | HD | This flag is used to indicate the state of the head at Interrupt. |
| $\mathrm{D}_{1}$ | Unit Select 1 | US 1 | These flags are used to indicate a Drive Unit Number at Interrupt. |
| $\mathrm{D}_{0}$ | Unit Select 0 | USO |  |
| STATUS REGISTER 1 |  |  |  |
| $\mathrm{D}_{7}$ | End of Cylinder | EN | When the FDC tries to access a Sector beyond the final Sector of a Cylinder, this flag is set. |
| $\mathrm{D}_{6}$ |  |  | Not used. This bit is always 0 (low). |
| $\mathrm{D}_{5}$ | Data Error | DE | When the FDC detects a CRC error in either the ID field or the data field, this flag is set. |
| $\mathrm{D}_{4}$ | Over Run | OR | If the FDC is not serviced by the main-systems during data transfers, within a certain time interval, this flag is set. |
| $\mathrm{D}_{3}$ |  |  | Not used. This bit always 0 (low). |
| $\mathrm{D}_{2}$ | No Data | ND | During execution of READ DATA, WRITE DELETED DATA or SCAN Command, if the FDC cannot find the Sector specified in the IDR Register, this flag is set. |
|  |  |  | During executing the READ ID Command, if the FDC cannot read the ID field without an error, then this flag is set. |
|  |  |  | During the execution of the READ A Cylinder Command, if the starting sector cannot be found, then this flag is set. |

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Table 12. Status Register (Continued)

| Bit |  |  | Description |
| :---: | :---: | :---: | :---: |
| No. | Name | Symbol |  |
| STATUS REGISTER 1 (Continued) |  |  |  |
| $\mathrm{D}_{1}$ | Not Writable | NW | During execution of WRITE DATA, WRITE DELETED DATA or Format A Cylinder Command, if the FDC detects a write protect signal from the FDD, then this flag is set. |
| $\mathrm{D}_{0}$ | Missing Address Mark | MA | If the FDC cannot detect the ID Address Mark after encountering the index hole twice, then this flag is set. |
|  |  |  | If the FDC cannot detect the Data Address Mark or Deleted Data Address Mark, this flag is set. Also at the same time, the MD (Missing Address Mark in Data Field) of Status Register 2 is set. |
| STATUS REGISTER 2 |  |  |  |
| $\mathrm{D}_{7}$ |  |  | Not used. This bit is always 0 (low). |
| $\mathrm{D}_{6}$ | Control Mark | CM | During executing the READ DATA or SCAN Command, if the FDC encounters a Sector which contains a Deleted Data Address Mark, this flag is set. |
| $\mathrm{D}_{5}$ | Data Error in Data Field | DD | If the FDC detects a CRC error in the data field then this flag is set. |
| $\mathrm{D}_{4}$ | Wrong Cylinder | wc | This bit is related with the ND bit, and when the contents of C on the medium is different from that stored in the IDR, this flag is set. |
| $\mathrm{D}_{3}$ | Scan Equal Hit | SH | During execution, the SCAN Command, if the condition of "equal" is satisfied, this flag is set. |
| $\mathrm{D}_{2}$ | Scan Not Satisfied | SN | During executing the SCAN Command, if the FDC cannot find a Sector on the cylinder which meets the condition, then this flag is set. |
| $\mathrm{D}_{1}$ | Bad Cylinder | BC | This bit is related with the ND bit, and when the content of C on the medium is different from that stored in the IDR and the content of C is $F F$, then this flag is set. |
| $\mathrm{D}_{0}$ | Missing <br> Address <br> Mark in Data <br> Field | MD | When data is read from the medium, if the FDC cannot find a Data Address Mark or Deleted Data Address Mark, then this flag is set. |
| STATUS REGISTER 3 |  |  |  |
| $\mathrm{D}_{7}$ | Fault | FT | This bit is used to indicate the status of the Fault signal from the FDD. |
| $\mathrm{D}_{6}$ | Write Protected | WP | This bit is used to indicate the status of the Write Protected signal from the FDD. |
| $\mathrm{D}_{5}$ | Ready | RDY | This bit is used to indicate the status of the Ready signal from the FDD. |
| $\mathrm{D}_{4}$ | Track 0 | T0 | This bit is used to indicate the status of the Track 0 signal from the FDD. |
| $\mathrm{D}_{3}$ | Two Side | TS | This bit is used to indicate the status of the Two Side signal from the FDD. |
| $\mathrm{D}_{2}$ | Head Address | HD | This bit is used to indicate the status of Side Select signal to the FDD. |
| $\mathrm{D}_{1}$ | Unit Select 1 | US 1 | This bit is used to indicate the status of the Unit Select 1 signal to the FDD. |
| $\mathrm{D}_{0}$ | Unit Select 0 | uso | This bit is used to indicate the status of the Unit Select 0 signal to the FDD. |

8272A

## ABSOLUTE MAXIMUM RATINGS*

Operating Temperature . . . . . . . . . . . . . $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$
Storage Temperature ........... $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$

Supply Voltage VCC ................... . . 0.5 to $+7 V$
Power Dissipation 1 Watt
*Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
${ }^{*} \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$
D.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=+5 \mathrm{~V} \pm 10 \%$

| Symbol | Parameter | Limits |  | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max |  |  |
| $\mathrm{V}_{\text {IL }}$ | Input Low Voltage | -0.5 | 0.8 | V |  |
| $V_{1 H}$ | Input High Voltage | 2.0 | $\mathrm{V}_{\text {CC }}+0.5$ | V |  |
| $\mathrm{V}_{\mathrm{OL}}$ | Output Low Voltage |  | 0.45 | V | $\mathrm{loL}=2.0 \mathrm{~mA}$ |
| $\mathrm{V}_{\mathrm{OH}}$ | Output High Voltage | 2.4 | $\mathrm{V}_{\text {cc }}$ | V | $\mathrm{lOH}=-400 \mu \mathrm{~A}$ |
| lac | $V_{\text {cc }}$ Supply Current |  | 120 | mA |  |
| IIL | Input Load Current (All Input Pins) |  | $\begin{gathered} 10 \\ -10 \\ \hline \end{gathered}$ | $\begin{aligned} & \mu \mathrm{A} \\ & \mu \mathrm{~A} \end{aligned}$ | $\begin{aligned} & V_{\mathbb{I N}}=V_{C C} \\ & V_{\mathbb{I N}}=0 V \end{aligned}$ |
| Loh | High Level Output Leakage Current |  | 10 | $\mu \mathrm{A}$ | $V_{\text {OUT }}=V_{\text {CC }}$ |
| Iofl | Output Float Leakage Current |  | $\pm 10$ | $\mu \mathrm{A}$ | $0.45 \mathrm{C} \leq \mathrm{V}_{\text {OUT }} \leq \mathrm{V}_{\text {CC }}$ |

CAPACITANCE $T_{A}=25^{\circ} \mathrm{C}, \mathrm{f}_{\mathrm{c}}=1 \mathrm{MHz}, \mathrm{V}_{\mathrm{CC}}=0 \mathrm{~V}$

| Symbol | Parameter | Limits |  | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max |  |  |
| $\mathrm{C}_{1 \times(\phi)}$ | Clock Input Capacitance |  | 20 | pF | All Pins Except Pin Under Test Tied to AC Ground |
| $\mathrm{C}_{\mathrm{IN}}$ | Input Capacitance |  | 10 | pF |  |
| $\mathrm{Cl}_{1 / \mathrm{O}}$ | Input/Output Capacitance |  | 20 | pF |  |

A.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}, \mathrm{V}_{C C}=+5.0 \mathrm{~V} \pm 10 \%$

| Symbol | Parameter | Typ(1) | Min | Max | Unit | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CLOCK TIMING |  |  |  |  |  |  |
| $\mathrm{t}_{\mathrm{Cr}}$ | Clock Period |  | 120 | 500 | ns | (Note 5) |
| ${ }^{\text {t }} \mathrm{CH}$ | Clock High Period |  | 40 |  | ns | (Note 4, 5) |
| $t_{\text {fast }}$ | Reset Width |  | 14 |  | tor |  |
| READ CYCLE |  |  |  |  |  |  |
| $t_{\text {AR }}$ | Select Setup to RD $\downarrow$ |  | 0 |  | ns |  |
| $t_{\text {fa }}$ | Select Hold from RD $\uparrow$ |  | 0 |  | ns |  |
| $t_{\text {RR }}$ | RD Pulse Width |  | 250 |  | ns |  |
| $\mathrm{t}_{\text {RD }}$ | Data Delay from RD $\downarrow$ |  |  | 200 | ns |  |
| tbF | Output Float Delay |  | 20 | 100 | ns |  |
| WRITE CYCLE |  |  |  |  |  |  |
| taw | Select Setup to WR $\downarrow$ |  | 0 |  | ns |  |
| IWA | Select Hold from WF $\uparrow$ |  | 0 |  | ns |  |
| ${ }^{\text {tww }}$ | WA Pulse Width |  | 250 |  | ns |  |
| tow | Data Setup to WR $\uparrow$ |  | 150 |  | ns |  |
| two | Data Hold from WR $\uparrow$ |  | 5 |  | ns |  |
| INTERRUPTS |  |  |  |  |  |  |
| til | INT Delay from RD $\uparrow$ |  |  | 500 | ns | (Note 8) |
| IW1 | INT Delay from WR $\uparrow$ |  |  | 500 | ns | (Note 6) |
| DMA |  |  |  |  |  |  |
| tracy | DRQ Cycle Period |  | 13 |  | $\mu 8$ | (Note 6) |
| takRo | DACK $\downarrow$ to DRQ $\downarrow$ |  |  | 200 | ns |  |
| tror | DRQ $\uparrow$ to $\mathrm{FD} \downarrow$ |  | 800 |  | ns | (Note 8) |
| trow | DRQ $\uparrow$ to WR $\downarrow$ |  | 250 |  | n8 | (Note 6) |
| trorw | DRQ $\uparrow$ to $\overline{\text { D }} \uparrow$ or WR $\uparrow$ |  |  | 12 | $\mu 8$ | (Note 8) |

A.C. CHARACTERISTICS $T_{A}=0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=+5.0 \mathrm{~V} \pm 10 \%$ (Continued)

| Symbol | Parameter | Typ(1) | Min | Max | Unit | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| FDD INTERFACE |  |  |  |  |  |  |
| twCy | WCK Cycle Time | $\begin{aligned} & 2 \text { or } 4 \\ & 1 \text { or } 2 \end{aligned}$ |  |  | $\mu \mathrm{S}$ | $\begin{aligned} & \text { MFM }=0 \\ & \text { MFM }=1 \text { (Note 2) } \end{aligned}$ |
| IWCH | WCK High Time | 250 | 80 | 350 | ns |  |
| $\mathrm{t}_{\mathrm{CP}}$ | Pre-Shift Delay from WCK $\uparrow$ |  | 20 | 100 | ns |  |
| $\mathrm{t}_{\mathrm{CD}}$ | WDA Delay from WCK $\uparrow$ |  | 20 | 100 | ns |  |
| twDD | Write Data Width |  | ${ }_{\text {twCH }}$ - 50 |  | ns |  |
| twe | WE $\uparrow$ to $W C K \uparrow$ or WE $\downarrow$ to WCK $\downarrow$ Delay |  | 20 | 100 | ns |  |
| twwcy | Window Cycle Time | $\begin{aligned} & 2 \\ & 1 \end{aligned}$ |  |  | $\mu \mathrm{S}$ | $\begin{aligned} & M M=0 \\ & M F M=1 \end{aligned}$ |
| twro | Window Setup to RDD $\uparrow$ |  | 15 |  | ns |  |
| thow | Window Hold from RDD $\downarrow$ |  | 15 |  | ns |  |
| trid | RDD Active Time (HIGH) |  | 40 |  | ns |  |
| FDD SEEK/DIRECTION/STEP |  |  |  |  |  |  |
| tus | US ${ }_{0,1}$ Setup to RW/SEEK $\uparrow$ |  | 12 |  | $\mu \mathrm{s}$ | (Note 6) |
| ${ }^{\text {t }}$ SU | US ${ }_{0,1}$ Hold after $\overline{\text { FW/SEEK } \downarrow}$ |  | 15 |  | $\mu \mathrm{s}$ | (Note 6) |
| tso | RW/SEEK Setup to LCT/DIR |  | 7 |  | $\mu \mathrm{s}$ | (Note 6) |
| tos | RW/SEEK Hold from LCT/DIR |  | 30 |  | $\mu \mathrm{s}$ | (Note 6) |
| tost | LCT/DIR Setup to FR/STEP $\uparrow$ |  | 1 |  | $\mu \mathrm{S}$ | (Note 6) |
| tstd | LCT/DIR Hold from FR/STEP $\downarrow$ |  | 24 |  | $\mu s$ | (Note 6) |
| tSTU | DS ${ }_{2,1}$ Hold from FR/Step $\downarrow$ |  | 5 |  | $\mu \mathrm{S}$ | (Note 6) |
| tstp | STEP Active Time (High) | 5 |  |  | $\mu \mathrm{s}$ | (Note 6) |
| tsc | STEP Cycle Time |  | 33 |  | $\mu \mathrm{s}$ | (Note 3, 6) |
| $t_{\text {F } R}$ | FAULT RESET Active Time (High) |  | 8 | 10 | $\mu \mathrm{s}$ | (Note 6) |
| LID ${ }^{\text {d }}$ | INDEX Pulse Width | 10 |  |  | $\mathrm{tcr}^{\text {cr }}$ |  |
| ${ }_{\text {t }}$ C | Terminal Count Width |  | 1 |  | tcr |  |

## NOTES:

1. Typical values for $T_{A}=25^{\circ} \mathrm{C}$ and nominal supply voltage.
2. The former values are used for standard floppy and the latter values are used for mini-floppies.
3. $\mathrm{tsC}_{\mathrm{SC}}=33 \mu \mathrm{~s} \mathrm{~min}$. is for different drive units. In the case of same unit, tsc can be ranged from 1 ms to 16 ms with 8 MHz clock period, and 2 ms to 32 ms with 4 MHz clock, under software control.
4. From 2.0 V to +2.0 V .
5. At 4 MHz , the clock duty cycle may range from $16 \%$ to $76 \%$. Using an 8 MHz clock the duty cycle can range from $32 \%$ to 52\%. Duty cycle is defined as: D.C. $=100$ ( $\mathrm{I} \mathbf{C H} \div$ tcr) with typical rise and fall times of 5 ns .
6. The specified vatues listed are for an 8 MHz clock period. Multiply timings by 2 when using a 4 MHz clock period.
A.C. TESTING INPUT, OUTPUT WAVEFORM

A.C. Testing: Inputs are driven at 2.4 V for a Logic " 1 " and 0.45 V for a Logic "0". Timing measurements are made at 2.0 V for a Logic " 1 " and 0.8 V for a Logic " 0 ".

## WAVEFORMS

PROCESSOR READ OPERATION


WAVEFORMS (Continued)
PROCESSOR WRITE OPERATION


DMA OPERATION


WAVEFORMS (Continued)
CLOCK TIMING


FDD WRITE OPERATION


WAVEFORMS (Continued)

## SEEK OPERATION



IMDEX

Tandy Video II
Custom IC
Part Number 8079020

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## IRPROVED TANDY 1000 VIDEO CONTROLLER

## VIDEO CONTROLLER FEATURES

1) 100 \% software compatible with the current Tandy 1000 video controller design.
2) Provides international support with alternate character sets and video mode control locking.
3) 128 K of Video/System memory.
4) External character ROM for all 200 line and 350 line modes.
5) Supports $720 \times 350$ Monochrome text and $720 \times 348$ Hercules graphics.

Improved Tandy 1000 Video Controller Pin-Out

| PIN* | NAME | DESCRIPTION | DRIVE | PIN* | - NAME | DESCRIPTION |  | DRIVE |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | MEMR- | Mem read | IN | 51 | CFNT3 | Font dat/adr | 3 | 2MA |
| 2 | MEMW- | Mem write | IN | 52 | CFNT4 | Font dat/adr | 4 | 2MA |
| 3 | GND | GROUND | IN | 53 | GND | GROUND |  | IN |
| 4 | IOR- | I/O Read | IN | 54 | CFNT5 | Font dat/adr | 5 | 2MA |
| 5 | IOW- | I/O Write | IN | 55 | CFNT6 | Font dat/adr | 6 | 2MA |
| 6 | BHE- | CPU CTL | IN | 56 | CFNT7 | Font dat/adr | 7 | 2MA |
| 7 | RESET | Sys Reset | IN | 57 | CFNT8 | Cfont add |  | 2MA |
| 8 | ADRLTCH | Add latch | IN | 58 | CFNT9 | Cfont add |  | 2MA |
| 9 | VIDWT | Video wait | 8MA | 59 | CFNTl0 | Cfont add |  | 2MA |
| 10 | XMDO | Memory A/D | 4MA | 60 | CFNT11 | Cfont add |  | 2MA |
| 11 | XMD 1 | Memory A/D | 4MA | 61 | CFNTl2 | Cfont add |  | 2MA |
| 12 | XMD2 | Memory A/D | 4MA | 62 | CFNT13 | Cfont add |  | 2MA |
| 13 | XMD 3 | Memory A/D | 4MA | 63 | N/C | NO CONNECT |  | 2MA |
| 14 | XMD4 | Memory A/D | 4MA | 64 | CLTH | Cfont latch |  | 2MA |
| 15 | vCC | +5 Volts | IN | 65 | VCC | + 5 Volts |  | IN |
| 16 | XMD5 | Memory A/D | 4MA | 66 | 14.3 | SYS CLOCK |  | 4MA |
| 17 | XMD6 | Memory A/D | 4MA | 67 | RFSH | REFRESH IN |  | IN |
| 18 | XMD7 | Memory A/D | 4MA | 68 | D0 | CPU Data |  | 4MA |
| 19 | YMD0 | Memory A/D | 4MA | 69 | D1 | CPU Data |  | 4MA |
| 20 | YMD1 | Memory A/D | 4MA | 70 | D2 | CPU Data |  | 4MA |
| 21 | YMD2 | Memory A/D | 4MA | 71 | D3 | CPU Data |  | 4MA |
| 22 | YMD3 | Memory A/D | 4MA | 72 | D4 | CPU Data |  | 4MA |
| 23 | YMD4 | Memory A/D | 4MA | 73 | D5 | CPU Data |  | 4MA |
| 24 | YMD5 | Memory A/D | 4MA | 74 | D6 | CPU Data |  | 4MA |
| 25 | YMD6 | Memory A/D | 4MA | 75 | D7 | CPU Data |  | 4MA |
| 26 | YMD7 | Memory A/D | 4MA | 76 | N/C | No Connect |  |  |
| 27 | OEXY- | RAM CTL | 4MA | 77 | DOTCLK | DOT CLOCK |  | 4MA |
| 28 | GND | Ground | IN | 78 | GND | Ground |  | IN |
| 29 | RAS- | RAM CTL | 8MA | 79 | 25 | 25.175 |  | IN |
| 30 | CAS- | RAM CTL | 8MA | 80 | 28MHZ | Clock |  | IN |
| 31 | MWEX- | RAM Write | 4MA | 81 | AO/MDO | CPU A/D |  | 4MA |
| 32 | MWEY- | RAM Write | 4MA | 82 | Al/MDI | CPU A/D |  | 4MA |
| 33 | MEMIOS- | Mem \& I/O sel | 2MA | 83 | A2/MD2 | CPU A/D |  | 4MA |
| 34 | N/C | NO CONNECT |  | 84 | A3/MD3 | CPU A/D |  | 4MA |
| 35 | N/C | NO CONNECT |  | 85 | A4/MD4 | CPU A/D |  | 4MA |
| 36 | OUTR | RED VIDEO | 4MA | 86 | A5/MD5 | CPU A/D |  | 4MA |
| 37 | OUTG | GREEN/MONO | 4MA | 87 | A6/MD6 | CPU A/D |  | 4MA |
| 38 | OUTB | BLUE VIDEO | 4MA | 88 | A7/MD7 | CPU A/D |  | 4MA |
| 39 | OUTI | INTENSITY | 4MA | 89 | A8/MD8 | CPU A/D |  | 4MA |
| 40 | OUTHSYNC | Horz Sync | 4MA | 90 | A9/MD9 | CPU A/D |  | 4MA |
| 41 | OUTVSYNC | Vert Sync | 4MA | 91 A | Al0/MD10 | CPU A/D |  | 4MA |
| 42 | N/C | NO CONNECT |  | 92 A | All/MDl 1 | CPU A/D |  | 4MA |
| 43 | N/C | NO CONNECT |  | 93 A | Al2/MD12 | CPU A/D |  | 4MA |
| 44 | N/C | NO CONNECT |  | 94 A | Al 3/MDl 3 | CPU A/D |  | 4MA |
| 45 | N/C | NO CONNECT |  | 95 A | A14/MD14 | CPU A/D |  | 4MA |
| 46 | FRMOE* | ROM Enable | 2MA | 96 A | Al5/MD15 | CPU A/D |  | 4MA |
| 47 | N/C | NO CONNECT | 2MA | 97 | Al 6 | CPU Address |  | IN |
| 48 | CFNTO | Font dat/adr 0 | 2MA | 98 | Al7 | CPU Address |  | IN |
| 49 | CFNT1 | Font dat/adr 1 | 2MA | 99 | Al 8 | CPU Address |  | IN |
| 50 | CFNT2 | Font dat/adr 2 | 2MA | 100 | A19 | CPU Address |  | IN |

## General Description

The new improved Tandy 1000 video controller chip is designed to operate with one of two types of monitors, an RGBI 200 line Color monitor or a Monochrome 350 line monitor. This custom controller implements all of the video logic for the Tandy 1000. Figure l shows a block diagram of this custom video controller chip.

If an RGBI 200 line monitor is used, the display system supports up to 16 colors. These sixteen colors are defined by combinations of the $R, G, B$, and $I$ bits as shown in the chart below.


## OPERATING MODES

The operating modes supported by the Tandy 1000 video controller may be grouped in two categories: Alphanumeric and Graphic. A list of these modes is shown in Table 1A.

## ALPHANUAERIC MODES

The Alphanumeric modes have two basic types of operation:
80 character and 40 character. In both modes the character font tables are stored in a separate character ROM and consist of one or two pages of 256 characters. Two bytes of data are used to define each character on the screen. The even address is the character code and is used to address the character font pattern stored in ROM. The odd address is the attribute byte, that defines the foreground and the background color of the character. The following chart shows how the attribute byte controls the colors.


TANDY 1000 IMPROVED VIDEO CONTROLLER MODES

| BIOS <br> MODE | TYPE | COLORS | ALPHA FORMAT | BUFFER START | BOX <br> SIZE | MONITOR | RESOL | UTION | TYPE |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0/1 | ALPHA | 16 | $40 \times 25$ | B8000 | $8 \times 9$ | CM-5/11 | 320 X | 225 | CGA |
| 2/3 | ALPHA | 16 | $80 \times 25$ | B8000 | $8 \times 9$ | CM-5/11 | 640 X | 225 | CGA |
| 4/5 | GRAPH ICS | 4 | $40 \times 25$ | B8000 | $8 \times 8$ | CM-5/11 | 320 x | 200 | CGA |
| 6 | GRAPHICS | 2 | $80 \times 25$ | B8000 | 8x8 | CM-5/11 | 640 X | 200 | CGA |
| 7 | ALPHA | 4 | $80 \times 25$ | B0000 | $9 \times 14$ | VM-5 | 720 X | 350 | MDA |
| 7H | GRAPHICS | 2 | $80 \times 25$ | B0000 | $9 \times 14$ | VM-5 | $720 \times$ | 348 | MDA |
| 8 | GRAPHICS | 16 | $20 \times 25$ | B8000 | $8 \times 8$ | CM-5/11 | 160 X | 200 | PCjr |
| 9 | GRAPHICS | 16 | $40 \times 25$ | B8000 | $8 \times 8$ | CM-5/11 | 320 X | 200 | PCjr |
| A | GRAPHICS | 4 | $80 \times 25$ | B8000 | $8 \times 8$ | CM-5/11 | 640 | 200 | PCjr |
| B | RESERVED | Used | by the | BIOS to | load the | color ch | aract | font |  |
| C | RESERVED | Used | by the | bIOS to | load the | mono cha | acter | font |  |
| E | GRAPHICS | 16 | $80 \times 25$ | A0000 | $8 \times 8$ | CM-5/11 | 640 X | 200 | TDA |

TABLE 1A

To take advantage of the type of monitor used, the Tandy 1000 supports two different types of character box sizes depending on the type of monitor that is used. With a standard 200 line RGBI monitor, the Tandy 1000 video controller supports an $8 \times 9$ character box. On the 350 lin monochrome monitor, a $9 \times 14$ character box is used.

## GRAPH IC MODES

The Tandy 1000 video controller supports a variety of graphics modes.

## GRAPH ICS MEHORY USAGE

*200 Line Low or Medium Resolution Graphics Memory uses either 2 or 4 banks of 8000 bytes. In either case, pixel information for the display's upper left corner is found a address B8000.


| Offset <br> From <br> B8000 | <------160 Bytes------> |  |
| :---: | :---: | :---: |
| 0000 |  | 00 Scans |
|  |  | (0,4,8,... 196) |
| 1F3F |  |  |
|  |  |  |
| 2000 |  | 01 Scans |
|  |  | (1,5,9,...,197) |
| 3F3F |  |  |
|  |  |  |
| 4000 |  | 10 Scans |
|  |  | (2,6,10,...,198) |
| 5F3F |  |  |
|  |  |  |
| 6000 |  | 11 Scans |
|  |  | (3,7,11,...,199) |
| 7F3F |  |  |



The 2 Color Medium Resolution $640 \times 200$ Graphics mode may require a high resolution monitor for proper operation. Available in the IBM PC and IBM PCjr, this mode has the following characteristics:

Contains a maximum of 200 rows of 640 PELE. Can display 2 of 16 possible colors. Requires 16 k bytes of read/write memory. Formats 8 eELs per byte for each byte in the following manner:

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PAD | PA O | PAD | PA O | PA O | PA O | PA O | PAD |
| \|First | \|Secund | \|Third | \|Forth | \|Fifth | \|Sixth | \|Synth | \| Eghth |
| \| |  |  |  |  |  |  |  |

4 COLOR MEDIUM RESOLUTION $640 \times 200$ GRAPHICS MODE (A)
The 4 Color Medium Resolution $640 \times 200$ Graphics mode may require a high resolution monitor for proper operation. Only supported on the IBM PCjr, this mode has the following characteristics:

Contains a maximum of 200 rows of 640 eELs
Can display 4 of 16 possible colors
Each pixel selects 1 of 4 colors
Requires 32 K bytes of read/write memory Formats 8 eELs per two bytes (l even byte and 1 odd byte) in the following manner:

EVEN BYTES



## 16 COLOR MEDIUM RESOLUTION $320 \times 200$ GRAPGICS MODE (9)

The 16 Color Medium Resolution $320 \times 200$ Graphics mode works with all types of display devices. This mode is available on the IBM PCjr only and has the following characteristics:

Contains a maximum of 200 rows of 320 PELs
Can display 16 of 16 possible colors
Each pixel selects 1 of 16 colors
Requires 32 K bytes of read/write memory
Formats 2 PELs per byte in the following manner:

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PA3 | PA2 | PA1 | PA0 | PA3 | PA2 | PA1 | PA0 |



16 COLOR LOW RESOLUTION $160 \times 200$ GRAPHICS MODE (8)
The 16 Color Medium Resolution $160 \times 200$ Graphics mode works with all types of display devices. This mode is available on the IBM PCjr only and has the following characteristics:

Contains a maximum of 200 rows of 160 PELs Can display 16 of 16 possible colors Each pixel selects 1 of 16 colors Requires 16 K bytes of read/write memory Formats 2 pELs per byte in the following manner:

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PA3 | PA2 | PA1 | PA0 | PA3 | PA2 | PA1 | PA0 |



## 4 COLOR MEDIUM RESOLUTION 160 X 200 GRAPHICS MODE (4/5)

The 16 Color Medium Resolution $320 \times 200$ Graphics mode works with all types of display devices. This mode is available on the IBM PC and IBM PCjr modes and has the following characteristics:

Contains a maximum of 200 rows of 320 PELs
Can display 4 of 16 possible colors
Each pixel selects 1 of 4 colors
Requires 16 K bytes of read/write memory
Formats 4 pELs per byte in the following manner:


HERCULES GRAPHICS (7B)
2 Color Hercules graphics works with a 350 line Monochrone display. This mode has the following characteristics:

720 PELs by 348 Rows
2 color Monochrome Graphics requires $32 k$ bytes of RAM Formats 8 PELs per bytes in the following manner :

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PAO | PAO | PAO | PAO | PAO | PAO | PAO | PAO |


| First | Second | Third | Forth | Fifth | Sixth | Seventh | Eighth |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Dsply | Dsply | Dsply | Dsply | Dsply | Dsply | Dsply | Dsply |
| PEL | PEL | PEL | PEL | PEL | PEL | PEL | PEL |
|  |  |  |  |  |  |  |  |

The 16 Color High Resolution $640 \times 200$ Graphics mode works with a 200 line high resolution RGBI monitor. This mode has the following characteristics:

Contains 200 rows of 640 PELs Can display 16 of 256 K possible colors Requires 64 k bytes of read/write memory Formats 2 pELs per byte for each byte in the following manner:

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PA3 | PA2 | PA1 | PA0 | PA3 | PA2 | PA1 | PA0 |


| First | Second |
| :---: | :---: |
| Display | Display |
| PEL |  |

## VIDEO I/O MAP





## CRTC CONTROLLER AND MISCELLANEOUS CONTROL REGISTERS

The CRTC and Miscellaneous control registers are accessed by two I/O commands. The two I/O commands function by first writing the desired index value to address hex 3D4 (3B4), and then writing the data to address hex 3D5 (3B5).

Index Register, Hex 3D4 (3B4): This register is read and write, and points to the specific data register addressed through hex 3D5 (3B5).


The following is a list of the data registers and their functions.

Index
(Hex) Register Description
$00 \quad$ Horizontal Total

02 Start Horizontal Sync
03 Sync Pulse Width
04 Vertical Total
05 Vertical Total Adjust
06 Vertical Characters Displayed
07 Start Vertical Sync
08 Reserved
09 Scan Lines per Character
0A Cursor Start
OB Cursor End
$0 \mathrm{C} \quad$ Start of Screen High
OD Start of Screen Low
OE Cursor Position High
0F Cursor Position Low
10 Mode Control
12 Character Generator Interface and Sync Polarity, or Display Sense
13 Character Font Pointer
21 Test Mode Register 2

HORIZONTAL TOTAL REGISTER, INDEX 00:
This register contains the total number of characters in the horizontal scan interval. The number consists of the total of the displayed and non-displayed characters. minus one. This register determines the frequency of the 'horizontal sync' signal.

HORIZONTAL DISPLAYED REGISTER, INDEX 01:
This register determines the total number of characters to be displayed during a horizontal line. This register must be loaded with a value that is less than the horizontal total register.

HORIZONTAL SYNC POSITION REGISTER, INDEX 02:
This register specifies the character position count at which the 'horizontal sync' signal becomes active. The specified value -1 must be programmed.

SYNC WIDTE REGISTER, INDEX 03:
This register specifies the pulse widths of the horizontal and vertical synchronization signals. The horizontal pulse width is programmed in units of character clocks. The vertical pulse width is programmed in units of the horizontal synchronization period. This register is programmed to match the display specifications.


Figure 1-31. Sync Pulse Width Register
VERTICAL TOTAL RBGISTER, IADEX 04:
This register contains the 8 bits for the total number of horizontal scan lines in the vertical scan interval. The total number consists of both the displayed and nondisplayed scan lines minus 1. This register and the Vertical Total Adjust register determine the frequency of the 'vertical sync' signal.

VERTICAL TOTAL ADJUST REGISTER, INDEX 05:
This register is used to adjust the total number of horizontal scan lines in the vertical scanning interval. It allows for an odd number of horizontal lines (525 for 60 Hz) 。


## VERTICAL DISPLAYED REGISTER. IRDEX 06:

This register contains the 8 least-significant bits for the number of horizontal scan lines displayed during the vertical scan interval. The ninth bit is the inversion of bit 6 of the Mode Control register.

VEEPICAL SYIC POSITION REGISTER, INDEX 07:
This register contains the 8 bits for the vertical scan line count. It determines when the 'vertical sync' signal becomes active. The specified value -1 must be programmed.

SCAN LINES PER CHARACTER REGISTER, INDEX 09:
This register determines the number of horizontal scan lines in a character row.


## CURSOR START REGISTER, INDEX OA:

Bits 3 through 0 in this register determine the horizontal scan line count at which the cursor outputbecomes active.


Figure 1-34. Cursor Start Register

## CURSOR END REGISTER, INDEX OB:

This register determines the horizontal scan line count when the cursor output becomes inactive.


START OF SCREEN HIGH REGISTER, INDEX OC:
This register contains the 6 most-significant bits for the starting memory address of the video display buffer. Fourteen address bits determine the starting address. This register is initialized to a value of hex 00 .

START OF SCREEN LOW REGISTER, INDEX OD:
This register, together with the Start of Screen High register, gives the starting address of the display buffer. For all modes, this register is initialized to a value of hex 00.

CURSOR POSITION HIGH REGISTER, INDEX OE:
This register contains the 4 most-significant bits for the cursor location.


## CURSOR POSITION LOW REGISTER, INDEX OF:

This register contains the 8 least-significant bits for the location of the cursor. A value of hex 00 in both of these registers will locate the cursor in the upper left-hand corner. The cursor is not supported in any graphics mode.

MODE CONTROL REGISTER, INDEX 10:


Figure 1-37. Mode Control, Write

## WRITE

BIT 7 When set to 1 , the inhibit write bit prevents any writes to the horizontal and vertical registers.


Figure 1-38. Mode Control, Read

CHARACTER GENERATOR INTERFACE AND SYNC POLARITY REGISTER, IADEX 12:

The register controls the character font ROM.


Figure 1-40. Character Generator Interface and Sync Polarity Register

BIT 5 This bit selects the font page that is used as the font table. When set to 1 , font page 1 is selected; when clear to 0 , font page 0 is selected.

BIT 4 When this bit is set to 1,512 character codes are displayable in the text modes. Bit 3 of the attribute byte then determines the font page when displaying the character. When this bit is set to 1 , only eight foreground colors are supported. When this bit is cleared to 0 , only 256 character codes are displayed, and bit 5 of this register determines the active font.
See font table, end of video section.

CHARACTER FONT REGISTER, INDEX 13:
Bit 4 of this register is used to select between the 200 line RGBI character set and the 350 line monochrome character set. A 0 in bit 4 selects the 200 line character set.

## TEST MODE REGISTER 1, INDEX 20

This register is reserved for future test purposes and should be cleared to zero at all times.

## TEST MODE REGISTER 2, INDEX 21

This register is used only during manufacturing tests and must be cleared to zero at all other times.


## VIDEO ARRAY REGISTERS

The following registers can be accessed by writing their Hex address to 3DA and their Hex Data to 3DE.


ARRAY PALETTE MASK REGISTER


ARRAY BORDER COLOR


## ARRAY MODE CONTROL REGISTER




## ARRAY PALETTE REGISTERS

There are sixteen 4 bit wide palette registers implemented by a $16 x 4$ bit RAM. These registers are Read/Write. Their address in the video array are from 10 - lF Hex, and can be used to re-define any color.

To load the palette, write the hex address to the video Array Register at $3 D A$ hex. Then, the new palette color is written to 3DE hex.

Palette address 10 hex is accessed whenever the color code from memory is a hex 0 , and address 11 hes is accessed whenever the color code from memory is a hex 1 , and so forth. A description of the color codes can be found in the "Available Colors Table" at the beginning of this section.

Note: The palette address can be 'masked' by using the palette mask register.

The following is a description of the register's bit functions:

Bit Number Function

| 0 | Blue |
| :--- | :---: |
| 1 | Green |
| 2 | Red |
| 3 | Intensity |

When loading the palette, the video is 'disabled' and the color viewed on the screen is the data contained in the register being addressed by the processor.

When the program has completed loading the palette，it must change the hex address to some address less than 10 hex for video to be＇enabled，again．

If a programmer does not wish a user to see the adverse effects of loading the palette，the palette should be loaded during the vertical retrace time．The program must modify the palette and change the address to less than 10 hex within the vertical retrace time．A vertical retrace inter rupt and a status bit are provided to facilitate this procedure．

In two color modes，the palette is defined by using one bit， PAO，with the following logic：


```
I PAO Function I
I===============================================1 I
I 0 Palette Register 0 I
I 1 Palette Register 1 I
```



In four color modes，the palette is defined by using two bits，PAl and PAO，with the following logic：

| I | PA1 | PAO | Function |  |  | I |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ＝$=$ | ＝ | ＝＝＝＝＝＝＝＝ | ＝＝ミニニ＝ |  |  |
| I | 0 | 0 | Palette | Register 0 | 0 | I |
| I | 0 | 1 | Palette | Register 1 | 1 | I |
| I | 1 | 0 | Palette | Register 2 | 2 | 1 |
| I | 1 | 1 | Palette | Register 3 | 3 | I |

In sixteen color modes, the palette is defined by using four bits, PA3, PA2, PAl and PAO, with the following logic:

| IPA3 | PA2 | PA1 | PAO | Function | I |
| :---: | :---: | :---: | :---: | :---: | :---: |
| I 0 | 0 | 0 | 0 | Palette Register 0 | I |
| I 0 | 0 | 0 | 1 | Palette Register 1 | I |
| 10 | 0 | 1 | 0 | Palette Register 2 | I |
| 10 | 0 | 1 | 1 | Palette Register 3 | I |
| I 0 | 1 | 0 | 0 | Palette Register 4 | I |
| 10 | 1 | 0 | 1 | Palette Register 5 | I |
| I 0 | 1 | 1 | 0 | Palette Register 6 | I |
| 10 | 1 | 1 | 1 | Palette Register 7 | 1 |
| I 1 | 0 | 0 | 0 | Palette Register 8 | I |
| 11 | 0 | 0 | 1 | Palette Register 9 | I |
| I 1 | 0 | 1 | 0 | Palette Register 10 | 1 |
| 11 | 0 | 1 | 1 | Palette Register 11 | 1 |
| I 1 | 1 | 0 | 0 | Palette Register 12 | I |
| 11 | 1 | 0 | 1 | Palette Register 13 | I |
| 11 | 1 | 1 | 0 | Palette Register 14 | I |
| 11 | 1 | 1 | 1 | Palette Register 15 | I |



## sid be Color Select (seep. II)





CONTROL BIT PROGRAMMING CHART


## CRTC PROGRAMMING TABLE

Mode 4,5,6,8 actual
 as indicated (framuideo inthatization tabliel.


## VIDEO/SYSTEM MEMORY ADDRESS MAP

| I | FFE8 BITS | I | VIDEO/SYSTEM MEMORY | I | VIDEO/SYST | EEM MEM | I |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| I | 321 | I | START ADDRESS | I | ADDRESS | RANGE | I |
| I | 000 | I | 00000 | I | 00000 - | 1 FFFF | I |
| I | 001 | I | 20000 | I | 20000 - | 3FFFF | I |
| 1 | 010 | 1 | 40000 | I | 40000 - | 5FFFF | I |
| I | 011 | I | 60000 | I | 60000 - | 7FFFF | I |
| 1 | 100 | 1 | 80000 | I | 80000 - | 9 FFFF | I |
| I | 111 | I | - |  | No System | Memory | I |

VIDEO MEMORY ADDRESSING MODES


Notes on video memory:
A write to port 65 h with bit 2 clear disables access to video memory through addresses $A D 000$ BFFFFh. It does nit disable access through lower addresses if that has been eriabled at port FFEEC. forts 3B4-3DFh are disabled, but port FFE8n is not.

If bit 6 or 7 at port FFEAh is cleared, access to main memory in the range $50000-$ 9FFFFh is disabled. These addresses curespond to the 128 k video memory upgrade.

Bits 1-3 at port FFE8 8 determine whether video memory is addressable at locations below Arexooh, and if so, at which address. If these bits are 1006, video memory is addressed at $80 X O X$-9FFFFh, as well as at AXADOBFFFFh. If lIb, video memory is not addressable below ADOXXG.





ASCII code IBM character Candy character (decimal) (font (font 1)
$.243 \leq 3 / 4$
.244
.245 integral sigh $\quad 20$ )
J (lower half of $\xi$ (same as integral sign) 2.1)
246
.247

$$
\div
$$

$\approx$ (approx. equal) , (cedilla)
.249
$\begin{array}{ll}\text { - (degree) } & 0 \text { (degree) } \\ \text { - (center dot, } & \cdots \text { (dieresis) }\end{array}$
$\div$ large)
.250 - (center dot, (center dot, small) small, one scan line higher)
. 251
$\sqrt{ }$ (radical) ${ }^{1}$ ( 1 superscript)

- 252 (a superscript ${ }^{3}$ (3 superscript)
-note tables
say $\eta^{(\text {eta })}$
should be here)
253


JACKSBORO SPECIFICATION jmp 05-26-88

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### 1.0 GENERAL

### 1.1 Functional Description

The PSSJ Tandy ASIC is contained in a 68 pin PLCC package, and comprises the Printer port, a Serial (RS232) port, the Sound function, and the Joystick function of the Tandy 1000 computers.
2.0 PIN LIST
PIN NAME PIN NO. DRIVE DESCRIPTION

| VCC | 1,35 | -- | Power inputs |
| :--- | :--- | :--- | :--- |
| VBB | 59 | -- | Analog Power input |

GND 18,52 -- Grounds

RST 25 TTL in System reset signal,

CLK14M 2 TTL in Clock signal input, $14.31313 \mathrm{MHz}, 50 \%$ duty cycle.

CLR2IN 37
TTL in Clock signal input, either 24 MHz or $1.8432 \mathrm{MHz}, 50 \%$ duty cycle.

IODO - IOD7 14,15,16,17 DS1218, Eight bit peripheral data

19,20,21,22

IOR- 1

IOW- 11

| A0 - A2, A7 | $6,7,8,9$ |
| :--- | :--- |
| CS0-CS2 | $3,4,5$ |
| PINT | 12 |

bus intended to drive 5 XT type I/O slots, as well as all on board peripherals.

TTL in CPU/DMA I/O Read signal, active low. System control line.

DS1218 CPU/DMA I/O Write signal. active low. System control line.

TTL in System address lines.
TWL in Address decode inputs.
2 mA TS Printer Interrupt, tristate.

| PIN NAME | PIN NO. | DRIVE | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| SINT | 13 | 2 mA TS | Serial Interrupt, tristate. |
| PPITIM | 68 | TTL in | Low frequency sound input. |
| AUDIO_IN | 55 | An in. | Analog audio input, 1 V p-p. |
| SND_OUT | 57 | An out | Analog audio output, 2 V $p-p$. |
| GAIN_OUT | 56 | An out | Analog audio output, 2 V $p-p$ |
| DRQ | 23 | 2 mA TS | Data request for DMA operations, tristate. |
| TC | 27 | TTL in | Terminal Count input. |
| DACK1 | 26 | TTL in | Data acknowledge for DMA ops. |
| WAIT- | 24 | 2 mA OD | Sound chip wait output, open drain. |
| JPOS1 - JPOS4 | 60,61,62,63 | DS1218 | Digital joystick position input. |
| JSW1 - JSW4 | 64,65,66,67 | DS1218 | Digital joystick switch inputs. |
| DAC_OUT | 58 | An out | Analog DAC output for external integration, comparison with joystick voltages. |
| PD0 - PD7 | $\begin{aligned} & 51,50,49,48 \\ & 47,46,45,44 \end{aligned}$ | $\begin{aligned} & \text { DS } 1218 \\ & 4 \mathrm{~mA} \mathrm{TS} \end{aligned}$ | Printer data inputs/outputs. |
| INIT | 40 | 4 mA OD | Printer initialization output. |
| AFXT- | 39 | 4 mA OD | Printer auto feed output. |
| STROBE- | 38 | 4 mA OD | Printer strobe output. |
| ACK- | 41 | TTL in | Printer acknowledge input. |


| PIN NAME | PIN NO. | DRIVE | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| PE | 43 | TTL in | Printer paper empty input. |
| SLCTIN- | 53 | TTL in | Printer select input. |
| BUSY- | 42 | TTL in | Printer busy input. |
| FAULT- | 54 | TTL in | Printer fault input. |
| DTR- | 36 | 2 mA | RS232 data terminal ready output. |
| RTS- | 33 | 2 mA | RS232 request to send output. |
| TXD- | 34 | 2 mA | RS232 transmit data output. |
| RI- | 29 | TTL in | RS232 ring indicator input. |
| DCD- | 30 | TTL in | RS232 carrier detect input. |
| DSR- | 28 | TTL in | RS232 data set ready input. |
| CTS- | 32 | TTL in | RS232 clear to send input. |
| RXD- | 31 | TTL in | RS232 receive data input. |

### 3.0 ABSOLUTE MAXIMUM RATINGS

|  |  | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: |
| Storage Temperature: | -65 |  | 150 | degrees $C$ |
| Operating Temperature: | 0 | 25 | 55 | degrees $C$ |
| All output pins | -0.5 |  | 7.0 | volts DC |
| All input pins | -0.5 |  | 7.0 | volts DC |
| Power Supply (Vcc) | -0.5 |  | 7.0 | volts DC |
| Power dissipation |  |  | 700 | milliwatts |

4.0 D. C. ELECTRICAL CEARACTERISTICS

### 4.1 Inputs

Leakage current

## Vih (TTL in) <br> Vih (DS1218)

Vil
Min_TYp__Max $\qquad$ Units

Input capacitance
$+/-10$ uA
2.0 Vcct.5 volts DC
2.1 Vcc+.5 volts DC
-0.5 0.8 volts DC
10
pF
4.2 PDO - PD7, INIT, /AFXT, /STROBE
$\qquad$

| Iol | 4 |  | mA |
| :--- | :--- | :--- | :--- |
| Vol |  | 0.4 | volts DC |
| Ioh | 1 |  | mA |
| Voh | 2.4 |  | volts DC |
| Capacitive load | 100 |  | pF |

4.3 /WAIT

Iol
Vol
Capacitive load

Min_Typ_Max $\qquad$ Units

4
100
mA
volts DC pF
4.4 DRQ, /TXD, /DTR, /RTS, PINT, SINT
Iol
Vol
Ioh
Voh
Capacitive load

| Min_Typ_Max___ | Units |  |
| :--- | :--- | :--- |
| 2 | 0.4 | mA |
|  | volts DC |  |
| 1 |  | mA |
| 2.4 |  | volts DC |
| 40 |  | pF |

### 4.5 IODO - IOD7

Iol
Vol
Ioh
Voh
Capacitive load

Min__Typ__Max $\qquad$ Units

## 8

2 mA
2.4

100 volts DC mA volts DC pF

### 5.0 AC CHARACTERISTICS

Parameter
Tav (Address Valid)
Tah (Address Hold)

Trpw (Read Pulse Width)
Twpw (Write Pulse Width)
Min_TYP_Max__Units

Tdsu (Data Setup (Write))

| -15 |  | nSec |
| :--- | :--- | :--- |
| 30 |  | nSec |
| 120 |  | nSec |
| 125 | nSec |  |
| 65 | 100 | nSec |
| 10 | 30 | nSec |
| 25 |  | nSec |
|  |  | nSec |



I/O Write Cycle


I/O Read Cycle
6.0 Modifications to the 76496

### 6.1 Extra Bit of Division by each channel.

When clocked by a 3.579545 MHz signal, the lowest frequency generated by the 76496 (with its 10 bit dividers) is 109.24 Hz . It is desired to be able to generate lower frequencies. An extra bit of division will allow frequencies down to 54.62 Hz , or an octave lower than the lowest note currently available. Since there is an extra bit in the frequency update register (second byte), it makes sense to implement this feature here. However, to maintain backwards compatibility, since it is not known what is programmed in this bit, there needs to be a way of defeating the extra bit of division. Therefore, there is a signal (SEDE), which enables the extra bit for all three channels. This bit defaults to a logic zero (low) on reset. When it is set, by writing to port C4 with bit 6 high, the extra divider will be enabled.

### 6.2 Synchronization of frequency dividers.

The current 76496 design loads each divider when initially written to, with no provision for synchronization of the dividers. This is a handicap when programming frequencies of low integer relationships to each other, because it is not possible to guarantee the phase of the signals. Therefore, if synchronization is desired, it is enabled by writing to port C4, with bit 5 set (which defaults to reset). When this bit is high, any write to a frequency register of the new sound channel will not only load its divider, but reload the dividers in the present 76496.

### 6.3 Minimum Wait State Generation

The 32 wait states generated by the 76496 need to be reduced. The chip must be guaranteed to latch the data written in the same time allotted for the 8250A megacell. Any wait states generated should only apply to a successive write (not the first in a series). All write timing should be referenced to the rising edge of the IOW- strobe.
.vote: " $76496^{\prime \prime}=$ Texas Instruments SNTC496 3-voice fore and 1-roice noise generator chip, used in the IBM PC-Jr. The $A S S J$ is register-compatible w th the 76496 .

### 7.0 Software Specification

Port CO - C3 Write
Access 76496 megacell


Ort C4 Read
Readback all bits except bit 3. In addition: ]

bit $7=$ SAD- $=$| Successive Approximation |
| :--- |
| done. Useful when polling |
| instead of DMA for |
| successive approximation. |

bit $3=\quad$| DMA interrupt has occurred. |
| :--- |

| To clear the interrupt it |
| :--- |
| is necessary bring DICL |
| low, then back high. |

Port C5 Write
Direct write to DAC (DF1,0 $=11$ bin). Pulse width and waveshape (DF1,0 = 01 bin).

sampling output:
FD- Fl control samples/sec
Port C6
R/W
SAMPI-3 control volume
Frequency LSB for DAC sound channel.


Port C7 R/W
Amplitude/frequency MSN for DAC sound channel.

$$
7 \quad 6
$$

65
$5 \quad 4$
4 32 2 1 10


SAMP3 SAMP2 SAMP1 res Fill F10 F9 F8

The amplitude will be programmable in 7 levels, with approximately 3 dB per level. The maximum level ('li') will closely approximate that in the existing sound chip. A value of '000' will result in no output. This level control also applies to the raw DAC output when outputting digitized sound.

The ramp will count up the five MSB's of the DAC. The triangle will count up the four MSB's of the DAC for the first half of the wave, then count them back down for the second half. The frequency range of the DAC as a sound channel will have the same upper limit and a lower limit of one octave lower than the new frequency range of the sound chip (down to 27.3 Hz .). Obviously, the bit programming order of the frequency is different. The actual frequency will be 111.86 kHz divided by the number programmed into the sound frequency register (s).

Port 200 - 207 WR -- Clear Joystick DAC counter
A write to port 20X, where $X=0$ to 7, will clear a free-running counter, and load a value of 16 into the l2-bit divider. The eight bit free-running counter will be clocked by the 3.58 MHz signal divided by 24 , or 149.1 KHz . The output of the eight bit counter will drive the DAC to produce a stairstep wave, which simulates a ramp for use by the joystick comparators. When the counter reaches a count of 255 , it will stop until port $20 x$ is written to again.

The elapsed time for the complete ramp will be approximately 1.7 milliseconds, closely approximating the elapsed time of the current randy 1000 Joystick circuitry.

Bit 4 at port C7 is used to determine the MR C version. If ale to set and clear the bit, old versionelse new, See DST.

## Port 200-207 RD -- Joystick Status

The data read at port 20X, where $\mathrm{X}=0$ to 7 , will be the outputs of the joystick position comparators and the states of the joystick pushbuttons, in the same manner as the current Tandy 1000 Joystick circuitry.

## Planar Control

Port 65 contains three bits which are used to enable the printer interface (bit 1), the printer output (bit 7), and the serial port (bit 4). These bits are all enabled (set high) on reset, and must be cleared by software to disable the appropriate function. The printer output enable function is logically "ored" with the current Tandy 1000 printer output enable bit, so that either one will enable the printer output buffer.

Additional control is available at port FFEB. Bit DO selects whether the serial clock is divided by 13 or 1 . Bit Dl must be high to enable the joystick function, and bit $D 2$ must be high to enable the sound chip functions. Bits D1 and D2 default to high on power up.

## I/O Map Summary

The following ports are utilized in the PSSJ part:

| PORT | $\mathrm{R} / \mathrm{W}$ | BITS | FUNCTION |
| :--- | :--- | :--- | :--- |
| $===========================================\pi==================$ |  |  |  |
| 0061 | W | 4 | Sound Chip Enable |
| 0065 | $\mathrm{R} / \mathrm{W}$ | $1,4,7$ | Planar Control |
| $00 \mathrm{CO}-00 \mathrm{C} 3$ | W | all | Sound Chip Data |
| $00 \mathrm{C} 4-00 \mathrm{C} 7$ | $\mathrm{R} / \mathrm{W}$ | all | DAC Functions |
| $0200-0207$ | $\mathrm{R} / \mathrm{W}$ | all | Joystick Function |
| $0378-037 \mathrm{~A}$ | $\mathrm{R} / \mathrm{W}$ | all | Printer Interface |
| $03 \mathrm{FB-03FF}$ | $\mathrm{R} / \mathrm{W}$ | all | Serial Interface |
| FFEB | $\mathrm{R} / \mathrm{W}$ | $0,1,2$ | UART clock select,JSE,DSE |

## Floppy Disk Support Chip Specification

## Floppy Disk Support Chip Specification Contents

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Pin Description ..... 2
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Floppy Disk Support Logic
Tandy Part \#8041404
January 29, 1987
1.0 General Description
1.1 The Tandy Part \#8041404 - Floppy Disk Support Logic: -Generates the clock to the 765 Floppy Disk Controller. -Generates the write clock to the Floppy Disk. -Generates step pulses, track 0 indicator, DMA request, and FDC interrupt signals.

| $1--$ | CLR16M | +5V | --24 |
| ---: | :--- | ---: | :--- |
| $2--$ | WCK | SWITCH | --23 |
| $3--$ | FDCCLR | INT+ | --22 |
| $4--$ | RDDATA* | DMA/INTE | --21 |
| $5--$ | RDD | DRQ | --20 |
| $6--$ | RDW | FDCINT | --19 |
| $7--$ | FRES/S | FDCDMRQ* | --18 |
| $8--$ | RW*/SEEK | PSO | --17 |
| $9--$ | TRKO* | PS1 | --16 |
| $10--$ | F/TRKO | WRD | --15 |
| $11--$ | STEP* | WRE | -14 |
| $12--$ | GND | WRDATA* | --13 |

Figure 1. Pin Assignment
1.2 DESCRIPTION OF PINS:

| PIN \# | PIN NAME | TYPE | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| 1 | CLK16M | INPUT | Frequency $=16.0000$ Tolerance $=100 \mathrm{pmm}$ |
| 2 | WCK | OUTPUT | If SWITCH $=0$, period $=2$ us, 250 ns pulse <br> If $\operatorname{SWITCH}=1$, period $=1$ us, 250 ns pulse |
| 3 | FDCCLK | OUTPUT | If $\operatorname{SWITCH}=0$, then CLKl6M/4 <br> If SWITCH $=1$, then CLKl6M/2 |
| 4 | RDDATA | INPUT | Serial data from FDD |
| 5 | RDD | OUTPUT | Serial data from FDC |
| 6 | RDW | OUTPUT | Read Data Window |
| 7 | FRES/S | INPUT | Step pulses to move head to another cylinder |
| 8 | RW*/SEEK | INPUT | Specifies seek mode when high |
| 9 | TRK0* | INPUT | From FDD, indicating head is on track 0 |
| 10 | F/TRK0 | OUTPUT | To FDC, indicating head is on track 0 |
| 11 | STEP* | OUTPUT | Moves head of FDD |
| 12 | GND |  | Ground |
| 13 | WRDATA* | OUTPUT | Serial Data to FDD |
| 14 | WRE | INPUT | Write Enable |
| 15 | WRD | INPUT | Serial Data from FDC |
| 16 | PS1 | INPUT | Write precompensation status |
| 17 | PS0 | INPUT | Write precompensation status |
| 18 | FDCDMRQ* | OUTPUT | DRQ delayed by 1.0 usec. |
| 19 | FDCINT | OUTPUT | Interrupt request |
| 20 | DRQ | INPUT | FDC DMA Request |
| 21 | DMA/INTE | INPUT | DMA request and FDC interrupt enable |
| 22 | INT+ | INPUT | Interrupt request generated by FDC |
| 23 | SWITCH | INPUT | $\begin{aligned} & 0=\text { low density drive } \\ & 1=\text { high density drive } \end{aligned}$ |
| 24 | +5v |  | +5 Volts |



BLOCK DIAGRAM


### 4.0 AC CHARACTERISTICS

### 4.1 FDCCLK Timing

## Parameter

$\mathrm{t}_{\mathrm{H}}$
$\mathrm{t}_{\mathrm{R}}, \mathrm{t}_{\mathrm{F}}$
$\mathrm{t}_{\mathrm{L}}^{\mathrm{L}}$
4.2 WCK Timing
$t_{H}$
$t_{R}, t_{F}$
$t_{L}$
$t_{C Y}$
4.3 WRDATA* Timing

4.4 DMA/INTERRUPT Timing

4.5 CONTROL Timing

| Parameter | Min. Typ. | Max. |
| :---: | :---: | :---: |
| $\mathrm{T}_{\mathrm{L}}-\mathrm{FT}_{\mathrm{H}}$ |  | 30 |
| $\mathrm{T}_{\mathrm{H}}^{\mathrm{L}} \mathrm{FT}_{\text {I }}$ |  | 30 |
| $\mathrm{RS}_{L}-\mathrm{FI}_{L}$ |  | 30 |
| $\mathrm{F}_{\mathrm{H}}{ }^{-} \mathrm{S}_{\mathrm{S}} \mathrm{L}$ |  | 30 |
| $\mathrm{F}_{\mathrm{L}}-\mathrm{S}_{\mathrm{H}}$ |  | 30 |
| $\mathrm{RS}_{\mathrm{L}}-\mathrm{S}_{\mathrm{H}}$ |  | 30 |

4.6 DATA SEPARATOR Timing

| $\mathrm{RDA}_{W}$ | 200 | 350 | 550 | nSec |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{RDA}_{\mathrm{L}}-\mathrm{RDD}_{\mathrm{H}}$ | 188 |  | 313 | nSec |
| RDD ${ }_{\text {W }}$ | 240 | 250 | 260 | nsec |
| $\mathrm{RDD}^{\mathrm{H}}$-RDW ${ }_{\text {c }}$ | 850 | 875 | 900 | nSec |
| RDW ${ }^{(N D)}{ }_{\text {W }}$ |  | 2.0 |  | $\mu \mathrm{Sec}$ |

" A "

| RDA $_{S}$ | 3062 | nsec |
| :--- | ---: | ---: |
| RDW $_{\mathbf{C}}-$ RDD $_{H}$ | 15 | nsec |

"B"
$\stackrel{R D A}{S}_{\text {RDW }_{C}}^{- \text {RDD }_{H}}$
4812
1938
nSec nsec
"C"
$\begin{array}{lr}\mathrm{RDA}_{\mathrm{S}} & 5062 \\ \mathrm{RDW}_{\mathrm{C}}-\mathrm{RDD}_{\mathrm{H}} & 15\end{array}$
nSec nSec

## FDSL AC TIMING




FIG. 4 DMA/INTERRUPT TIMING.


FIG. 5 CONTROL LOGIC TIMING.


FIG. 6 DATA SEPARATOR TIMING.

```
    KFIT CUSTOM CHIP
(KEYBOARD, FLOPPY SUPPORT, INTERRUPT, TIMER)
```

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TANDY PART \#: 8079019

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## FUNCTIONAL DESCRIPTIONS

This Tandy KFIT custom IC consists of the following functional blocks:

- Programmable Peripheral Interface (PPI)
- Keyboard Interface Logic
- Floppy Disk Interface Logic
- Programmable Interrupt (equivalent to Intel 8259A) and sharing interrupt logic
- Programmable Timer (equivalent to Intel 8254-2 and Clock Divider
- Address Decoding Logic

Programmable Peripheral Interface

This section of the KFIT custom integrated circuit replaces the Intel 8255 A that was used on the original Tandy 1000 computer. On the block diagram for this section of logic, the 8255A is represented by three 74 LS 244 buffers addressed by read $A(0060)$, read $B$ ( 0061 ), read $C$ ( 0062 ). Also the two latches addressed by write $B(0061)$, write $C$ (0062) which are part of the original 8255A logic.

## Keyboard Interface Logic

This section of the KFIT custom integrated circuit is design to support Tandy 1000 keyboard or Tandy 101 enhanced keyboard. KYBDTYP signal is used to select Tandy 1000 keyboard when is LOW or Tandy 101 enhanced keyboard when is HIGH. The KYBDTYP is being read in to port FFEB (hex) bit 7. The KBDDATA - keyboard data is serial data bit stream and then is converted to 8 bits parallel data by 74LS322. The serial data is entered in the LOW to HIGH transition of the KBCLK.

## Floppy Disk Interface Logic

This section of the KFIT custom integrated circuit is design to support Floppy Disk Digital Output Register (DOR) function. This register is mapped in address 03F2 hex - data bit 0 to 7 (write only) to generate drive select DSOB,DS1B,DS2B; FDCRST (FDC reset) DMA/I and MTRONB (motor ON) signal. The DMA/I signal is used to disable FDCINTI, FDCDMRQ, and FDACKI signals for allowing the used of external FDC controller.

## Programmable Interval Timer

This section of the KFIT custom integrated circuit is equivalent to an Intel 8254-5 and is designed to use with the Tandy 1000 TX . It is organized as three independent l6-bit counters, each with a clock of 1.19 MHZ . The 1.19 MHZ clock is generated from 14 MHZ divided by 12. All modes of operation are software programmable.

## Programmable Interrupt

This section of the KTIF custom integrated circuit is equivalent to an Intel 8259A that capable of handling eight-vector priority interrupt, individual request mask and programmable interrupt modes. This circuit generates INTR output signal for the CPU. In addition, the sharing interrupt logics are implemented in the design for $I R Q 1$ (between keyboard and real time clock interrupt)

## Address Decoding Logic

This section contains 3 to 8 address decode to generate Programmable Interrupt Chip select, Programmable Interval Timer chip select, Floppy Disk chip select (FDCCHP*) and Programmable Peripheral Interface address of three decoded address $A, B$ and $C$. (see IO signal definition). The FDC port is enabled by Planar register-port 0065 hex bit 3 when bit 3 is HIGH.

## BLOCK DIAGRAM



## INPUT/OUTPUT PIN DESCRIPTIONS



| \# | Signal Output Current | Pin <br> Number | Type | Descriptions |
| :---: | :---: | :---: | :---: | :---: |
| 18 | 14 MHZ | 27 | I | $\begin{aligned} & \text { Clock signal } \\ & 14.318 \mathrm{MHZ} \end{aligned}$ |
| 19 | $\begin{aligned} & \text { BUSY }{ }^{8 m a} \\ & \text { (O.C., Pull_up) * } \end{aligned}$ | 58 | 0 | Keyboard busy When High |
| 20 | KYBDTYP <br> (Pull-up) | 61 | I | Keyboard type select. When High, selects IBM PC keyboard. When Low selects Tandy keyboard |
| 21 | PPITM 2ma | 12 | 0 | Programmable <br> Peripheral <br> Interface Timer <br> output signal <br> for sound generator. |
| 22 | KBDDATA 8ma (3-state) | 60 | I/O | Input data signal from keyboard. <br> In the IBM PC keyboard this pin is used as an output to hold the data Low. |
| 23 | $\begin{aligned} & \text { KBDCLK } \\ & \text { (3-state) } \end{aligned}$ | 59 | 1/0 | Input clock signal from keyboard. In the IBM PC keyboard this pin is used as an output to hold the clock LOW. |
| 24 | $\begin{array}{cc} \text { DS0B } & 8 \mathrm{ma} \\ \text { (O.C. Pull_up*) } \end{array}$ | 57 | 0 | Drive select signal When Low. |
| 25 | $\begin{array}{cc} \text { DSlB } & 8 \mathrm{ma} \\ \text { (O.C. Pull_up*) } \end{array}$ | 56 | 0 | Drive select signal when is LOW. |
| 26 | $\begin{array}{cc} \text { DS2B } & \text { 8ma } \\ \text { (O.C. Pull_up*) } \end{array}$ | 55 | 0 | Drive select signal when is LOW. |
| 27 | $\begin{aligned} & \text { DCB } \\ & \text { (Pull_up) } \end{aligned}$ | 38 | I | Disk change signal when is LOW |


| \# | Signal | Output Current | Pin <br> Number | Type | Descriptions |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 28 | DRATE (O.C.) | 16 ma | 54 | 0 | Data rate select signal. When is LOW, 500 kbps is selected. When is HIGH 250 kbps is selected. |
| 29 | FDCRST | 4ma | 46 | 0 | FDC reset signal to the FDC controller when is HIGH. |
| 30 | RESET |  | 66 | I | System reset input signal when is HIGH. |
| 31 | MTRONB (O.C.) | 16 ma | 53 | 0 | Floppy disk motor ON output signal when is LOW. |
| 32 | FDHSELI |  | 42 | I | Head select input signal from floppy disk controller. |
| 33 | $\begin{aligned} & \text { FDHSELOB } \\ & \text { (O.C.) } \end{aligned}$ | 16 ma | 49 | 0 | Head select input signal for floppy drives when is LOW. |
| 34 | FDWREI |  | 41 | I | Write enable input signal from floppy disk controller. |
| 35 | FDWREOB (O.C.) | 16 ma | 50 | 0 | Write enable output signal for floppy drives when is LOW |
| 36 | FDDIRI |  | 40 | I | Head travel direction input signal from FDC controller. |
| 37 | $\begin{aligned} & \text { FDD IROB } \\ & \text { (O.C.) } \end{aligned}$ | 16 ma | 51 | 0 | Head travel direction for floppy drive. |
| 38 | FDCCHPB | 4ma | 47 | 0 | FDC chip select output signal or FDC controller when is LOW. |


| * | Signal | Output Current | Pin <br> Number | Type | Descriptions |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 39 | FDACKIB |  | 44 | I | ```FDC controller acknowledge output signal when is LOW.``` |
| 40 | FDACKOB | 2ma | 45 | 0 | FDC controller acknowledge output signal when is LOW. |
| 41 | FDCINT |  | 39 | 1 | ```Floppy disk interrupt input signal when is HIGH.``` |
| 42 | FDCDMRQ |  | 43 | I | Floppy disk service request input signal to DMA when is LOW |
| 43 | FDCDRQ | 4ma | 48 | 0 | Floppy disk service request output signal to the DMA when is LOW |
| 44 | $\begin{aligned} & \text { I RQ2 } \\ & \text { (S.T. } \end{aligned}$ | up) | 5 | I | Interrupt request 2 input signal |
| 45 | $\begin{aligned} & \text { IRQ3 } \\ & \text { (S.T. } \end{aligned}$ | 1_up) | 6 | I | Interrupt request 3 input signal |
| 46 | $\begin{aligned} & \text { I RQ4 } \\ & \text { (S.T. } \end{aligned}$ | 1_up) | 7 | I | Interrupt request 4 input signal |
| 47 | $\begin{aligned} & \text { IRQ5 } \\ & \text { (S.T. } \end{aligned}$ | _up) | 8 | I | Interrupt request 5 input signal |
| 48 | $\begin{aligned} & \text { IRQ6 } \\ & \text { (3-state } \end{aligned}$ | Pull_up) | 10 | I | Interrupt request 6 input/output signal |
| 49 | $\begin{aligned} & \text { IRQ7 } \\ & \text { (S.T. Pu } \end{aligned}$ | L_up) | 9 | I | Interrupt request 7 input signal |


| \# | Signal | Output Current | Pin <br> Number | Type | Descriptions |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 50 | INTAB |  | 4 | I | Interrupt acknowledge signal. This signal is used to enable interrupt vector data onto the data bus by a sequence of interrupt acknowledge pulses issued by the CPU |
| 51 | INTR | 2ma | 11 | 0 | Interrupt request signal. This signal is used to interrupt the CPU when HIGH |
| 52 | RTC I NTB (S.T. P | 1_up) | 2 | I | Real time clock interrupt signal from the Real Time Clock device when LOW. |
| 53 | NOVDI |  | 26 | I | NOV RAM data in sigñal |
| 54 | NOVCE | 2ma | 23 | 0 | NOV RAM chip enable signal |
| 55 | NOVDO | 2ma | 25 | 0 | NOV RAM data out sigñal |
| 56 | NOVCK | 2 ma | 24 | 0 | NOV_RAM clock |
| 57 | RFRSHB |  | 3 | I | DMA acknowledge signal from 8237. This signal is active HIGH |
| 58 | REFREQ | 2 ma | 13 | 0 | DMA Request signal fol 8237. This signal is active HIGH |
| 59 | COL/MON |  | 68 | I | Input configuration control signal |
| 60 | PARCHK |  | 67 | I | Parity Check input signal |


\# Signal | OutputCurrent Nin <br>  Typer Descriptions |
| :--- |


| 61 | VCC | 1 | Power | supply |
| :---: | :---: | :---: | :---: | :---: |
| 62 | VCC | 35 | Power | supply |
| 63 | GND | 52 | Groun |  |
| 64 | GND | 18 | Groun |  |
| 65 | not used | 62 |  |  |
| 66 | not used | 63 |  |  |
| 67 | not used | 64 |  |  |
| 68 | not used | 65 |  |  |
| Notes:- O.C. = Open Collector |  |  |  |  |
| - 3-State $=$ Tris State |  |  |  |  |
| - S.T. = Schmitt Trigger |  |  |  |  |
| - * = Max. $\quad \begin{aligned} & \text { ( } 6 \mathrm{ma}, \mathrm{Min.}=0.4 \mathrm{ma} \text { sinking current. } \\ & \text { These signals must have external termination. }\end{aligned}$ |  |  |  |  |
|  |  |  |  |  |

## I/O MAPS

I/O Signal Definition:

| C | B | A | Address Range Hex | Function |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0020-0027 | Interrupts |
| 0 | 0 | 1 | 0040-0047 | Timer |
|  |  |  | 00C0 - 00C7 | Sound |
| 0 | 1 | 0 | 0060-0067 | PPI |
|  |  |  | 0065 | Planar Register |
| 0 | 1 | 1 | 03F0-03F7 | Floppy |
| 1 | 0 | 0 | 0200-0207 | Joystick |
| 1 | 0 | 1 | 0378-037F | Printer |
|  |  |  | 037 C | novram |
|  |  |  | 03F8-03FF | Serial |
| 1 | 1 | 0 | FFE8 - FFEF | Non IBM compatible |
| 1 | 1 | 1 | - | Inactive |

## Register Definition:



| Address Range Hex | Bit | Description |
| :---: | :---: | :---: |
| 0062 - Port C |  | Read/Write |
|  | 0 | $\mathrm{R} / \mathrm{W}$ not used |
|  | 1 | R/W not used |
|  | 2 | R/W not used |
|  | 3 | R/W $0=$ slow speed |
|  | 4 | Read NOVDI |
|  | 5 | Read output Timer \#2 |
|  | 6 | Read 0=color |
|  | 7 | Read l=Parity check |
| 0063-0064 |  | Port not used |
| Planar Control |  |  |
| 0065 | 0 | Planar Register Read/Write |
|  | 1 | Reserved |
|  | 2 | Reserved |
|  | 3 | l=FDC chip select enable |
|  | 4 | Reserved |
|  | 5 | Reserved |
|  | 6 | Reserved |
|  | 7 | Reserved |
| 0066 |  | Not Used |
| 0067 |  | Port D not used |
| Non Volatile Memory Access |  |  |
| 037C |  | Non-volatile memory write only |
|  | 0 | NOVDO |
|  | 1 | NOVCE |
|  | 2 | NOVCLK |
|  | 3 | Reserved |
|  | 4 | Reserved |
|  | 5 | Reserved |
|  | 6 | Reserved |
|  | 7 | Reserved |

Floppy Disk Control

| Address Range Hex | Bit | Description |
| :---: | :---: | :---: |
| 03 FO |  | Not used |
| 03 Fl |  | FDC Mode Control |
|  | 0 | Not used |
|  | 1 | Write - Drive Select switch $0=0-0$ 1-1 |
|  |  | $1=0-1 \quad 1-0$ |
|  | 2 | Not used |
|  | 3 | Not used |
|  | 4 | Not used |
|  | 5 | Not used |
|  | 6 | Not used |
|  | 7 | Not used |
| 03 F 2 |  | FDC Digital Output Register (DOR) |
|  |  | Write Only |
|  |  | DS0 DS1 DS2 |
|  | 0 | Write - 0 - 10 |
|  | 1 | Write - 0 0 1 |
|  | 2 | Write - FDC reset |
|  | 3 | Write - Enable DMA Req/Int. |
|  | 4 | Write - Drive 0 Motor ON |
|  | 5 | Write - Drive 1 Motor ON |
|  | 6 | Not used |
|  | 7 | Not used |
| 03 F 3 |  | Not used |
| 03 F 4 |  | FDC chip select |
| 03 F 5 |  | FDC chip select |
| 03 F 6 |  | Not used |
| $03 \mathrm{F7}$ |  | FDC Data Rate Selection |
|  | 0 | Not used |
|  | 1 | Write - Data Rate |
|  |  | $0=500 \mathrm{~K}$ bits per second |
|  |  | $1=250 \mathrm{~K}$ bits per second |
|  | 2 | Not used |
|  | 3 | Not used |
|  | 4 | Not used |
|  | 5 | Not used |
|  | 6 | Not used |
|  | 7 | $0=$ Disk Change |

## System Configuration Register

| FFEB |  | $\begin{aligned} & 0 \\ & 1 \\ & 2 \\ & 3 \\ & 4 \\ & 5 \\ & 6 \end{aligned}$ <br> 7 | Non IBM Compatible Read/Write <br> Reserved <br> Reserved <br> Reserved <br> Reserved <br> Reserved <br> Read l=Keyboard Interrupt <br> Read l=Real Time Clock <br> Interrupt <br> Write $1=$ Enable Real Time clock <br> Interrupt <br> Read Keyboard Select <br> $0=T a n d y$ Keyboard <br> l=101 Enhanced Keyboard |
| :---: | :---: | :---: | :---: |
| Summary on the active/float data |  |  | bits. (READ ONLY) |
| Address | Net Name | Active Bits | Float Bits |
| 0065 | CSEN | XD3 | XD0-XD2, XD4-XD7 |
| 03 F 7 | FDMDRDB | XD7 | XDO-XD6 |
| FFEB | CDENRDB | XD5 | XD0-XD4 |
|  |  | XD6 |  |
|  |  | XD7 |  |

ELECTRICAL SPECIFICATIONS


## KEYBOARD TIMING SPECIPICATIONS



I/O Write Cycle


I/O Read Cycle
Address Cs.


## FLOPPY DISK TIMING SPECIFICATIONS:




I/O Write Cycle
Address Cs.
PROGRAMMABLE INTERRUPT TIMING AND DESCRIPTIONS
Must meet Intel 8259A. Any differences must be specified.
PROGRAMMABLE TIMER TIMING AND DESCRIPTIONS
Must meet Intel 8254-5.
Any differences must be specified.

## ADDRESS PORT EQUATIONS



## Dallas Semiconductor TimeChip

## FEATURES

- TimeChip keeps track of hundreaths of seconds, seconds, minutes, hours, days, date of the month, months, and years
- Adjusts for months with fewer than 31 days
- Leap year automatically corrected
- No address space required
- Provides nonvolatile controller functions for battery backing up RAM
- Supports redundant batteries for highrel applications
- Uses a 32.768 KHz watch crystal
- Full $10 \%$ operating range
- Operating temperature range $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$
- Space saving 16 -pin DIP package


## PIN CONNECTIONS



## PIN NAMES

Pins 1 \& 2- $X_{1}, X_{2}-32.768 \mathrm{KHz}$ Crystal

| 3 | $\overline{W E}$ | En |
| :---: | :---: | :---: |
| Pin 4 | - BAT1 | - Battery 1 Input |
| Pins 5 \& 8 | - GND | - Ground |
| Pin 6 | D | - Data In |
| Pin 7 | Q | - Data Out |
| Pin 9 | $\cdot \frac{R O M /}{\text { RAM }}$ | - ROM-RAM Select |
| Pin 10 | - CEO | - Chip Enable Out |
| Pin 11 | - CEI | - Chip Enable Input |
| Pin 12 | - $\quad$ E | - Output Enable |
| Pin 13 | RST | - Reset |
| Pin 14 | - BAT2 | - Battery 2 Input |
| Pin 15 | - Vcco | - +5V Output |
| Pin 16 | - VCCI | - +5V DC Input |

## DESCRIPTION

The DS1215 is a combination of a CMOS timekeeper and a nonvolatile memory controller. In the absence of power, an external battery maintains the timekeeping operation and provides power for a CMOS static RAM. The watch provides hundredths of seconds, seconds, minutes, hours, day, date, month, and year information, while the nonvolatile controller supplies all the necessary support circuitry to convert a CMOS RAM to a nonvolatile memory. The DS1215 can be interfaced with elther RAM or ROM without leaving gaps in memory.
The last date of the month is automatically adjusted for months with less than 31 days, including correction for leap year every four years. The watch operates in one of two formats: a 12-hour mode with an AM/PM indicator, or a 24 -hour mode.

The nonvolatile memory controller portion of the circuit is designed to handle power fail detection, memory write protection, and battery redundancy. In short, the controller changes standard CMOS memories into nonvolatile memories, and provides continuous power to the TimeChip. Alternatively the TimeChip can be used with ROM memory by controlling the Chip Enable Output signal (CEO) while the TimeChip is being accessed.

## OPERATION

The block diagram of Figure 3 illustrates the main elements of the TimeChip. Communication with the TimeChip is established by pattern recognition of a serial bit stream of 64 bits which must be matched by executing 64 consecutive write cycles containing the proper data on Data In (D). All accesses which occur prior to recognition of the 64 -bit pattern are directed to memory via the Chip Enable Output pin (CEO).
After recognition is established, the next 64 read or write cycles either extract or update data in the TimeChip and Chip Enable Output remains high during this time, disabling the connected memory.
Data transfer to and from the timekeeping function is accomplished with a serial bit stream under control of chip enable ( $\overline{\mathrm{CEI}})$, output enable ( $\overline{\mathrm{OE}}$ ), and write enable ( $\overline{\mathrm{WE}}$ ). Initially, a read cycle using the $\overline{\mathrm{CEI}}$ and $\overline{O E}$ control of the TimeChip starts the pattern recognition sequence by moving a pointer to the first bit of the 64 -bit comparison register. Next, 64 consecutive write cycles are executed using the $\overline{C E I}$ and $\overline{W E}$ control of the TimeChip. These 64 write cycles are used only to gain access to the TimeChip.
When the first write cycle is executed, it is ccmpared to bit 1 of the 64 -bit comparison register. If a match is found, the pointer increments to the next location of the comparison register and awaits the next write cycle. If a match is not found, the pointer does not advance and all subsequent write cycles are ignored. If a read cycle occurs at any time during pattern recognition, the present sequence is aborted and the comparison register pointer is reset. Pattern recognition continues for a total of 64 write cycles as described above until all the bits in the comparison register have been matched. (This bit pattern is shown in Figure 1). With a correct match for 64 bits, the TimeChip is enabled and data transfer to or from the timekeeping registers may proceed. The next 64 cycles will cause the TimeChip to either receive data on $D$, or transmit data on $Q$, depending on the level of $\overline{O E}$ pin or the $\overline{W E}$ pin. Cycles to other locations outside the memory block can be interleaved with CEI cycles without interrupting the pattern recognition sequence or data transfer sequence to the TimeChip.
A $32,786 \mathrm{~Hz}$ quartz crystal, Seiko part no. DS-VT-200 or equivalent, can be directly connected to the DS1215 via pins 1 and $2\left(X_{1}, X_{2}\right)$. The crystal selected for use should have a specified load capacitance ( $\mathrm{C}_{\mathrm{L}}$ ) of 6 pF .

## NONVOLATILE CONTROLLER OPERATION

The operation of the nonvolatile controller circuits within the TimeChip is determined by the level of the ROM/RAM select pin. When ROM/RAM is connected to ground, the controller is set in the RAM mode and performs the circuit functions required to make static CMOS RAM and the timekeeping function nonvolatile. First a switch is provided to direct power from the battery inputs or $\mathrm{V}_{\mathrm{CCI}}$ to $\mathrm{V}_{\mathrm{CCO}}$ with a maximum voltage drop of 0.2 volts. The $\mathrm{V}_{\mathrm{CCO}}$ output pin is used to supply uninterrupted power to CMOS static RAM. The DS1215 also performs redundant battery control for high reliability. On power fail the battery with the highest voltage is automatically switched to $\mathrm{V}_{\mathrm{CC}}$. If only one battery is used in the system, the unused battery input should be connected to ground. The DS1215 provides the function of safeguarding the TimeChip and RAM data by power fail detection and write protection. Power fail detection occurs when $\mathrm{V}_{\mathrm{CC}}$ falls below VTP which is equal to $1.26 \times \mathrm{V}_{\text {BAT }}$. The DS1215 constantly monitors the $\mathrm{V}_{\mathrm{CC}}$ supply pin. When $\mathrm{V}_{\mathrm{CC}}$ is less than VTP, a comparator outputs a power fail signal to the control logic. The power fail signal forces the chip enable output (CEO) to $\mathrm{V}_{\mathrm{C}}$ I or $\mathrm{V}_{\mathrm{BAT}}-0.2$ volts for external RAM write protection. During nominal supply conditions, CEO will track CEI with a maximum propagation delay of 20 ns . Internally, the DS1215 aborts any data transfer in progress without changing any of the TimeChip registers and prevents future access until $\mathrm{V}_{\mathrm{CCI}}$ exceeds VTP. A typical RAM/TimeChip interface is illustrated in Figure 4.
When the ROM/RAM pin is connected to $\mathrm{V}_{C C O}$, the controler is set in the ROM mode. Since ROM is a read-only device which retains data in the absence of power, battery backup and write protection is not required. As a result, the chip enable logic will not force CEO high when power fails. However, the TimeChip does retain the same internal nonvolatility and write protection as described in the RAM mode. In addition, the chip enable output is set at a low level on power fail as VCCI falls below the level of VBAT. A typical ROM/TimeChip interface is illustrated in Figure 5.

## TIMECNIP COMPARISON REGISTER DEFINITION Figure 1

Byte 0


Byte 1


Byte 2


Byte 3


Byte 4


Byte 5


Byte 6
 A3

Byte 7

| 0 | 1 | 0 | 1 | 1 | 1 | 0 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $5 C$ |  |  |  |  |  |  |  |

## Moter

The pattern recognition in Hex is C5, 3A, A3, 5C, C5, 3A, A3, 5C. The odds of this pattern being accidentally duplicated and causing inadvertent entry to the TimeChip is less than 1 in $10^{: 9}$.

## TIMECHIP REOISTER IMFORMATION

The TimeChip information is contained in 8 registers of 8 bits each which are sequentially accessed one bit at a time after the 64-bit pattern recognition sequence has been completed. When updating the TimeChip registers, each must be handled in groups of 8 bits. Writing and reading individual bits within a register could produce erroneous results. These readl write registers are defined in Figure 2.
Data contained in the TimeChip registers are not binary coded decimal format (BCD) in 12 -hour mode. Reading and writing the registers is always accomplished by stepping through all 8 registers, starting with bit 0 of register 0 and ending with bit 7 of register 7.

## TIMECHIP REOISTER DEFINITION



## AM-PM/12/24 MODE:

Bit 7 of the hours register is defined as the 12 - or 24 -hour mode select bit. When high, the 12 -hour mode is selected. In the 12 -hour mode, bit 5 is the AM/PM bit with logic high being PM. In the 24 -hour mode, bit 5 is the second $\mathbf{1 0}$-hour bit (20-23 hours).

## OScILLATOR AND RESET BITS

Bits 4 and 5 of the day register are used to control the reset and oscillator functions. Bit 4 controls the reset pin (Pin 13). When the reset bit is set to logical 1, the reset input pin is ignored. When the reset bit is set to logical $\mathbf{0}$, a low input on the reset pin will cause the TimeChip to abort data transfer without changing data in the timekeeping registers. Reset operates independently of all other inputs. Bit 5 controls the oscillator. When set to Logic 0 the oscillator turns on and the watch becomes operational.

## 2140 aits

Registers $1,2,3,4,5$, and 6 contain one or more bits which will always read logical 0 . When writing these locations, either a logical 1 or 0 is acceptable.

TIMECNIP BLOCK DIAGRAM Figure 3


## RAMITIMECHIP INTERFACE Figure 4



## ROM/TIMECHIP INTERFACE Figure 5



## ABsOLUTE MAXIMUM RATINOS*

Voltage on any Pin Relative to Ground -1.0 V to +7.0 V
Operating Temperature $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$
Storage Temperature $\quad-55^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C}$
Soldering Temperature $260^{\circ} \mathrm{C}$ for 10 Sec
*This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operation sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

RECOMMENDED D.C. OPERATINQ CONDITIONS $10^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ )

| PARAMETER | SYMBOL | MIN | TYP | max | UNITS | NOTES |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply Voltage | $V_{C C}$ | 4.5 | 5.0 | 5.5 | V | 1 |
| Logic 1 | VIH | 2.2 |  | $\mathrm{V}_{\text {CC }}+0.3$ | v | 1 |
| Logic 0 | VIL | -0.3 |  | + 0.8 | V | 1 |
| VBAT1 or VBAT2 Battery Voltage | VBAT | 2.5 |  | 3.7 | V | 7 |

D.C. ELECTRICAL GHARACTERISTICS
$10^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=4.5$ to 5.5 V$)$

| Supply Current | ICCI |  |  | 5 | mA | 6 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply Current | ICCO1 |  |  | 80 | mA | 8 |
| VCCO = VCCI 0.2 |  |  |  |  |  |  |

$\left(0^{\circ} \mathrm{C}\right.$ to $\left.70^{\circ} \mathrm{C}, \mathrm{V} C \mathrm{CC} 4.5 \mathrm{~V}\right)$

| CEO Output | VOH1 | VCCI or <br> VBAT-0.2 |  |  | V | 9 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| VBAT1 or VBAT2 <br> Battery Current | IBAT |  |  | 1 | $\mu \mathrm{~A}$ | 6 |
| Battery Backup Current <br> QVCO $=\mathrm{V}_{\text {BAT }}-0.2 \mathrm{~V}$ | ICCO2 |  |  | 10 | $\mu \mathrm{~A}$ | 10 |

CAPACITANCE ( $\mathrm{t}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ )

| PARAMETER | SYMBOL | MIN | UNITS | NOTES |
| :--- | :---: | :---: | :---: | :---: |
| Input <br> Capacitance | $\mathrm{C}_{\text {IN }}$ | 5 | pF |  |
| Output |  |  |  |  |

A.C. ELECTRICAL CHARACTERISTICS ROM $/ \overline{\operatorname{RAM}}=\mathrm{GND}\left(0^{\circ} \mathrm{C}\right.$ to $70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=4.5$ to 5.5 V$)$

| Parameter | SYMBOL | MIN | TYP | max | UNITS | NOTES |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Read Cycle Time | $t_{\text {R }}$ | 250 |  |  | ns |  |
| $\overline{C E I}$ Access Time | tco |  |  | 200 | ns |  |
| $\overline{\mathrm{OE}}$ Access Time | toe |  |  | 100 | ns |  |
| $\overline{\text { CEI To Output Low } Z 2}$ | tcoe | 10 |  |  | ns |  |
| $\overline{\text { OE }}$ To Output Low $Z$ | toee | 10 |  |  | ns |  |
| $\overline{\text { CEI To Output High } Z}$ | tod |  |  | 100 | ns |  |
| $\overline{O E}$ To Output High Z | todo |  |  | 100 | ns |  |
| Read Recovery | trR | 50 |  |  | ns |  |
| Write Cycle | twc | 250 |  |  | ns |  |
| Write Pulse Width | twp | 170 |  |  | ns |  |
| Write Recovery | tWR | 50 |  |  | ns | 4 |
| Data Set Up | tDS | 100 |  |  | ns | 5 |
| Data Hold Time | tDH | 10 |  |  | ns | 5 |
| CEI Pulse Width | tcw | 170 |  |  | ns |  |
| $\overline{\text { RST Pulse Width }}$ | trist | 200 |  |  | ns |  |
| $\overline{C E I}$ Propagation Delay | tPD | 5 | 10 | 20 | ns | 2, 3 |
| $\overline{\text { CEI High to Power Fail }}$ | tPF |  |  | 0 | ns |  |


| Recovery at Power Up | tREC $^{4}$ |  |  | 2 | ms |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| VCC Slew Rate $4.5-3.0 \mathrm{~V}$ | $\mathrm{t}_{\mathrm{F}}$ | 0 |  |  | ms |  |

TIMINQ DIAORAM-READ CYCLE TO TIMECHIP ROM/ $\overline{R A M}=$ GND


TIMING DIAGRAM-WRITE CYCLE TO TIMECHIP ROM/ $\overline{R A M}=$ GND

A.C. ELECTRICAL CHARACTERISTIGS ROM/ $\overline{R A M}=V_{C C O}\left(0^{\circ} \mathrm{C}\right.$ to $\left.70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V} \pm 10 \%\right)$

| PARAMETER | SYMBOL | MIN | TYP | max | UNITS | NOTES |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Read Cycle Time | $\mathrm{t}_{\mathrm{RC}}$ | 250 |  |  | ns |  |
| $\overline{C E I}$ Access Time | tCO |  |  | 200 | ns |  |
| $\overline{\mathrm{OE}}$ Access Time | toE |  |  | 200 | ns |  |
| $\overline{\text { CEI }}$ to Output in Low Z | tcoe | 10 |  |  | ns |  |
| $\overline{\mathrm{OE}}$ to Output in Low Z | toEE | 10 |  |  | ns |  |
| $\overline{\mathrm{CEI}}$ to Output in High Z | tOD |  |  | 100 | ns |  |
| $\overline{\mathrm{OE}}$ to Output in High 2 | tODO |  |  | 100 | ns |  |
| Address Set Up Time | tAS | 20 |  |  | ns |  |
| Address Hold Time | ${ }^{\text {taH }}$ |  |  | 10 | ns |  |
| Read Recovery | tRR | 50 |  |  | ns |  |
| Write Cycle Time | twC | 250 |  |  | ns |  |
| $\overline{\mathrm{CE}}$ Pulse Width | tcw | 170 |  |  | ns |  |
| $\overline{\text { OE Pulse Width }}$ | tow | 170 |  |  | ns |  |
| Write Recovery | tWR | 50 |  |  | ns | 4 |
| Data Set Up Time | tDS | 100 |  |  | ns | 5 |
| Data Hold Time | tDH | 10 |  |  | ns | 5 |
| $\overline{\text { RST Pulse Width }}$ | tRST | 200 |  |  | ns |  |
| $\overline{C E I}$ Propagation Delay | tPD | 5 | 10 | 20 | ns | 2,3 |
| $\overline{\text { CEI High to Power Fail }}$ | tPF |  |  | 0 | ns |  |

$\left(0^{\circ} \mathrm{C}\right.$ to $\left.70^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}<4.5 \mathrm{~V}\right)$

| Recovery at Power Up | t REC |  |  | 2 | ms |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Vcc Slew Rate 4.5-3V | $\mathrm{t}_{\mathrm{F}}$ | 0 |  |  | ms |  |

TIMING DIAQRAM-READ CYCLE ROM/ $\overline{\operatorname{RAM}}=\mathrm{VCCO}$


TIMING DIAQRAM-WRITE CYCLE $\quad$ ROM $/ \overline{R A M}=V_{C C O}$


## TIMING DIAGRAM-POWER DOWN



## TIMING DIAGRAM-POWER UP



TIMANG DIAGRAM-RESET FOR TIMECHIP


## nOTES

1. All voltages are referenced to ground.
2. Measured with load shown in Figure 6.
3. Input pulse rise and fall times equal 10 ns .
4. tWR is a function of the latter occurring edge of $W E$ or $\overline{C E}$ in RAM mode or $\overline{O E}$ or $\overline{C E}$ in ROM mode.
5. tDH and tDS $^{2}$ are functions of the first occurring edge of $\overline{W E}$ or $\overline{C E}$ in RAM mode or $\overline{O E}$ or $\overline{C E}$ in ROM mode.
6. Measured without RAM connected.
7. Trip point voltage for power fail detect.
$V_{T P}=1.26 \times V_{\text {BAT }} \quad$ For $10 \%$ operation $V_{B A T}=3.5 \mathrm{~V}$ max.; for $5 \%$ operation $V_{B A T}=3.7 \mathrm{~V}$ max.
8. ICCO1 is the maximum average load current the DS1215 can supply to memory.
9. Applies to $\overline{\mathrm{CEO}}$ with the ROM/ $\overline{\mathrm{RAM}}$ pin grounded. When the ROM/RAM pin is connected to VCCO, CEO will go to a low level as $\mathrm{V}_{\mathrm{CCI}}$ falls below $\mathrm{V}_{\text {BAT }}$.
10. I $\mathrm{CCO}_{2}$ is the maximum average load current which the DS1215 can supply to memory in the battery backup mode.
11. Applies to all input pins except $\overline{\operatorname{RST}}$. $\overline{\mathrm{RST}}$ is pulled internally to VCCI -

## OUTPUT LOAD Figure 6

D.U.T.

DS1215
TimeChip


| DIM. | INCHES |  |
| :---: | :---: | :---: |
|  | MIN. | MAX. |
| A | .740 | .780 |
| B | .240 | .260 |
| C | .120 | .140 |
| D | .290 | .310 |
| E | .020 | .030 |
| F | .110 | .130 |
| G | .090 | .110 |
| H | .300 | .350 |
| J | .008 | .012 |
| K | .015 | .021 |



OFIAMIDMA CONTFIDL GHIF

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Tandy $1000 T L$ DFAM/DMA Control IG comsista of the following functional blocks:

- FOM/DFAM Decode Latch and Control signals
- Fage Fegisters
- Memory Address Multiplexer


## ROM/DRAM Decode Latch and Control signal

The FOM/DFAM Decode Latch block contains the circuitry ta decode the CFU's Address bus Al7 - Azs and to provide the necessary latched signals for controllimg ROME and DFAMS. CFU's Address Al7 - A23 together with MCO - MCZ determine which sepment (bank) of memory is being selected based on one of six possible memory configurations. ©see memory map figure 1.). The memory configurations ramging from 25e kila bytes to 1.5 mega bytes.

Additional support for memory refresh is also provided in this block. During a Fefresh cyele, the assertion of REFFESH: will cause the circuitry to igmore the current address inputs and
 MEMCYC and FRCLK signals are used for controlling the start of a memory access and timing for all FAS×is, MAO - MAS and CAS×*s signals. (see Timing Diagram figure z.).

FUMCS* is decoded from Al7 - A23 and latched by ALE when HLDA is active. The ROM address ranges from oEOOOOh to oFFFFFH for the low address, EEOOOOh to EFFFFFh and FEOOOOh to FFFFFFh for high address. This output signal is asserted when any one of the three address ranges is detected and REFRESHi: is inactive.

The two outputs LMEGCSi and MDEEN: are intended to be used as memory buffer enable signal. LMEGCSi is active whenever any memory access is made to an address below oloogon or when FEFRESH: is active. MDEEN: is memory tata bus buffer and becomes active whenever CAS×*s or FDMCS\% is active.

The ROM/DFAM Decode's internal latch is controlled by two input. signals - HLDA and ALE. During a CFU Memory Cycle, ALE will enable the RAM Decode latch and allow the input from decoder to be transferred to the output pins. When ALE goes inactive, the decoder outputs is latched for the remainder of the cyele. Asserting HLDA will enable the decoder outputs to the output pins and force fomCsi inactive. This block has one output signal that is unlatched. This output, AFlEs, is intended to be used by external circuitry as an indication that a le-bit memory transfer is taking place.

## Fage Fiegister

At the time the EqGOTA-5 - OMA Controller takes control of the aburess bus, the first operation comes in two bytes. The first. byte is a lower address bus esao through saty that is fut directly on the s address bus by DMA controller. The eecond byte is a upper address (SAE through SAls) that is on its date outputs, to be latetied in the $74 A L S S 7 E$ internally by Aduress Strobe (AS) signal from DMA Gontroller.

Two 4 by 4 registers are useg to perform Fage Fiegister function. During DMA Eus Cycle, the read functian is controlled by the DMA Fequest Acknowledge signals - DACtza and DACtsi in conjunctiom with HLDA: Ematies the FFage Fiegister to be output as the upper address (SAle and Al7 tromugh Azs). The write function is controlled by EAO to SAS in conjunction with FGFEGWFis to latch data bits - XDO to XD7 into Fage Fegister.

## Address Multiplexer

The Memory Address Multiplexer is used to provide the Fow Address or Columin Address and refresh counter that is required by Dynamic Fams. Additionally, it provides the drive and buffering capability for memory address bus MaO to MAE. Ma7 is generated from SAO or SAS which is multiplexed ty REFFESH: signal. The addresses for the memory are multiplexed as shown below.

Equation For Multiplexed Memory Addresses

## Fiow Address (First)

| MAO | SAl | EAS |
| :---: | :---: | :---: |
| MAI | SAZ | SAl0 |
| MAZ | 5 A | SAl1 |
| MAS | SA4 | SAl2 |
| MA4 | SA5 | SA13 |
| MAS | EAE | SAl4 |
| MAE | SA7 | SA15 |
| MA7 | SAE/SAO | SAlE |
| MAS | SAl7 | SA1S |

## 2. MEMOFIY CONFIGURATION

| DFT104 | Mce | MC1 | MCO | Emino | Efus 1 | EANS | CONTFOL | ADDRESS FANGE |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | ¢ | 0 | 0 | 12 ck |  | 12 Ec | Fas 1 | (000000-01FFFF) |
|  |  |  |  |  |  | Fasz | (020000-09FFFF) |
| 2 | 0 | 0 | 1 | 512 c |  |  |  | Reso | (000000-07FFFF) |
| 3 | 0 | 1 | 1 | 512 c |  |  | Faso | (000000-07FFFF) |
|  |  |  |  |  | 12 Ek |  | Fas 1 | (080000-09FFFF) |
| 4 | 0 | 1 | 0 | 512 k |  |  | Faso | (000000-07FFFF) |
|  |  |  |  |  | 12 E |  | Fasi | (0EFFFF-OGFFFF) |
|  |  |  |  |  |  | i2ct | Fase | (100000-17FFFF) |
| 5 | 1 | 1 | 1 | 5126 |  |  | Faso | (000000-07FFFF) |
|  |  |  |  |  | 12 cs |  | Ras! | (080000-09FFFF) |
|  |  |  |  |  |  | 512k | Fasz | (100000-17FFFF) |
| $\epsilon$ | 1 | 1 | 0 | 512 K |  |  | Faso | (000000-07FFFF) |
|  |  |  |  |  | 12 EK |  | Rasl | (080000-09FFFF) |
|  |  |  |  |  |  | 512 c | Fesz | (100000-17FFFF) |
|  |  |  |  |  |  | 512 | Ras3 | (180000-1FFFFF) |
| 7 | 1 | 0 | 1 | NOT | DEFINED |  |  |  |
| 8 | , | 0 | 0 | NOT | DEF INED |  |  |  |

FIGUFE i. MEFTGFY CONFIGURATION.

## 3. DMA Control Loqic Equations



```
    +/1a23 &/1a22 &/1a21 &/1a20 &/1a19 % mcl
    +refresh;
/Erasl =/1a23 &/1az2 &/1a21 &/1a2O &/1alS &/1alS &/1al7 &/mco
    & /mcl & /mc2
    +/1a23 & /1a22 &/1a21 &/1a20 & lal9 &/1a!日 &/1al7 & mcl
    +refresh;
/Eras2 = /1a23 &/1a22 &/1a21 &/1a20 &/1al9 &/1alS & lal7 &/mco
    & /mcl & /mc2
    +/1a23 &/1a22 &/1a21 & 1520 &/1al9 &/1als &/1al7 & /mco & mel
        & /mcz
    +/1a23 & /1a22 & /la21 & 1a20 & /la19 & mcl & mcZ
    +refresh;
/Eras3 = /1a23 & /la22 &/1a21 & lazO & lal马 &/mco & mcl & mcz
    +refresh;
/Eras = /1a23 &/1a22 &/la2T &/1a20 &/1a19 &/1a18 &/mc2 &/mcl
        & /mco & /refresh
    +/la23 & /1a22 & /1a21 & /la20 & /1al9 & /mc2 & mco & /refresh
    +/la23 & /la22 & /1a2l & /la20 & /la19 & mct & irefrest,
    +/la23 & /la2z & /laZl &/1a20 & lalG & /lalS &/Lal7 & mcl
        & /refresh
    +/1a23 &/la2z &/1az1 & lazo &/1al9 & /la1S &/lal7 &/mcz & mcl
        & /mco & /refresh
    +/la23 & /1a22 & /la21 & la20 &/1alЭ & mc2 & mcl & /refresh
    +/1a23 & /1a22 &/1a21 & la20 & mcz & mcl & /mco & /refresh;
/Blmeg = /la23 & /laz2 &/laZl & /lazo+refresh;
/Eromcs =/1a23 &/la22 &/1a21 &/1a20 & lalS & lats & lal7 &/refresh
    +1a23 & 1a22 & 1a21 & la20 & la19& lals & la17 &/refresh
    +1a23 & la22 & la2l & /1a20 & lal9 & lalg & lal7 & /refresti;
```

5


Figure 2. Block Diogram


FIGURE 3. PIN CONFIGURATION

| FIN\# | PIN NAME | TYFE | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| 9 | SAO | INFUT | CFU ADDRESS LINE |
| 8 | 5 Al | INFUT | CFU ADDRESS LINE |
| 7 | SAZ | INFUT | CFU ADDFESS LINE |
| $E$ | SAS | INFUT | CFU ADDRESS LINE |
| 5 | SA4 | INFUT | CFU ADDRESS LINE |
| 4 | 5 A 5 | INFUT | CFU ADDFESS LINE |
| 3 | SAE | INFUT | CFU ADDFESS LINE |
| 2 | SA7 | INFUTT | CFU ADORESS LINE |
| 1 | SAE | INFUT / OUTFUT | CFU ADDFESS LINE |
| 68 | SAF | INFUT/OUTFUT | CFU ADDFESS LINE |
| $E 7$ | SAlO | INFUT/OUTFUT | CFU ADDFESS LINE |
| 6 | EAl1 | TNFUT / DUTFUT | CFU ADDRESS LINE |
| 65 | SA12 | INFUT/OUTFUT | CFU ADDFESS LINE |
| 64 | 5 Al 3 | INFUT/OUTFUT | CFU ADDFESS LINE |
| $E 3$ | SA14 | INFUT/GUTFUT | CFU ADDFESE LINE |
| 62 | SA15 | INFUT/ OUTFUT | CFU ADDFESS LINE |
| 60 | SAlE | INFUT/OUTFUT | CFU ADDRESS LINE |
| 59 | Al7 | INFUT/GUTFUT | CPU ADDRESS LINE |
| 59 | A1S | INFUT/DUTFUT | CFU ADDFESS LINE |
| 57 | A19 | INFUT/DUTFUT | CFU ADDFESS LINE |
| $5 E$ | A20 | INFUT/DUTFUT | CFU ADDFESS LINE |
| 55 | A 21 | INFUT / OUTFUT | CFU ADDRESS LINE |
| 54 | A22 | INFUT/OUTFUT | CFU ADDFESE LINE |
| 53 | A23 | INFUT/GUTFUT | CFU ADDRESE LINE |
| 45 | ALE | INFUT | ADDFESS LATCH ENAELE Active HIGH - to latch generated RAS/CAS/LMEG. |
| 43 | FEFFES 5 : | INFUT | FEFFESH - Active LOW to initiate a refresh cycle for Dynamic RAMs. |
| 14 | DACK2* | INFUT | E2S7 Channel 2 DMA ACKNOWLEDGE |
| 16 | DACK3: | INFUT | 3237 Channel 3 DMA ACKNOWLEDGE |
| 13 | HLDA | INFUT | HOLD ACKNOWLEDGE Active HIGH - it indicates that the DMA has the system bus. |
| 44 | FGREGWR* | INPUT | FAGE FEGISTEF WFIITEactive LOW to perform I/O WFITE cycle to the DMA Fage Fiegister. |
| 46 | MEMCYC | INFUT | MEMOFY CYCLE - Active HIGH to initiate FAS/CAS/MAO-MAS outputs. |
| 19 | AS | INFUT | ADORESS STROEE - Active HIGH it latches the adidres 1 ines SAE-SAls Do-07 into external lateti for DMA cycle |


| PIN\# | PIN NAME | TVPE | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| 10 | XBHE* | INFUT | EYTE HIGH ENAELE To enable the high memory data bytes Ds-D15 |
| 47 | FRCLK | INFUT | $1 E M H 2$ CLOCK. |
| 11 | MCO | INFUT | MEMORY CONFIGURATION SELECT LSE. |
| 12 | MC1 | INFUT | MEMORY CONFIGURATION SELECT. |
| 42 | MCL | INFUT | MEMORY CONFIG. SELECT MSE |
| 20 | MAO | OUTFUT | MULTIFLEXED MEMORY ADDFESS - addressing required for DFAM memory. |
| 22 | MAI | DUTFUT | MULTIFLEXED MEMOFY ADDFESS - adyressing required for DFAM memory. |
| 24 | MAZ | UUTFUT | MULTIFLEXED MEMORY ADDFESS - addressing required for DRAM memery. |
| 26 | MA3 | OUTFUT | MULTIFLEXED MEMORY ADDRESS - addressing required for DFAM memory. |
| 29 | MA4 | OUTFUT | MULTIPLEXED MEMORY ADDRESS - addressing required for DRAM memory. |
| 31 | MAS | QUTFUT | MULTIFLEXED MEMOFY ADDRESS - addressing required for DRAM memory. |
| 33 | MAE | OUTFUT | MULTIFLEXED MEMDFY ADDFESS - addressing required for DRAM memory. |
| 35 | MA7 | OUTFUT | MULTIFLEXED MEMDFY ADDRESS - addressing required for DRAM memory. |
| 37 | MAS | OUTFUT | MULTIFLEXED MEMORY ADDFESS - addressing required for DFAM memory. |
| 17 | CASH* | OUTPUT | COLUMN ADDRESS STROBE HIGH - Active LOW, it's used to select the high data byte MDE-MD15 for a DRAM access cycle. |


| PIN\# | PIN NAME | TYPE | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| 15 | CASL* | OUTFUT | COLUMN ADDRESS STROEE LOW - Active LOW, it's used to select low data tyte MDO-MD7 for a DRAM |
| 36 | RAEO* | OUTFUT | access cycle. <br> ROW ADDFESS STROEE o - <br> Active Low, it's used <br> for selecting DFAM <br> Eanko. |
| 39 | FASI: | OUTFUT | FOW ADDRESS STROBE 1 Active LOW, it's used for selecting DFAM Eank 1 |
| 40 | FAS2* | Qutput | FOW ADDRESS STROBE 2 Active LOW, it's used for selecting DRAM Eank 2 |
| 41 | RASS* | GUTFUT | FOW AdDress stroee 3 Active Low, it's used for selecting DRAM Eank 2 |
| 51 | LMEGCS* | QUTPUT | LOWER MEGABYTE CHIP SELECT - Active LOW,it indicates that bellow 1 megabyte |
| 50 | MDEEN* | GUTFUT | MEMOFY DATA BUS ENABLE Active Low it is used to enable the data bus buffer |
| 49 | AF15* | OUTPUT | AFle Active Low - it. signals the control logic (CFU CNTL) that the memory cycle is a 16 bit 1 wait state cycle. |
| 46 | ROMCS* | OUTFUT | ROM CHIF SELECT Active LOW |
| 21 | xDo | Infut | DATA EUS o for the Deripheral bus. |
| 23 | XDI | Infut | DATA EUS $/$ for the peripheral bus. |
| 25 | XD2 | INFUT | DATA EUS 2 for the peripheral bus. |
| 28 | xDS | Infut | DATA EUS 3 for the peripheral bus. |
| 30 | XD4 | Infut | DATA Bus 4 for the peripheral bus. |
| 32 | XD5 | INPUT | DATA EUS 5 for the peripheral bus. |
| 34 | XDE | INFUT | DATA EUS 6 for the peripheral bus. |
| 36 | XD7 | INPUT | DATA EUS 7 for the peripheral bus. |
| $\begin{aligned} & 18,52 \\ & 27,61 \end{aligned}$ | vCc GND |  | $+5 V$ FOWER SUPFLY GROUND |

## 7. TIMING PPECIFICATIONS

| SYM | PARAMETER | MIN | MAX | UNITS | REMARKS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| T1 | FFRCLK Perigu | E\%5 | 250 | กธer |  |
| Tla | PRCLK Low Time | 25 | 125 | กรec |  |
| T1b | FRCLK High time | 25 | 125 | nsec |  |
| T2 | Ale Active Delav | $E$ | 5 | ทisec |  |
| TS | ALE Inactive Delav | 5 | 25 | - 5 ¢c |  |
| T4 | Address Delay | 1 | 60 | nэec |  |
| TS | Latched Address From ALE Active | - | 23 | กระ¢ |  |
| TE | MEMCYC Active Delay From $\downarrow$ FRCLK | 5 | 25 | nsec |  |
| T7 | MEMCYC Inactive Delav From $\downarrow$ PRCLK | 5 | 35 | กระ¢ |  |
| TE | RASx: Active Delav From $\downarrow$ FRRCLK | 5 | 45 | กธะ | at 100pf |
| TS | RASx: Inactive Delay From $\uparrow$ ORCLK | 5 | 30 | nsec |  |
| T10 | RASx* Frectiarye Time | 110 |  | กระ¢ |  |
| T11 | FAEx* Hold Time | 80 |  | nsec |  |
| T12 | CASx* Active Delav Time From $\uparrow$ FRCLK | 5 | 40 | . nsec | at 1ESpf |
| T13 | CASx* Inactive Delav Time From $\downarrow$ MEMCYC | 5 | 30 | nsec |  |
| T14 | Row Adidress Setud Time | 5 |  | nsec |  |
| T15 | Memory Autress Delay Time From SAl to SAle | $\square$ | 50 | กรセ¢ | at 200pf |
| T15 | Memory Address Delav From ALE | 8 | 50 | nsec |  |
| T17 | Memorv Ardress Statile Delav From 4 PRCLK |  | 50 | nsec | at 200pf |
| T1s | Row Address Hold Time | 20 |  | กระ¢ |  |
| T19 | Column Address Setup Time | 5 |  | nsec |  |
| T20 | Column Address Hold Time | 30 |  | nsec |  |
| T21 | Column Address Hold Time Referenced To RAS $\times$ * | 120 |  | nsec |  |
| T22 | ALE $\uparrow$ to MDEEN $\uparrow \downarrow$ A/I |  | 50 | nsec |  |
| T23 | ALE $\uparrow$ to LMEGCS* $\downarrow \uparrow$ A/I |  | 50 | nsec |  |
| T24 | ALE $\uparrow$ to ROMCS* $\downarrow$ t $\mathrm{A} / \mathrm{I}$ |  | 50 | nsec | $\begin{aligned} & \text { CPUHLDA } \\ & \text { HIGH } \end{aligned}$ |


| SYM | PARAMETER | MIN | MAX | UNITS | FEMARKS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| T25 | A17-A2s $\uparrow \downarrow$ to AF1E* $\downarrow \uparrow$ A/ 1 |  | 50 | nsec |  |
| TE | CFUHLDA $\downarrow$ to MDEEN $\uparrow$ I |  | 50 | nsec |  |
| T27 | CFUHLDAft to FOMCSt $\downarrow \uparrow$ ¢/I |  | 50 | nsec |  |
| T28 | CFUHLDAT $\downarrow$ to SAS-GA!S $\downarrow \uparrow$ |  | 50 | ก5ec |  |
| T29 | CFUHLDA¢ to SA1E,A17-A23 $\downarrow$ |  | 50 | nser |  |
| TSO | DACK2*, DACKO末 $\uparrow \downarrow$ to SAle, $\uparrow \downarrow$ A17-A23 |  | 50 | nsec |  |
| T3: |  |  |  |  |  |
| $T 2$ | DACEx- to AFIEA, MDEENF |  |  |  |  |
| T33 | LMEGCSi: |  | 80 | M5ec |  |
| T34 | REFRESH: $\uparrow \downarrow$ to MA> $\downarrow \downarrow / \mathrm{T}$ |  | 50 | nsec | at 200pf |
| T35 | XDO-XD7 Setup to AS $\downarrow$ | 100 |  | nsec |  |
| TSE | XDO-XD7 Setup to FGFEGUFF* | 100 |  | nsec |  |
| T 37 | SAO-SAS Setup to FGREGWF\% | 100 |  | nsec |  |
| TSE | FEFFESHi: $1 \uparrow$ to AFIE: |  | 50 | ทรec |  |
| TS9 | FEFRESH: $\downarrow$ i to MDEEN: |  | 50 | nsec |  |




## 9. ELECTRICAL SPECIFICATIONS

ABSOLUTE MAX RATINGS (NON-DFERATING, VSS=O.OV)

|  | MIN |  | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: |
| ETORAGE TEMPEFATURE | $-E .5$ |  | $+150$ | Degrees C. |
| VOLTAGE ON ANY FIN W.F.T GROUND | -0. 5 |  | 7.0 | Volts |
| OPERATING ELECTRICAL SPECIFICATIONS: |  |  |  |  |
| OPERATING AMEIENT | MIN | TYF | MAX | UNITS |
| AIF TEMF. FiANGE | 0 | 25 | 70 | Degrees $C$ |
| FOWER SUFFLIES |  |  |  |  |
| vCC | 4.5 | 5.0 | 5.5 | Volts |
| VSS | 0 | 0 | 0 | Volts |
|  | MIN | TYP | MAX | UNITS |
| LEAKAGE CURFENT |  |  |  |  |
| Vin $=0.0 \mathrm{~V}$ |  | 20 |  | Microamps |
| $V_{\text {in }}=5.0 \mathrm{~V}$ | $-20$ |  |  | Microamps |

## INPUT VOLTAGES

LOGIC "O" (Vil)
0.8
voIts
LDGIC "1" (Viti) 2.0
volts

## OUTPUT VOLTAGES CURFENT LDADING

```
LOGIC "O" (vol) 0.4 volt.s
MA[O]-MA[\varepsilon] ($ Bma(min)
SXAS-SXA1E,Al7-A2S (0 4marmin)
RAS>* (6 4ma(mim)
CASX未 & E ma(min)
Others must be able
to SINK minimum o 2ma
LOGIC "1" (VOH) 2.4 valts
MA[O]-MA[B],
SXAB-SXA1E,A17-AZS Ema
CAS<<来,RAS<<* @ 4ma
CASX* @ S ma
Others must be able
to DFIVE minimum o Zma
```


## INPUT CAPACITANCE

All infuts 0.0 \& Vin 5.010
picofarads

DUTPUT CAFACITANCE
MA[O]-MA[E]
200
Ficofarads
CAS
165
FiAS 100
SAB-SAlE 150
All other outputs
50
picafarads
Ficofarads Picofarads

Ficロfarads
CPU Control Chip

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## Functional Description

The Tandy 1000 TL CPU Control IC consists of the following function blocks:

-     - Clock Generation, Ready and Reset Control Logic
- Bus Control Logic
-     - Data Conversion Logic
-     - Wait State Control Logic
-     - DMA Arbitration Logic
- Numerical Co-processor Control Logic
- A20Gate and CPU Reset Logic


## Clock Generation, Ready and Reset Control Logic

The clock generation logic includes the clock inputs used to derive all of the system clocks. The 16 MHz clock input is used to generate the CPU clock (PRCLK), and is twice the CPU operation frequency of 8 MHz . The 28.6 MHz clock input is used to generate the 14.31818 MHz clock (OSC) and other clocks required by the system.

Three clocks are generated from the 28.6 MHz clock input. These are OSC, 3.58 MHz , and the 1.19 MHz clock outputs. The 28.6 MHz signal input is divided by 2 to generate the OSC output which is 14.31818 MHz . The 28.6 MHz clock is divided by 8 to generate the 3.58 MHz clock which is used by the sound generator circuit. The 28.6 MHz clock is also divided by 24 to generate the 1.19 MHz clock that is used by the Interval Timer 8254.

All other system clocks are generated from the 16 MHz clock input. PRCLK is used to drive the CPU, the Co-processor, and the DRAM/DMA Control IC. PRCLK is output as 16 MHz in the 8 MHz (FAST) mode and is divided by 2 to 8 MHz in the 4 MHz (SLOW) mode. PRCLK output buffer has sufficient drive capability to meet the 3.8 volt minimum Vih requirement of the $80286 / 80287$ clock inputs.

SCLK and DMACLK are system clocks generated from 16 MHz . SCLK is generated by dividing PRCLK by 2 and DMACLK is generated by dividing PRCLK by 4. Both SCLK and DMACLK are held in a LOW state immediately following a reset and will not start operation until the CPU issues the first bus cycle by asserting S1* LOW. Immediately following $S l^{*}$ asserted LOW, SCLK and DMACLK will make their first LOW to HIGH transition at the falling edge of PRCLK during the start of Tc. This will synchronize SCLK and DMACLK with PRCLK. Refer to timing diagrams for an illustration of the startup of SCLK and DMACLK.

A ready signal (READY*) is generated by the CPU Control IC to allow the CPU to operated with slower devices such as peripherals and slow memory. READY* is synchronized with PRCLR in this control block by the control logic of READY* located in the wait State Control Logic.

Two reset output signals are generated by the CPU Control IC to provide a reset to the main system and a separate reset to the CPU. RESET which is active HIGH is a synchronized reset and is used for reseting the main system. RESCPU is dedicated to the CPU for proper reseting of the 80286. Both RESET and RESCPU are generated when RES* input is asserted LOW to indicate a power-on reset or when RES* is asserted LOW by a reset switch. RESCPU is also generated when a Shutdown condition of the CPU is detected. When a Shutdown condition is detected, RESCPU is asserted HIGH for 16 PRCLK cycles and then negated to assure proper CPU operation. RESCPU can also be generated by doing an I/O write to Port 068 with bit $2=0$. This will generated RESCPU to reset the CPU and can be used to jump from protected mode to real time mode of the CPU.

The RES* input is buffered internal to the CPU Control IC with a Schmitt trigger input buffer. This signal should be held in a LOW state during power-up for at least 10 msec or until all operating voltages have reached their specified range of operation.

## Bus Control Logic

The Bus Control Logic of the CPU Control IC generates several Memory and $I / O$ command and control signals that are issued by the 80286. There are two types of control signals that are generated as output signals. The command outputs are the first type of control signals generated which are decoded from the CPU status inputs $M I / O^{*}, S l^{*}, S O^{*}$. The generated output command signals are MEMR*, MEMW*, IOR*, IOW*, INTA* which determine which type of cycle is to be performed. The second type of signals are the control signals which is ALE, DT/R*, DSDEN0*, DSDEN1*, and MEMCYC. These signals latch the address from the CPU, control the direction and enabling of the data bus buffers, and determine the start of an on board memory cycle. Table l. contains a list of the command output signals that are generated from the CPU status line inputs.

| HI/O* S1* SO* |  |  |  |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Type of Cycle |
| 0 | 0 | 1 | Interrupt Acknowledge |
| 0 | 1 | 0 | I/O Read |
| 0 | 1 | 1 | None; Idle |
| 1 | 0 | 0 | Halt or Shutdown |
| 1 | 0 | 1 | Memory Read |
| 1 | 1 | 0 | Memory Write |
| 1 | 1 | 1 | None; Idle |

Table 1. CPO Control Signal Generation
The state machine that controls the output timing of the command and control sigrals has three bus states, which are the Idle state (Ti), the Status state (Ts), and the command state (Tc). The Idle state (Ti) is generated when the CPU is not actively issuing a bus cycle. During an Idle state, all command and control output signals are in an inactive condition. The beginning of a bus cycle is detected when the CPU asserts Sl* or So*. The state machine will start a Status state (Ts) and will assert ALE to an active HIGH state until the end of Ts. At the end of the status state, the state machine enters the command state (Tc) and will assert the decoded command issued by the CPU. The command signal may be delayed by one half bus cycle or one PRCLK clock cycle, if the conditions exist to produce a command delay. A Command delay will be generated on all $\mathrm{I} / \mathrm{O}$ cycles, all Memory cycles in which both AF16* and MEMCSl6* are negated are inactive HIGH, and on all INTA* cycles.

For the control of the data bus buffer, three control signals are generated which are DT/R*, DSDEN0*, and DSDENl*. DT/R* is used to control the direction of the data bus to and from the CPU. DSDENO* and DSDEN1* are enable signals for the data bus buffers. DSDEN0* is qualified with AO to control the lower 8 bits of the data bus (DO-D7). DSDEN1* is qualified with BHE* to control the upper 8 data bits (D8-D15). Control of the upper and lower data bits independently is required in the AT type architecture to control the 8 to 16 bit data conversions.

ALE is a control signal used to latch the address line from the CPU so that the address will remain stable throughout the complete command cycle. ALE as generated when either $\mathrm{SO}^{*}$ or Sl * is asserted from the CPU to start a bus cycle. This allows the address latches to be enabled as early as possible to provide sufficient address setup time to the on board memory array. ALE is negated at the start of the command state (Tc) to latch the address while they are guaranteed from the CPU.

The last control signal generated is MEMCYC which is used to control the start of a memory access to the on board memory array. MEMCYC is asserted HIGH at the middle of the Status state (Ts) and is held active during the complete bus cycle. MEMCYC is negated at the end of a bus cycle or the end of the command state (TC).

## Bus Conversion Logic

This block of the CPU Control IC provides the logic to control the 16 bit transfers to and from an 8 bit device or memory. The conversion logic detects when a conversion is required, asserts a wait to the CPU by asserting the READY* line to a HIGH or inactive state, and then generates the required control signals to the data buffers to perform the conversion. A conversion is required during all 16 bit transfers to all $1 / O$ devices, the Interrupt Controller, and 8 bit memory. An 8 bit memory cycle is detected by sampling the state of AF16* and/or MEMCS16* at the beginning of a memory cycle. If both are inactive HIGH then the conversion logic acknowledges it as an 8 bit memory cycle. The control signals that are generated are DIRHLB (Direction High Low Byte), ENHLB (Enable High Low Byte), and CNTLOFF (Control off). ENHLB enables the conversion buffer during the conversion cycle and DIRHLB determines the direction of the buffer. CNTLOFF is used to latch the lower 8 bits (D0-D7) during a 16 bit read from an 8 bit device. SAO is also controlled and generated by the conversion logic so the even byte is read first then SAO is toggled to a HIGH state so the odd byte will be read.

## Wait State Control Logic

The Wait State Control Logic is used to allow slow memory and peripheral devices to be used with a faster CPU. Wait states are controlled by the CPU Control IC by negating the READY* line to the CPU to inactive state. The CPU will not end the cycle in progress until the READY* line is reasserted to an active LOW state. The READY* line is negated at the middle of the Status state (Ts) to an inactive HIGH to guarantee recognition of a wait state by the CPU. The READY* line is reasserted at the middle of the Command state or wait State to end the existing cycle.

There are several default wait state cycles that are generated by the CPU Control IC for control of all bus cycles. During a 16 bit memory cycle which is determined by the assertion of AFl6* or MEMCS16*, a default of one wait state is automatically inserted. During a 8 bit memory cycle which is determined when both AFl6* and MEMCS16* are negated, a default of four wait states are automatically inserted. All $1 / O$ cycles have a default of four wait states inserted to guarantee timing compatibility with existing $I / O$ channel add in boards. All of the default wait state are the same in the 8 MHz (FAST) mode and the 4 MHz (SLOW) mode except for the $I / O$ bus cycles. During an $I / O$ cycles in the 4 MHz (SLOW) mode, a default of only two wait states are inserted. This allows only the memory cycles to be effected to slow down the program speed without degrading the performance of I/O accesses.

All bus cycles can be extended above the default number of wait states by negating the IOCHRDY (I/O Channel Ready) to inactive LOW. The CPU Control IC will continue to insert wait states to the CPU until IOCHRDY is released allowing it to be asserted. IOCHRDY should not be held LOW for more that 15 usec. because Refresh to the DRAMs will be inhibited.

## DHA Arbitration Logic

The DMA Arbitration Logic in the CPU Control IC control the arbitration of the bus between the CPU and the DMA Controller. The arbitration logic detects when the DMA is requesting the bus when HRQ (Hold Request) is asserted HIGH. After HRQ is recognized, the arbitration logic will assert CPUHRQ to the CPU to request release of the bus. The CPU will respond to CPUHRQ after finishing the cycle in progress by suspending operation and asserting CPUHLDA to the CPU Control IC. The arbitration Logic will then tri-state the output command signals and generate HLDA to the DMA, allowing it to take control over the bus.

All DMA cycles have a default of one wait state automatically inserted to maintain compatibility with existing I/O channel boards and peripheral devices. The DMA cycle can be extended by and I/O device by applying a LOW state to the IOCHRDY signal. The DMA Controller will be held in a wait state until IOCHRDY is returned to a HIGH state. Refer to Timing Diagrams for Timing specifications of a DMA cycle and DMA Arbitration operation.

The CPU Control IC also controls two functions required in the 80286 architecture during a DMA cycle. In order to guarantee the integrity of an Interrupt Acknowledge (INTA) cycle, which requires two bus cycles to complete, a DMA arbitration inhibit is included in the arbitration logic. After an INTA cycle has started, a DMA cycle will be inhibited until the CPU performs a memory write cycle. The memory write cycle will normally be executed after the INTA cycle due the stack operation of the CPU required to service an interrupt routine. The inhibit will guarantee the a DMA cycle can not be allowed until both bus cycles of an INTA cycle is complete.

## Numerical Co-processor Control Logic

The Numerical Co-processor Control Logic provides an interface for the 80286 CPU to control an 80287 Co-processor. This logic controls the decoding required to select and reset the 80287 , control of the BUSY* and ERROR* signals from the 80287 to the CPU, and generating the interrupt signal during an error condition.

The input signal 287CS* is an I/O decode at 0FO-OFFh I/O address. 287CS* is used by the CPU Control IC to generate the Numerical Co-processor Chip Select (NPCS*), and the reset signal (RES287*) to the 80287. Refer to the I/O decode table for details about the further internal decode done by the CPU Control IC.

When the 80287 receives a command to perform a task, it will output the BUSY* signal which is input to the CPU Control IC. The CPU Control IC will then assert BUSY287* to the CPU. During normal operation of the 80287, when it finishes the task, it will negate BUSY* which will in turn negate BUSY287* to the CPU. If the ERROR input from the 80287 is asserted during this busy time, which indicates a 80287 error, BUSY287* output is latched and INT287* is asserted LOW to generate an interrupt to the CPU. Both BUSY287* and INT287* will remain latched in an active LOW state until they are cleared by writing to the I/O address 0FOh or OFlh. These signals are cleared after a system reset.

RES287* is generated to reset the 80287. It is generated by writing to $1 / O$ address 0 Flh or when a system reset is issued.

## A20 and CPU Reset Control Logic

The CPU Control IC incorporates the logic required to control the Address line $A 20$ and to reset the CPU. In the PC architecture, A20 must be held in a LOW state so the some programs that do wrapping will function correctly. After a reset, the output signal GA20 is help in a LOW state. If a program needs to address above the 1 MEG limit, OUT to $I / O$ address 068 h with data bit 1 to a HIGH or "I" state will allow A20 to be muxed to GA20.

The CPU can reset itself by doing an I/O wtrie to address 068h with bit 2 at a LOW or " 0 " state. After a reset has occurred, by Reading $I / O$ address 068 , it can be determined it was a power up reset or if the CPU reset itself. Refer to the $I / O$ address map for further details.

| Address |  |  |
| :--- | :--- | :--- | :--- |

## Pin Configuration



| Pin* | Pin Name | Type | Description |
| :---: | :---: | :---: | :---: |
| 62 | 16 MHz | I | 16 MHZ Clock Input |
| 63 | PRCLK | 0 | 80286 Processor Clock - 16:8 MHz |
| 64 | SCLK | 0 | System Clock - 8:4 MHZ |
| 65 | DMACLK | 0 | DMA Controller Clock - $4: 2 \mathrm{MHz}$ |
| 60 | 28.6MHZ | I | 28.63636 MHZ Clock Input |
| 59 | OSC | 0 | 14.31818 MHZ Clock |
| 58 | 3.58MHZ | 0 | 3.5795 MHz Clock for sound chip |
| 57 | 1.19MHZ | 0 | 1.19 MHz Clock for Interval Timer chip 8254. |
| 18 | MI/O | I | Memory Input/Output from the CPU. Indicates a memory access when HIGH and a I/O access when LOW. Used to generate the memory and $I / O$ command signals for the system. |
| 17 | S1* | I | Status Line 1 from the CPU |
| 16 | S0* | I | Status Line 0 from the CPU |
| 19 | READY* | 0 | Ready signal to the 80286 Processor Indicates the current bus be completed when LOW. |
| 68 | CPUHRQ | 0 | CPU hold request signal for the CPU. Indicates DMA transfers by the DMA controller when HIGH. It is also active during refresh cycles. |
| 4 | CPUHLDA | I | Hold Acknowledge signal from CPU. Indicates the CPU granting a DMA cycle to the DMA controller when HIGH. And causes all command signals to be tri-stated provided the CNTLOFF output is LOW |
| 33 | MEMR* | 0 | Memory Read Command output signal instructs a memory device to place data on the bus when LOW. It is also active during refresh. $13$ |


| 32 | MEMW* | 1/0 | Memory Write Command Input/Output signal. Instructs a memory device to read the data on the bus when LOW. |
| :---: | :---: | :---: | :---: |
| 22 | IOR* | 1/0 | Input/Output Read signal. Indicates a Read cycle is performed with an I/O device or port when LOW. |
| 21 | IOW* | 1/0 | Input/Output Write signal. <br> Indicates a Read cycle is performed with an I/O device or port when LOW. |
| 20 | INTA* | 0 | Inter rupt Acknowledge for the Interrupt Controller. It is used by the Interrupt controller to output the interrupt vector onto the data bus when LOW. |
| 55 | ALE | 0 | Address Latch Enable. It is used to hold the address during bus cycle when HIGH. |
| 56 | MEMCYC | 0 | Memory cycle signal. It is used to generate memory control signal, RAS, MUX and CAS when HIGH. |
| 36 | RES* | I | Reset signal. It is connected to the power good and used to reset the system when LOW. |
| 37 | RESET | 0 | Reset output signal. It is a synchronized reset signal for general system reset when HIGH. |
| 34 | RESCPU | 0 | CPU Reset output signal. It is used to reset CPU when HIGH. |
| 47 | XBHE* (SBHE*) | 1/0 | Bus High Enable Input/Output signal. It is used to enable the high byte data bus signals when LOW |
| 45 | SAO | 0 | System Address 0. |
| 8 | A0 | I | Address 0 input signal from CPU. It is used to generate the enable signal for the data bus. |


| 9 | Al | I | Address 1 input signal from CPU. It is used to detect the SHUT DOWN condition of the CPU. |
| :---: | :---: | :---: | :---: |
| 10 | SA3 | I | System Address 3 input signal. It is used to generate the chip select and reset signal for 80287 . |
| 11 | SA4 | I | System Address 4 input signal. It is used to generate the chip select. |
| 12 | A20 | I | Address 20 from the CPU. It is used to generate GA20 signal when CPUHLDA LOW. |
| 46 | GA2 0 | 0 | Gated Address 20. Address 20 is being negated by XD1 and PORTI (internal). When XDI is LOW, Gated A20 on the CPU address bus is forced LOW. When XDl is HIGH, GA20 is transmited as Address 20. |
| 7 | PPICS* | I | Programable Peripheral Interface Chip Select input signal. It is used to generate Chip Select Signal signal for the peripheral interface device when is LOW. |
| 6 | 287CS* | I | 287 Chip select input signal. It is used for generating the Numerical processor Select NPCS for 80287 when is HIGH. |
| 5 | HRQ | I | Hold Request input signal from DMA controller. It is used to generate the CPU Hold Request Signal when is HIGH. |
| 67 | HLDA | 0 | Hold Acknowledge output signal for DMA Controller. It is used to provide Hold Acknowledge for DMA controller when is HIGH. |
| 3 | DMAAEN | I | DMA Address Enable input Signal from DMA Controller. It is used to provide enable signal for any $1 / 0$ device during DMA Access to the system memory when is HIGH. |


| 2 | DMAMR* | I | DMA Memory Read signal from DMA controller. It is used to generate Memory Read - MEMR* signal when is LOW. |
| :---: | :---: | :---: | :---: |
| 66 | DMARDY | 0 | DMA Ready output signal for DMA Controller. It is used to extend memory Read and write cycles from the DMA Controller for slower memory or I/O device when is HIGH. |
| 52 | BUSY* | I | Busy input signal from 80287. It indicates that 80287 is currently executing a command when is LOW. |
| 53 | ERROR* | I | Error input signal from 80287. It indicates an unmasked error condition exists. |
| 51 | BUSY287* | 0 | Busy 80287 output signal for the CPU. It indicates to the processor the operating condition of the 80287 when is LOW. |
| 50 | INT287* | 0 | Interrupt 80287 output signal for the Interrupt controller. It is the interrupt request from 80287 |
| 49 | RES287 | 0 | Reset 80287 output signal for the 80287. It is used to reset to the 80287 when is HIGH. |
| 48 | NPCS* | 0 | Numerical Processor Chip Select output signal for the 80287. It is used to select the 80287 device when is LOW. |
| 38 | DT/R* | 0 | Data Transmit/Received output signal. It is used to determine the data direction to and from local data bus. It is a write bus cycle when is HIGH and read bus cycle when is LOW. |
| 39 | DSDEN0* | 0 | Data Strobe Data Enable 0 output. It is used to enable the data transceivers connected to the low byte (D0-D7) when is LOW. |
| 40 | DSDEN1* | 0 | Data Strobe Data Enable l output signal. It is used to enable the data transceivers connected to the high byte (D8-D15) data bus when is LOW. $16$ |


| 41 | ENHLB* | 0 | Enable High To Low byte output signal. It is used to perform the high to low conversion when is LOW. |
| :---: | :---: | :---: | :---: |
| 42 | CNTLOFF | 0 | Control Off output signal. It is used to enable the low data bus latch during byte accesses when is LOW. |
| 43 | DIRHLD | 0 | Direction High to Low byte output signal. It is used to perform high to low byte conversion when is HIGH. |
| 13 | AF16* | I | AF16* output signal. It is used to control the 16 bit memory accesses and to inhibit the command delays for memory accesses by $I / O$ device when is LOW. |
| 14 | MEMCS16* | I | Memory Chip Select input signal. It is used to inhibit command delays when 16 bit memory accesses are made when is LOW. |
| 54 | IOCHRDY | I | I/O Channel Ready input signal from I/O device. It is used generate wait states in $1 / O$ or memory accesses by I/O device when is HIGH. |
| 15 | VIDWAIT* | I | Video Wait input signal from video controller. It is used to generate Video delay ready signal when is HIGH. |
| 31 | D0 | I/O | Data Bus Bit 0 of the peripheral data bus. |
| 30 | D1 | I/O | Data Bus Bit 1 of the peripheral data bus. |
| 29 | D2 | I/O | Data Bus Bit 2 of the peripheral data bus. |
| 28 | D3 | I/O | Data Bus Bit 3 of the peripheral data bus. |
| 26 | D4 | I/O | Data Bus Bit 4 of the peripheral data bus. |


| 25 | D5 | I/O | Data Bus Bit 5 of the peripheral <br> data bus. |
| :--- | :--- | :--- | :--- |
| 24 | D6 | I/O | Data Bus Bit 6 of the peripheral <br> data bus. |
| 1,35 | VCC | I/O | Data Bus Bit 7 of the peripheral <br> data bus. |
| 27,44 | VSS | GWR | Power Supply. <br> 61 |

## ELECTRICAL SPECIFICATIONS



CURRENT LOADING AND CAPACITANCE OF EACH OUTPUT At Vol $=0.45$ volts, Voh $=2.4$ volts

| PIN NAME | MIN | UNITS |  |
| :---: | :---: | :---: | :---: |

OUTPUT

| RESCPU, $\overline{\text { NPCS }}, \overline{\text { BUSY287 }} \overline{\text { INT2 }} 87$ RES287, CPUHRQ, HLDA, DIRHLB CNTLORF, ENHLB, DMARDY | 2 | ma | 20pf |
| :---: | :---: | :---: | :---: |
| RESET, GA20 | 2 | ma | 50pf |
| READY, MEMCYC, DT/ $\bar{R}, \overline{D S D E N O}$ DSDENI | 4 | ma | 20pf |
| SCLK, DMACLK, OSC, 1.19MHZ 3.58 MHZ | 4 | ma | 50pf |
| PRCLR, INTA | 8 | ma | 50pf |
| ALE | 8 | ma | 50pf |
| $\overline{\text { MEMR }}$ | 8 | ma | 80 pf |

## INPUT/OUTPUT

| XD0 - XD7 | 2 | ma | 80pf |
| :---: | :---: | :---: | :---: |
| $\overline{\text { XBEE }}$ | 4 | ma | 50pf |
| MEMW | 8 | ma | 80pf |
| $\overline{\text { IOR, }} \overline{\text { IOW }}$ | 8 | ma | 115pf |
| SAO | 4 | ma | 120pf |

Timing Diagram Specifications

| ISYM! DESCRIPTION | ! MIN ! MAX |  | !UNITS INOTES |  |
| :---: | :---: | :---: | :---: | :---: |
| $!==$ ! |  |  |  |  |
| $!$ ! ! | $!$ |  | $!$ | $!$ |
| 11 !PRCLK Period | 162 | 250 | !ns | $!$ |
| $!1$ | ! |  | $!$ | ! |
| 12 IPRCLK Low Time | 125 | 125 | Ins | $!$ |
| $!$ ! ! | ! |  | ! | ! |
| ! 3 IPRCLK High Time | $!25!$ | 125 | !ns | 1 |
| 1 ! - 1 | 1 |  | $!$ | ! |
| 14 ! $\overline{S O}, \overline{S 1}, \mathrm{M} / \overline{\text { IO }}$ Setup time to PRCLK | 120 |  | Ins | $!$ |
|  | ! |  | $!$ | $!$ |
| $!5!\overline{S O}, \overline{S 1}, M / \overline{I O}$ Hold time to PRCLK | 111 |  | Ins | 1 |
| $!!$ | ! |  | $!$ | $!$ |
| 16 1ALE active from $\overline{\text { S0 }}$, $\overline{\text { S1 }}$ - | 15 ! | 25 | Ins | 1 |
| I | + |  |  | $!$ |
| 17 IALE inactive from PRCLK | 151 | 25 | Ins | $!$ |
| ! | 1 |  |  | $!$ |
| 17A 1OSC delay from 28.6 MHz | 1 | 30 | Ins | 1 |
| 8 ! MEMCYC active from PRCLK! | $5!$ | 25 | Ins | $!$ |
| 8 IMEMCYC active from PrCLK | 5 |  | , | ! |
| 18A 13.85 MHZ delay from OSC | 1 ! | 15 | Ins | 1 |
| $!$ ! ! | ! |  | ! | ! |
| ! 9 IMEMCYC inactive delay from PRCLR $\downarrow$ ! | $8!$ | 35 | Ins | $!$ |
| $!$ ! ! |  |  |  | $!$ - |
| 19A 11.19 MHZ delay from OSC | 1 ! | 15 | Ins | 1 |
| $!$ ! ${ }^{\text {! }}$ |  |  |  | ! |
| 110 !Command active delay from PRCLR \& | $5!$ | 45 | ! ns | $!$ |
| ! | 1 ! |  |  | $!$ |
| !11 ! Command inactive delay from PRCLK! | 51 | 45 | !ns | 1 |
| $!12!$ ! $\bar{R}$ active delay from PRCLK | 51 | 50 | Ins | ! Read |
|  | 5 |  | , | , |
| $113 \mathrm{DT} / \bar{R}$ inactive delay from PRCLK | 10 | 40 | Ins | 1 Read |
| 1 l | $10!$ |  | In | 1 |
| !14 [DSDENO, 1 active delay from PRCLK | 10 | 50 | !ns | 1 Read |
| 1- |  |  | ! |  |
| !15 1DSDENO, 1 inactive delay from DT/R ! | $3!$ |  | Ins | 1 Read |
| $!$ ! ${ }^{\text {d }}$ | 1 |  | $!$ |  |
| !16 IDSDENO, 1 inactive delay from PRCLK1 | 8 ! | 50 | Ins | 1 Read |
| ! | ! |  |  |  |
| 117 1DT/ $\overline{\mathrm{R}}$ inactive delay from DSDEN0, 1 | $3!$ |  | Ins | $!$ Read |
| 1 ! | , |  |  |  |
| ! 18 !DSDENO, 1 active delay from PRCLK ! ! | 81 | 60 | Ins | ! Write |
| $!19$ ! $\overline{\text { DSDENO, } 1}$ inactive delay from PRCLK ${ }^{!}$ | 8 ! | 50 | ! $n$ s | ! Write |


| 1SYM! DESCRIPTION | !MIN |  | !UNITS ! NOTES |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $!==1$ |  |  |  |  |  |
| , |  |  | ! | $!$ |  |
| $120 \backslash \overline{\text { AF16 }}$, MEMCS16 ${ }^{\text {M }}$ setup time to ALE | 135 |  | Ins | $!$ |  |
| ! |  |  | $!$ | $!$ |  |
| 121 1/ $\overline{\text { AF16 }}$, MEMCS16 hold time from ALE | 11 |  | Ins | 1 |  |
| 1 | $!$ |  | ! | ! |  |
| 122 1PRCLK delay from 16 MHz | 1 ! | 50 | Ins | 1 |  |
| $!$ | $!$ |  |  | 1 |  |
| 123 [ $\overline{\text { RSET }}$ hold time from PRCLK $\downarrow$ | $!10$ |  | !ns | 1 |  |
|  |  |  | $!$ | $!$ |  |
| 124 IVSET setup time to PRCLK | ! 25 |  | Ins | $!$ |  |
| ! | $!$ ! |  | $!$ | $!$ |  |
| 125 !RESET A/I delay from PRCLK | 15 | 35 | Ins | $!$ |  |
| 1 | $!$ ! |  | $!$ | ! |  |
| $!26$ !SCLK delay from PRCLK | $!1$ | 25 | ! ns | $!$ |  |
| ! | $!$ ! |  | ! | $!$ |  |
| $\underline{27}$ IDMACLK delay from SCLK | $!$ | 15 | !ns | $!$ |  |
| ! | $!$ |  | ! | $!$ |  |
| 128 LRESCPU inactive delay from PRCLK | $!5!$ | 50 | Ins | 1 |  |
| $!29$ ! RESCPU active delay from SCLK $\dagger$ | 5 | 20 | Ins | $!$ |  |
|  | I |  |  | ! |  |
| 130 1HRQ setup to DMACLK ${ }^{1}$ | ! 15 ! |  | !ns |  | 1 |
|  | , |  |  |  |  |
| 131 IHRQ holdtime from DMACLK $\dagger$ | 111 |  | Ins | 1 | 1 |
| $1$ | ! |  |  | $!$ |  |
| 132 ICPU日RQ A/I delay from CPUHLDA | $!$ | 30 | Ins | $!$ |  |
| $!$ | 1 |  | $!$ | 1 |  |
| 133 IHLDA active delay from SCLK | $1 \quad 1$ | 25 | !ns | 1 |  |
| ! | ! |  | ! | $!$ |  |
| $!34$ ! HLDA A/I delay from CPUHLDA | $!$ | 30 | !ns | $!$ |  |
|  | $!$ |  | $!$ | $!$ |  |
| !35 IHLDA inactive delay from DMACLK ${ }^{1}$ | $!$ | 30 | Ins | , |  |
|  | ! |  | , | , |  |
| 136 IDMAMR setup to DMACLK 9 | ! 15 ! |  | Ins | 1 |  |
| $!$ | $!$ ! |  |  | $!$ |  |
| 137 ICMD tristate elay from CPUHLDA | $!$ | 30 | ! ns | , | 2 |
| ! ${ }^{137}$ | ! |  | $!$ | $!$ |  |
| 138 \MEMR active delay from DMACLK ${ }^{\text {¢ }}$ | ! | 25 | Ins | $!$ |  |
| 1 (Due to DMAMR) | $!$ |  | $!$ | . |  |
|  | 1 |  | ! | I |  |
| 139 ! $\overline{\text { MEMR }}$ inactive delay from DMAMR | $!$ | 25 | Ins | $!$ |  |
|  | ! |  | ! | $!$ |  |
| 140 ! $\overline{\text { MEMR }}$ tri-state delay from DMACLK ${ }^{\text {f }}$ | ! | 25 | !ns | $!$ |  |
| $!$ (Due to DMAMR) | $!$ |  |  |  |  |
| $!$ | $!$ ! |  | ! | 1 |  |
| 141 !CMD active delay from CPUHLDA | $!1$ | 30 | !ns | , |  |
| 1 | $!$ ! |  |  | $!$ |  |
| 142 IDMARDY inactive delay | $!\quad!$ | 30 | !ns | $!$ |  |
| 1 from DMAMR, $\overline{\text { IOR }}$ | $!\quad!$ |  | ! | $!$ |  |
| 1 ! | $!$ |  |  | $!$ |  |
| 143 IDMARDY active delay from DMACLK | $1!$ | 30 | Ins | $!$ |  |




1. Setup and Hold times are required to guarantee recognition of signa' at clock edge.
2. CMD = MEMR,MEMW,IOR and IOW.
$A / I=$ Active and Inactive

## Power Supply

1000TL POWER SUPPLIES

1000TL 67 WATT SINGLE INPUT POWER SUPPLY

# 1000TL 67 Watt Single Input Power Supply Contents 

SectionOperating CharacteristicsBlock DiagramPage2
Theory of Operation ..... 3

OPERATING CHARACTERISTICS

|  | MINIMUM | TYPICAL | MAXIMUM | UNITS |
| :---: | :---: | :---: | :---: | :---: |
| Operating Voltage Range | 90 | 120 | 135 | VAC |
| Line Frequency | 47 | 50/60 | 63 | Hz |
| Output Voltages |  |  |  |  |
| Vol | 4.85 | 5.00 | 5.15 | V |
| Vo2 | 11.40 | 12.00 | 12.60 | V |
| Vo3 | -13.20 | -12.00 | -10.80 | V |
| Output Loads |  |  |  |  |
| Iol | 1.25 | - | 7.0 | A |
| Io2 | 0.15 | - | 2.4 | A |
| Io 3 | 0 | - | 0.25 | A |
| Over Current Protection |  |  |  |  |
| Current Limit ICLI | - | - | 14.0 | A |
| ICL2 | - | - | 4.8 | A |
| ICL 3 | - | - | 1.0 | A |
| Over Voltage Protection |  |  |  |  |
| Crowbar | 5.8 | - | 6.8 | V |
| Output Noise |  |  |  |  |
| Vol | - | - | 50 | nV P-P |
| Vo2 | - | - | 100 | mV P-P |
| Vo3 | - | - | 150 | mV P-P |
| Efficiency | 63 | 65 | - | \% |
| Holdup Time |  |  |  |  |
| Full Load at Nominal Line | 16 | - | - | mSec. |
| Insulation Resistance |  |  |  |  |
| Input to Output | 7 | 1000 | - | M ohms |
| Input to Ground | 7 | 1000 | - | M ohms |
| Isolation |  |  |  |  |
| Input to Ground | 1.7 | - | - | KVDC |



## AC Input Circuit

This circuit is composed of an AC Power Switch, a fuse, a line filter, and an inrush current limiting circuit and rectifying smoothing circuit. The inrush current limiting circuit controls the charging current to electrolytic capacitors when power is ON. The line filter reduces noise that leaks from the power source to the AC line or that returns from the unit to the power souce; it satisfies the specifications of noise regulations.

## Control Circuit \& Power Converter Circuit

This circuit is a self oscillation switching system, generally called an R.C.C. (Ringing Choke Converter). The R.C.C. circuit does not fix the oscillating frequency. Whenever input voltage is high or the load becomes light, the oscillating frequency will be high.

The current through R2 supplies transistor Ql's base, then Q1 turns ON. When transistor $Q 1$ is $O n$, the $Q 1$ current excites the transformer $T l$ and voltage rises in the bias coil of $T l(5-6)$ which leads transistor $Q 1$ positive bias, then transistor Ql turns ON.

When transistor Q1 turns ON, collector current charges the energy to primary inductance of transformer $T 1$ (l-3). Increasing the collector current of transistor Ql to the point of:

$$
\begin{aligned}
& \mathrm{I} \\
& C=I_{B} . \text { hfe }
\end{aligned}
$$

Then, transistor Ql immediately turns OFF. In a moment, transformer $T l$ will have negative voltage which will be supplied to the secondary circuit through a rectifier. A Short Circuit protector is provided to protect transistor $Q 1$ from excess amounts of current when the secondary circuit becomes shorted. When transistor $Q 2$ detects the voltage drop at Rl2, the collector of $Q 2$ shorts the base and emitter of Q1. Then Ql stops working so that the circuit protects Ql from over current.

The over current protector in the -12 V line is provided by the three terminal positive voltage regulator ICl (built-in current fold back protection ), which protects Ql against excessive current from the -12V line.

5v Output Voltage Detecting Circuit
The circuit detects the change of output load current compared with the output voltage and AC line input voltage, which feeds back to the control circuit through a photo coupler PHCl to keep the output voltage stable. The Photo coupler isolates the primary and secondary circuits.

## Over-Voltage Protection

When the +5 output voltage rises, between 5.8 V to 6.8 V , a control signal turns on the photo coupler PHC2 (Photo Thyristor) with the current of zener diode (Dll) and stops oscillation by turning on Q3, which turns off Q1 in the switching circuit.

In the case of stopped oscillation, correct the cause of the failure, and reinput the power. The overvoltage protection circuit will reset automatically under good conditions.

The Photo Thyristor isolates the primary and secondary circuits.

1000TL 67 WATT DUAL INPUT POWER SUPPLY

# 1000TL 67 Watt Dual Input Power Supply Contents 

SectionOperating Characteristics$\frac{\text { page }}{1}$Block Diagram2
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| OPERATING CHARACTERISTICS |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | Minimum | TYPICAL | maximum | UNITS |
| Operating Voltage Range | $\begin{array}{r} 90 \\ 198 \end{array}$ | $\begin{aligned} & 120 \\ & 240 \end{aligned}$ | $\begin{aligned} & 135 \\ & 264 \end{aligned}$ | VAC |
| Line Frequency | 47 | 50/60 | 63 | Hz |
| Output Voltages |  |  |  |  |
| Vol | 4.85 | 5.00 | 5.15 | V |
| Vo2 | 11.40 | 12.00 | 12.60 | V |
| vo3 | -13.20 | -12.00 | -10.80 | v |
| Output Loads |  |  |  |  |
| Iol | 1.25 | - | 7.0 | A |
| Io2 | 0.15 | - | 2.4 | A |
| Io3 | 0 | - | 0.25 | A |
| Over Current Protection |  |  |  |  |
| Current Limit ICLI | - | - | 14.0 | A |
| ICL2 | - | - | 4.8 | A |
| ICL3 | - | - | 1.0 | A |
| Over Voltage Protection |  |  |  |  |
| Crowbar | 5.8 | - | 6.8 | $v$ |
| Output Noise |  |  |  |  |
| Vol | - | - | 50 | mV P-P |
| Vo2 | - | - | 100 | mV P-P |
| vo3 | - | - | 150 | IV P-P |
| Efficiency | 63 | 65 | - | \% |
| Holdup Time |  |  |  |  |
| Full Load at Nominal Line | 16 | - | - | mSec. |
| Insulation Resistance |  |  |  |  |
| Input to Output | 7 | 1000 | - | M ohms |
| Input to Ground | 7 | 1000 | - | M ohms |
| Isolation |  |  |  |  |
| Input to Ground | 1.25 | - | - | kvac |
| Input to Output | 3.75 | - | - | KVAC |



## Theory of Operation

## AC Input Circuit

This circuit is composed of an $A C$ Power Switch, a fuse, a line filter, and an inrush current limiting circuit and rectifying smoothing circuit. The inrush current limiting circuit controls the charging current to electrolytic capacitors when power is $O N$. The line filter reduces noise that leaks from the power source to the AC line or that returns from the unit to the power souce; it satisfies the specifications of noise regulations.

## Control Circuit \& Power Converter Circuit

This circuit is a self oscillation switching system, generally called an R.C.C. (Ringing Choke Converter). The R.C.C. circuit does not $f i x$ the oscillating frequency. Whenever input voltage is high or the load becomes light, the oscillating frequency will be high.

The current through R4 and R5 supplies transistor Ql's base, then Ql turns ON. When transistor Ql is On, the Ql current excites the transformer $T 1$ and voltage rises in the bias coil of $T 1(2-3)$ which leads transistor Ql positive bias, then transistor Ql turns ON .

When transistor Ql turns ON, collector current charges the energy to primary inductance of transformer $T l(4-6)$. Increasing the collector current of transistor $Q 1$ to the point of:

$$
I_{C}>I_{B} . \text { hfe }
$$

Then, transistor Ql immediately turns OFF. In a moment, transformer $T l$ will have negative voltage which will be supplied to the secondary circuit through a rectifier. A Short Circuit protector is provided to protect transistor Ql from excess amounts of current when the secondary circuit becomes shorted. When transistor Q2 detects the voltage drop at R13, the collector of $Q 2$ shorts the base and emitter of Q1. Then Q1 stops working so that the circuit protects Ql from over current.

The over current protector in the -12 V line is provided by the three terminal positive voltage regulators IC2, IC3 (built-in current fold back protection ), which protects $Q 1$ against excessive current from the -12 V line.

## 5v Output Voltage Detecting Circuit

The circuit detects the change of output load current compared with the output voltage and AC line input voltage, which feeds back to the control circuit through a photo coupler PHCl to keep the output voltage stable. The Photo coupler isolates the primary and secondary circuits.

Over-Voltage Protection
When the +5 output voltage rises, between 5.8 V to 6.8 V , a control signal turns on the photo coupler PHC2 (Photo Thyristor) with the current of zener diode (Dll) and stops oscillation by turning on Q3, which turns off Ql in the switching circuit.

In the case of stopped oscillation, correct the cause of the failure, and reinput the power. The overvoltage protection circuit will reset automatically under good conditions.

The Photo Thyristor isolates the primary and secondary circuits.

model no. 8790084

Keyboard

### 1.0 GENERAL

The keyboard is a direct, plug-compatible replacement for the Enhanced Keyboard for the IBM PC, XT, and AT personal computer. No software modification or special interface is needed by the user.

### 2.0 SCOPE

This specification describes the functional, mechanical, electrical, environmental, and reliability characteristics of the FUJITSU N860-4703-T Keyboard assembly.

The keyboard is encoded in such a way as to produce a unique output code for each key that is pressed and/or released. The communication with the host computer is a synchronous serial link. The Key Layout, Switch Encoding and Serial Communication are all compatible with the IBM PC, XT and AT.
3.0 MECHANICAL SPECIFICATION
3.1 Key Layout, Legends, and Colors

Figure 1 shows keytop layout, appropriate legends and keytop colors. The keys are numbered fram left to right starting with the spacebar row (Row A) and ending with the function key row (Row F).

## 



FIGURE 1

### 3.2 KEYSWITCH

3.2.1 Total Travel: $0.150 ",+/-0.020$ " $(3.8,+/-0.5 \mathrm{~mm})$
3.2.2 FORCE: All keyswitches shall utilize a 2.0 ounce $(+/-0.9$ oz) operating force. This is accomplished using both a rubber keyswitch membrane and springs.
3.2.3 BREAKOVER FEEDBACK: The keyswitches utilize a tactile feedback to assure the operator that the key has been fully pressed.

### 4.0 FUNCTIONAL REQUIREMENTS

### 4.1 SCAN CODES

The keyboard generates a unique Hex scan code for each keyswitch that is pressed (make code) and released (break code). For the AT Mode, the break code is the same as the make code preceded by "FO" Hex. Example: The make code for the "ESC" key is 76 Hex and the break code is two bytes, F0 76 Hex. For the XT Mode, the break code is 80 Hex, plus the make code. Example: The make code for the "ESC" key is 01 Hex and the break code is 81 Hex. The keyswitch-to-scan-code (make and break codes) assignments, Standard ASCII Codes, and Extended ASCII Codes are listed on the following pages.

## KEYBOARD SCAN CODES

| Key | Key <br> Descript. | AT Mode |  | XT Mode |  | Standard ASCII (Scancode/ASCII code) |  |  |  | Extended ASCII <br> (Scancode/ASCII code) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  | Code | Code | Code | Code | Norm | Shift | Ctrl | Alt | Norm | Shift | Ctrl | Alt |
| 1 | Esc | 76 | F076 | 01 | 81 | 011B | 011B | 011B | ----* | 011B | 011B | 011B | 0100 |
| 2 | F1 | 05 | F005 | 3B | BB | 3800 | 5400 | SE00 | 6800 | 3B00 | 5400 | 5E00 | 6800 |
| 3 | F2 | 06 | F006 | 3C | BC | $3 \mathrm{C00}$ | 5500 | SFO0 | 6900 | 3 COO | 5500 | 5F00 | 6900 |
| 4 | F3 | 04 | F004 | 3D | BD | 3D00 | 5600 | 6000 | 6A00 | 3D00 | 5600 | 6000 | 6A00 |
| 5 | F4 | 0 C | FOOC | 3E | BE | 3E00 | 5700 | 6100 | 6B00 | 3E00 | 5700 | 6100 | 6B00 |
| 6 | F5 | 03 | F003 | 3F | BF | 3 F 00 | 5800 | 6200 | $6 \mathrm{C00}$ | 3 FOO | 5800 | 6200 | $6 \mathrm{C00}$ |
| 7 | F6 | OB | FOOB | 40 | C0 | 4000 | 5900 | 6300 | 6D00 | 4000 | 5900 | 6300 | 6D00 |
| 8 | F7 | 83 | F083 | 41 | C1 | 4100 | 5A00 | 6400 | 6E00 | 4100 | 5A00 | 6400 | 6E00 |
| 9 | F8 | 0A | F00A | 42 | C2 | 4200 | 5B00 | 6500 | 6 F 00 | 4200 | 5B00 | 6500 | 6F00 |
| 10 | F9 | 01 | F001 | 43 | C3 | 4300 | SC00 | 6600 | 7000 | 4300 | 5 COO | 6600 | 7000 |
| 11 | F10 | 09 | P009 | 44 | C4 | 4400 | SD00 | 6700 | 7100 | 4400 | 5D00 | 6700 | 7100 |
| 12 | F11 | 78 | F078 | 57 | D7 | ----- | ---- | $\cdots$ | ----. | 8500 | 8700 | 8900 | 8B00 |
| 13 | F12 | 07 | F007 | 58 | D8 | $\cdots$ |  | 720 | -...- |  | 8800 | 8A00 | 8 COO |
| 14 | Print Scrn | E07C | E0F07C | E02AE037 | E0B7EOAA | $\mathrm{Note}_{2}{ }_{2}$ | Note | 7200 |  | $\mathrm{Note}_{2}{ }_{2}$ | Note ${ }_{2}$ | 7200 |  |
| 15 | Scroll Lock Pause Break | ${ }^{7 E}$ | F07E E1F014F077 | ${ }^{46}$ E11D45 | C6 | Note ${ }^{2}$ Note | Note Note | --.... ${ }^{4}$ | Note ${ }^{2}$ Note | $\mathrm{Note}^{2}$ | $\mathrm{Note}^{2}$ | $\cdots{ }^{-1}$ | $\mathrm{Note}^{2}$ |
| 16 | Pause Break | E11477 | E1F014F077 | E11D45 | E19DC5 | Note | Note | Note | Note ${ }^{\text {a }}$ | Note ${ }^{3}$ | Note ${ }^{3}$ | Note ${ }^{4}$ | Note ${ }^{3}$ |
| 17 | ~or 1 | OE | F00E | 2 B | ${ }^{\text {AB }}$ | 2960 | 297E | -...- | ---- | 6000 | 7E00 | $\cdots$ | 2900 |
| 18 | ! or 1 | 16 | F016 | 02 | 82 | 0231 | 0221 | $\cdots$ | 7800 | 0231 | 0221 | ---. | 7800 |

Table 1

## KEYBOARD SCAN CODES

| Key | Key <br> Descri | AT Mode |  | XT Mode |  | Standard ASCII (Scancode/ASCII code) |  |  |  | Extended ASCII (Scancode/ASCll code) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Make | Break | Make | Break |  |  |  |  |  |  |  |  |
|  |  | Code | Code | Code | Code | Norm | Shift | Ctrl | All | Norm. | Shift | Ctrl | Alt |
| 19 | @ or 2 | 1E | F01E | 03 | 83 | 0332 | 0340 | 0300 | 7900 | 0332 | 0340 | 0300 | 7900 |
| 20 | \# or 3 | 26 | P026 | 04 | 84 | 0433 | 0423 | .---- | 7A00 | 0433 | 0423 | ...... | 7A00 |
| 21 | \$ or 4 | 25 | F025 | 05 | 85 | 0534 | 0524 | ----- | 7B00 | 0534 | 0524 | --.-- | 7B00 |
| 22 | \% or 5 | 2E | F02E | 06 | 86 | 0635 | 0625 | -.-.- | 7C00 | 0635 | 0625 | --.-- | $7 \mathrm{C00}$ |
| 23 | ${ }^{\text {or }} 6$ | 36 | F036 | 07 | 87 | 0736 | 075E | 071E | 7D00 | 0736 | 075E | 071E | 7D00 |
| 24 | \& or 7 | 3D | F03D | 08 | 88 | 0837 | 0826 | --.-. | 7E00 | 0837 | 0826 | ....- | 7E00 |
| 25 | * or 8 | 3E | F03E | 09 | 89 | 0938 | 092A | ----- | 7 F 00 | 0938 | 092A | ----- | 7F00 |
| 26 | ( or 9 | 46 | F046 | 0A | 8A | 0^39 | 0A28 | -..-- | 8000 | 0A39 | 0A28 | ----- | 8000 |
| 27 | ) or 0 | 45 | F045 | OB | 8 B | 0 B 34 | 0B29 | $\cdots$ | 8100 | 0B34 | 0B29 | -..-- | 8100 |
| 28 | - or- | 4E | F04E | 0C | 8 C | 0C2D | 0C5F | 0C1F | 8200 | 0C2D | 0 CSF | 0C1F | 8200 |
| 29 | + or $=$ | 55 | F055 | 0D | 8 D | 0D3D | 0D2B | --..- | 8300 | 0D3D | 0D2B | $\cdots$ | 8300 |
| 30 | Backspace | 66 | F066 | OE | 8E | 0E08 | 0E08 | 0E7F | .-... | 0F08 | 0E08 | 0E7F | 0E00 |
| 31 | Insert | E070 | E0F070 | E02AE052 | E0D2F0AA | 5200 | 5200 | ----- | $\cdots$ | 52E0 | 52F30 | 92F0 | A200 |
| 32 | Home | E06C | E0F06C | E02AE047 | EOC7E0AA | 4700 | 4700 | 7700 | ----. | 47E0 | 47E0 | 77E0 | 9700 |
| 33 | Pg Up | E07D | E0F07D | L02AE049 | E0C9E0AA | 49005 | 4900 | 8400 |  | $49 \mathrm{E} 0_{5}$ | 49 E 05 | 84E0 | 99005 |
| 34 | Num Lock | 77 | 1077 | 45 | C5 | Note ${ }^{5}$ | Note ${ }^{5}$ | ----. | Note ${ }^{5}$ | Note ${ }^{5}$ | Note ${ }^{5}$ | --.- | Note ${ }^{5}$ |
| 35 | 1 | E04A | 120104A | E035 | E0B5 | 352F | 352F | ---.- | ---- | F02F | E02F | 9500 | A400 |
| 36 | - | 7 C | F07C | 37 | B7 | 372A | 372A | ..... | ..... | 372A | 372A | 9600 | 3700 |
| 37 | - | 7B | F07B | 4A | CA | 4A2D | 4A2D | --.-- | ----- | 4A2D) | 4A2D | 8 E 00 | 4A00 |

## KEYBOARD SCAN CODES

| Key | Key Descript. | AT Mode |  |
| :---: | :---: | :---: | :---: |
|  |  |  | Break |
|  |  | Code | Code |
| 38 | Tab | OD | FOOD |
| 39 | Q or q | 15 | F015 |
| 40 | W orw | 1D | F01D |
| 41 | Eore | 24 | F024 |
| 42 | Rort | 2D | F02D |
| 43 | Tort | 2 C | F02C |
| 44 | Y ory | 35 | P035 |
| 45 | Uoru | 3 C | F03C |
| 46 | Iori | 43 | F043 |
| 47 | Ooro | 44 | F044 |
| 48 | Porp | 4D | F04D |
| 49 | \{ or [ | 54 | D054 |
| 50 | \} or 1 | 5B | FOSB |
| 51 | \|or 1 | 5D | FOSD |
| 52 | Delete | E071 | E0F071 |
| 53 | End | E069 | E0F069 |
| 54 | Page Down | E07A | E0F07A |
| 55 | 7 or Home | 6C | F06C |
| 56 | 8 | 75 | F075 |
| 57 | 9 or Page Up | 7D | F07D |


| XTMode |  | Standard ASCII (Scancode/ASCII code) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Make | Break |  |  |  |  |
| Code | Code | Norm. | Shift | Cirl | Alt |
| OF | 8 F | 0F90 | OFO0 | ----- |  |
| 10 | 90 | 1071 | 1051 | 1011 | 1000 |
| 11 | 91 | 1177 | 1157 | 1117 | 1100 |
| 12 | 92 | 1265 | 1245 | 1205 | 1200 |
| 13 | 93 | 1372 | 1352 | 1312 | 1300 |
| 14 | 94 | 1474 | 1454 | 1414 | 1400 |
| 15 | 95 | 1579 | 1559 | 1519 | 1500 |
| 16 | 96 | 1675 | 1655 | 1615 | 1600 |
| 17 | 97 | 1769 | 1749 | 1709 | 1700 |
| 18 | 98 | 186F | 184P | 180 F | 1800 |
| 19 | 99 | 1970 | 1950 | 1910 | 1900 |
| 1A | 9A | 1A5B | 1A7B | 1A1B | ----. |
| 1B | 9B | 1BSD | 1B7D | 1B1D | ----- |
| 2B | AB | 2BSC | 2B7C | 2BIC | -...- |
| E02AE053 | E0D3E0AA | 5300 | 5300 | ----- | --... |
| E02AE04F | EOCFEOAA | 4F00 | 4F00 | 7500 | ----- |
| E02AE051 | EODIE0AA | 5100 | 5100 | 7600 | ----- |
| 47 | C7 | 4700 | 4737 | 7700 | Note ${ }^{6}$ |
| 48 | C8 | 4800 | 4838 | -..- | Note ${ }^{6}$ |
| 49 | C9 | 4900 | 4939 | 8400 | Note ${ }^{6}$ |


| (Scancode/ASCII code) |  |  |  |
| :---: | :---: | :---: | :---: |
| Norm | Shift | Ctrl | Alt |
| 0F09 | 0F00 | 9400 | A500 |
| 1071 | 1051 | 1011 | 1000 |
| 1177 | 1157 | 1117 | 1100 |
| 1265 | 1245 | 1205 | 1200 |
| 1372 | 1352 | 1312 | 1300 |
| 1474 | 1454 | 1414 | 1400 |
| 1579 | 1559 | 1519 | 1500 |
| 1675 | 1655 | 1615 | 1600 |
| 1769 | 1749 | 1709 | 1700 |
| 186F | 184F | 180F | 1800 |
| 1970 | 1950 | 1910 | 1900 |
| 1ASB | 1A7B | 1A1B | 1 A 00 |
| 1B5D | 1B7D | 1B1D | 1B00 |
| 2B5C | 2B7C | 2B1C | 2B00 |
| 53E0 | 53E0 | 93E0 | A300 |
| 4FE0 | 4FE0 | 75E0 | 9 FOO |
| S1E0 | S1E0 | 76E0 | A100 |
| 4700 | 4737 | 7700 | Note ${ }^{6}$ |
| 4800 | 4838 | 8D00 | Note ${ }^{6}$ |
| 4900 | 4939 | 8400 | Note ${ }^{6}$ |

## KEYBOARD SCAN CODES

| Key | Key Descript. | AT Mode |  |
| :---: | :---: | :---: | :---: |
|  |  | Make | Break |
|  |  | Code | Code |
| 58 | + | 79 | F079 |
| 59 | Caps Lock | 58 | F058 |
| 60 | A ora | 1 C | F01C |
| 61 | Sors | 1B | F01B |
| 62 | D ord | 23 | F023 |
| 63 | Forf | 2B | F02B |
| 64 | Gorg | 34 | I034 |
| 65 | Horh | 33 | 1 O 033 |
| 66 | J or j | 3B | F03E |
| 67 | K ork | 42 | F042 |
| 68 | Lorl | 4B | F04B |
| 69 | : or; | 4C | 104 C |
| 70 | "or' | 52 | F052 |
| 71 | Enter | 5A | 105A |
| 72 | 4 | 6 B | F06B |
| 73 | 5 | 73 | F073 |
| 74 | 6 | 74 | F074 |


| XT Mode |  |
| :---: | :---: |
| Make | Break |
| Code | Code |
| 4E | CE |
| 3A | BA |
| 1E | 9E |
| 1F | 9F |
| 20 | A0 |
| 21 | AI |
| 22 | A2 |
| 23 | A3 |
| 24 | A4 |
| 25 | A5 |
| 26 | A6 |
| 27 | A7 |
| 28 | A8 |
| 1 C | 9 C |
| 4B | CB |
| 4C | CC |
| 4D | CD |


| Standard ASCII (Scancode/ASCII code) |  |  |  | $\begin{gathered} \text { Extended ASCII } \\ \text { (Scancode/ASCII code) } \end{gathered}$ |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Norm. | Shift | Ctrl | Alt | Norm | Shift | Cirl | Alt |
| $4 \mathrm{E}_{2} \mathrm{~B}_{7}$ | $4 \mathrm{E} 2 \mathrm{~B}_{7}$ | ---- |  | $4 \mathrm{E} 2 \mathrm{~B}_{7}$ | $4 \mathrm{E}_{2} \mathrm{~B}_{7}$ | 9000 | $4 \mathrm{E} 00_{7}$ |
| Note | Note | --..- | Note ${ }^{7}$ | Note | Note | -..- | Note ${ }^{7}$ |
| 1 E 61 | 1E41 | 1E01 | 1E00 | 1E61 | 1 E41 | 1E01 | 1E00 |
| 1 F73 | 1 F53 | 1 F 13 | 1 F 00 | 1F73 | 1F53 | 1F13 | 1 F 00 |
| 2064 | 2044 | 2004 | 2000 | 2064 | 2044 | 2004 | 2000 |
| 2166 | 2146 | 2106 | 2100 | 2166 | 2146 | 2106 | 2100 |
| 2267 | 2347 | 2207 | 2200 | 2267 | 2247 | 2207 | 2200 |
| 2368 | 2348 | 2308 | 2300 | 2368 | 2348 | 2308 | 2300 |
| 246A | 244A | 240A | 2400 | 246A | 244A | 240A | 2400 |
| 256 B | 254B | 250B | 2500 | 256B | 254B | 250B | 2500 |
| 266C | 264C | 260 C | 2600 | 266C | 264C | 260C | 2600 |
| 273B | 273A | --." | ---- | 273B | 273A | .... | 2700 |
| 2827 | 2822 | ---- | ---- | 2827 | 2822 | --- | 2800 |
| 1C0D | 1COD | 1C0A | $\cdots$ | 1C0D | 1C0D | 1C0A | 1 COO |
| 4B00 | 4B34 | 7300 | Note ${ }_{6}$ | 4 BOO | 4B34 | 7300 | Note ${ }_{6}$ |
| ----- | 4 C 35 | --- | Note ${ }^{6}$ | 4COO | 4C35 | 8 FOO | Note ${ }^{6}$ |
| 4D00 | 4D36 | 7400 | Note ${ }^{6}$ | 4D00 | 4D36 | 7400 | Note ${ }^{6}$ |

## KEYBOARD SCAN CODES

| Key | Key Descript. | AT Mode |  | XT Mode |  | Standard ASCII (Scancode/ASCII code) |  |  |  | $\begin{gathered} \text { Extended ASCII } \\ \text { (Scancode/ASCII code) } \end{gathered}$ |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Make | Break | Make | Break |  |  |  |  |  |  |  |  |
|  |  | Code | Code | Code | Code | Norm | Shift | Ctrl | Alt | Norm | Shift | Ctrl | Alt |
| 76 | Left Shift | 12 | F012 | 2A | AA | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ |
| 77 | Z orz | 1A | F01A | 2C | AC | 2C7A | 2C5A | 2C1A | $2 \mathrm{C00}$ | 2C7A | 2C5A | 2C1A | $2 \mathrm{C00}$ |
| 78 | X or x | 22 | F022 | 2D | AD | 2D78 | 2D58 | 2D18 | 2000 | 2 D 78 | 2D58 | 2D18 | 2000 |
| 79 | Corc | 21 | F021 | 2E | AE | 2E63 | 2E43 | 2E03 | 2E00 | 2E63 | 2E43 | 2E03 | 2E00 |
| 80 | Vorv | 2A | F02A | 2F | AF | 2F76 | 2F56 | 2F16 | 2F00 | 2F76 | 2F56 | 2F16 | 2F00 |
| 81 | B orb | 32 | F032 | 30 | B0 | 3062 | 3042 | 3002 | 3000 | 3062 | 3042 | 3002 | 3000 |
| 82 | Norn | 31 | F031 | 31 | B1 | 316E | 314E | 310E | 3100 | 316E | 314E | 310E | 3100 |
| 83 | Morm | 3A | F03A | 32 | B2 | 326D | 324D | 320D | 3200 | 326D | 324D | 320D | 3200 |
| 84 | < or, | 41 | F041 | 33 | B3 | 332C | 333C | ----- | ----- | 332C | 333C | ----- | 3300 |
| 85 | $>$ or. | 49 | F049 | 34 | B4 | 342E | 343E | ---.. | ----- | 342E | 343E | ---- | 3400 |
| 86 | ? or $/$ | 4A | F04A | 35 | BS | $352 \mathrm{~F}_{8}$ | $3^{353 F_{8}^{8}}$ | -----8 | $\cdots$ |  |  |  |  |
| 87 | Right Shift | 59 | F059 | 36 | B6 | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ |
| 88 | Up Arrow | E075 | FOF075 | E02AF048 | E0C8F0AA | 4800 | 4800 | ----- | --- 6 | 48E0 | 48E0 | 8DE0 | 9800 |
| 89 | 1 or End | 69 | F069 | 4F | CF | 4F00 | 4F31 | 7500 | Note ${ }_{6}$ | 4F00 | 4F31 | 7500 | Note ${ }_{6}^{6}$ |
| 90 | 2 | 72 | F072 | 50 | D0 | 5000 | 5032 | $\cdots$ | Note ${ }_{6}$ | 5000 | 5032 | 9100 | Note ${ }_{6}$ |
| 91 | 3 or Pg Dn | 7A | F07A | 51 | D1 | 51009 | 51339 | 76009 | Note ${ }_{9}$ | ${ }_{5100}$ | 5133 | 76009 | Note ${ }_{9}$ |
| 92 | Left Ctrl | 14 | F014 | 1D | 9D | Note ${ }^{10}$ | Note | Note 9 | Note | Note ${ }^{9}$ | Note ${ }^{9}$ | Note ${ }^{9}$ | Note ${ }^{9}$ |
| 93 | Left Alt | 11 | F011 | 38 | B8 | Note ${ }^{10}$ | Note ${ }^{10}$ | Note ${ }^{10}$ | Note ${ }^{10}$ | Note ${ }^{10}$ | Note ${ }^{10}$ | Note ${ }^{10}$ | Note ${ }^{10}$ |
| 94 | Space | 29 | F029 | 39 | B9 | 3920 | 3920 | 3920 | 3920 | 3920 | 3920 | 3920 | 3920 |

## KEYBOARD SCAN CODES

| Key <br> \# | Key |  |
| :--- | :--- | :--- | :--- |
| Descript. |  |  | | AT Mode |
| :---: | :--- | :--- |
| Make |
| Code | | Break |
| :---: |
| Code |


| XT Mode |  | Standard ASCII (Scancode/ASCII code) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Make | Break |  |  |  |  |
| Code | Code | Norm | Shift | CtHL | Alt |
| E038 | E0B8 | Note ${ }_{9}^{10}$ | Note ${ }_{9} 10$ | Note9 ${ }^{10}$ | Note ${ }^{10}$ |
| EOID | E09D | Note ${ }^{\text {a }}$ | Note ${ }^{\text {a }}$ | Note ${ }^{9}$ | Note ${ }^{9}$ |
| E02AE04B | EOCBE0AA | 4B00 | 4B00 | 7300 | --.-. |
| E02AE050 | EODOEDAA | 5000 | 5000 | --- | ----- |
| E024E04D | EOCDEOAA | 4D00 | 4000 | 7400 |  |
| 52 | D2 | 5200 | 5230 | ---- | Note ${ }^{6}$ |
| 53 | D3 | 5300 | 532 E | ---- | ---- |
| E01C | E09C | 1C0D | 1COD | 1C0A | ---- |


| Extended ASCII <br> (Scancode/ASCII code) |  |  |  |
| :---: | :---: | :---: | :---: |
| Norm | Shift | Cul | Alt |
| Note ${ }_{9}^{10}$ | Note ${ }^{10}$ | Note ${ }^{10}$ | Note ${ }^{10}$ |
| Note ${ }^{9}$ | Note ${ }^{\text {a }}$ | Note | Note ${ }^{9}$ |
| 4BE0 | 4BED | 73E0 | 9 B 00 |
| 50E0 | 50 EO | 91 E 0 | A000 |
| 4DE0 | 4DEO | 74 E 0 | 9 D 00 |
| 5200 | 523H | 9200 | Note ${ }^{6}$ |
| 5300 | 532E | 9300 |  |
| E00D | E00D | E00A | A600 |

## NOTES

Note1 -INT OSH is invoked and a screen dump is performed
Note2 -the scroll lock active bit is toggled
Note3 -the pause state is initiated
Note4 -INT 1BH is invoked
Note5 -the numlock active bit is toggled
Note6 -ALT num pad generates raw ascii code of typed number
Note 7 -the caps lock active bit is toggled
Note8 -hold shift lock active until key is released
Note9 -hold control shift active until key is released
Note10-hold alternate shift active until key is released

### 5.0.1 COMMUNICATION MODE 1 (PC/XT Mode)

The keyboard communicates with the host computer using a synchronous serial protocol at approximately 9600 BPS. One start bit, eight data bits, no stop bit, and no parity are used to make up the nine bit data word. When no communications are in progress, the keyboard holds the data line low and the clock line high.

Before starting a transmission, the keyboard lowers the clock line as a Request To Send (RTS). The state of the data line is then checked. If the host system is holding the data line low, transmissions are disabled. The keyboard will retain the keycode for the pressed key in its buffer until the clock and data lines return to the idle state. The keyboard then resumes scanning the keyboard matrix until transmissions are enabled.

When transmissions are enabled, 100 to 250 microseconds after the keyboard drives the clock line low, the keyboard transmits its data in the previously described format. Data is valid during the time the clock line is high and for a minimum of 10 microseconds after the falling edge of the clock line. See Figure 2 for a timing diagram.
< KEYBOARD DATA OUTPUT-XT MODE >



FIGURE 2

### 5.0.2 COMMUNICATION MODE 2 (AT MODE)

The keyboard comminicates with the host computer using a synchronous serial protocol at approximately 9600BPS. One start bit, eight data bits, odd parity, and one stop bit form the eleven bit data word. This communication is bi-directional, with the keyboard clocking all data transfers. When no communications are in progress, the data and clock lines are high, indicating an idle state.
<KEYBOARD DATA OUTPUT-AT MODE > Figure 3 A


T1 ; $104 \mu \mathrm{~s} \pm 20 \%$
T2; $20 \mu \mathrm{~s}$ MIN.
T3 ; $20 \mu \mathrm{~s}$ MIN.
T4; $35 \mu \mathrm{~s} \pm 20 \%$
<KEYBOARD DATA INPUT-AT MODE > Figure 3B


T5 ; $60 \mu \mathrm{~s}$ MIN. (WAITING TIME FOR HOST DATA OUTPUT) T6; $5 \mu \mathrm{~s}$ MIN.

FIGURE 3

Before starting a transmission, the keyboard checks the status of the clock and data lines. Transmissions are disabled if the clock line is low and the code for the pressed key is held in the keyboard buffer until transmissions are enabled. If the clock line is high and the data line is low, the host system is sending a Request To Send (RTS), and the code for the pressed key is stored in the buffer.

When transmissions are enabled and no RTS is detected from the host system, both clock and data line will be high. The keyboard starts a transmission by sending a low start bit, followed by the rest of the data word. Data is valid 20 microseconds minimum prior to the falling edge of the clock. See Figure 3 for timing diagram.

During the transmission of a data word, the keyboard periodically checks the state of the clock line. If the clock line is low during these checks prior to the rising edge of the parity bit, a data collision occurs. When a data collision occurs, the keyboard stops transmitting, returns the code for the pressed key to the keyboard buffer, and prepares to respond to actions requested by the host system.

### 5.1 COMMANDS FROM THE KEYBOARD TO THE HOST SYSTEM

Keyboard Buffer Overran - AT and XI Modes
This buffer can store up to sixteen codes for pressed keys. When this buffer is full and the seventeenth code is received, this code is replaced by an Overrun code. The following chart shows the Overrun Codes for each mode.

```
XT Mode FF Hex
AT Mode 00 Hex
```

Codes received after this code are lost until the keyboard clears additional space in the keyboard buffer.

Self Test passed - AA Hex AT and XI Modes
The keyboard issues this command upon successful completion of the keyboard self test. The self test consists of the following.

## XT MODE

1. Memory is cleared.
2. Keyboard buffer is cleared.
3. ROM checksum is read and compared.
4. RAM is tested.
5. Self test completion code is output.
6. AA Hex is output to indicate successful self test.
7. FC Hex is output to indicate a defect in the self test.

## AT MODE

1. ROM checksum is read and compared.
2. RAM is tested.
3. AA Hex is output to indicate successful self test.
4. FC Hex is output to indicate a defect in the self test.

This self test is initiated by the host system reset (Ctrl, Alt, Delete) or by Power-on Reset. Upon successful completion of the Self Test during Power-on Reset, the keyboard is set to XT mode if the keyboard detects a low level on the data line for more than 10 microseconds after 5 microseconds from the falling edge of the clock line. If this condition is not met, the keyboard is placed in the AT Mode, and Typematic Rate and Delay are set to the following defaults:

Typematic Rate 10.9 cps
Delay 500 milliseconds

ECHO - EE Hex AT and XT Modes
The Echo Command (EE Hex) is sent in response to an Echo Command from the host system instead of the nomal Acknowledge (ACK) for diagnostic purposes.

Acknowledge - FA Hex AT Mode
The keyboard sends an Acknowledge (FA Hex) in response to a valid cormand from the host system, with the exceptions of the Resend and Echo commands.

Resend -- FE Hex AT and XT Modes
The keyboard issues a Resend (FE Hex) in response to inputs which have parity errors, framing errors, or invalid data received from the host system.

KEYBOARD BUFFER OVERRUN - AT and XT Modes
When the 16-character keyboard buffer receives the 17th character, an overflow condition occurs. This condition is communicated to the host system by transmitting the Keyboard buffer Overrun (FF Hex for XT Mode, 00 Hex for AT Mode) to the host system.

### 5.2 COMMANDS FROM THE HOST SYSTEM TO THE KEYBOARD

Prior to sending commands to the keyboard, the host system must first check to see if the keyboard is sending data. If the keyboard is transmitting, and the data is past the parity bit, the host system must accept the data prior to initiating its own transmission.

If the keyboard's data has not yet reached the tenth clock pulse (Parity Bit), or is not transmitting data, the host system assumes control by lowering the clock line for a minimum of 60 microseconds, then releasing the clock line after clamping the data line low to indicate a start bit. The keyboard will respond with an RTS within 5 microseconds by clocking the start bit into the keyboard. The keyboard continues to clock data as shown in the timing diagram (Figure 3B). The host system must ensure that the data is valid prior to the rising edge and after the falling edge of the keyboard clock pulse.

After the parity bit, the host system should raise the data line to indicate a stop bit. The keyboard checks for a logical high stop bit, then clamps the data line low prior to clock in the stop bit. This signals the host system that the keyboard received the data correctly (Acknowledge). If the host system has not raised the data line to indicate a stop bit, a framing error results and the keyboard continues to clock data until the data line is raised by the host system. Upon receiving either a framing or parity error, the keyboard issues a RESEND to the host system.

All commands from the host system require a response from the keyboard. The keyboard will respond to these commands within 20 microseconds.

The following commands may be sent to the keyboard at any time, following the protocol described for the AT Mode. These commands are valid only in the AT Mode. During the reset command, the keyboard will not respond within 20 microseconds as described above.

RESET - FF Hex
Upon receiving this command, the keyboard transmits an
Acknowledge to the host system. The keyboard then waits for the host system to accept the Acknowledge response. The host system will accept the Acknowledge by raising the clock and data lines for a minimum of 500 mic roseconds.

The keyboard then executes the self test routine similar to the power-on Reset, and is placed in its default state.

RESEND - FE Hex
Upon receiving this command, the keyboard will transmit the last byte of data sent to the host system.

SET DEFAULT - F6 Hex
This command resets the keyboard to the Power-Up default state. The keyboard responds with an Acknowledge, clears the output buffer, sets the scanset to AT Mode, sets the default typematic rate and delay, and continues to scan the matrix.

DEFAULT DISABLE - F5 Hex
This command is similar to the Set Default command, except the keyboard stops scanning the matrix and waits for further instructions to be sent by the host system.

ENABLE - F4 Hex
Upon receipt of this command the keyboard responds with an Acknowledge, clears the output buffer, and starts scanning the matrix.

This command consists of one command byte and one parameter byte. The keyboard Acknowledges the command byte, stops scanning the matrix, and waits for the parameter byte. Upon receiving the parameter byte, the keyboard sends an Acknowledge, sets the typematic rate and delay as indicated in Table 2, and continues scanning the matrix.

If another command is received instead of the parameter byte, the set typematic rate/delay function ends with no change to the existing rate or delay parameters. The new command is processed and the keyboard continues scanning the matrix.

The parameter byte consists of an eight-bit word with bit 7 (most significant bit) always being set. Bits 0-4 set the typematic rate and bits $5-6$ set the delay.

Example
$\begin{array}{lllllllll}1 & 0 & 1 & 0 & 1 & 0 & 1 & 1 & \text { Parameter Byte } \\ 7 & 6 & 5 & 4 & 3 & 2 & 1 & 0 & \text { Bit }\end{array}$
$\begin{array}{lllllllll}7 & 6 & 5 & 4 & 3 & 2 & 1 & 0 & B i t ~ N u m b e r ~\end{array}$
The above example shows the default 10.9 cps typematic rate and 500 microsecond delay. Bits 0-4 (010111) correspond to the rate of 10.9 cps shown on Table 2 ii and bits 5-6 (01) correspond to the 500 microsecond delay shown in Table 2 i.

## See Table 2 for Typematic Rate/Delay values other than the default settings.

i) Delay

| Bit | Delay |
| :---: | :---: |
| 65 | ms |
| 00 | 250 |
| 01 | 500 |
| 10 | 750 |
| 10 | 1000 |

ii) Rate

| Bit    <br> 4 3 2 1 0 |  |  |  |  | Rate <br> (CPS) |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 30.0 |
| 0 | 0 | 0 | 0 | 1 | 26.7 |
| 0 | 0 | 0 | 1 | 0 | 24.0 |
| 0 | 0 | 0 | 1 | 1 | 21.8 |
| 0 | 0 | 1 | 0 | 0 | 20.0 |
| 0 | 0 | 1 | 0 | 1 | 18.5 |
| 0 | 0 | 1 | 1 | 0 | 17.1 |
| 0 | 0 | 1 | 1 | 1 | 16.0 |
| 0 | 1 | 0 | 0 | 0 | 15.0 |
| 0 | 1 | 0 | 0 | 1 | 13.3 |
| 0 | 1 | 0 | 1 | 0 | 12.0 |
| 0 | 1 | 0 | 1 | 1 | 10.9 |
| 0 | 1 | 1 | 0 | 0 | 10.0 |
| 0 | 1 | 1 | 0 | 1 | 9.2 |
| 0 | 1 | 1 | 1 | 0 | 8.6 |
| 0 | 1 | 1 | 1 | 1 | 8.0 |


| Bit     <br> 4 3 2 1 0 |  |  |  |  | Rate <br> (CPS $)$ |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 1 | 0 | 0 | 0 | 0 | 7.5 |
| 1 | 0 | 0 | 0 | 1 | 6.7 |
| 1 | 0 | 0 | 1 | 0 | 6.0 |
| 1 | 0 | 0 | 1 | 1 | 5.5 |
| 1 | 0 | 1 | 0 | 0 | 5.0 |
| 1 | 0 | 1 | 0 | 1 | 4.6 |
| 1 | 0 | 1 | 1 | 0 | 4.3 |
| 1 | 0 | 1 | 1 | 1 | 4.0 |
| 1 | 1 | 0 | 0 | 0 | 3.7 |
| 1 | 1 | 0 | 0 | 1 | 3.3 |
| 1 | 1 | 0 | 1 | 0 | 3.0 |
| 1 | 1 | 0 | 1 | 1 | 2.7 |
| 1 | 1 | 1 | 0 | 0 | 2.5 |
| 1 | 1 | 1 | 0 | 1 | 2.3 |
| 1 | 1 | 1 | 1 | 0 | 2.1 |
| 1 | 1 | 1 | 1 | 1 | 2.0 |

TABLE 2

ECHO - FE Hex
This command is provided for diagnostic purposes. The keyboard shall respond with EE Hex, instead of Acknowledge, and continue scanning the matrix.

## SET/RESET STATUS INDICATORS - ED Hex

This command consists of one command byte and one parameter byte. The keyboard Acknowledges the cormand byte, stops scanning the matrix, and waits for the parameter byte. Upon receiving the parameter byte, the keyboard sends an Acknowledge, sets the status indicators, and starts scanning the matrix.

If another command is received instead of the parameter byte, the keyboard disregards the Set/Reset Status Indicators command without changing the present status of the indicators, processes the new command, and starts scanning the matrix.

The parameter byte is an eight-bit word with bits 3-7 always set to low. Bit 0 is the Scroll Lock Indicator, bit 1 is the Num Lock Indicator, and bit 2 is the Caps Lock Indicator. A high in each individual bit indicates that Indicator is active and the indicator lamp should be on.

Example
$\begin{array}{lllllllll}0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & \text { PARAMETER BYTTE } \\ 7 & 6 & 5 & 4 & 3 & 2 & 1 & 0 & \text { BIT NUMBER }\end{array}$
The above example shows the power-on default of the Tandy 3000 NL. Bit 1 is high indicating Num Lock is active and the Num Lock indicator lamp is on.

READ KEYBOARD ID - F2 Hex
This command causes the keyboard to return two identification bytes AB83 Hex. The keyboard responds with an Acknowledge to the command and stops scanning the matrix. The keyboard then transmits the keyboard ID AB83 Hex and resumes scanning the matrix.

SET/READ SCAN SET - FO Hex
This command is used to select one of three Scan Sets or to tell the host system which Scan Set is currently being used. This command consists of a command byte and a parameter byte. Upon receiving this command, the keyboard sends an Acknowledge to the host system and waits for the parameter byte. When the keyboard receives the parameter byte, it responds with an Acknowledge.

A parameter byte of 00 Hex will cause the keyboard to transmit the Hex value for the Scan Set currently in use. A parameter byte of 01 Hex selects the Scan Set 1 (XI Scan Codes), 02 Hex selects Scan Set 2 (Default AT Scan Codes), and 03 Hex selects Scan Set 3 (Special AT Scan Codes - See Table 3 for Scan Codes and Default Key State Information). The keyboard resumes scanning the matrix.

TABLE 3
KEY DESCRIPTION
MAKE CODE
EREAR CODE
DEFAULT KEY STATE

| Esc | 08 | F0 08 | Make only |
| :---: | :---: | :---: | :---: |
| Fl | 07 | F0 07 | Make Only |
| F2 | 0 F | FO 0F | Make Only |
| F3 | 17 | F0 17 | Make only |
| F4 | $1 F$ | FO 1F | Make only |
| F5 | 27 | F0 27 | Make only |
| F6 | 2 F | F0 2F | Make only |
| F7 | 37 | F0 37 | Make only |
| F8 | 3 F | F0 3F | Make only |
| F9 | 47 | F0 47 | Make only |
| F10 | 4 F | F0 4F | Make Only |
| Fll | 56 | F0 56 | Make Only |
| F12 | 5 E | F0 5E | Make Only |
| Print Scrn | 57 | F0 57 | Make Only |
| Scroll Lock | 5 F | F0 5F | Make Only |
| Pause or Break | 62 | F0 62 | Make only |
|  | OE | F0 OE | Typematic |
| 1 | 16 | F0 16 | Typematic |
| 2 | 1 L | FO 1E | Typematic |
| 3 | 26 | FO 26 | Typematic |
| 4 | 25 | F0 25 | Typematic |
| 5 | 2E | FO 2E | Typematic |
| 6 | 36 | F0 36 | Typenatic |
| 7 | 3D | FO 3D | Typematic |
| 8 | 3E | F0 3E | Typematic |
| 9 | 46 | F0 46 | Typematic |
| 0 | 45 | F0 45 | Typematic |
| - | 4 E | F0 4E | Typenatic |
| = | 55 | F0 55 | Typematic |
| Backspace | 66 | F0 66 | Make only |
| Insert cursor pad | 67 | FO 67 | Make only |
| Home cursor pad | 6E | FO 6E | Make only |
| page Up cursor pad | 6F | F0 6F | Make only |
| Num Lock | 76 | F0 76 | Make Only |
| / number pad | 77 | F0 77 | Make only |
| * number pad | 7E | FO 7E | Make only |
| - number pad | 84 | F0 84 | Make only |
| Tab | OD | FO OD | Typematic |
| q | 15 | F0 15 | Typematic |
| w | 10 | F0 1d | Typematic |
| e | 24 | F0 24 | Typematic |
| $r$ | 2D | F0 2D | Typematic |
| t | 2 C | F0 2C | Typematic |
| y | 35 | F0 35 | Typematic |
| u | 3 C | F0 3C | Typematic |
| i | 43 | F0 43 | Typematic |
| $\bigcirc$ | 44 | F0 44 | Typematic |
| p | 4D | F0 4D | Typematic |
| [ | 54 | F0 54 | Typematic |
| ] | 5B | F0 5B | Typematic |


| $\backslash$ | 5 C | FO 5C | Typematic |
| :---: | :---: | :---: | :---: |
| Delete cursor pad | 64 | F0 64 | Typematic |
| Hame cursor pad | 65 | F0 65 | Make only |
| Page Up cursor pad | 6 D |  | Make only |
| 7 number pad | 6 C |  | Make only |
| 8 number pad | 75 | F0 75 | Make Only |
| 9 number pad | 7 D | F0 7D | Make Only |
| + number pad | 7 C | F0 7c | Make only |
| Caps Lock | 14 | F0 14 | Make/Break |
| a | 1 C | F0 1c | Typematic |
| s | 18 | F0 1B | Typematic |
| d | 23 | F0 23 | Typematic |
| $f$ | 2B | F0 2B | Typematic |
| g | 34 | F0 34 | Typematic |
| h | 33 | F0 33 | Typematic |
| j | 3B | F0 3B | Typematic |
| k | 42 | F0 42 | Typematic |
| 1 | 4B | FO 4B | Typematic |
| ; | 4 C | FO 4C | Typematic |
| - | 52 | F0 52 | Typematic |
| Enter | 5A | FO 5A | Typematic |
| 4 number pad | 6B | FO 6B | Make only |
| 5 number pad | 73 | F0 73 | Make only |
| 6 number pad | 74 | FO 74 | Make only |
| Left Shift | 12 | F0 12 | Make/Break |
| 2 | 1A | FO 1A | Typematic |
| $\mathbf{x}$ | 22 | FO 22 | Typematic |
| c | 21 | FO 21 | Typematic |
| v | 2A | FO 2A | Typematic |
| b | 32 | FO 32 | Typematic |
| n | 31 | F0 31 | Typematic |
| m | 3A | F0 3A | Typematic |
| , | 41 | F0 41 | typematic |
|  | 49 | F0 49 | Typematic |
| Right Shift | 59 | F0 59 | Make/Break |
| Up Arrow cursor pad | 63 | F0 63 | Typematic |
| 1 number pad | 69 | F0 69 | Make only |
| 2 number pad | 72 | F0 72 | Make only |
| 3 number pad | 7A | F0 7A | Make only |
| Enter number pad | 79 | F0 79 | Make only |
| Left Ctrl | 11 | F0 11 | Make/Break |
| Left Alt | 19 | F0 19 | Make/Break |
| Spacebar | 29 | F0 29 | Typematic |
| Right Alt | 39 | F0 39 | Make only |
| Right Ctrl | 58 | F0 58 | Make only |
| Left Arrow cursor | 61 | FO 61 | Typematic |
| Down Arrow | 60 | FO 60 | Typematic |
| Right Arrow | 6A | FO 6A | Typematic |
| Ins number pad | 70 | F0 70 | Make only |
| Del number pad | 71 | F0 71 | Make only |

SET ALL KEYS - TYPEMATIC - F7 Hex

SET ALL KEYS - MAKE/BREAK - F8 Hex

SET ALU KEYS - MAKE ONLY - F9 Hex

SET ALL KEYS - TYPEMATIC/MAKE/BREAK - FA Hex
These commands affect Scan Set 3 only, but may be sent using any Scan Set. The keyboard responds with an Acknowledge, clears the output buffer, sets all keys to the function requested by the command, and continues scanning the matrix if it was previously enabled.

SET SINGLE KEY -- TYPEMATIC -- FB Hex

SET SINGLE KEY - MAKE/BREAK — FC Hex

SET SINGLE KEY -- MAKE ONLY -- FD Hex
These commands consist of a command byte and a parameter byte. The keyboard responds to the command byte with an Acknowledge and waits for the parameter byte. The parameter byte is the scan code from Scan Set 3 for the key to be changed. Upon receiving the parameter byte, the keyboard sets the selected key to the function selected by the command byte, and continues to scan the matrix if it was previously enabled. These commands affect only Scan set 3 operation, but may be sent using any Scan Set.

### 5.3 KEY ROLLOVER

The keyboard incorporates N-Key Rollover in software to avoid loss of keystroke data during high speed entry. N-Key Rollover is defined as all keys pressed and released will be output in the proper sequence. However, when the keyboard detects more than four keys pressed during a scan of the matrix, the keyboard does not output the keycodes until one or more of the keys are released. If the released key was not properly detected as a pressed key, an error condition occurs and the keyboard issues an buffer overrun code to the host system.

### 5.4 AUTOREPEAT

The power-on default condition will cause the last key pressed to repeat at 10.9 characters-per-second after a 500 millisecond delay. This may be changed by the system when the keyboard is using the AT Communications Mode.

### 5.5 BUFFERING

The keyboard is capable of storing 16 scan codes in a first in/ first out (FIFO) circular buffer. When the buffer overflows, the last code is replaced by a Hex 00, in AT Mode and a Hex FF, in XT Mode.

### 5.6 STATUS INDICATORS

Three LED Status Indicators are provided: Num Lock, Caps Lock, and Scroll Lock.

These indicators are located in the keytop of each respective key. The keyboard will power up with all indicators OFF, except when the host system (such as the Tandy 3000 NL ) sets them to ON.

### 6.0 ELECTRICAL REQUIREMENTS

The interface consists of two bi-directional lines, clock and data, which are controlled by 74ISl25 equivalent buffers. The keyboard side is terminated by 2200 Ohm resistors. All voltage levels are TIL compatible and the keyboard drivers are capable of sinking 20 mA minimum including the current sourced by the pullup resistors on the keyboard.

### 6.1 CONNECTOR

The connector is a 5-pin DIN connector. Connections are shown in the following table.

Table 4

| PIN \# | SIGNAL |
| :---: | :--- |
|  |  |
| 1 | CLOCK |
| 2 | DATA |
| 3 | NO CONNECIION |
| 4 | LOGIC GROUND |
| 5 | +5 VOLTS DC |

6.2 CHASSIS GROUND
Chassis ground is isolated from logic ground.
6.3 POWER REQUIREMENTS
The keyboard requires 5 Volts $D C,+/-5 \%$, at 500 milliamps (max).
7.0 ENVIRONMENTAL REQUIREMENTS
7.1 TEMPERATURE
OPERATING 0 to 50 degrees $C$ NON-OPERATING. . . . . . . . . . . . - 20 to 60 degrees C
7.2 RELATIVE HUMIDITY
$20 \%$ to $90 \%$ non-condensing
7.3 SHOCK
Operating and non-operating ..... 10G 11 mS duration
7.4 VIBRATION
operating and non-operating .....
.0 .3 mm amplitude 10 to55 Hz
8.0 RELIIABILITY
8.1 SWITCH LIFE
Switch life of the keyboard is a minimum of 20 million cycles.


## Port 3

Port 3 is an 8 -bit quasi-bidirectional $/ \mathrm{O}$ port. It also contains the interrupt, timer. serial port and $\overline{\text { RD }}$ and WR pins that are used by various options. The output latch corresponding to a special function must be programmed to a one (1) for that function to
operate. Port 3 can sink/source one TTL load. The special functions are assigned to the pins of Port 3. as follows:
-RXD/data (P3.0). Serial port's receiver data input (asynchronous) or data inputoutput (synchronous).

- TXD/clock (P3.1). Serial port's transmitter data output (asynchronous) or clock output (synchronous).
- INTO (P3.2). Interrupt 0 input or gate control input for counter 0 .
- $\overline{\text { INT 1 }}$ (P3.3). Interrupt 1 input or gate control input for counter 1.
- 70 (P3.4). Input to counter 0.
- T1 (P3.5). input to counter $t$.
- $\overline{W R}$ (P3.6). The write control signal latches the data byte from Port 0 into the External Data Memory.
- $\overline{\mathrm{KD}}$ (P3.7). The read control signal enables External Data Mernory to Port 0.


## RST/VPD

A low to high transition on this pin fat approximately 3V) resets the 8051. If $V_{P D}$ is held within its spec (approximately +5 V ), while $\mathrm{V}_{\mathrm{CC}}$ drops below spec. $V_{P D}$ will provide standby power to the RAM. When $V_{P O}$ is low, the RANTs current is drawn from VCC A small internal resistor permits power-on reset using only a capacitor connected to $V_{\text {CC }}$.

## ALE/PROG

Provides Address Latch Enable output used for latching the address into external memory during normal operation. Receives the program pulse input during EPROM programming.

## SEN

The Program Store Enable output is a control signal that enables the external Program Memory to the bus during normal fetch operations.

## EANDO

When held at a TTL high level. the 8051 executes instructions from the internal ROM/EPROM when the PC is less than 4096. When held at a TTL low level. the 8051 fetches all instuctions from external Program Memory. The pin also receives the 21V EPROM programming supply voltage.
ITAL
Input to the oscillator's high gain amplifier. A crystal or external source can be used.
XTAL2
Output from the oscillator's amplifier. Required when

[^1]$\qquad$

$\qquad$
-
$\square$


Zs
Circuit ground potential.
KC
*5V power supply during operation. programming and verification.

## Port 0

Port 0 is an 8-bit open drain bidirectional WO port.
$n$ is also the multiplexed tow-order address and data bus when using external memory. It is used for data input and output during programming and verificaion. Port 0 can sinkisource two ITL loads.

## Port 1

Port 1 is an 8-bit quasi-bidirectional I/O port. It is used for the low-order address byte during programming and verification. Port 1 can sintisource one TTL load.
Port 2
Port 2 is an 8 -bit quasi-bidirectional $V O$ port. it also emits the high-order 8 bits of address when accessing external memory. It is used for the high-order address and the control signals during programming and verification. Port 2 can sink/sourte one Th load.

## PIN DESCRIPTION


$\qquad$
a crystal is used.
1 I_

${ }^{-1}$








Disk Drive

# SONY 

MECHATRONIC PRODUCTS GROUP
PART NO. : 00-0065
REVISION : 12-88

ISSUE DATE : May 10, 88

## Product Specifications

MODEL NAME :
8790144 (MP-F11W-70D)

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2. Specifications

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1. Description

This document describes the specifications of the soNy 3.5" Micro Floppy Disk Drive, the MP-F11W-70D that is designed for general applications, even for lap-top computers. It features a low profile, a low power cunsumption, single power voltage, a light weight, ruggedness, high reliability and easiness to use.
2. Specifications
2.1 Configuration

The drive consists of Read/Write heads, head positioning mechanism, disk motor, interface logic circuit and Read/Write circuit.
2.2 Physical Dimensions

The detailed physical dimensions are shown in figure 2.1. The main dimensions are:

1) Height $: 25.4 \mathrm{~mm}$ ( 1.0 inch)
2) Width : 101.6 mm (4.0 inches)
3) Depth : 150.0 mm ( 5.9 inches) (excluding a front panel thickness)
2.3 Weight

Weight : $425 \mathrm{~g} \quad$ (0.94 pound)
2.4 Performance
2.4.1 Recording Capacity (unformatted, MFM)
1.0 Mbytes/cisk 0.5 Mbytes/surface
2.4.2 Transfer Rate

Burst transfer rate : $250 \mathrm{Kbits} / \mathrm{sec}$ for MFM


```
2.4.3 Access Time
    a. Track to Track Slew Rate : 3 msec min.
    b. Head Settling Time : }15\textrm{msec}\mathrm{ max.
    c. Motor Start Time : 500 msec max.
2.4.4 Functional
    a. Rotation Speed : 300 rpm
    The continuous speed variation is within +l.5%.
    The instantanuous speed variation is within }\pm1.5%\mathrm{ .
    b. Recording Density : 8717 BPI
        (Side 1, Track 79)
    c. Track Density : 135 TPI
    d. Number of Cylinders : 80
    e. Number of Tracks : 160
    f. R/W Heads : 2
```


### 2.4.5 Reliability

```
a. Mean Time Between Failures (MTBF) : 30,000 POH
b. Mean Time to Repair (MTTR) : 30 minutes
c. Preventive Maintenance (PM) : Not Required
d. Components life : 5 years or \(15,000 \mathrm{POH}\)
e. Error Rate :
1. Soft Read Error : Less than 1 per \(10^{9}\) bits read
2. Hard Read Error : Less than 1 per \(10^{12}\) bits read
3. Seek Error \(:\) Less than 1 per \(10^{6}\) seeks
```

```
2.5 Input Power Requirements
    2.5.1 Power Consumption
    Standby 0.1 W max.
    Operation (read/write mode) l.l W typ.
    2.5.2 Supply Voltages
    Voltage Max. Ripple
    Current
    +5.0V \pml0% 0.1Vpp 20 mA max. (Standby)
                                    220 mA Typ. (Read/Write)
                                    680 mA max. (Motor starts)
                                    890 mA max. (Step during motor rotates)
2.5.3 Current Profile
    See Figure 2.2
```



Figure 2.2. Current Profile

## Page 7

### 2.6 Environmental Limits

### 2.6.1 Temperature Range

| Operating | $: 5^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}$ ambient $\left(40^{\circ} \mathrm{F}\right.$ to $\left.122^{\circ} \mathrm{F}\right)$ |  |
| :--- | :--- | :--- |
| Transportation $:$ | $-40^{\circ} \mathrm{C}$ to $60^{\circ} \mathrm{C}$ | $\left(-40^{\circ} \mathrm{F}\right.$ to $\left.140^{\circ} \mathrm{F}\right)$ |
| Storage | $:-20^{\circ} \mathrm{C}$ to $60^{\circ} \mathrm{C}$ | $\left(-20^{\circ} \mathrm{F}\right.$ to $\left.140^{\circ} \mathrm{F}\right)$ |

### 2.6.2 Humidity Range

Operating : 8\% to $80 \%$ relative humidity with a wet bulb temperature of $29^{\circ} \mathrm{C}$ ( 85 F ) and no condensation.

Transportation
and Storage : 5\% to $95 \%$ relative humidity with a wet bulb temperature of $29^{\circ} \mathrm{C}$ (85F) and no condensation.

### 2.6.3 Vibration

Operating : The unit can perform Read/Write operations without errors at continuous vibration from 10 to 500 Hz at a maximum of 0.5 G along each of the two mutually perpendicular axes.

Transportation
and Storage : The unit can withstand continuous vibration from 10 to 500 Hz with a maximum level of 2.0 G along each of the three mutually perpendicular axes without any degradation of any characteristics below the performance specifications.
2.6.4 Shock

Operating : The unit can withstand a shock of 5.0G shock for 11 msec with a $1 / 2$ sine wave shape in each of the two mutually perpendicular axis while performing normal read/write functions without damage or any loss of data.

Transportation
and Storage : The unit when unpacked can witbstand an 11 msec with a $1 / 2$ sine wave shock of 70 G on any of the three mutually perpendicular axis.

### 2.6.5 Orientation

The drive can be set horizontal or vertical including a toploading. The detail orientations are shown in Figure 2.3.

3. Signal Interface
3.1 Connector and Pin Assignments
3.1.1 Signal connector

$$
\begin{array}{ll}
\text { Receptacle } & : 3 \text { M } 3414-6500 \times x \text { or equivalent } \\
\text { Cable } & : 3 M 3365 / 34 \text { or equivalent }
\end{array}
$$

### 3.1.2 Signal Connector Pin Assignment

| PIN | SIGNAL DESCRIPTION | PIN | SIGNAL DESCRIPTION |
| ---: | :---: | ---: | :---: |
| 1 | N.C. | 2 | N.C. |
| 3 | N.C. | 4 | N.C. |
| 5 | $+5 V$ | 6 | DRIVE SELECT 3 |
| 7 | $+5 V$ | 8 | INDEX |
| 9 | $+5 V$ | 10 | DRIVE SELECT 0 |
| 11 | $+5 V$ | 14 | DRIVE SELECT 1 |
| 13 | RETURN | 16 | MOTOR ON |
| 15 | RETURN | 18 | DIRECTION |
| 17 | RETURN | 20 | STEP |
| 19 | RETURN | 22 | WRITE DATA |
| 21 | RETURN | 24 | WRITE GATE |
| 23 | RETURN | 26 | TRACK O0 |
| 25 | RETURN | 28 | WRITE PROTECT |
| 27 | RETURN | 30 | READ DATA |
| 29 | N.C. | 32 | HEAD SELECT |
| 31 | N.C. | 34 | DISK CHANGE |
| 33 | N.C. |  |  |

3.2 DC Characteristics of Interface Signals
3.2.1 Output Signal from Drive

C-MOS Open Drain Driver is used for the Output.
Output voltage (VOH) : Open
Output voltage (VOL) : 0.4 V max. Output current (IOL) : 48 mA max.
3.2.2 Input Signal to Drive

```
Input voltage (VIH) : 2.2 V min. Input voltage (VIL) : 0.8 V max.
```


### 3.2.3 Recommended Circuit for Signal Interface

The detail interface circuits are shown in figure 3.1.
A 1 K ohm pull-up resister can be put on each output line from the drive. The recommended cable length is 1.5 m (4.92ft.) or less.

Recommended driver ICs : 7406, 7438 or equivalent.


Figure 3.1. Interface Circuits

### 3.3 Signal Definitions

### 3.3.1 DRIVE SELECT 0,1,2,3

The SELECT lines are used to enable or disable all other interface lines except a MOTOR ON line. When the SELECT line is true (low), the drive is enabled and is considered active. When the SELECT line is false (high), all controlled inputs except the MOTOR ON line are ignored and all output lines are disabled.
N.B. IN USE (LED) lamp

When a drive is selected, the IN USE lamp on the selected drive is turned on, and when a drive is not selected, it is turned off.

### 3.3.2 MOTOR ON

When this input is true (low), the spindle motor will start to run. When this line is made false (high), the spindle motor will decelerate and stop. The spindle motor will not rotate until a disk is inserted even if the MOTOR ON signal is low (true). If the MOTOR ON signal becomes false during either a write or erase operation, the disk motor will not stop rotating until both the ERASE GATE signal and the WRITE GATE signal become high (false).

### 3.3.3 STEP

When a drive is selected, a true (low) pulse on this line will cause the read/write heads to move to the adjacent track. The direction of the head movement is determined by the DIRECTION input at the trailing edge of the pulse. The step operation can be performed even if there is no disk inserted in the drive.

### 3.3.4 DIRECTION

When a drive is selected, a false (high) level on this input will cause a STEP input to move the read/write head away from the disk spindle. A true (low) level will cause a STEP input to move the read/write head toward the drive spindle.

### 3.3.5 HEAD SELECT

When a drive is selected, a true (low) level on this input will cause Head 1 (upper) to be selected. A false (high) level on this input will cause Head 0 (lower) to be selected. If the HEAD SELECT signal changes during either write or erase operation, the head will not be changed until both ERASE GATE and WRITE GATE signal become high (false).

### 3.3.6 WRITE GATE

When a drive is selected and this line is made true (low), the write current circuit is enabled and information in the WRITE DATA input may be written.

### 3.3.7 WRITE DATA

If a drive is selected, and the WRITE GATE is true (low), a true pulse (low) on the WRITE DATA line will cause a bit to be written on the disk. Pulses on this line will be neglected when WRITE GATE is false (high). No precompensation is required.

### 3.3.8 INDEX

When the drive is selected, a true (low) pulse is generated on this line by each revolution of the spindle.

### 3.3.9 TRACK 00

This line is true (low) when the drive is selected and the Read/Write head is positioned on track 00.
3.3.10 WRITE PROTECT

If a write-protect disk or no disk is inserted while a drive is selected, this line will be true (low) and the drive will not be able to write data.

### 3.3.11 READ DATA

When the drive is selected, a true (low) pulse is generated on this line every time a bit is detected.

### 3.3.12 DISK CHANGE

This line is true (low) whenever a disk is removed from the drive. The line will remain true (low) until both the following conditions have been met:

1. A disk is inserted.
2. A STEP pulse has been received when the drive is selected.
3.4 Timing Requirements
3.4.1 Head Access


| T 1 | $: 0.5$ usec min. |
| ---: | :--- |
| T 2 | $: 0.5$ usec min. |
| T 3 | $: 3.0 \mathrm{msec}$ min. |
| T 4 | $: 0.5$ usec min. |
| $\mathrm{T} 5: 0.5$ usec min. |  |
| T 6 | $: 18$ msec min. |
| T 7 | $: 0.5$ usec min. |

### 3.4.2 TRACK 00 Signal



Tl : 2.9 msec max.
T2 : 2.9 msec max.

### 3.4.3 Write Data Timing



*DISR IN, the disk-in sensor signal inside the drive, is low when a disk is inserted in the drive.

### 3.4.4 Read Data Timing



*DISR IN, the disk-in sensor signal inside the drive, is low when a disk is inserted in the drive.

### 3.4.5 Index Pulse



Tl* : 197 msec min., 203 msec max.
T2 : 1.5 msec min., 1.7 msec max.
*When the disk motor rotation is at the steady state.
3.4.6 Disk Change


T1 : 0.5 usec max. $T 2: 1.0$ usec max.
*DISR IN, che disk-in sensor signal inside the drive, is low when a disk is inserted in the drive.

### 3.5 Power On and Power off Requirements

3.5.1 Data Protection

Turning power on or off will not cause any damage to recorded data on the disk as long as the drive is not in the midst of writing when the power is shut off or supplied.
3.5.2 Power Supply Sequencing

When the power is turned on, no special power supply sequencing is required. When the power is turned off, the power supply must fall monotonically to zero volt.
3.6 Power-On Reset Timing

A drive will be ready in 100 msec after power on. It takes up to 100 msec to reset the control IC of the drive.
4. Safety

The drive will meet the following product safety regulations:
U.L. 478
C.S.A. C.22.2, No. 154
U.L. 94V-0 for Front Bezel

## 5. Power On Initialization

In order to reduce the peak current requirement when drives are used in a daisy chain, the MP-FIIW-70D has been designed not to seek track 00 automatically. If all the drives connected in the daisy chain sought trach 00 simultaneously, this would place a significant power drain on the host system.
Thus, the host system must perform the following routine just after power on in order to reset the track counter inside the drive.

Power on initialization


# SONY 

MECHATRONIC PRODUCTS GROUP
PART NO. : 061
REVISION : 11-88
ISSUE DATE : Mar. 4, 88

## OEM MANUAL

## MODEL NAME:

MP-F11W / MP-F17W

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## 1. INTRODUCTION

## 1-1. Purpose

This material provides the information necessary to interface the MP-F11W/MP-F17W Micro Floppydisk drive to a floppy disk controller.

## 1-2. General description

The SONY Micro Floppydisk Drive represents a technological break through offering extreme lightweight, compactness, $(101.6 \mathrm{~mm}$ wide by 25.4 mm high by 150.0 mm deep, weighting 425 g ) that provides a versatile data storage component for the systems designer.

SONY's leadership in high-density recording techniques, perfected in video technology, enabled SONY engineers to develop the 3.5" Micro Floppydisk standard. Yet an unformatted, storage capacity of 1 MBytes (MP-F11W, double density), 2MBytes (MP-F17W high density) in 135 tracks per inch provides that of conventional $5.25^{\prime \prime}$ disks or more.

A semi-rigid protective shell provides protection unique to the Micro Floppydisk. When the disk is inserted into or taken out of the drive, its shutter automatically opens and closes to protect the disk from dust, dirt, fingerprints and other foreign objects that might degrade performance. A metal centering hub allows positioning with greater ease and more positive accuracy in over 30,000 disk interchanges.

The SONY proprietary read/write and tunnel erase head, developed using video techniques, is positioned by a precision stepper motor assembly, providing fast access while maintaining high positioning accuracy. High coercivity media accomplishes high data integrity with the SONY high density head.

The SONY Micro Floppydisk drive is interface signal-compatible with conventional $5^{\prime \prime}$ floppydisk drives. Accordingly, popular FDC chips such as Western Digital FD1WDs, NEC $\mu$ PD765 series can be used.

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Whether your application is small business systems, test instruments, personal computers or any related application, you will find that the Micro Floppydisk drive will offer a whole new range of possibilities.

The Micro Floppydisk drive offers the following features :

* 3.5 " floppy disk media with automatic shutter mechanism
* Large capacity ........................... 1MB(MP-F11W) / 2MB(MP-F17W)
* Small footprint
* Light weight
* Low power consumption
* High track density .................... 135 TPI
* High reliability ........................... MTBF 30,000POHs
* High transfer rate .................... $\quad 250 \mathrm{Kbps}(1 \mathrm{MB}), \quad 500 \mathrm{Kbps}(2 \mathrm{MB})$

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2. MECHANISM

The mechanism mounted on the chassis consists of a cassette compartment mechanism, a disk chucking and disk rotation mechanism, magnetic heads and carriage, a head positioning mechanism, some detectors, and so on.

## 2-1. Chassis

The chassis made of aluminum diecast is strong, highly durable, and precision-manufactured. This is the base structure for mounting most of mechanisms and printed circuit board.

## 2-2. Cassette compartment mechanism

The cassette compartment mechanism precisely positions the cassette on chassis by one touch operation. The mechanism is designed that the cassette is ejected easily as well.

2-3. Disk chucking and disk rotation mechanism
The disk chucking mechanism is mounted on the top of the thin brushless direct drive DC motor. The disk chucking mechanism precisely positions a disk and drives it.

The disk rotation mechanism directly mouted in a chassis consists of a printed circuit board for drive control, a rotor magnet, a rotor yoke, a stator yoke, and so on.

CENTERHOLE


Fig. 2-1 Chucking Mechanism

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2-4. Magnetic heads and head carriage
The tunnel erase type head with narrow track width is developed to attain a 135 TPI track density. The head is designed for higher Read signal quality. The head surface is treated for less the disk wear and the head wear. The head comprises a read/write gap to read and write data and two erase gaps to erase the recorded track edge immediately after recording. The heads and the carriage are very impotant parts in the MFDD, so they are especially precision-assembled.


Fig. 2-2 Magnetic Head

## 2-5. Head positioning mechanism

The head positioning mechanism consists of a stepping motor, a lead screw, a bearing, and a guide shaft. The head carriage is held with the guide shaft and the lead screw with stepping motor. The 4 -phase stepping motor rotates 2 micro steps ( $18^{\circ} \mathrm{X} 2$ ) per track.

## 2-6. Detection mechanism

## a) Write protect

The micro switch is mounted on the printed circuit board to detect the position of the write protect tab on the disk. If a write-protected disk is inserted and the micro switch is not pushed down, the recording/erasing power is not supplied and the data is protected from an erroneous writing command.
b) Track 00

The mechanism consists of a photo-interrupter to detect the outermost track position (Track 00). The head carriage is not moved more outer track, after the MFDD detects the Track 00 . So, the head and head positioning mechanism do not get damaged.

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c) Index

Since the MFDD is designed to make the spindle motor and the disk be fixed in the same corresponding position, the Index pluse is generated by the spindle motor circuit.
d) Disk-in

If the cassette pushes the micro switch on the printed circuit board, it is detected that the cassette is inserted.
e) Density select (Only for MP-F17Ws)

The micro switch is mounted on the printed circuit board to detect the hole of the high density disk. If a high density cassette is inserted and the micro switch is not pushed, the MFDD works in high density mode. If a nomal density disk is inserted and the micro switch is pushed down, the MFDD works in normal density mode.

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## 3. ELECTRONICS

The electronics of the MFDD consists of a read/write circuit, a control circuit and some drivers, and so forth. The electronic circuits are mounted on two printed circuit boards (Motor and Logic boards).

## 3-1. Read/write circuit

The read/write circuit consists of a read circuit, a write circuit, and a power on detection circuit. These circuits are mostly included in a read/write amplifier LSI(bipolar) on the Logic board.

## 3-2. Control circuit

The control circuit consists of an input and output interface circuit, a spindle motor control circuit, a stepping motor control circuit, a sensor circuit, and a function control circuit. These circuits are mostly included in a control LSI(C-MOS) on the Logic board.

## 3-3. Spindle motor driver circuit

This circuit consists of a three phase spindle motor driver, a rotation speed detector, a rotation phase detector, and a current limiter. These circuits are mostly included in a spindle motor driver IC(bipolar) on the Motor board.

3-4. Stepping motor driver circuit
This circuit consists of two bridge driver circuits. The IC is made by the bipolar process, which is mounted in the Logic board.

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Fig. 3-1 General Block Diagram

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## 4. SPECIFICATIONS

## 4-1. Drive performance

Table 4-1 lists performance specifications for the SONY Micro Floppydisk drive.

TABLE 4-1 Performance Specifications

| Specification | MP-F11W | MP-F17W (2MB mode) |
| :---: | :---: | :---: |
| CAPACITY <br> Unformatted per disk Unformatted per track Recording density Burst transfer rate | 1MB for MFM <br> 6.25 KB for MFM <br> 8,717BPI <br> 250Kbits/sec | 2MB for MFM <br> 12.5KB for MFM <br> 17,434BPI <br> $500 \mathrm{Kbits} / \mathrm{sec}$ |
| ACCESS TTME <br> Track to track <br> Setting Time | 3msec <br> 15 msec |  |
| FUNCTIONAL <br> Rotational speed Track density Cylinders Tracks R/W heads Encoding method | 300 r 1351 80 160 2 MEN |  |
| POWER CONSUMPTION Read/Write mode Standby mode $\qquad$ | $220 \mathrm{~mA}$ |  |


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## 4-2. Dimensional data

Table 4-2 lists the dimensional data for the Micro Floppydisk drive.
TABLE 4-2 Dimensional Data

| Physical Dimension | Value |
| :--- | :--- |
| Height | 25.4 mm |
| Width | 101.6 mm |
| Depth | 150.0 mm |
| Weight | 425 g |


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## 5. INTERFACE DESCRIPTION

## 5-1. Host system interface

The SONY Micro Floppydisk drive is compatible with conventional floppy disk controllers. The interface consists of 15 signal lines for data, control and hand-shaking (see Figure 5-1, details will appear in a product specification of each mode).


Fig. 5-1 Interface Signal Diagram

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## 5-2. Signal interface

The drive has 15 interface signals. Ten signals are input to the drive, and six are output.

Table 5-2 MNEMONIC

| SIGNAL DESCRIPTION | MNEMONIC |
| :--- | :--- |
| DRIVE SELECT 0 | SELECT 0 |
| DRIVE SELECT 1 | SELECT 1 |
| DRIVE SELECT 2 | SELECT 2 |
| DRIVE SELECT 3 | SELECT 3 |
| DIRECTION SELECT | DIRTN |
| STEP | STEP |
| WRITE DATA | WRTDATA |
| WRITE GATE | WRTGATE |
| HEAD SELECT | HDSL |
| INDEX | INDEX |
| TRACK00 | TRK00 |
| WRITE PROTECT | WRTPRT |
| READ DATA | RDDATA |
| MOTOR ON | MTON |
| DISKCHANGE | DSKCHG |


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## 6. EJECT BUTTON, INDICATOR AND SELECT SWITCH

## 6-1. Eject button

The eject button (see Fig. 6-1) is used to remove a disk cartridge from the unit. Depression of the eject button causes the disk cartridge in the unit to be ejected.

## 6-2. Indicator

The activity indicator (see Fig. 6-1) indicates that a drive is selected.


Fig. 6-1 Front Bezel

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## 6-3. Drive select switch

Drive Select switch located next to a power connector on the rear side is used to designate drive0 up to drive3 in a daisy chain application. Usually, this switch is set to drive 1 in shipping from the drive factory.


Fig. 6-2 Drive Select Switch

## SECTION 4 <br> PART REPLACEMENT

4-1 COVER ASS'Y RE PLACEMENT
4-1-1 Removal
a. Insert your finger into the rear right side portion of the Cover Ass'y as shown in Fig. 4-1, and dislocate the Cover Ass'y from the hook while applying some force to the direction marked with arrow. The whole Cover Ass'y can be taken out.

## 4-1-2 Installation

a. Match the projection located in the front side of the Cover Ass'y with square hole of the front panel ass'y and then put the rear of the Cover Ass'y into the proper position. (Refer to Fig. 4-1)


Fig. 4-1 Cover Ass'y Replacement

## 4-2 FRONT PANEL ASS'Y AND EJECT BUTTON RE PLACEMENT

## 4-2-1 Removal

a. Remove the cover ass'y. (Refer to 4-1)
b. Push slightly the plastic hinges located on each of right and left sides with ( - ) shaped screw driver through the square hole of upper cover, and pull it slowly toward you. (Refer to Fig. 4-2)
c. Dislocate the Eject Button from the hook of the slide plate.

## 4-2-2 Installation

a. Hang the Eject Button to the hook of slide plate as shown Fig. 4-2.
b. Install the Front Panel Ass'y into the metal frame by sliding in the hook located on each side of the Front Panel Ass'y.
c. Make sure that protuberances of LED and the Eject Button are properly located in the recess and square hole of the Eront Panel Ass'y.
d. Install the cover ass'y. (Refer to 4-1)


Fig. 4-2 Front Panel Ass'y Replacement 4-3 LG-2 MOUNTED BOARD RE PLACEMENT

4-3-1 Removal
a. Remove the cover ass'y. (Refer to 4-1)
b. Disconnect the all connectors. (CN 102 for stepping motor, CN 104 for 00 sensor, CN 105 and CN 106 for head carriage ass'y)
c. Desolder the jumper cable (CN1) on MT-2


Fig. 4-3 LG-2 Mounted Board Replacement
d. Remove the three screws ( $52.6 \times 4$ ) securing the LG-2 Mounted Board and then remove the LG-2 Mounted Board. (Refer to Eig 4-3)

## 4-3-2 Installation

2. Solder the jumper cable to the CN 1 of MT-2 mounted board.
b. Install the LG-2 Mounted Board with three screws (B2.6x4).
c. Connect the all connectors.
d. Perform the Radial Alignment and TRK00 sensor Adjustment. (Refer to 5-1)
c. Install the cover ass'y. (Refer to 4-1)

4-4 MT-2 MOUNTED BOARD RE PLACEMENT

## 44-1 Removal

a. Desolder the jumper cable on MT-2 Mounted Board with soldering iron.
b. Remove the four serews ( $\mathrm{P} 2.6 \times 4$ ) securing the Stator Yoke Ass'y and then remove the Stator Yoke Ass'y. (Refer to Eig. 4-4)
c. Disconnect the connenctor CN2 (hall IC ass'y) and then remove the MT-2 Mounced Board.


Fig. 4-4 MT-2 Mounted Board Replacement

## 4-2 2 Installation

a. Connece the connector CN2 (hall IC ass'y) onto the MT-2 Mounted Board.
b. Solder the Jumper Cable to CN1 of MT-2 Mounted Board.
c. Match the hole of the MT-2 Mounted Board with the emboss of the chassis ass'y.
d. Install the Stator Yoke Ass'y on MT-2 Mounted Board with four screws (P2.6×4), not to damage to six motor coils for disk motor. (Refer to Fig. 4-4)

## 4-5 HALL IC ASS'Y RE PLACEMENT

## 4-5-1 Removal

a. Remove the four screws ( $\mathrm{P} 2.6 \times 4$ ) securing the seator yoke ass'y and then remove the stator yoke ass'y. (Refer to Eig. 4-4)
b. Disconnect the connector CN2 (Hall IC Ass'y).
c. Remove the screw (PSW2x5) securing the Hall IC Ass'y and then remove the Hall IC Ass'y. (Refer to Fig. 4-5)


Fig. 4-5 Hall IC Ass'y Replacement

## 4-5-2 Installation

a. Easten loosely the Hall IC Ass'y with a screw (PSW2×5). But this screw must not be tightened for later adjustment.
b. Connect the connector CN2 (Hall IC Ass'y) onto the MT- 2 mounted board.
c. Match the hole of the MT-2 mounted board with the emboss of the chassis ass'y.
d. Install the stator yoke ass'y on MT- 2 mounted board with four screws (P2.6x4), not to damage to six motor coils for disk motor. (Refer to Fig. 4-4)
e. Perform the Index phase adjustment, (Refer to 5-3)

## 4-6 CASSE TTE HOLDER ASS'Y RE PLACEME NT 4-6-1 Removal

a. Connect the drive to the MFD Function Checker (Refer to Fig. 2-14), move the head to the TRKOO and then disconnect the drive from MFD Function Checker.
b. Remove the cover ass'y. (Refer to 4-1)
c. Remove the front panel ass'y and eject button. (Refer to 4-2)
d. Be careful not to apply the excessive force to the head carriage ass'y. While pushing the eject lever, take the Cassette Holder Ass'y. (Refer to Fig. 4-6)


Fig. 4-6 Cassette Holder Ass'y Replacement

## 4-6-2 Installation

a. Insert carefully the Cassette Holder Ass'y underncath the arm of the head carriage ass'y, then set the holder into the location shown by axrow as shown in Fig. 4-6, while pushing the eject lever.
b. Install the front panel ass'y and eject button. (Refer to 4-2)
c. Install the cover ass'y. (Refer to 4-1)

4-7 SLIDE PLATE ASS'Y RE PLACEMENT
4-7-1 Removal
a. Remove the cover ass'y. (Refer to 4-1)
b. Remove the front panel ass'y and eject button. (Refer to 4-2)
c. Remove the cassette holder ass'y. (Refer to 4-6)
d. Slide the trigger arm to set the Disk In-mode as shown in Fig. 4-7.
e. Remove the one end of Tension Springs an chassis ass'y and then take the Slide Plate Ass'y. (Refer to Eig. 4-7)


Fig. 4.7 Slide Plate Ass'y Replacement

## 4-7-2 Installation

a. While pushing the trigger arm, set the Slide Plate Ass'y into the location shown in Fig. 4-7.
b. Hang the Tension Springs on hook of chassis ass'y.
c. Push the Slide Plate Ass'y, while pushing the trigger arm back side. (This stays "disk-in mode")
d. Install the cassette holder ass'y. (Refer to 4-6)
e. Install the front panel ass'y and eject button. (Refer to 4-2)
f. Install the cover ass'y. (Refer to 4-1)
g. Make the head clean. (Refer to 5-4)

## 4-8 00 SE NSOR RE PLACEMENT

## 4-8-1 Removal

a. Connect the drive to the MFD Function Checker as shown in Fig. 2-14, move the head to TRK 79 and then disconnect the drive from the MFD Function Checker.
b Remove the cover ass'y. (Refer to 4-1)
c. Disconnect the connector CN 104 (00 Sensor) from the LG-2 mounted board.
d. Remove the screw (PSW2.6×5) securing the 00 Sensor and remove 00 Sensor. (Refer to Fig.


Fig. 4-8 00 Sensor Replacement

## 4-8-2 hastallation

a. Install the 00 Sensor with a screw (PSW 2.6× 5).
b. Connect the CN 104 connector onto the LG-2 mounted board.
c. Perform the radial alignoment and TRK00 sensor adjustment. (Refer to 5-1)
d. Install the cover ass'y. (Refer to $4-1$ )

## 4-9 HEAD CARRIAGE ASS'Y RE PLACEMENT

## 4-9-1 Removal

a. Remove the cover ass'y. (Refer to 4-1)
b. Remove the front panel ass'y and eject button. (Refer to 4-2)
c. Remove the cassette hoider ass'y. (Refer to 4-6)
d. Disconnect the connectors CN 105 and CN 106 (flexible boards) from the LG-2 mounted board.
e. Remove the screw ( $\mathrm{P} 2.6 \times 4$ ) securing the Guide Retainer and then remove the Guide Retainer, Head Carriage Ass'y and Slide Guide Shaft. (Refer to Fig. 4-9)
4-9-2 Installation
Note: Apply Sony oil to the Guide Shaft before installation. Apply Sony oil to the openings of Head Carriage Ass'y using the bamboo stick.
a. Pass the Guide Shaft through the opening of Head Carriage Ass'y.
Note: The spring of Head Carriage Ass'y, that is located in around shaft hole, should be installed inside so that the Guide Shaft is forced outwardly by the spring force, as shown in Fig. 4-9.
b. Put the Head Carriage Ass'y and the Slide Guide Shaft in place, and install the guide retainet with a screw ( $\mathrm{P} 2.6 \times 4$ ). (Refer to Fig. 4-9)
c. Connect the flexible boards to CN 105 and CN 106 on the LG-2 mounted board.


Fig. 4-9 Head Carriage Ass'y Replacement
d. Install the cassette holder ass'y. (Refer to 4-6)
e. Install the front panel ass'y and eject button. (Refer so 4-2)
f. Pertorm the radial alignment and TRK00 sensor adjustment. (Refer to 5-1)
g. Perform the head compliance. (Refer to 5-2)
h. Install the cover ass'y. (Refer to 4-1)
i. Make the head clean. (Refer to 5-4)

## 4-10 STE PPING MOTOR ASS'Y (ROTOR ASS'Y AND STATOR ASS'Y) RE PLACEMENT

## 4-10-1 Removal

a. Remove the cover ass'y. (Refer to 4-1)
b. Remove the front panel ass'y and eject button. (Refer to 4-2)
c. Remove the cassette holder ass'y. (Refer to 4-6)
d. Disconnecr the connector CN 102 (Stator Ass'y) from the LG-2 mounted board.
e. Remove the three screws (B2.6x4) securing LG-2 mounted board so that removal of Stator Ass'y can be easily performed.
f. Remove the two screws (PSW2.6x5) securing the Stator Ass'y and then remove the Stator Ass'y and steel ball. (Refer to Fig. 4-10)
g. Wipe away the grease applied around lead screw with soft cloth before removal of Stacor Ass'y and Rotor Ass'y, not to leave the grease in the chass'y hole during the removal.


Fig. 4-10 Stepping Motor Ass'y (Stator Ass'y and Rotor Ass'y/ Replacement
h. While twisting Rotor Ass'y, separate the Rotor Ass'y from the needle pin of head carriage ass'y.
i. Take the steel ball from the hole of chassis ass'y.

4-10-2 Inseallation
Note: The stepping motor must be replaced with the whole ass'y, since the wrong combination (Rotor Ass'y and Stator Ass'y) in the ass'y causes the mulfunction.

Note: Apply Molykote Grease (EM10L) (same quantity of match tip) on whole area of lead screw and two steel balls before the installation.
a. Insert the steel ball into hole of the chassis ass'y.
b. While lifting the head carriage ass'y a little, insert the lead screw of the Rotor Ass'y between the needle and plate spring of head carriage ass'y,
c. Insert the steel ball into hole of Rotor Ass'y.
d. Fasten loosely the Stator Ass'y with two screws (PSW2.6x5). But these screws must not be rightened for later adjustment.
e. Connect the connector CN102 (Rotor Ass'y) onto the LG-2 mounted board.
f. Install the cassette holder ass'y. (Refer to 4-6)
g. Install the front panel ass'y and eject button. (Refer to 4-2)
h. Perform the radial alignment and TRK00 sensor adjustment. (Refer to 5-1)
i. Install the cover ass'y. (Refer to 4-1)
j. Make the head clean. (Refer to 5-4)

## 6-6-2 Chip parts replacement procedure

This unit uses chip components such as carbon resistor, ceramic capacitor, transistor and diode in some circuits. It also uses IC's of flat-pack type. As the appearance of carbon resistor and ceramic capacitor are identical, destinguishment of each can be possible by visual check of reference address of silk-screen print on the printed circuit board. As the shape of transistor and diode are same, they also are distinguished by the reference address of silk-screen print. Tools:

Soldering iron: 20W
af possible, use soldering tip with heatcontroller of $270 \pm 10^{\circ} \mathrm{C}$ )
Desoldering metal braid ("SOLDER TAUL" or equivalent)
Solder (of 0.6 mm dia. is recommended.)

## Tweezers

## Soldering Conditioas:

Tip temperature; $270 \pm 10^{\circ} \mathrm{C}$
Solder within 2 sec . per an electrode
Higher temperature or longer tip application than specified may be damaged to the chip component.
(1) Resistor and capacitor
2. Add heat onto the chip-part by the sop of soldering iron tip and slide the chip-part aside when the solder is melted.
b. Confirm visually with care that there is no pattern peeling, damage, and/or bridge where the part was removed or its surrounding.
c. Presolder the pattern into thin where the part was removed.
d. Place a new chip-part onto the pattern and solder both sides.

CAUTION: Do not use the chip-part again once used.

(2) Transistoc and diode
a. Cut the leads of the semiconductor part to be removed with a cutter.
b. Remove the each pin of semiconductor from the pattern by tweezers while heating the pin by soldering iron.
c. Confirm visually with care that there is no pattern peeling, damage, and/or bridge where the part was removed or its surrounding.
d. Presolder the pattern into thin where the part was removed.
e. Place a new semiconductor onto the pattern and solder the leads.

(3) IC (Flat-pack type)
a. Cut the leads of the IC to be cemoved with a cutter.
b. Remove the each pin of IC from the pateem by iweezers while heating the pin by soldering ifon.
c. Confirm visually with care that there is no pattern peeling, damage, and/or bridge where the part was removed or its surrounding.

d. Presolder the partern into thin where the part was cemoved.
e. Place a new IC onto the pattern and solder it.
f. Confirm by a tester that each conduction between IC's terminal and copper partern is surely made.
g. If not, resolder the portion.


## 6-2. MECHAMICAL PARTS LIST

Note: 1. Parts printed in Bold-Face type are normally stocked for replacement purposes. The remaining parts shown in this list are not normally required for routine service work. Orders for parts not shown in Bald-riace trpe will be processed, but allow for additional delivery time.
2. The screws and washers may be supplied with the substitution that is similar to ones listed because of part-standardisation program in SONY.

| No. | Parte No. | Description |
| :---: | :---: | :---: |
| 1 | 4-8010-280-4 | Hend Carriage Ass'y (0m) |
| 2 | 4-8010-253-A | Steppint Motor $\operatorname{Ass}^{\prime \prime} \mathrm{y}$ |
| 3 | 4-8010-283-4 | Cassette Holder Ase'y |
| 4 | 4-613-114-01 | Tention Spring |
| 5 | 4-8010-234-A | Slide Plate Ass'y |
| 6 | 4-613-132-01 | Teamion Spring |
| 7 | A-8051-404-A | LG-2 (1T) Mounted Board |
| 8 | A-8051-349-A | MT-2 (1A) Mounted Board |
| 9 | A-8080-284-A | Hall IC Ase'y |
| 10 | 8-759-404-25 | Em8458-b |
| 11 | $x-4613-102-1$ | Stator Yoke Ase'y |
| 12 | 1-808-058-11 | 00 seamer |
| 13 | 4-606-001-11 | Slide Guide Shaft |
| 14 | 4-613-103-01 | Triserer Art |
| 15 | 4-613-104-01 | spring |
| 16 | 4-613-105-01 | Guide Retainer |
| 17 | I-4613-105-1 | Bracket Ass'y (Left) |
| 18 | x-4613-106-1 | Bracket Ase'y (Riglt) |
| 19 | 4-613-145-02 | Cover |
| 20 | 3-701-428-01 | Special seren + 2.684 |
|  | 4-613-717-01 | Serev IL PSW $2 \times 5$ |
|  | 4-614-322-01 | Screv LR PSW 2.6x5 |
|  | 7-621-259-25 | 8crev +P 2.6x4 |
|  | 7-621-775-20 | Screw $+12.6 \times 5$ |
|  | 7-621-759-45 | Screv +PSW 2.6m6 |
|  | 7-623-923-11 | Hylone Washer 2.6 |

## 6-3. OVER ALL DIAGRAM



## 6-4. CIRCUIT DIAGRAM

## 6-4. Circuit Diagram on LG-2 (1T






- Component Side -



## - Pattern Side -



## 6-6-1. ELECTRIC PARTS IIST

Note: 1. All capacitors are in micto farads unless otherwise specified.
2. All inductors are in micro henrics unless otherwise specified.
3. All resistors are in olims.
4. "Cilip" stands for chip component.

Ref. No. Parts No.
Description
Ref. No. Parts No.

## RESISTORS

| R101 | 1-216-049-00 | METAL Clip | 1 K | 5\% 1/10W |
| :---: | :---: | :---: | :---: | :---: |
| R105 | 1-216-049-00 | METAL CHIP | 1K | $5 \%$ 1/10W |
| $R 106$ | 1-216-049-00 | METAL CIIP | 1 K | 5\% 1/10w |
| R107 | 1-216-049-00 | METAL CHIP | 1K | $5 \% 1 / 10 \mathrm{~W}$ |
| R108 | 1-216-049-00 | METAI. CHIP | 1K | 5\% 1/10W |
| R109 | 1-216-049-00 | METAL CHIP | 1 K | 5\% 1/10W |
| R110 | 1-216-049-00 | METAL CIIP | 1 K | 5\% 1/10W |
| R111 | 1-216-049-00 | metal chip | 1 K | $5 \% 1 / 10 \mathrm{~W}$ |
| R112 | 1-216-049-00 | METAL CIIP | 1K | $571 / 10 \mathrm{~W}$ |
| R114 | 1-216-049-00 | METAL CHIP | 1K | 5\% 1/10W |
| Ri 15 | 1-216-049-00 | METAL Clip | 1 K | $571 / 10 \mathrm{~W}$ |
| R116 | 1-216-049-00 | METAI. CaIP | 1K | $5 \%$ 1/10w |
| R117 | 1-216-049-00 | METAL CIIP | IK | 5\% 1/10W |
| R118 | 1-216-049-00 | METAL CIIP | 1 K | 5\% 1/10w |
| R119 | 1-216-049-00 | metal Cila | 1 K | 5\% 1/10w |
| R120 | 1-216-049-00 | METAL CHIP | 1 K | 5\% 1/10W |
| R124 | 1-216-073-00 | METAL CHIP | 10K | 5\% 1/10W |
| R125 | 1-216-045-00 | Metal Chip | 680 | 5\% 1/10w |
| R126 | 1-216-073-00 | METAL CHIP | 10K | 5\% 1/10w |
| R127 | $1-216-045-00$ | METAL CHIP | 680 | 5\% 1/10w |
| R128 | 1-216-085-00 | METAL CHIP | 33K | 5\% 1/10w |
| 1129 | 1-216-085-00 | METAL CIIP | 33K | $5 \%$ 1/10w |
| R130 | 1-216-119-00 | METAL CHIP | 820K | $5 \% 1 / 10 \mathrm{~W}$ |
| R131 | 1-216-073-00 | METAL CHIP | 10K | 5\% 1/10W |
| R132 | 1-216-073-00 | METAL CHIP | 10K | 52 1/10w |
| R133 | 1-216-049-00 | METAL CHIP | IK | 5\% 1/10w |
| R134 | 1-216-097-00 | METAL CHIP | 100K | 5\% 1/10w |
| R136 | 1-216-059-00 | METAL CHIP | 2.7 K | 5\% 1/10w |
| R138 | 1-216-059-00 | METAL CHIP | 2.7 K | $5 \%$ 1/10W |
| R139 | 1-216-166-00 | METAL CHIP | 47 | $5 \% 1 / 8 \mathrm{H}$ |
| R140 | 1-216-166-00 | METAL CHIP | 47 | 5\% 1/8w |
| 8141 | 1-216-166-00 | METAL CHIP | 47 | $5 \% ~ 1 / 8 W$ |
| R142 | 1-216-166-00 | METAL CIIP | 47 | $5 \%$ 1/8w |
| R143 | 1-216-166-00 | Metal. Chip | 47 | $5 \% \quad 1 / 8 \mathrm{~W}$ |
| R145 | 1-216-017-00 | METAL CHIP | 47 | $521 / 10 \mathrm{~W}$ |
| R1 47 | 1-216-049-00 | Metal Chip | IK | 52 1/10W |
| R148 | 1-216-041-00 | METAL CIIP | 470 | $5 \% 1 / 10 \mathrm{~W}$ |
| R149 | 1-216-041-00 | METAL CHIP | 470 | $5 \% ~ 1 / 10 \mathrm{~W}$ |
| R150 | 1-216-063-00 | METAL CHIP | 3.9K | $5 \% 1 / 10 \mathrm{~W}$ |
| R151 | 1-216-067-00 | METAL CHIP | 5.6K | $5 \%$ 1/10w |
| R152 | 1-216-085-00 | Metal. CHIP | 33K | $521 / 10 \mathrm{~W}$ |
| R154 | 1-216-748-11 | METAL CIIP | 39K | 1\% 1/10W |
| R155 | 1-216-091-00 | METAL CHIP | 56K | $5 \% 1 / 10 \mathrm{~W}$ |
| R1 56 | 1-216-061-00 | METAL CHIP | 3.3K | 5\% 1/10W |
| R158 | 1-216-075-00 | METAL CHIP | 12K | 5\% 1/10W |
| R161 | 1-216-043-00 | METAL. CHIP | 560 | 5\% 1/10W |
| R163 | 1-216-073-00 | METAL CHIP | 10K | 5\% 1/10w |
| R164 | 1-216-180-00 | METAL CRIP | 180 | 57 1/8W |
| R166 | 1-216-089-00 | METAL CHIP | 47K | $531 / 10 \mathrm{~W}$ |
| R168 | 1-216-089-00 | METAL CIIIP | 47K | $5 \% 1 / 10 \mathrm{~W}$ |
| R170 | 1-216-089-00 | METAL CHIP | 47K | 5\% $1 / 10 \mathrm{~W}$ |


| Ref. N | No. Parts No. | Description |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R176 | 1-216-073-00 | metal. Chlp | 10K | 5\% 1/ | 1/10W |
| R177 | 1-216-073-00 | METAL. CHIP | 10K | 5\% 1/ | 1/10W |
| R180 | 1-216-073-00 | HETAI. CHIP | 10K | $5 \% 1 /$ | 1/10w |
| RJ105 | 1-216-295-00 | metal. Chtp | 0 | 5\% 1/ | 1/10W |
| R3110 | 1-216-295-00 | METAL CHIP | 0 | 5\% 1/ | 1/10W |
| RJIl3 | 1-216-295-00 | METAL CIIIP | 0 | 5\% 1/ | 1/10W |
| SLA105 | 1-216-295-00 | METAL CHIP | 0 | 5\% 1/ | 1/10w |
| SLB102 | 1-216-295-00 | METAL CHIP | 0 | 5\% 1/ | 1/10w |
| SWITCA |  |  |  |  |  |
| S101 | $1-554-644-00$ | SWITCII, SLID |  |  |  |
| OSCILIATOR |  |  |  |  |  |
| $\times 101$ | 1-567-912-11 | OSCILIATOR, | ceramic | (4.91M | 1M) |
| IC LINR |  |  |  |  |  |
| PS101 | 1-532-727-11 | IC LINK |  |  |  |
| MT-2 (1A) MOUNTED BOARD |  |  |  |  |  |
| CAPACITORS |  |  |  |  |  |
| C4 | 1-1 24-26\|-00 | EL.ECT | 10 | 20\% | \% Sov |
| C5 | 1-163-017-00 | CERAMIC CHIP | 0.0047 | 10\% | \% 50 V |
| C6 | 1-163-037-11 | CERAMIC CHIP | 0.022 | 10\% | \% 25V |
| C7 | 1-163-081-00 | CERAMIC CHIP | 0.22 |  | 25 V |
| C8 | 1-163-038-00 | CERAMIC CIIIP | 0.1 |  | 25 V |
| C9 | 1-162-638-11 | CERAMIC CHIP | 1 |  | 16 V |
| Clo | 1-162-638-11 | CERAMIC CHIP | 1 |  | 16 V |
| C1I | 1-162-638-11 | CERAMIC CHIP | 1 |  | 16 V |
| C] 2 | 1-163-038-00 | CERAMIC CIIIP | 0.1 |  | 25 V |
| C13 | 1-163-038-00 | CERAMIC CIIP | 0.1 |  | 25 V |
| COMNECTOR |  |  |  |  |  |
| CN2 | $1-566-427-11$ | PIN, CONNECT | OR 3P |  |  |
| DIODE |  |  |  |  |  |
| D1 | 8-719-904-92 | GL-9HY2 (Yel | 10w) |  |  |
| IC |  |  |  |  |  |
| IC] | 8-759-305-25 | [1A] 3440MS |  |  |  |
| RESISTORS |  |  |  |  |  |
| R1 | 1-216-035-00 | METAl. CHIP | 270 | $521 / 1$ | /10w |
| R2 | 1-216-184-00 | METAL CHIP | 270 | 5\% 1/ | 1/8N |
| R4 | 1-216-073-00 | METAL CHIP | 10K | $5 \% 1 / 1$ | /10w |
| R5 | 1-216-073-00 | metal chip | 10K | $5 \% 1 / 1$ | /10W |
| R6 | 1-216-081-00 | hetal Chip | 22K | 5\% 1/1 | /10W |
| R7 | 1-216-095-00 | netal CEIP | 82K | $5 \% 1 / 1$ | /10w |
| R9 | 1-216-142-00 | HETAL CHIP | 4.7 | $5 \% 1 /$ | 1/8W |
| R10 | 1-216-142-00 | NETAL CHIP | 4.7 | $5 \%$ 1/ | 1/8W |
| R11 | 1-216-142-00 | METAL CHIP | 4.7 | $5 \%$ 1/ | 1/8W |
| R12 | 1-216-298-00 | RETAL CHIP | 2.2 | $5 \%$ 1/1 | /10w |
| R13 | 1-216-298-00 | METAL CHIP | 2.2 | 5\% 1/1 | /10w |
| R14 | 1-216-073-00 | NETAL CHIP | 10K | 5\% 1/1 | /10W |
| R15 | 1-216-073-00 | METAL CHIP | 10x | $5 \% 1 / 1$ | /10w |
| R16 | 1-216-298-00 | METAL CHIP | 2.2 | $5 \% 1 / 1$ | /10W |
| R17 | 1-216-298-00 | HETAL CHIP | 2.2 | $5 \% 1 / 1$ | /10w |
| R18 | 1-216-298-00 | hetal chip | 2.2 | 5\% 1/11 | /10W |
| RJ2 | 1-216-296-00 | METAL CIIIP | 0 | $5 \%$ 1/ | 1/8w |


|  | No. Parts No. | Description |  |  |
| :---: | :---: | :---: | :---: | :---: |
| R.J3 | 1-216-295-00 | METAL. CHIP | 0 | 5\% 1/10w |
| RJ5 | 1-216-295-00 | METAI. CIIP | 0 | 5\% 1/10W |
| RJ7 | 1-216-296-00 | METAL. CHIP | 0 | 5\% 1/8W |
| RJ8 | 1-216-295-00 | METAL. CHIP | 0 | 5\% 1/10W |
| SLAI | 1-216-295-00 | METAL CHIP | 0 | 5\% 1/10w |
| SHITCHES |  |  |  |  |
| S1 | 1-571-237-11 | SWrqck, MlC |  |  |
| S2 | 1-571-237-11 | SWITCH, MIC |  |  |

## Software

## Software

## Software Contents

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## BIOS Services

## Device I/O Services <br> Introduction

The BIOS (Basic Input/Output System) is the lowest-level interface between other software (application programs and the operating system itself) and the hardware. The BIOS routines provide various device input/output services as well as bootstrap and print screen and other services. Some of the services that BIOS provides are not available through the operating system, such as the graphics routines.
All calls to the BIOS are made through software interrupts (that is, by means of assembly language "INT $x$ " instructions). Each I/O device is provided with a software interrupt, which transfers execution to the routine.

Entry parameters to BIOS routines are normally passed in CPU registers. Similarly, exit parameters are generally returned from these routines to the caller in CPU registers. To insure BIOS compatibility with other machines, the register usage and conventions are, for the most part, identical.
The following pages describe the entry and exit requirements for each BIOS routine. To execute a BIOS call, load the registers as indicated under the "Entry Conditions" banner. (Register AH will contain the function number in cases where a single interrupt can perform more than one operation.) Then issue the interrupt given for the call. The following example can be used to read a character from the keyboard:

MOV AH, 0
INT 16H

Upon return, AL contains the ASCII character and AH the keyboard scan code.

Note: All registers except those used to return parameters to the caller are saved and restored by the BIOS routines.

Following is a quick reference list of software interrupts for all device I/O and system status services.

Service
Video Display
Equipment
Memory Size
Diskette
Serial Communications
System Services
Keyboard
Line Printer
Bootstrap Loader
System Clock

## Software Interrupts

10 hex ( 16 dec )
11 hex ( 17 dec )
$12 \mathrm{hex}(18 \mathrm{dec})$
13 hex ( 19 dec )
14 hex ( 20 dec )
15 hex (21dec)
16 hex ( 22 dec )
17 hex ( 23 dec )
19 hex ( 25 dec )
1Ahex (26 dec)
Keyboard
16 hex ( 22 dec )
Function SummaryAH $=0$ : $\quad$ Read keyboard (destructive with wait)
$\mathrm{AH}=1: \quad$ Scan keyboard (nondestructive, no wait)
$\mathrm{AH}=2: \quad$ Get current shift status
$\mathrm{AH}=5: \quad$ Store ASCII character and scan code inkeyboard buffer
$\mathrm{AH}=10 \mathrm{H}$ : Extended keyboard read
$\mathrm{AH}=11 \mathrm{H}:$ Extended ASCII status
$\mathrm{AH}=12 \mathrm{H}: \quad$ Extended shift status read
Function Descriptions
Read KeyboardRead the next character typed at the keyboard. Return the ASCII valueof the character and the keyboard scan code, removing the entry fromthe keyboard buffer (destructive read).
Entry Conditions
$\mathrm{AH}=\mathbf{0}$
Exit Conditions
$\mathrm{AL}=$ ASCII value of character
$\mathrm{AH}=$ keyboard scan code

## Scan Keyboard

Set up the zero flag ( Z flag) to indicate whether a character is available to read from the keyboard or not. If a character is available, return the ASCII value of the character and the keyboard scan code. The entry remains in the keyboard buffer (non-destructive read).

## Entry Conditions

$\mathrm{AH}=1$

## Exit Conditions

Z = no character available
$\mathrm{NZ}=\mathrm{a}$ character is available, in which case:
$\mathrm{AL}=$ ASCII value of character
$\mathrm{AH}=$ keyboard scan code

## Get Shift Status

Return the current shift status.

## Entry Conditions

$\mathrm{AH}=2$

## Exit Conditions

AL = current shift status (bit settings: set = true, reset=false)
Bit $0=$ RIGHT SHIFT key depressed
Bit 1 = LEFT SHIFT key depressed
Bit $2=$ CTRL (control) key depressed
Bit 3 = ALT (alternate mode) key depressed
Bit $4=$ SCROLL state active
Bit 5 = NUMBER lock engaged
Bit 6 = CAPS lock engaged
Bit $7=$ INSERT state active
4.

## Store ASCII Character

## Entry Conditions

$\mathrm{AH}=5$
$\mathrm{CL}=\mathrm{ASCII}$ character
$\mathrm{CH}=$ Scan Code

## Exit Conditions

AL $=00$ : Successful
AL $=01$ : Buffer full
[C] = Operation failed

## Extended Keyboard Read

## Entry Conditions

$\mathrm{AH}=10 \mathrm{H}$

## Exit Conditions

$\mathrm{AL}=$ ASCII value of character
$\mathrm{AH}=$ keyboard scan code

## Extended ASCII Status

## Entry Conditions

$\mathrm{AH}=11 \mathrm{H}$

## Exit Conditions

$\mathrm{Z} \quad=$ No character is available
$\mathrm{NZ}=$ A character is available, in which case:
$\mathrm{AL}=$ ASCII value of character
$\mathrm{AH}=$ keyboard scan code

```
Extended Shift Status Read
Entry Conditions
AH}=12
```


## Exit Conditions

AL $\quad=$ shift status (bit settings: set $=$ true, reset $=$ false $)$

```
Bit 7 = INSERT active
Bit \(6=\) CAPS LOCK active
Bit \(5=\) NUM LOCK active
Bit 4 = SCROLL LOCK active
Bit 3 = ALT pressed
Bit 2 = CTRL pressed
Bit 1 = LEFT SHIFT pressed
Bit 0 = RIGHT SHIFT pressed
AH = extended shift status (bit settings: set = true, reset \(=\) false)
Bit 7 = SYS REQ pressed
Bit \(6=\) CAPS LOCK active
Bit \(5=\) NUM LOCK active
Bit \(4=\) SCROLL LOCK active
Bit 3 = RIGHT ALT active
Bit \(2=\) RIGHT CTRL active
Bit 1 = LEFT ALT active
Bit 0 = LEFT CTRL active
```


## Video Display

These routines provide an interface for the video display - the output half of the console (CON) device. MS-DOS considers the video display to be the default standard output (STDOUT) device.

## Software Interrupts

10 hex ( 16 dec )

## Function Summary Table Supported Video BIOS Calls

INT 10H

| AH $=00$ | Set Video Mode |
| :--- | :--- |
| $\mathbf{A H}=01$ | Set Cursor Type |
| $\mathbf{A H}=02$ | Set Cursor Position |
| $\mathbf{A H}=03$ | Read Cursor Position |
| $\mathbf{A H}=05$ | Select Active Display Page |
| $\mathbf{A H}=06$ | Scroll Active Page Up |
| $\mathbf{A H}=07$ | Scroll Active Page Down |
| $\mathbf{A H}=08$ | Read Attribute/Character at <br> Current Cursor Position |
| $\mathbf{A H}=09$ | Write Attribute/Character at <br>  <br> Current Cursor Position |
| $\mathbf{A H}=\mathbf{0 A}$ | Write Character Only at <br> Current Cursor Position |
| $\mathbf{A H}=\mathbf{0 B}$ | Set Color Palette |
| $\mathbf{A H}=\mathbf{0 C}$ | Write Dot |
| $\mathbf{A H}=\mathbf{0 D}$ | Read Dot |
| $\mathbf{A H}=\mathbf{0 E}$ | Write TrY to Active Display |
| $\mathbf{A H}=\mathbf{0 F}$ | Current Video State |

## INT 10H

| $\mathrm{AH}=10$ | Color Palette Interface |
| :---: | :--- |
| $\mathrm{AL}=00$ | Set Individual Register |
| $\mathrm{AL}=01$ | Set Border Color |
| $\mathrm{AL}=02$ | Set All Palette Registers and Border |
| $\mathrm{AH}=13$ | Write String |
| $\mathrm{AL}=00$ | Write Character String |
| $\mathrm{AL}=01$ | Write Character String and |
|  | Move Cursor |
| $\mathrm{AL}=02$ | Write Character and Attribute Strings |
| $\mathrm{AL}=03$ | Write Character and Attribute Strings |
|  | and Move Cursor |

## Function Descriptions <br> Set CRT Mode

## Entry Conditions

$$
\begin{aligned}
\mathrm{AH} & =0 \\
\mathrm{AL} & =\text { mode value, as follows: }
\end{aligned}
$$

[^2]Note: If the high order bit of the AL register is 1 , then the video buffer is not cleared.

## Set Cursor Type

Set the cursor type and attribute.

## Entry Conditions

$\mathrm{AH}=1$
$\mathrm{CH}=$ bit values:
Bits 5-6 = an invisible or erratically blinking cursor
Bits 5-6 $=0$ : produces a visible, blinking cursor
Bits 4-0 $=$ start line for cursor within character cell
$\mathrm{CL}=$ bit values:
Bits 4-0 $=$ end line for cursor within character cell

## Set Cursor Position

Write (set) cursor position.

## Entry Conditions

$\mathrm{AH}=2$
$\mathrm{BH} \quad=$ page number (must be 0 for graphics modes)
$\mathrm{DH}=\operatorname{row}(0=$ top row $)$
DL $\quad=$ column $(0=$ leftmost column $)$

## Get Cursor Position

Read (get) cursor position.

## Entry Conditions

$\mathrm{AH}=3$
$\mathrm{BH}=$ page number (must be 0 for graphics modes)

## Exit Conditions

$\mathrm{DH}=$ row of current cursor position $(0=$ top row $)$
DL $=$ column of current cursor position ( $0=$ leftmost column)
$\mathrm{CX}=$ cursor type currently set [1]:
See previous "Set Cursor Type" ( $\mathrm{AH}=1$ ).

## Select Active Page

Select active display page (valid in alpha mode only).

## Entry Conditions

$\mathrm{AH}=5$
$\mathrm{AL}=0$ through 7: new page value for modes 0,1
$\mathrm{AL}=0$ through 3: new page values for modes 2,3
$\mathrm{AL}=80 \mathrm{H}$ : read CRT/CPU page registers
$A L=81 \mathrm{H}$ : set CPU page register to value in BL
$\mathrm{AL}=82 \mathrm{H}$ : set CRT page register to value in BH
$\mathrm{AL}=83 \mathrm{H}$ : set CRT and CPU page registers in BH and BL

## Exit Conditions

If Bit 7 of $\mathrm{AL}=1$ upon entry, then:
$\mathrm{BH}=$ contents of CRT page register
$\mathrm{BL}=$ contents of CPU page register

## Scroll Up

Scroll active page up.

## Entry Conditions

$\mathrm{AH}=6$
AL $=$ numbers of lines to scroll. The number of lines that will be left blank at the bottom of the window. ( $0=$ blank entire window)
CH = row of upper left corner of scroll window
CL = column of upper left comer of scroll window
DH = row of lower right corner of scroll window
DL $\quad=$ column of lower right corner of scroll window
BH $=$ attribute (alpha modes) or color (graphics modes) to be used on blank line

## Attributes

Color modes.
Foreground color:
Bit $0=$ blue
Bit $1=$ green
Bit $2=$ red
Bit $3=$ intensity
All bits off $=$ black
Background color:
Bit $4=$ blue
Bit $5=$ green
Bit $6=$ red
Bit $7=\quad$ blink
All bits off $=\quad$ white

## Scroll Down

Scroll active page down.

## Entry Conditions

$\mathrm{AH}=7$
AL $\quad=$ number of lines to scroll $(0=$ blank entire window $)$
$\mathrm{CH}=$ row of upper left corner of scroll window
CL = column of upper left comer of scroll window
DH $=$ row of lower right corner of scroll window
DL = column of lower right corner of scroll window
BH $=$ attribute (alpha modes) of color (graphics modes) to be used on blank line. See "Scroll Up" ( $\mathrm{AH}=6$ ) for attribute values and "Set Color Palette" (AH = 11) for color values.

## Read Attribute or Color/Character

Read a character and its attribute or color at the current cursor position.

## Entry Conditions

$\mathrm{AH}=8$
BH $=$ display page number (not used in graphics modes)

## Exit Conditions

$\mathrm{AL}=$ character read
AH $=$ attribute of character (alpha modes only)

## Write Attribute or Color/Character

Write a character and its attribute or color at the current cursor position.

## Entry Conditions

$\mathrm{AH}=9$
BH $=$ display page number (not used in graphics modes)
CX = number of characters to write
AL = character to write
BL $=$ attribute of character (for alpha modes) or color of character (for graphics modes. If Bit 7 of BL is set, the color of the character is XOR'ed with the color value). See "Scroll Up" $(\mathrm{AH}=6)$ for attribute values and "Set Color Palette" $(\mathrm{AH}=0 \mathrm{BH})$ for color values.

## Write Character Only

Write character only at current cursor position.

## Entry Conditions

$\mathrm{AH}=\mathbf{0 A H}$
BH $=$ display page number (valid for alpha modes only)
CX = number of characters to write
$\mathrm{AL}=$ character to write
BL $=$ color of character (graphics mode)

## Set Color Palette

Select the color palette.

## Entry Conditions

$\mathrm{AH}=0 \mathrm{BH}$
BH $=0$ : Set background color ( $0-15$ ) to color value in BL.

BL = color value:

| $1=$ blue | 5 | $=$ magenta | $9=$ light blue |
| :--- | :--- | ---: | :--- |
| 2 | $=$ green | $6=$ yellow | $10=$ light green |
| 3 | $=$ cyan | $7=$ light magenta |  |
| 4 | $=$ red | $8=$ dark grey | $11=$ light cyan |
|  | $12=$ light red | $15=$ white |  |

or
BH = 1: Set default palette to the number (0 or 1) in BL.
In black and white modes:
BL $=0: 1$ for white
$\mathrm{BL}=1: 1$ for black
In 4 color graphics modes:
BL $\quad=0: \quad(1=$ green, $2=$ red, $3=$ yellow $)$
BL $=1: \quad(1=$ cyan, $2=$ magenta, $3=$ white $)$
In 16 color graphics modes:

| $1=$ blue | $5=$ magenta | $9=$ light blue | $13=$ light magenta |
| :--- | :--- | :--- | :--- |
| $2=$ green | $6=$ yellow | $10=$ light green | $14=$ yellow |
| $3=$ cyan | $7=$ light grey | $11=$ light cyan | $15=$ white |
| $4=$ red | $8=$ dark grey | $12=$ light red |  |

Note: For alpha modes, Palette Entry 0 indicates the border color. For graphics modes, Palette Entry 0 indicates the border and the background color.

## Write Dot

Write a pixel (dot).
Entry Conditions
$\mathrm{AH}=0 \mathrm{CH}$
$\mathrm{DX}=$ row number
CX = column number
$\mathrm{AL}=$ color value (When Bit 7 of AL is set, the resultant color value of the dot is the exclusive OR of the current dot color value and the value in AL.)

## Read Dot

Read a pixel (dot).
Entry Conditions
$\mathrm{AH}=0 \mathrm{DH}$
$\mathrm{DX}=$ row number
CX = column number

## Exit Conditions

AL = color value of dot read

## Write TTY

Write a character in teletype fashion. (Control characters are interpreted in the normal manner.)

## Entry Conditions

$\mathrm{AH}=0 \mathrm{EH}$
$\mathrm{AL}=$ character to write
BL $=$ foreground color (graphics mode)
BH $=$ display page (alpha modes)

## Get CRT Mode

Get the current video mode.
Entry Conditions
$\mathrm{AH}=\mathbf{0 F H}$

## Exit Conditions

AL = current video mode. See the previous "Set CTR Mode" ( $\mathrm{AH}=0$ ) for values
$\mathrm{AH}=$ number of columns on screen
BH = current active display page

## Set Palette Registers

Sets palette registers.

## Entry Conditions

$\mathrm{AH}=10 \mathrm{H}$
$\mathbf{A L}=0$ : Set Palette register
$\mathrm{BL}=$ number of palette register ( $0-15$ ) to set
$\mathrm{BH}=$ color value to store
$\mathrm{AL}=1$ : Set border color register
$\mathrm{BH}=$ color value to store
AL $=2$ : Set palette color value to store and border registers

ES:DX points to a 17 -byte list.
Bytes 0-15 $=$ values for palette registers 0-15
Byte $16=$ value for border register

## Write String

Display a string of characters on screen.

## Entry Conditions

$\mathrm{AH}=13 \mathrm{H}$
$\mathrm{ES}: \mathrm{BP}=$ pointer to start of string
$\mathrm{CX}=$ length of string (attributes do not count)
$\mathrm{DX}=$ starting cursor position $(\mathrm{DH}=$ row, $\mathrm{DL}=$ column $)$
$\mathrm{BH}=$ page number (for text modes)
BL = attribute for characters (graphics modes)
$\mathrm{AL}=00$ : Characters only string, cursor not updated
$=01$ : Characters only string, cursor updated
$=02:$ Character, attribute alternating string, cursor not updated
$=03$ : Character, attribute alternating string, cursor updated

Control characters interpreted as w/function DEL.

## Serial Communications

These routines provide asynchronous byte stream I/O from and to the RS-232C serial communications port. This device is labeled the auxiliary (AUX) I/O device in the device list maintained by MS-DOS.

## Software Interrupts

14 hex ( 20 dec )

## Function Summary

$\mathrm{AH}=0$ : Reset Comm port
$\mathrm{AH}=1$ : Transmit character
$\mathrm{AH}=2:$ Receive character
AH $=3$ : Get current Comm status
DX $=$ communication port number (0 or 1 )

## Function Descriptions <br> Reset Comm Port

Reset (or initialize) the communication port according to the parameters in AL, DL, and DH.

## Entry Conditions

$\mathrm{AH}=0$
AL $=$ RS-232C parameters, as follows:
DX = port number (0 or 1 )

| 7 | 6 | 5 | 4 | 3 | 2 |
| :--- | :--- | :--- | :--- | :---: | :---: |$|$| 1 |
| :---: |
| Baud Rate |


| 000 | $=110$ baud $00=$ none | 0 | $=1$ bit | $10=7$ bits |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 001 | $=150$ baud $01=$ odd | 1 | $=2$ bits | $11=8$ bits |  |
| 010 | $=300$ baud $11=$ even |  |  |  |  |
| 011 | $=600$ baud |  |  |  |  |
| 100 | $=1200$ baud |  |  |  |  |
| 101 | $=2400$ baud |  |  |  |  |
| 110 | $=4800$ baud |  |  |  |  |
| 111 | $=9600$ baud |  |  |  |  |

## Exit Conditions

$\mathrm{AX}=R S$-232C status; See the following"Get Current Comm Status" ( $\mathrm{AH}=3$ )

## Transmit Character

Transmit (output) the character in AL (which is preserved).

## Entry Conditions

$\mathrm{AH}=1$
$\mathrm{AL}=$ character to transmit
DX = port number (0 or 1 )

## Exit Conditions

$\mathrm{AH}=$ RS-232C status; See the following "Get Current Comm Status" ( $\mathrm{AH}=3$ ). If Bit 7 is set, the routine was unable to transmit the character because of a timeout error.)
AL is preserved

## Receive Character

Receive (input) a character in AL (wait for a character, if necessary). On exit, AH will contain the RS-232 status, except that only the error bits ( $1,2,3,4,7$ ) can be set; the timeout bit (7), if set, indicates that data set ready was not received and the bits in AH are not meaningful. Thus, AH is non-zero only when an error occurred.

## Entry Conditions

$\mathrm{AH}=2$
DX = port number (0 or 1 )

## Exit Conditions

$\mathrm{AL}=$ character received
$\mathrm{AH}=$ RS-232C status; See the following "Get Current Comm Status" ( $\mathrm{AH}=3$ )

## Get Current Comm Status

Read the communication status into AX.

## Entry Conditions

$\mathrm{AH}=3$
DX $=$ port number $(0$ or 1$)$

## Exit Conditions

$\mathrm{AH}=R S-232 C$ status, as follows (set = true):
Bit $0=$ data ready
Bit $1=$ overrun error
Bit $2=$ parity error
Bit $3=$ framing error
Bit $4=$ break detect
Bit $5=$ transmitter holding register empty
Bit $\mathcal{C}=$ transmitter shift register empty
Bit $7^{\circ}=$ timeout occurred
AL $\quad$ modem status, as follows (set = true):
Bit $0=$ delta clear to send
Bit $1=$ delta data set ready
Bit $2=$ trailing edge ring detector
Bit $3=$ delta receive line signal detect
Bit $4=$ clear to send
Bit $5=$ data set ready
Bit $6=$ ring indicator
Bit $7=$ receive line signal detect

## Line Printer

These routines provide an interface to the parallel line printer. This device is labeled "PRN" in the device list maintained by the operating system.

## Software Interrupts

17 hex ( 23 dec )

Function Summary
$\mathrm{AH}=0$ : Print character
$\mathrm{AH}=1$ : Reset printer port
$\mathrm{AH}=2$ : Get current printer status

## Function Descriptions

Print a Character

## Entry Conditions

$\mathrm{AH}=0$
$\mathrm{AL}=$ character to print
DX $\quad=$ printer to be used (0-2)

## Exit Conditions

$\mathbf{0 A H}=$ printer status. See the following "Get Current Printer Status" (AH = 2)
(If Bit 0 is set, the character could not be printed because of a timeout error.)

## Reset Printer Port

Reset (or initialize) the printer port.
Entry Conditions
$\mathrm{AH}=1$
DX $\quad=$ printer to be used (0-2)

## Exit Conditions

$$
\begin{aligned}
\mathrm{AH}= & \text { printer status; See the following "Get Current } \\
& \text { Printer Status" }(\mathrm{AH}=2)
\end{aligned}
$$

## Get Current Printer Status

Read the printer status into AH.

## Entry Conditions

$\mathrm{AH}=2$

## Exit Conditions

DX $\quad=$ printer to be used (0-2)
$\mathrm{AH}=$ printer status as follows (set $=$ true):
Bit $0=$ timeout occurred
Bit $1=$ [unused]
Bit $2=$ [unused]
Bit $3=1 / O$ error
Bit $4=$ selected
Bit $5=$ out of paper
Bit 6 = acknowledge
Bit $7=$ not busy

## System Clock

These routines provide methods of reading and setting the clock maintained by the system. This device is labeled CLOCK in the device list of the operating system. An interface for setting the multiplexer for audio source is also provided.

## Software Interrupts

1A hex ( 26 dec )

## Function Summary

$\mathrm{AH}=0: \quad$ Get time of day
$\mathrm{AH}=1: \quad$ Set time of day
$\mathrm{AH}=2: \quad$ Read real-time clock
$\mathbf{A H}=3: \quad$ Set real-time clock
$\mathrm{AH}=4: \quad$ Read date from real-time clock
$\mathrm{AH}=5: \quad$ Set the date in the real-time clock
$\mathrm{AH}=80 \mathrm{H}$ : Set up sound multiplexer
The clock runs at the rate of $1,193,180 / 65,536$ per second (about 18.2 times per second).

## Function Descriptions <br> Get Time of Day

Get (read) the time of day in binary format.

## Entry Conditions

$\mathrm{AH}=0$

## Exit Conditions

$\mathrm{CX}=$ high (most significant) portion of the clock count
DX = low (least significant) portion of the clock count
$\mathrm{AL}=0$ of the clock was read or written (via $\mathrm{AH}=0,1$ ) within the current 24 -hour period; otherwise, $\mathrm{AL}=0$

## Set Time of Day

Set (write) the time of day using binary format.

## Entry Conditions

$\mathrm{AH}=1$
$\mathrm{CX}=$ high (most significant) portion of clock count
$\mathrm{DX}=$ low (least significant) portion of clock count

## Read Clock Time of Day

Read the time of day kept in the clock.

## Entry Conditions

## $\mathrm{AH}=2$

## Exit Conditions

$\mathrm{CH}=$ hours in BCD
$\mathrm{CL}=$ minutes in BCD
DH $=$ seconds in BCD

## Set Clock Time of Day

Set the time of day kept in the clock.

## Entry Conditions

$\mathrm{AH}=3$
$\mathrm{CH}=$ hours in BCD
$\mathrm{CL}=$ minutes in BCD
DH $=$ seconds in BCD

## Read Clock Date

Read the date kept in the clock.
Entry Conditions
$\mathrm{AH}=4$

## Exit Conditions

$\mathrm{CH}=$ century in BCD
CL = year in BCD
$\mathrm{DH}=$ month in BCD
DL = day in BCD

## Set Clock Date

Set the date kept in the clock.

## Entry Conditions

$\mathrm{AH}=5$
$\mathrm{CH}=$ century in BCD
CL = year in BCD
$\mathrm{DH}=$ month in BCD
DL = day in BCD

## Sound Multiplexer

Sets the multiplexer for audio source.

## Entry Conditions

$\mathrm{AH}=80$
$\mathrm{AL}=$ source of sound
$00=8253$ channel 2
02 = audio in
03 = complex sound generator chip

## Disk I/O Support for Diskette Only

## System Configuration

## Software Interrupt <br> 13 hex ( 19 dec )

## Function Summary

$\mathrm{AH}=0: \quad$ Reset diskette
$\mathrm{AH} \quad=1: \quad$ Return status of last diskette operation
AH $=2: \quad$ Read sector(s) from diskette
$\mathrm{AH}=3: \quad$ Write sector(s) to diskette
$\mathrm{AH}=4: \quad$ Verify sector(s) on diskette
$\mathrm{AH}=5 ; \quad$ Format track on diskette
$\mathrm{AH}=08 \mathrm{H}: \quad$ Read drive parameters
$\mathrm{AH}=15 \mathrm{H}: \quad$ Read DASD type
$\mathrm{AH}=16 \mathrm{H}:$ Diskette change line status

## Function Descriptions <br> Reset Diskette

Reset the diskette system. Resets associated hardware and recalibrates all diskette drives.

## Entry Conditions

$\mathrm{AH}=0$

## Exit Conditions

See the following "Exits From All Calls."

## Return Status of Last Diskette Operation

Returns the diskette status of the last operation in AH.

## Entry Conditions

$\mathrm{AH}=1$

## Exit Conditions

$\mathrm{AL}=$ status of the last operation. For values, see the following "Exits From All Calls."

## Read Sector(s) from Diskette

Read the desired sector(s) from the diskette into RAM.

## Entry Conditions

$\mathrm{AH}=2$
DL $=$ drive number ( $0-2$ if Tandy 1000 TL; 0-1 if Tandy 1000 SL)
DH = head number (0-1)
$\mathrm{CH}=$ track number (0-79)
CL $=$ sector number (1-9)
AL $=$ sector count (1-9)
ES:BX = pointer to disk buffer

## Exit Conditions

See the following "Exits from all Calls."
$\mathrm{AL}=$ number of sectors read

## Write Sector(s) to Diskette

Write the desired sector(s) from RAM to disk.

## Entry Conditions

$\mathrm{AH}=3$
DL = drive number ( $0-2$ if Tandy 1000 TL; 0-1 if Tandy 1000 SL)
$\mathrm{DH}=$ head number (0-1)
$\mathrm{CH}=$ track number (0-79)
CL $\quad=$ sector number (1-9)
$\mathrm{AL}=$ sector count (1-9)
$\mathrm{ES}: \mathrm{BX}=$ pointer to disk buffer

## Exit Conditions

See the following "Exits From All Calls."
AL $=$ number of sectors written

## Verify Sector(s) on Diskette

Verify the desired sector(s) are readable.

## Entry Conditions

$\mathrm{AH}=4$
DL $\begin{aligned}= & \text { drive number ( } 0-2 \text { if Tandy } 1000 \text { TL; 0-1 if Tandy } \\ & 1000 \mathrm{SL} \text { ) }\end{aligned}$
$\mathrm{DH}=$ head number (0-1)
$\mathrm{CH}=$ track number (0-79)
CL $\quad=$ sector number (1-9)
$\mathrm{AL}=$ sector count (1-9)

## Exit Conditions

See the following "Exits From All Calls."
$\mathrm{AL} \quad=$ number of sectors verified

## Format on Diskette

Format the desired track.

## Entry Conditions

$\mathrm{AH}=5$
AL $=$ sector count (1-9)
DL $=$ drive number ( $0-2$ if Tandy $1000 \mathrm{TL} ; 0-1$ if Tandy 1000 SL)
DH $=$ head number (0-1)
$\mathrm{CH}=$ track number (0-79)
CL $\quad=$ sector number (1-9)
34.
$\mathrm{ES}: \mathrm{BX}=$ pointer to a group of address fields for each track. Each address field is made up of 4 bytes. These are C, H, R, and N , where:
$\mathrm{C}=$ track number
$\mathrm{H}=$ head number
$\mathbf{R}=$ sector number
$\mathrm{N} \quad=$ the number of bytes per sector

$$
(00=128,01=256,02=512,03=1024)
$$

There is one entry for every sector on a given track.

## Exit Conditions

See the following "Exits From All Calls."

## Read Drive Parameters

Return the drive parameters.
Entry Conditions
$\mathrm{AH}=08 \mathrm{H}$
DL $\quad=$ drive number ( $0-2$ if Tandy 1000 TL; 0-1 if Tandy 1000 SL)

## Exit Conditions

$\mathrm{AX}=0$
$\mathrm{BH}=0$
CH = Maximum usable track number
CL = Maximum usable sector number
DH = Maximum usable head number
DL $=$ Number of diskette drives installed (0-2 if Tandy 1000TL; 0-1 if Tandy 1000 SL)
ES:Dl $=$ Pointer to diskette drive parameter table for the maximum media type supported on the specified drive

```
\(\mathbf{C F}=0:\) No error
CF = 1: Illegal parameter
```


## Read DASD Type

Return the change line status.

## Entry Conditions

$\mathrm{AH}=15 \mathrm{H}$
DL $=$ drive number ( $0-1$ if Tandy 10000 TL; 0-1 if Tandy 1000 SL)

## Exit Conditions

CF $\quad=1$ : Operation was not successful. Previous versions of the Tandy 1000 will return $C F=1$.
$\mathrm{AH}=1$ : Invalid command.
$\mathrm{CF}=0$ : Operation was successful
$\mathrm{AH}=0$ : Drive not present
$=1$ : Diskette, no change line available
= 2: Diskette, change line available

## Diskette Change Line Status

Return the status of the diskette change line.

## Entry Conditions

$\mathrm{AH}=16 \mathrm{H}$
DL $=$ drive number ( $0-2$ if Tandy 1000 TL; 0-1 if Tandy 1000 SL)

## Exit Conditions

$\mathrm{CF} \quad=0$ : If $\mathrm{AH}=0$
$\mathrm{CF}=1$ : If AH is non 0
$\mathrm{AH}=0$ : Diskette change signal not active
= 1: Invalid diskette parameter
= 6: Diskette change signal active
$=80$ : Diskette drive not ready (drive door is open)

## Exits From All Calls

$\mathrm{AH}=$ Status of operation, where set $=$ true
Error Code Condition
01H Illegal Function
02H Address Mark Not Found
03H Write Protect Error
04H Sector Not Found
06H Diskette Change Line Active
08H DMA Overrun
09H Attempt to DMA Across a 64 K Boundary
10H Bad CRC on Disk Read
$20 \mathrm{H} \quad$ Controller Failure
40H Seek Failure
80H Device Timeout, Device Failed to Respond
$[\mathrm{NC}]=$ operation successful ( $\mathrm{AH}=0$ )
$[\mathrm{C}] \quad=$ operation failed ( $\mathrm{AH}=$ error status)

## Equipment

This service returns the "equipment flag" (hardware configuration of the computer system) in the AX register.

## Software Interrupts

11 hex ( 17 dec )
The "equipment flag" returned in the AX register has the following meanings for each bit:
Reset $=$ the indicated equipment is not in the system
Set $\quad=$ the indicated equipment is in the system
Bit $0=$ diskette installed
Bit 1 = math coprocessor
Bits 2,3 always $=11$
Bits 4,5 initial video mode
$0140 \times 25$ Color
$10 \quad 80 \times 25$ Color
$1180 \times 25$ Monochrome
Bits 6,7 number of diskette drives (only if Bit $0=1$ )
$00 \quad 1$
$01 \quad 2$
103 (Tandy 1000 TL ONLY)
Bit $8 \quad 0=$ DMA present (always present)
1 = no DMA present
Bits 9, 10, 11 number of RS232 cards
Bit 12 game I/O adapter present (joystick)
Bit 13 not used
Bits 14,15 number of printers
38.

## Memory Size

This service returns the total number of kilobytes of RAM in the computer system (contiguous starting from Address 0 ) in the AX register. The maximum value returned is 640 .

## Software Interrupts

12 hex ( 18 dec )

## Bootstrap Loader

Track 0 , Sector 1 is read into Segment 0, Offset 7C00.
Control is then transferred as follows: $(C S)=0000 \mathrm{H}$
$(\mathrm{IP})=7 \mathrm{C} 00 \mathrm{H}$
(DL) - drive where bootstrap sector was read

## Software Interrupts

19 hex ( 25 dec )

## System Services

## Software Interrupts

15 hex (21dec)

## Function Summary

$\mathrm{AH}=\mathrm{COH}:$ Machine identification
$\mathrm{AH}=15 \mathrm{H}$ : Read and write EEPROM data should he 70 H , see below

## Function Descriptions <br> Machine Identification

The machine identification algorithm is the same as all previous Tandy 1000 's. As well, the Tandy 1000 SL and Tandy 1000 TL computers have a new BIOS call to further identify the machine.
All current and previous Tandy 1000 computers have the following machine identification:
Byte at address FFFF:E = FF hex (compatible with IBM PC)
Byte at address FC000:0 $=21$ hex (Tandy 1000 unique)

## Entry Conditions

$\mathrm{AH}=\mathrm{COH}$

## Exit Conditions

If CF $=0$
$\mathrm{ES}: \mathrm{BX}=$ pointer to machine identification data in ROM
DW 0003 Byte count of data that follows (always 3)
DB $x x$ Model ID
DB xx Submodel ID
DB $x x$ BIOS revision level
40.

IF CF $\quad=1$, the call is not supported (all previous versions of the Tandy 1000)

|  | Tandy 1000 SL | Tandy 1000 TL |
| :--- | :---: | :---: |
| Model ID | FF | FF |
| Submodel ID | 00 | 01 |
| BIOS revision level | $x x$ | $x x$ |

## Function Descriptions <br> Read From EEPROM

Read the 16 -bit value from the indicated EEPROM word.

## Entry Conditions

$\mathrm{AH}=70 \mathrm{H}$
$\mathrm{AL}=0$
$\mathrm{BL} \quad=$ word number to read $(0-63)$

## Exit Conditions

DX = word value
Carry flag set indicates EEPROM call not supported.

Write to EEPROM
Write a 16 -bit value to the indicated EEPROM word.
Entry Conditions
$\mathrm{AH}=70 \mathrm{H}$
$\mathrm{AL}=1$
BL $=$ word number to write (0-63)
DX $=$ word value to write
Exit Conditions
Carry Flag set indicates EEPROM call not supported.
Get ROM Page
Get ROM page addressed at EDO: $\varnothing$.
Entry Conditions:

$$
\begin{aligned}
& A B=7 \sigma H \\
& A L=2
\end{aligned}
$$

Exit Conditions:

$$
A C=\text { current ROM page }(\varnothing-6)
$$

Set ROM Page
Set ROM page addressed at EXOD: $\varnothing$.
Entry Conditions:

$$
\begin{aligned}
& A^{\prime} H=7 \varnothing H \\
& A C=3 \\
& D C=R O M \text { page }(\phi-6)
\end{aligned}
$$

Exit Conditions:
ROM page set if carry clear. Carry set if invalid ROM page specified.
42.

## Tandy 1000 SL and Tandy 1000 TL BIOS Sound Support

The BIOS in these computers has the same support for sound as all previous Tandy 1000 computers, as well as support for additional sound features. The API for this new BIOS support is defined in the following information.

## Software Interrupts

1A hex (26 dec)

## Function Summary

$\mathrm{AH}=81 \mathrm{H}$ : Get sound status
$\mathrm{AH}=82 \mathrm{H}$ : Input sound (from the microphone)
$\mathrm{AH}=83 \mathrm{H}: \quad$ Output sound (to the speaker)
$\mathrm{AH}=84 \mathrm{H}$ : Stop sound input and output

## Function Descriptions <br> Get Sound Status

Gets sound status.

## Entry Conditions

$$
\mathrm{AH}=81 \mathrm{H}
$$

## Exit Conditions

Not Busy:
$\mathrm{AX}=00 \mathrm{C} 4 \mathrm{H}$
$\mathrm{CF}=0$

Busy:
$\mathrm{AX}=00 \mathrm{C} 4 \mathrm{H}$
$\mathrm{CF}=1$

## Input Sound

Inputs sound from the microphone.
Entry Conditions
$\mathrm{AH}=82 \mathrm{H}$
ES:BX = buffer address
CX = buffer length
DX = transfer rate (1-4095, where 1 is the fastest transfer rate)

## Exit Conditions

Not Busy:
$\mathrm{AH}=0$
$\mathrm{CF}=0$
Busy:
$\mathbf{A H}=0$
$\mathrm{CF}=1$
44.

## Output Sound

Outputs sound to the speaker.

## Entry Conditions

AH $=83 \mathrm{H}: \quad$ Output sound (to the speaker)
ES: BX = buffer address
$\mathrm{CX}=$ buffer length
DX $=$ transfer rate (1-4095, where 1 is the fastest transfer rate)
$\mathrm{AL} \quad=$ volume $(0-7$, where $0=$ no sound $)$
Exit Conditions
Not Busy:
For sanyle rate $s$
$\mathrm{AH}=0$
$\mathrm{CF}=0$ samples/sec, the $A X$

CF $=0$
Busy:

$$
\begin{array}{ll}
\mathrm{AH} & =0 \\
\mathrm{CF} & =1
\end{array}
$$

value is:

$$
\Delta x=3.57 \cdot 10^{6}
$$

5

Stop Sound Input and Output
Stops sound input and output.

## Entry Conditions

$\mathrm{AH}=84 \mathrm{H}$
Notes: The transfer rate values in register DX are not the same for calls $\mathrm{AH}=82 \mathrm{H}$ and $\mathrm{AH}=83 \mathrm{H}$. To input a buffer of data with the $\mathrm{AH}=82 \mathrm{H}$ call with a given DX value, then play it back with the $\mathrm{AH}=83 \mathrm{H}$ call so that it sounds the same, set the DX value for output approximately 11.5 times as large as the DX value for input when run on a Tandy 1000 SL and approximately 10.0 times faster on a Tangy 1000 TL.

This BIOS call uses the DMA hardware to input and output the sound buffer. When functions $\mathrm{AH}=82 \mathrm{H}$ and $\mathrm{AH}=83 \mathrm{H}$ are called, the BIOS initiates the I/O and returns to the calling program immediately. When the DMA transfer is complete, the BIOS will receive a hardware interrupt and will execute a software INT 15 H with $\mathrm{AH}=91 \mathrm{H}$ and $\mathrm{AL}=\mathrm{FBH}$. If an application program needs to know when the data transfer is complete, it has to hook INT 15 H and watch for this event.

The BIOS call masks the hardware restriction of not being able to DMA across a 64 kilobyte memory address boundary from the calling program.

# Keyboard ASCII and Scan Codes Function Keys, Cursor Keypad, Numeric Keypad 

| $\begin{aligned} & \text { SCAN } \\ & \text { CODE } \end{aligned}$ | NORM CASE ASCII CODE |  | UPPER CASE ASCII CODE |  | CTRL CASEALT CASE ASCI CODE |  | ASCII CODE |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 38 | F1 | $\times 3 \mathrm{~B}$ | F11 | $\times 54$ | F21 | $\times 5 \mathrm{E}$ | F31 | $\times 68$ |
| 3C | F2 | $\times 3 \mathrm{C}$ | F12 | $\times 55$ | F22 | $\times 5 \mathrm{~F}$ | F32 | $\times 69$ |
| 3D | F3 | $\times 3 \mathrm{D}$ | F13 | $\times 56$ | F23 | $\times 60$ | F33 | $\times 6$ A |
| 3E | F4 | $\times 3 \mathrm{E}$ | F14 | $\times 57$ | F24 | $\times 61$ | F34 | $\times 68$ |
| 3 F | F5 | $\times 3 F$ | F15 | $\times 58$ | F25 | $\times 62$ | F35 | $\times 6 \mathrm{C}$ |
| 40 | F6 | $\times 40$ | F16 | $\times 59$ | F26 | $\times 63$ | F36 | $\times 6 \mathrm{D}$ |
| 41 | F7 | $\times 41$ | F17 | $\times 5 \mathrm{~A}$ | F27 | $\times 64$ | F37 | $\times 6 E$ |
| 42 | F8 | +42 | F18 | $\times 58$ | F28 | $\times 65$ | F38 | $\times 6 \mathrm{~F}$ |
| 43 | F9 | $\times 43$ | F19 | $\times 5 \mathrm{C}$ | F29 | $\times 66$ | F39 | $\times 70$ |
| 44 | F10 | $\times 44$ | F20 | $\times 5 \mathrm{D}$ | F30 | $\times 67$ | F40 | $\times 71$ |
| 57 | F11 | e8500 |  | e8700 |  | e8900 |  | e8800 |
| 58 | F12 | e8600 |  | e8800 |  | e8A00 |  | e8C00 |
| E037 | PrintScm* |  | PrintScrn* |  | CPrSc | $\times 72$ | SysRq* | -- |
| 46 | Scr Lock - |  | Scr Lock- |  | .. | -- | Scr Lock | -- |
| E145 | Pause* | .- | Pause* | -. |  |  | Pause* | - |
| E046 |  |  |  |  | Break* | $\times 00$ |  |  |
| E052 | Insert |  | $\times 52$ | -52E0 | - | -92EO | -- | - 1200 |
| E047 | Home |  | $\times 47$ | -47E0 | $\times 77$ | e77E0 | -- | e9700 |
| E049 | Page Up |  | $\times 49$ | -47E0 | $\times 84$ | e84E0 | -- | -9900 |
| E053 | Delete |  | $\times 53$ | -53E0 | -- | e93E0 | - | eA300 |
| E04F | End |  | $\times 45$ | e4FEO | x75 | e75E0 | -- | e9F00 |
| E051 | Page Down |  | $\times 51$ | -51E0 | $\times 76$ | -76EO | - | eA100 |
| E048 | Up |  | $\times 48$ | -48EO | - | e8DE0 | -- | e9800 |
| E04B | Left |  | $\times 4 \mathrm{~B}$ | e4BEO | $\times 73$ | e73E0 | - | -9800 |
| E050 | Down |  | $\times 50$ | e50EO | -- | e91E0 | -- | eA000 |
| E04D | Right |  | $\times 4 \mathrm{~d}$ | e4DEO | x74 | e74E0 | -- | e9D00 |
| 45 |  |  | Num Lock | -- | -- | -- | Num Lock | - |
| E035 | Num Lock-- |  |  | 2 F |  | e9500 |  | eA400 |
| 37 | Home | 2A | * | 2A |  | e9600 |  | e37FO |
| 47 |  | $\times 47$ | 7 | 37 | CliSc | $\times 77$ | $*$ | -- |
| 48 | UP Page Up | $\times 48$ | 8 | 38 |  | e8D00 | * | - |
| 49 |  | $\times 49$ | 9 | 39 | TOS | $\times 84$ | 7 | $\cdots$ |
| 4A | - | 2D | - | 2D |  | e8E00 |  | e4AFO |
| 4B | Left | $\times 48$ | 4 | 34 | LWord | $\times 73$ | 4 | -- |
| 4 C |  | e4CFO | 5 | 35 |  | e8F00 | 7 | -- |
| 40 | RIGHT | $\times 4 \mathrm{D}$ | 6 | 36 | RWord | x74 | $\Psi$ | - |
| $4 E$ | + | 2B | + | 2 B |  | -9000 |  | -4EFO |
| 4 F | End $\times$ | $\times 4 \mathrm{~F}$ | 1 | 31 | ErEOL | $\times 75$ | $*$ | -. |
| 50 | DOWN <br> Pg Dn | $\times 50$ | 2 | 32 |  | -9100 | $*$ | -- |
| 51 |  | $\times 51$ | 3 | 33 | ErEOS | x76 | 7 | -- |
| 52 | Ins | $\times 52$ | 0 | 30 |  | e9200 | $\Psi$ | - |
| 53 | Del | $\times 53$ |  | 2 E |  | e9300 | -- | -- |
| E01C | Enter ESC | OD | Enter | OD | LF | OA |  | eA600 |
| 01 |  | 1 B | ESC | 1B | ESC | 1B |  | c01FO |
| 02 | ESC | 31 | $!$ | 21 | -- | - | ALT1 | $\times 78$ |
| 03 | 2 | 32 | @ | 40 | NULL | 00 | ALT2 | $\times 79$ |
| 04 | 3 | 33 | * | 23 | -- | -- | ALT3 | x7A |
| 05 | 4 | 34 | \$ | 24 | - | -- | ALT4 | $\times 78$ |
| 08 | 5 | 35 | \% | 25 | -- | -- | ALT5 | $\times 7 \mathrm{C}$ |
| 07 | 6 | 36 | $\wedge$ | 5 E | RS | 1E | AlT6 | $\times 70$ |
| 08 | 7 | 37 | \& | 28 | -- | - | ALT7 | x7E |
| 09 | 8 | 38 | * | 2A | - | -- | ALT8 | x7F |


| $\begin{aligned} & \text { SCAN } \\ & \text { CODE } \end{aligned}$ | NORM CASE ASCII CODE |  | UPPER CASE ASCII CODE |  | CTRL CASEALT CASE ASCII CODE |  | ASCII CODE |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OA | 9 | 39 | 1 | 28 | -- | -- | ALT9 | $\times 80$ |
| OB | - | 30 | ) | 29 | -- | -- | ALTO | $\times 81$ |
| OC | - | 2 D | - | 5 F | us | 1F | ALT- | $\times 82$ |
| OD | = | 3D | + | 2B | - | -- | ALT $=$ | $\times 83$ |
| OE | BS | 08 | BS | 08 | DEL | 7 F |  | COEFO |
| OF | HT | 09 | BTab | xOF |  | e9400 |  | eA500 |
| 10 | 9 | 71 | Q | 51 | DC1 | 11 | ALTQ | $\times 10$ |
| 11 | w | 77 | W | 57 | ETB | 17 | ALTW | $\times 11$ |
| 12 | e | 65 | E | 45 | ENQ | 05 | ALTE | $\times 12$ |
| 13 | r | 72 | R | 52 | DC2 | 12 | ALTR | $\times 13$ |
| 14 | t | 74 | T | 54 | DC4 | 14 | ALTT | $\times 14$ |
| 15 | y | 79 | Y | 59 | EM | 19 | ALTY | $\times 15$ |
| 16 | u | 75 | U | 55 | NAK | 15 | ALTU | $\times 16$ |
| 17 | i | 69 | 1 | 49 | HT | 09 | ALTI | $\times 17$ |
| 18 | - | 6 F | 0 | 4F | SI | OF | ALTO | $\times 18$ |
| 19 | p | 70 | P | 50 | DLE | 10 | ALTP | $\times 19$ |
| 1 A | [ | 58 | \{ | 7B | ESC | 18 |  | e1AFO |
| 1 B | ] | 50 | \} | 70 | GS | 1 D |  | e1BFO |
| 1 C | Enter | OD | Enter | OD | LF | OA |  | e1CFO |
| 1D | Ctrl | -- | Ctrl | -. | Ctrl | -- | Ctrl | -- |
| E01D | CtrI | -- | CtH | -- | Ctri | .- | Ctr | -- |
| $1 E$ | a | 61 | A | 41 | SOH | 01 | ALTA | $\times 1 E$ |
| $1 F$ | s | 73 | S | 53 | DC3 | 13 | ALTS | $\times 1 F$ |
| 20 | d | 64 | D | 44 | EOT | 04 | ALTD | $\times 20$ |
| 21 | f | 66 | F | 46 | ACK | 06 | ALTF | $\times 21$ |
| 22 | g | 67 | G | 47 | 8EL | 07 | ALTG | $\times 22$ |
| 23 | h | 68 | H | 48 | BS | 08 | ALTH | $\times 23$ |
| 24 | j | 6A | $J$ | 4A | LF | OA | ALTJ | $\times 24$ |
| 25 | k | 6 B | K | 4B | $V T$ | OB | ALTK | $\times 25$ |
| 26 | 1 | 6 C | L | 4 C | FF | OC | ALTL | $\times 26$ |
| 27 | ; | 3B | : | 3A | -- | -- |  | e27FO |
| 28 |  | 27 | - | 22 | -- | - |  | e28F0 |
| 29 | ' | 60 | $\sim$ | $7 E$ | -- | - |  | e29F0 |
| 2A | LShift | -- | LShift | .- |  |  | LShift | -- |
| 2 B | 1 | 50 | 1 | 7 C | FS | 1 C |  | e2BFO |
| 2 C | 2 | 7A | 2 | 5A | SUB | 1 A | ALTZ | $\times 2 \mathrm{C}$ |
| 2D | $\times$ | 78 | X | 58 | CAN | 18 | ALTX | $\times 20$ |
| 2 E | c | 63 | C | 43 | ETX | 03 | ALTC | $\times 2 \mathrm{E}$ |
| 2 F | $v$ | 78 | $v$ | 56 | SYN | 16 | ALTV | $\times 2 \mathrm{~F}$ |
| 30 | b | 62 | B | 42 | STX | 02 | ALTB | $\times 30$ |
| 31 | $n$ | 6 E | N | 4 E | So | OE | ALTN | $\times 31$ |
| 32 | m | 60 | M | 4D | CR | OD | ALTM | $\times 32$ |
| 33 | , | 2 C | $<$ | 3 C | -. | D |  | e33F0 |
| 34 | - | 2 E | > | 3 E | -- | - |  | e34FO |
| 35 | 1 | $2 F$ | ? | 3F | $\cdots$ | - |  | e35FO |
| 36 | RShift | -- | RShift | -- | RShift | - | RShift | -- |
| 38 | Alt | -- | Alt | -. | Alt | - | Alt | -- |
| EO38 | Alt | -. | Alt | $\cdots$ | Alt | -. | Alt |  |
| 39 | SPACE | 20 | SPACE | 20 | SPACE | 20 | SPACE | 20 |
| 3 A | CapsLo |  | CapsLock | -- | -- | - | CapsLock | -- |
| $56+$ | 1 | 5C | \| | 7C | - | - | -. | .- |

48. 

## Keyboard Tables

These symbols have special meanings in the following tables:
-- Indicates that no ASCII code is generated for the key combination.
$x \quad$ Values preceded by x are extended ASCII codes. The keyboard driver returns a NULL ASCII code and the number in the table for the scan code.
e Values preceded by e are produced when you are using an enhanced BIOS. When using the BIOS Read Key function, these keys are either discarded or translated to a value compatible with older computers. When using the Enhanced Read Key function, $\mathrm{AH}=\mathbf{1 0 H}$, INT $\mathbf{1 6 H}$, these keys are returned to your program.
$+\quad$ A + in the scan code field denotes the extra key on the international version of the enhanced keyboard. This key is not available on the standard USA enhanced keyboard.
$¥ \quad$ The ALT key provides a way to generate the ASCII code of the decimal numbers in the range 1 to 255 . Hold down the aLT key while typing a number in that range on the numeric keypad. When you release the alt key, the character of the ASCII code you typed is generated and displayed.
BREAK Empties the keyboard queue and executes the keyboard break interrupt (INT 1BH). Places a NULL ASCII scan code in the keyboard queue.
PAUSE Delays system activity until you press another key.
PrtSC or Invokes the BIOS print screen function (INT 5H).
Print Scrn

CPrSc Tells MS-DOS to direct console output to both the printer and the console. A second CPrSc halts printer output.
SysRq Interrupts the current process and allows another program to take control, if supported. When the SysRq key is pressed, INT 15 H is invoked with $\mathrm{AX}=8500 \mathrm{H}$. When the key is released, INT 15 H is invoked with $\mathrm{AX}=8501 \mathrm{H}$.
Reset Restarts your computer.

# MS-DOS Memory Map 

| Hexadecimal |  |
| :--- | :--- |
| Starting Address |  |
| (Segment:Offset) | Description |
| 000:00 | BIOS Interrupt Vectors |
| 000:80 | Available Interrupt Vectors |
| $0040: 00^{1}$ | ROM BIOS Data Area |
| $0050: 00$ | MSDOS and BASIC Data Area |
| $0070: 00$ | I/O.SYS Drivers |
| $0190: 00^{2}$ | MS-DOS |
| 05B0:00 | Available to user |
| X800:00 | Video RAM in 32 K video modes |
| XC00:00 | Video RAM in 16K video mode |
| B800:00 | Video RAM Window (32K) |
| E000:00 | ROM Drive |
| F000:00 | Reserved for system ROM |
| FC00:00 | System BIOS ROM |

## Notes:

${ }^{1}$ Detailed description in following pages.
${ }^{2}$ Approximate address; subject to change.
${ }^{3} \mathrm{X}$ is defined as follows:

| Memory Size | X Value |
| :---: | :---: |
| 128 K | 1 |
| 256 K | 3 |
| 384 K | 5 |
| 512 K | 7 |
| 640 K | 9 |
| 768 K | B |

${ }^{4}$ Video memory accessed through the B800:0 window for all video modes.

## ROM BIOS Data Area

The following table gives the starting offset, and length of each BIOS device driver. This area is located at segment 40:00.

| Comm card address | 0000 |
| :--- | :--- |
| Printer addresses | 0008 |
| Devices installed | 0010 |
| Not used | 0012 |

Memory size 0013 (1 word)
I/O channel RAM size $0015 \quad 2$ (1 word)

KBD data area 001739
Disk data area 003E 11
Video data area 004930
Not used 0067
Clock data area $\quad 006 \mathrm{C}$
KBD Break \& Reset flags $0071 \quad 3$
Not used 0074
Printer timeout counter
Comm timeout counter
KBD extra data area
0078
007C 4 ( 1 byte per card)
0080
4 (2 words)

The structure and usage of the Video driver RAM data area is as follows:

## HEX Offset

| From Segment | Length and |
| :--- | :--- |
| 0040:0000 | Intended Use |


| 49 H | 1 byte - | current CRT mode (0-7) |
| :--- | :--- | :--- |
| 4 AH | 1 word - | screen column width |
| 4 CH | 1 word - | byte length of screen <br> 4 EH |
| 50 H | 8 word - | address/offset of beginning of <br> current display page |
| 60 H | 1 word -row/col coordinates of the <br> cursor for each of up to 8 <br> display pages <br> current cursor type (See <br> "Set Cursor Type" for correct <br> encoding) |  |

$62+1$ byte - current display paged word base address +4 of the CRT controller card

1 byte - copy of value written to the Mode Select Register
1 byte - current color palette setting

The equipment check BIOS call (INT 11H) and memory size BIOS call (INT 12 H ) return information from the following data areas:

HEX Offset

| From Segment | Length and |
| :---: | :--- |
| 0040:0000 | Intended Use |
| 10 H | Devices installed word |
| 13 H | Memory installed word |

The structure and usage of the diskette driver RAM data area is as follows:

| HEX Offset |  |  |
| :---: | :---: | :---: |
| From Segment 0040:0000 | Length and Intended Use |  |
| 3EH | 1 byte - | drive recalibration status - bit 3-0, if 0 then drive 3-0 needs recalibration before next Seek. Bit 7 indicates interrupt occurrence |
| 3FH | 1 byte - | motor status - Bit 3-0 drive 3-0 motor is on/off. Bit 7 current operation is write, requires delay |
| 40H | 1 byte - | motor turn off timeout counter (see Timer ISR) |
| 41H | 1 byte - | disk status - codes are defined as in this section |
| 42H | 7 bytes - | 7 bytes of status returned by the controller during result phase of operation |


| Value | Error Condition |
| :--- | :--- |
| 01H | Illegal Function |
| 02H | Address Mark Not Found |
| 03H | Write Protect Error |
| 04H | Sector Not Found |
| 06H | Diskette Change Line Active |
| 08H | DMA Overrun |
| 09H | Attempt to DMA Across a 64K Boundary |
| 10H | Bad CRC on Disk Read |
| 20H | Controller Failure |
| 40H | Seek Failure |
| 80H | Device Timeout, Device Failed to Respond |

The structure and usage of the RS232 driver RAM data area is as follows:
$\left.\begin{array}{ll}\begin{array}{ll}\text { HEX Offset } \\ \text { From Segment } \\ \text { 0040:0000 }\end{array} & \begin{array}{l}\text { Length and } \\ \text { Intended Use }\end{array} \\ 00 \mathrm{H} & 4 \text { words - }\end{array} \begin{array}{l}\text { Base address of each one of } 4 \\ \text { possible comm cards }\end{array}\right\}$

The structure and usage of the Keyboard driver RAM data area is as follows:
HEX Offset


1E

## Length and

 Intended Use1 byte- Keyboard shift state flag returned by function 02
Bits 7- INSERT state active, 6 - CAPS LOCK on/off, 5 - NUM LOCK on/off, 4 - SCROLL LOCK on/off, 3 - ALT key pressed 2 - CTRL key pressed 1 - Left SHIFT key pressed, 0 - Right SHIFT key pressed,
1byte- Secondary shift state flag, Bits INSERT key pressed, 6 - CAPS LOCK pressed, 5 - NUM LOCK pressed, 4 - SCROLL LOCK NUM LOCK pressed,
4 - SCROLL pressed,
4 - SCROLL LOCK pressed,
3 - Pause on/off, pressed,
3 - Pause on/off, 2,1,0 - not used
1 byte- Used to store ALT keypad entry
1 word- Pointer to beginning of the keyboard buffer
1 word- Pointer to end of the keyboard buffer

16- Keyboard buffer (enough for words)
15- Type ahead entries

The structure and usage of the clock service routine is as follows:

| HEX Offset <br> From Segment <br> 0040:0000 | Length and <br> Intended Use |
| :--- | :--- |
| 6 CH | 1 word $-\quad$Least significant 16 bits of <br> clock count |
| 6EH | 1 word - $\quad$Most significant 16 bits of <br> clock count |
| 70 H | 1 byte $-\quad$Twentyfour hour rollover <br> flag |

## Additional Data Area

HEX Offset<br>\section*{From Segment}<br>0040:0000



Bit 3: $0=$ Slow CPU speed mode

$$
1=\underset{ }{\text { Fast CPU speed }} \text { mode }
$$

Bit 4: $0=$ Internal color video support enabled
$1=\begin{aligned} & \text { Internal color } \\ & \text { video support }\end{aligned}$ disabled, external color video enabled

## HEX Offset <br> From Segment <br> 0040:0000

Bit 5: 0 = No external monochrome video installed
1 = External monochrome video installed
B6H $\quad 1$ byte Bit $0: 0=$ Drive C is $5-1 / 4^{\prime \prime}$
$1=$ Drive C is $3-1 / 2^{\prime \prime}$
40:C2 $\quad 1$ byte $\quad 01=$ ROM drive is $A:$
$02=$ ROM drive is $\mathrm{B}:$
03 = ROM drive is C:
oif $=$ ROM drive is $D$ :

```
thrs columa returned by
INT 15H function 4Fh
```


## KEYBOARD SCAN CODES



When $>$ shift key down: $\langle a / t\rangle$ overrides $\langle$ (entr1〉,


## this column returned by

 TNT 16 H function 0.0 HStandard ASCII
(Scancode/ASCII code)
 .
 this column returned by INT 164 function 10 it

## KEYBOARD SCAN CODES

| Key | Key | Hardware Kybrd |  |
| :---: | :---: | :---: | :---: |
| \# | Descript. | Make | Break |
|  |  | Code | Code |
| 19 | @ or 2 | 1 E | F01E |
| 20 | \# or 3 | 26 | F026 |
| 21 | \$ or 4 | 25 | F025 |
| 22 | \% or 5 | 2 E | F02E |
| 23 | ^ or 6 | 36 | F036 |
| 24 | \& or 7 | 3D | F03D |
| 25 | * or 8 | 3E | F03E |
| 26 | ( or 9 | 46 | F046 |
| 27 | ) or 0 | 45 | F045 |
| 28 | - or - | 4 E | F04E |
| 29 | + or = | 55 | F055 |
| 30 | Backspace | 66 | F066 |
| 31 | Insert | E070 | E0F070 |
| 32 | Home | E06C | E0F06C |
| 33 | Pg Up | E07D | E0F07D |
| 34 | Num Lock | 77 | F077 |
| 35 | 1 | E04A | E0F04A |
| 36 | * | 7 C | F07C |
| 37 | - | 7B | F07B |


| Kybrd Interrupt |  |
| :---: | :---: |
| Make | Break |
| Code | Code |
| 03 | 83 |
| 04 | 84 |
| 05 | 85 |
| 06 | 86 |
| 07 | 87 |
| 08 | 88 |
| 09 | 89 |
| 0A | 8A |
| 0B | 8B |
| ${ }^{0} \mathrm{C}$ | 8 C |
| OD | 8D |
| 0E | 8E |
| E02AE052 | E0D2E0AA |
| E02AE047 | E0C7EOAA |
| E02AE049 | E0C9E0AA |
| 45 | C5 |
| E035 | E0B5 |
| 37 | B7 |
| 4A | CA |

Standard ASCII
(Scancode/ASCII code)

| (Scancode/ASCII code) |  |  |  | (Scancode/ASCII code) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Norm. | Shift | Ctrl | Alt | Norm | Shift | Cirl | Alt |
| 0332 | 0340 | 0300 | 7900 | 0332 | 0340 | 0300 | 7900 |
| 0433 | 0423 | -.--- | 7A00 | 0433 | 0423 | -..-- | 7100 |
| 0534 | 0524 | ---- | 7800 | 0534 | 0524 | -..-- | 7B00 |
| (6)35 | 0625 | -...- | 7 COO | 0635 | 0625 | -..-- | 7 COO |
| 0736 | 075E | 071 E | 7 DOO | 0736 | 075 E | 071 E | 7D0) |
| 0837 | 0826 | -.--- | 7E00 | 0837 | 0826 | ----- | 7E00 |
| 0938 | 092A | ----- | 7F00 | 0938 | 092A | ----- | 7FOO) |
| 0A39 | 0A28 | -.... | 8000 | 0A39 | 0 A 28 | ----- | 8000 |
| 0B34 | 0829 | ---- | 8100 | 0 B 34 | (0B29 | ---- | 8100 |
| OC2D | OC5F | OCIF | 8200 | OC2D | 0 C 5 F | ()CIF | 8200 |
| 0D3D | 0D2B | ---- | 8300 | 0D3D | 0D2B | -- | 8300 |
| 0 O 88 | 0 E 08 | OE 7F | ----- | 0 E 08 | OE08 | OE7F | OEOO |
| 5200 | 5200 |  | ----- | 52 E 0 | 52E0 | 92E0 | A200 |
| 4700 | 4700 | 7700 | -...- | 47E0 | 47E0 | 77E0 | 9700 |
| 49005 | $4900{ }_{5}$ | 8400 |  | 49 EO 5 | 49 EO 5 | 84 E 0 | ${ }_{9900}{ }_{5}$ |
| Note ${ }^{5}$ | Note ${ }^{5}$ | ----- | Note ${ }^{5}$ | Note ${ }^{5}$ | Nute ${ }^{5}$ | ---- | Note ${ }^{5}$ mandart? |
| 352F | 352F | ----- | ----- | E02F | E02F | 9500 | A 400) 3 nenthandur. |
| 372A | 372A | ---- | ----- | 372A | 372A | 9600 | 3700 |
| 4A2D | 4A2D | --.-. | --.-- | 4A2D | 4A2D | 8 E 00 | 4 A 00 |

## KEYBOARD SCAN CODES

| Kcy | Kcy | Hardware Kybrd |  |
| :---: | :---: | :---: | :---: |
| \# | Descript. | Make | Break |
|  |  | Code | Code |
| 38 | Tab | OD | F00D |
| 39 | Q or q | 15 | F015 |
| 40 | W or w | 1D | F01D |
| 41 | E or c | 24 | F024 |
| 42 | R or ${ }^{\text {r }}$ | 2D | F02D |
| 43 | Tort | 2 C | F02C |
| 44 | Y or y | 35 | F035 |
| 45 | U or ${ }^{\text {u }}$ | 3 C | F03C |
| 46 | Iori | 43 | F043 |
| 47 | O oro | 44 | F044 |
| 48 | P or p | 4D | F04D |
| 49 | \{ or 1 | 54 | D054 |
| 50 | \} or ] | 5B | F05B |
| 51 | \| or 1 | 5D | F05D |
| 52 | Delete | E07! | E0F071 |
| 53 | End | E069 | E0F069 |
| 54 | Page Down | E07A | E0F07A |
| 55 | 7 or Home | 6 C | F06C |
| 56 | 8 | 75 | F075 |
| 57 | 9 or Page Up | 7D | F07D |


| Kybrd Interrupt |  | Standard ASCII |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Make | Break | (Scancode/ASCII code) |  |  |  |
| Code. | Code | Norm | Shift | Ctri | Alt |
| OF | 8 F | (0F90) | OFO) | ----- | ---- |
| 10 | 90 | 1071 | 1051 | 1011 | 1000 |
| 11 | 91 | 1177 | 1157 | 1117 | 1100 |
| 12 | 92 | 1265 | 1245 | 1205 | 1200 |
| 13 | 93 | 1372 | 1352 | 1312 | 1300 |
| 14 | 94 | 1474 | 1454 | 1414 | 1400 |
| 15 | 95 | 1579 | 1559 | 1519 | 1500 |
| 16 | 96 | 1675 | 1655 | 1615 | 1600 |
| 17 | 97 | 1769 | 1749 | 1709 | 1700 |
| 18 | 98 | 186F | 184F | 180F | 1800) |
| 19 | 99 | 1970 | 1950 | 1910 | 1900 |
| 1A | 9A | 1A5B | 1A7B | 1A1B | -.--- |
| 18 | 9 B | 1B5D | 1B7D | 1B1D |  |
| 2B | AB | 2B5C | 2B7C | 2B1C | ...-- |
| E02AE053 | E0D3E0AA | 5300 | 5300 | ----- | ----- |
| E02AE04F | EOCFE0AA | 4F00 | 4FO) | 7500 | $\cdots$ |
| E02AE051 | EODIE0AA | 5100 | 5100 | 7600 |  |
| 47 | C7 | 4700 | 4737 | 7700 | Note ${ }^{6}$ |
| 48 | C8 | 4800 | 4838 | ---- | Note ${ }^{6}$ |
| 49 | C9 | 4900 | 4939 | 8400 | Note ${ }^{6}$ |

## KEYBOARD SCAN CODES

| Key | Key | Hardw | Kybrd |
| :---: | :---: | :---: | :---: |
| \# | Descript. | Make | Break |
|  |  | Code | Code |
| 58 | + | 79 | F079 |
| 59 | Caps Lock | 58 | F058 |
| 60 | A or a | 1 C | Foic |
| 61 | S or s | 1 B | F01B |
| 62 | D ord | 23 | F023 |
| 63 | $F$ or $f$ | 2B | F02B |
| 64 | G or g | 34 | F034 |
| 65 | H or h | 33 | F033 |
| 66 | J or j | 3B | F03B |
| 67 | K or $k$ | 42 | F()42 |
| 68 | Lorl | 4B | F)4B |
| 69 | : or; | 4 C | F(04C |
| 70 | " or' | 52 | F052 |
| 71 | Enter | 5A | F05A |
| 72 | 4 | 6B | F06B |
| 73 | 5 | 73 | F073 |
| 74 | 6 | 74 | F074 |


| Kybrd Interrupt |  |
| :--- | :---: |
| Make | Break |
| Code | Code |
| $4 E$ | CE |
| 3A | BA |
| $1 E$ | $9 E$ |
| $1 F$ | $9 F$ |
| 20 | A0 |
| 21 | Al |
| 22 | A2 |
| 23 | A3 |
| 24 | A4 |
| 25 | A5 |
| 26 | A6 |
| 27 | A7 |
| 28 | A8 |
| IC | $9 C$ |
| $4 B$ | CB |
| $4 C$ | CC |
| $4 D$ | CD |

Standard ASCII
(Scancode/ASCII code)

| Norm | Shift | Ctrl | Alt | Norm | Shift | Cirl | Alt |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $4 E 2 B_{7}$ | $4 E 2 B_{7}$ | -...- | Note ${ }^{7}$ | $4 \mathrm{E} 2 \mathrm{~B}$ | $\begin{aligned} & \text { 4E2B } \\ & \text { Note } \end{aligned}$ | 9000 | $\begin{aligned} & 4 \mathrm{E} 0 \mathrm{O}_{7} \\ & \mathrm{Note} \end{aligned}$ |
| 1E61 | 1E41 | 1E01 | IE00 | 1E61 | \|E4| | 1E01 | IEOO |
| 1F73 | IF53 | IF13 | 1F00 | 1F73 | IF53 | 1F13 | IFO0) |
| 2064 | 2044 | 2004 | 2000 | 2064 | 2044 | 2004 | 2000 |
| 2166 | 2146 | 2106 | 2100 | 2166 | 2146 | 2106 | 2100 |
| 2267 | 2247 | 2207 | 2200 | 2267 | 2247 | 2207 | 2200) |
| 2368 | 2348 | 2308 | 2300 | 2368 | 2348 | 2308 | 2300 |
| 246A | 244A | 240A | 2400 | 246A | 244A | 240A | 2400) |
| 256B | 254B | 250B | 2500) | 25613 | 254B | 2508 | 2500 |
| 266C | 264C | 260C | 2600 | 266C | 264C | 260) | 2600 |
| 273B | 273A | -... | -.-- | 273B | 273A | ---- | 2700 |
| 2827 | 2822 | ---- | ---- | 2827 | 2822 | ---- | 2800 |
| 1C0D | 1C0D | IC0A | ---- | 1C0D | 1C0D | 1 COA | 1 COO |
| $4 B 00$ | 4B34 | 7300 | Note ${ }^{6}$ | 4 BOO | 4B34 | 7300 | Note ${ }^{6}$ |
|  | 4C35 | ... | Note ${ }^{6}$ | $4 \mathrm{C00}$ | 4 C 35 | $8 \mathrm{F00}$ | Note ${ }^{6}$ |
| 00 | 4D36 | 7400 | Note ${ }^{6}$ | 4D00 | 4 D 3 | 74 | Note ${ }^{6}$ |

## KEYBOARD SCAN CODES

| Key | Key | Hardware Kybrd |  |
| :---: | :---: | :---: | :---: |
| \# | Descript. | Make | Break |
|  |  | Code | Code |
| 76 | Left Shift | 12 | F012 |
| 77 | Z or z | IA | F01A |
| 78 | X or x | 22 | F022 |
| 79 | Cor c | 21 | F021 |
| 80 | V or v | 2A | F02A |
| 81 | B or b | 32 | F032 |
| 82 | N or n | 31 | F031 |
| 83 | M or m | 3A | F03A |
| 84 | < or, | 41 | F041 |
| 85 | > or. | 49 | F049 |
| 86 | ? or / | 4A | F04A |
| 87 | Right Shift | 59 | F059 |
| 88 | Up Arrow | E075 | E0F075 |
| 89 | 1 or End | 69 | F069 |
| 90 | 2 | 72 | F072 |
| 91 | 3 or Pg Dn | 7A | F07A |
| 92 | Left Ctrl | 14 | F014 |
| 93 | Left Alt | 11 | F011 |
| 94 | Space | 29 | F029 |


| Kybrd Interrupt |  |
| :--- | :--- |
| Make | Break |
| Code | Code |
| $2 A$ | AA |
| 2 C | AC |
| 2 D | AD |
| 2 E | AE |
| $2 F$ | AF |
| 30 | B0 |
| 31 | B1 |
| 32 | B2 |
| 33 | B3 |
| 34 | B4 |
| 35 | B5 |
| 36 | B6 |
| E02AE048 | E0C8E0AA |
| $4 F$ | CF |
| 50 | D0 |
| 51 | D1 |
| $1 D$ | $9 D$ |
| 38 | B8 |
| 39 | B9 |


| Standard ASCII (Scancode/ASCII code) |  |  |  | Extended ASCII (Scancode/ASCII code) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Norm | Shift | Ctrl | Alt | Norm | Shift | Cirl. | Alt |
| Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ |
| 2C7A | 2C5A | 2C1A | 2600 | 2C7A | 2C5A | 2C1A | 2 COO |
| 2D 78 | 2D58 | 2D18 | 2000 | 2D78 | 2D 58 | 2D 18 | 2000 |
| 2E63 | 2E43 | 2E03 | 2 EOO | 2E63 | 2E43 | 2E03 | 2E00 |
| 2F76 | 2F56 | 2F16 | 2 FOO | 2F76 | 2F56 | 2F16 | 2F00 |
| 3062 | 3042 | 3002 | 3000 | 3062 | 3042 | 3002 | 3000 |
| 316 E | 314E | 310 E | 3100 | 316 E | 314E | 310 E | 3100 |
| 326D | 324D | 320D | 3200 | 326D | 324D | 320D | 3200 |
| 332C | 333C | ----- | -.--- | 332C | 333C | ----- | 3300 |
| 342E | 343E | ----- | ----- | 342E | 343E | -..... | 3400 |
| $352 F_{8}$ | $353 \mathrm{~F}_{\mathrm{D}}$ |  |  | $352 \mathrm{~F}_{8}$ | $353 \mathrm{~F}_{8}$ |  |  |
| $\text { Note }{ }^{8}$ | $\text { Note }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ | $\text { Note }^{8}$ | $\text { Note }^{8}$ | Note ${ }^{8}$ | Note ${ }^{8}$ |
| 4800 | 4800 | ----- | - | 48E0 | 48E0 | 8DE0 | 9800 |
| 4F00 | 4F31 | 7500 | Note ${ }_{6}$ | 4F00 | 4F31 | 7500 | Note ${ }^{6}$ |
| 5000 | 5032 | -...- | Note ${ }_{6}$ | 5000 | 5032 | 9100 | Note ${ }_{6}$ |
| 51009 | ${ }_{5133} 9$ | ${ }_{7600}{ }_{9}$ | $\text { Note }_{9}^{6}$ |  | $5133_{9}$ | ${ }_{7600}^{9}$ | Note ${ }_{9}$ |
| Nute ${ }^{9}$ | Note ${ }^{9} 10$ | Note ${ }_{10}$ | Note | Note <br> 10 | Note ${ }^{9}$ | Note <br> 10 | Note ${ }_{10}$ |
| Note | Note ${ }^{\text {a }}$ | Note ${ }^{\text {a }}$ | Note | Note ${ }^{\text {a }}$ | Note ${ }^{10}$ | Note ${ }^{10}$ | Note ${ }^{10}$ |
| 3920 | 3920 | 3920 | 3920 | 3920 | 3920 | 3920 | 3920 |

## KEYBOARD SCAN CODES

| Key | Key | Hardw | Kybrd |
| :---: | :---: | :---: | :---: |
| \# | Descript. |  |  |
| 95 |  | E011 | E0F011 |
| 96 | Right Crrl | E014 | E0F014 |
| 97 | Left Arrow | E06B | E0F06B |
| 98 | Down Arrow | E072 | E0F072 |
| 99 | Right Arrow | E074 | E0F074 |
| 100 | 0 or Ins | 70 | F070 |
| 101 | . or Del | 71 | F071 |
| 102 | Enter | E05A | E0F05A |


| Kybrd Inte Make | rupt Break | Standard ASCII (Scancode/ASCII code) |  |  |  | Extended ASCII <br> (Scancode/ASCII code) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Code | Code | Norm | Shift | Cirl | Alt | Norm | Shift | Cirl |  |
| E038 | E0B8 | Note ${ }_{9}{ }^{10}$ | $\text { Note }{ }_{9}^{10}$ | $\operatorname{Note}_{9}^{10}$ | Notc ${ }_{9}{ }^{10}$ | $\text { Note }_{9}^{10}$ | Notc9 ${ }_{9}{ }^{10}$ | Note9 ${ }_{9}{ }^{10}$ | Note ${ }^{10}$ |
| EOID | E09D | Note | Note | Note | Note | Notc | Note | Note | Note |
| E02AE04B | eocbeoaa | 4B00 | $4 \mathrm{B00}$ | 7300 | ----- | 4BE0 | 4BE0 | 73 E 0 | $9 \mathrm{B00}$ |
| E02AE050 | E0DOE0AA | 5000 | 5000 | --- | ----- | 50 E 0 | 50E0 | 91 E 0 | A000) |
| E024E04D | EOCDEOAA | 4D00 | 4000 | 7400 | --.- 6 | 4DE0 | 4DE0 | 74 E 0 | 9 DOO |
| 52 | D2 | 5200 | 5230 | -... | Note ${ }^{6}$ | 5200 | 523H | 9200 | Note |
| 53 | D3 | 5300 | 532 E | ---- |  | 5300 | 532 E | 9300 |  |
| E01C | E09C | 1C0D | IC0D | 1C0A | .-. | EOOD | E00D | E00A | A600 |

## NOTES

Notel -INT O5H is invoked and a screen dump is performed
Note $2 \rightarrow$ he scroll lock active bit is toggled
Note3 —he pause state is initiated
Note4 -INT 1BH is invoked
Note5 —he numlock active bit is toggled
Note6 --ALT num pad generates raw ascii code of typed number
Note 7 -the caps lock active bit is toggled
Note8 - hold shift lock active until key is released
Note9 - hold control shift active until key is released
Note 10-hold alternate shift active until key is released


[^0]:    * CAN SUBSTITUTE LS FOR ALL HCT PARTS

[^1]:    $\qquad$

[^2]:    Alpha Modes
    $\mathrm{AL}=0: 40 \times 25$ black and white
    AL $=1: 40 \times 25$ color
    $\mathrm{AL}=2: 80 \times 25$ black and white
    $\mathrm{AL}=3: 80 \times 25$ color
    Graphics Modes
    $\mathrm{AL}=4: 320 \times 200$ color graphics
    $\mathrm{AL}=5: 320 \times 200$ black and white graphics with 4 shades
    $\mathrm{AL}=6: 640 \times 200$ black and white graphics
    with 2 shades
    $\mathrm{AL}=7:$ monochrome text
    Additional Modes
    $A L=8: 160 \times 200$ color graphics with 16 colors
    $\mathrm{AL}=9: 320 \times 200$ color graphics with 16 colors
    $\mathrm{AL}=\mathrm{A}: 640 \times 200$ color graphics
    with 4 colors

